

2A Ultra Low Dropout Linear Regulator

General Description

The uP8809 is an ultra low dropout linear regulator specifically designed to provide termination voltage for DDR memory system. Designed with low on-resistance NMOSFETs, this device is capable of sinking/sourcing up to 2A output current.

This device works with dual supplies, a control input for the control circuitry and a power input as low as 1.0V for providing current to output. The output voltage is tightly regulated to track reference voltage input within 20mV variation with fast to line/load transient.

Other features include chip shutdown function, soft-start, on-chip thermal protection, and bi-directional current limit protection. The uP8809 is available in PSOP-8L, FSOP-8L, WDFN3x3-8L packages with very low thermal resistance.

Applications

- Desktop PCs, Notebooks, and Workstations
- □ Graphic Cards
- Low Voltage Logic Supplies
- Microprocessor and Chipset Supplies
- Split Plane Microprocessor Supplies
- Advanced Graphics Cards Supplies
- SoundCards and Auxiliary Power Supplies
- SMPS Post Regulators

Features

- Generate Termination Voltage for DDR Memory Interface
 - Output Voltage Traces Reference Input
 - Capable of Sinking/Sourcing 2A
 - Integrated Low R_{DS(ON)} MOSFETs
 - Excellent Line and Load Regulation
- Low External Part Count
- Bidirectional Current Limit Protection
- On-Chip Thermal Protection
- Support Pure MLCC
- RoHS Compliant and Halogen Free

Ordering Information

Order Number	Package Type	Top Marking
uP8809PSW8	PSOP-8L	uP8809P
uP8809PSF8	FSOP-8L	uP8809P
uP8809PDD8	WDFN3x3-8L	uP8809P

Status:

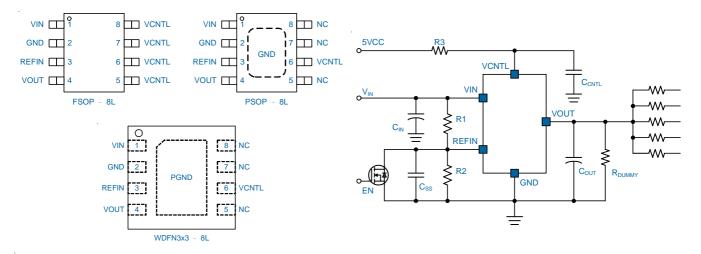
In Production: uP8809PSW8

Others: Please check the sample/production availability with uPI representatives.

Note:

- (1) Please check the sample/production availability with uPI representatives.
- (2) uPI products are compatible with the current IPC/JEDEC J-STD-020 requirement. They are halogen-free, RoHS compliant and 100% matte tin (Sn) plating that are suitable for use in SnPb or Pb-free soldering processes.

Pin Configuration & Typical Application Circuit

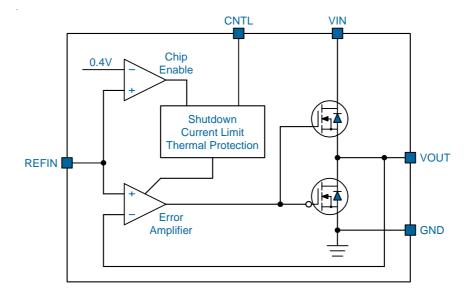




Functional Pin Description

No.	Name	Pin Function
1	VIN	Input Voltage. This is the drain input to the power device that supplies current to the output pin. Large bulk capacitors with low ESR should be placed physically close to this pin to prevent the input rail from dropping during large load transient. A 10uF ceramic capacitor is recommended at this pin. $V_{\mathbb{N}}$ cannot be forced higher than $V_{\mathbb{CNTL}}$ otherwise the current limit function may be false triggered and disable the output voltage.
2	GND	Ground.
3	REFIN	Reference Voltage Input. This pin is the non-inverting input of the error amplifier. The output voltage is regulated to track the reference voltage input. This pin is also monitored by the shutdown comparator. Pulling this pin lower than 0.15V shuts down this device.
4	VOUT	Output Voltage. This pin is power output of the device. A pull low resistance exists when the device is disabled by pulling low the EN pin. To maintain adequate transient response to large load change, typical value of 1000uF AI electrolytic capacitor with 10uF ceramic capacitors are recommended to reduce the effects of current transients on VOUT.
5, 7, 8	NC	Not Internally Connected.
6	VCNTL	Supply Input for Control Circuit. This pin provides bias voltage to the control circuitry and driver for the pass transistor. The driving capability of output current is proportioned to the V_{VCNTL} . For the device to regulate, the voltage on this pin must be at least 2.0V greater than the output voltage, and no less than $V_{CNTL_MIN.}$ V_{CNTL} input voltage must be ready before V_{IN} input voltage.
Exposed Pad	GND	Ground. The exposed pad acts the dominant power dissipation path and should be soldered to well design PCB pads as described in the <i>Application Informations Chapter</i> .

Functional Block Diagram





Functional Description

Definitions

Some important terminologies for LDO are specified below.

Dropout Voltage

The input/output voltage differential at which the regulator output no longer maintains regulation against further reductions in input voltage. Measured when the output drops 2% below its nominal value. Dropout voltage is affected by junction temperature, load current and minimum input supply requirements.

Line Regulation

The change in output voltage for a change in input voltage. The measurement is made under the conditions of low dissipation or by using pulse techniques such that average chip temperature is not significantly affected.

Load Regulation

The change in output voltage for a change in load current at constant chip temperature. The measurement is made under the conditions of low dissipation or by using pulse techniques such that average chip temperature is not significantly affected.

Maximum Power Dissipation

The maximum total device dissipation for which the regulator will operate within specifications.

Quiescent Bias Current

Current which is used to operate the regulator chip and is not delivered to the load.

The quiescent current I_Q is defined as the supply current used by the regulator itself that does not pass into the load. It typically includes all bias currents required by the LDO and any drive current for the pass transistor.

General

The uP8809 is an ultra low dropout linear regulator specifically designed to provide termination voltage for DDR memory system. Designed with low on-resistance NMOSFETs, this device is capable of sinking/sourcing up to 2A output current.

This device works with dual supplies, a control input for the control circuitry and a power input as low as 1.0V for providing current to output. The output voltage is tightly regulated to track reference voltage input within 20mV variation with fast to line/load transient.

Other features include chip shutdown function, soft-start, on-chip thermal protection, and bi-directional current limit protection.

Power On Reset

The uP8809 mainly consists of power on reset and chip enable, pass transistors, current limit, error amplifier and

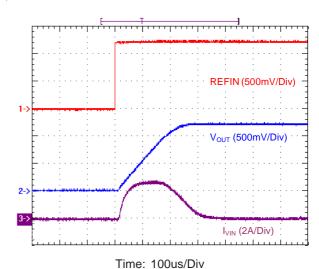
temperature protection as shown in *Functional Block Diagram*. The uP8809 continuously monitors control input and power input for power on reset (POR) to ensure the device can work properly. The typical POR rising levels are 2.7V and 0.6V for control input and power input respectively.

Chip Enable and Soft Start

Once POR is granted, the uP8809 is ready for normal operation. The REFIN pin is a dual-function input pin: reference input and shutdown control input. A singal level transistor is adequate to pull this pin lower than 0.15V and shuts down the device reducing the shutdown current below 50uA.

The uP8809 tightly regulates the output voltage to track V_{REFIN} if it is higher than 0.4V. The output voltage ramp up/down speed is limited as 5.7mV/us to limit inrusht current form power input as shown in Figure 1. The inrush current to charge/discharge the output capacitor is calculated as:

$$I_{INRUSH} = 5.7 \text{mV/us x C}_{OUT}$$



 $V_{IN} = 2.5V, V_{CNTL} = 5V, C_{OUT} = 470uF$

Figure 1. Chip Enable and Soft Start

A 100uF output capacitor will demand 0.5V input current during softstart. If the output capacitor is larger than 470uF, current limit function may be activated and limit the inrush current to about 2.2A. Make sure the power input is capable of delivering inrush current with selected output capacitor.

Output Voltage Regulation

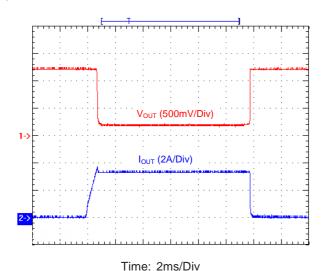
The output voltage is tightly regulated to track the reference voltage applied at REFIN pin. The error signal is amplified to control the gates of NMOSFET for sourcing current from VIN and singing current GND respectively. Since the gate voltage is provided by the control input V_{CNTL} , it is highly recommended the control input is 2V higher than the output voltage to achieve tight regulation and fast transient response.



Functional Description

Current Limit

The uP8809 monitors sourcing and sinking ouput currents. The output currents are limited to a safe level by reducing the gate voltages of the NMOSFET during output overload or short circuit. The output voltage is reduced if the load continuously demands current higher than the current limit level. The output voltage is re-built up when overload or short circuit conditions are removed.



 $V_{IN} = 2.5V, V_{CNTL} = 5V, C_{OUT} = 1.25V$

Figure 1. VOUT Output Source Current Limit

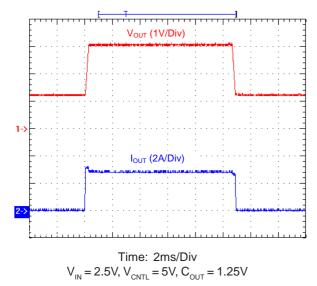


Figure 2. VOUT Output Sink Current Limit

Thermal Protection

The uP8809 implements a thermal shutdown function to protect the device from damage when output overload or short circuit. Both NMOSFETs are turned off when the juction temperature exceeds about 170°C, allowing the junction temperature to cool down. The output voltage is re-built up when the junction temperature reduces by 40°C, resulting in a pulsed output during continuous thermal over load conditions.

The thermal protection is designed to protect the device from damage during abnormal operation. It is highly recommended to keep the maximum junction temperature under 150°C during normal operation for maximum realiability.



	Absolute Maximum Rating
(Note 1)	_
	0.3V to +7V
	0.3V to +7V
Other Pins	0.3V to (V _{CNTL} + 0.3V)
Storage Temperature Range	
Junction Temperature	150°C
Lead Temperature (Soldering, 10 sec)	260°C
ESD Rating (Note 2)	
HBM (Human Body Mode)	2kV
MM (Machine Mode)	200V
<u> </u>	Thermal Information
	Thomas in ormation
Package Thermal Resistance (Note 3)	
	50°C/W
	5°C/W
	90°C/W
	35°C/W
	60°C/W
	5°C/W
Power Dissipation, $P_D @ T_A = 25^{\circ}C$	
	2.0W
FSOP-8L	1.11W
WDFN3x3-8L	1.67W
	Recommended Operation Conditions
(Note 2)	•
Operating Junction Temperature Range	
Operating Ambient Temperature Range	
	+2.8V to +6.0V
	+1.0V to V _{CNTL}
	m Ratings" may cause permanent damage to the device

- **Note 1.** Stresses listed as the above "Absolute Maximum Ratings" may cause permanent damage to the device. These are for stress ratings. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may remain possibility to affect device reliability.
- **Note 2.** Devices are ESD sensitive. Handling precaution recommended.
- **Note 3.** θ_{JA} is measured in the natural convection at $T_A = 25^{\circ}\text{C}$ on a low effective thermal conductivity test board of JEDEC 51-3 thermal measurement standard.
- Note 4. The device is not guaranteed to function outside its operating conditions.



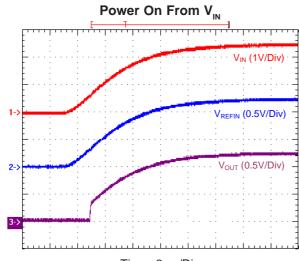
Electrical Characteristics

 $(V_{CNTL} = 5V, T_A = 25^{\circ}C, unless otherwise specified)$

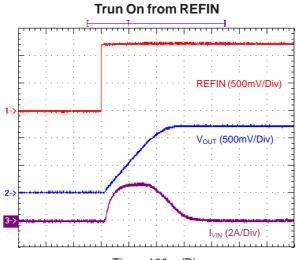
Parameter	Symbol	Test Conditions	Min	Тур	Max	Units
Supply Input Voltage						
POR Threshold	V _{CNTLRTH}	V _{CNTL} Rising		2.7	2.8	V
POR Hysteresis	V _{CNTLHYS}	V _{CNTL} Falling		0.2		V
Control Input Current in Shutdown	I CNTL_SD	$V_{CNTL} = V_{IN} = 5.0V, I_{OUT} = 0A, V_{REFIN} = 0V$	-	50	90	uA
Control Input Current	I _{CNTL}	$V_{CNTL} = 5.0V, V_{IN} = 1.8V, V_{REFIN} = 0.9V, $ $I_{OUT} = 0A.$		0.5	1.5	mA
Output Voltage						
Output Voltage Offset	V _{os}	$V_{CNTL} = 5.0V, V_{IN} = 1.8V, V_{REFIN} = 0.9V, $ $I_{OUT} = 0A.$	-20		+20	mV
Load Regulation	ΔV_{LOAD}	-1A < I _{OUT} < 1A	-20		+20	mV
Dropout Voltage		V _{CNTL} = 4.5V, I _{OUT} = 1.5A, VOUT = 1.25V		225	525	mV
	V _{DROPOUT}	V _{CNTL} = 4.5V, I _{OUT} = 1.0A, VOUT = 1.25V		150	350	mV
Protection	Protection					
Current Limit	 LIMIT_SOURCE	$V_{IN} = 1.8V$, $V_{OUT} = 0.9V$, Sourcing	2.3			А
	I _{LIMIT_SINK}	$V_{IN} = 1.8V$, $V_{OUT} = 0.9V$, Sinking	2			
Short Circuit	I _{SC_VIN}	VOUT short circuit to VIN, Sinking	1			Α
Short Circuit	I _{SC_GND}	VOUT short circuit to GND, Sourcing	1.5			Α
REFIN Enable/Shutdown						
REFIN Logic High Threshold	V _{ENH}	REFIN rising to enable the device	0.4			V
REFIN Logic Low Threshold	V _{ENL}	REFIN falling to disable the device			0.15	V
Thermal Protection						
Thermal Shutdown Temperature	T _{SD}			160		οС
Thermal Shutdown Hysteresis	T _{SDHYS}			30		οС



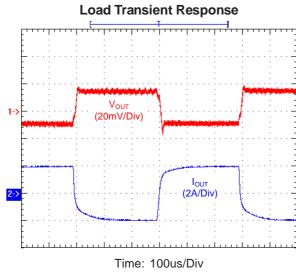
Typical Operation Characteristics



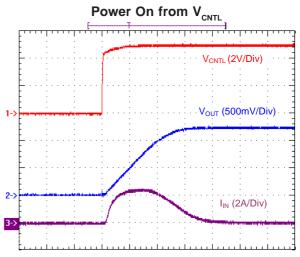
 $\label{eq:continuous} Time: 2ms/Div \\ V_{CNTL} = 5V, \, V_{REFIN} = 1.25V, \, C_{OUT} = 470uF$



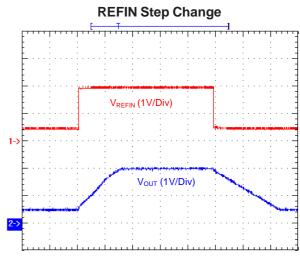
Time: 100us/Div $V_{IN} = 2.5V$, $V_{CNTL} = 5V$, $C_{OUT} = 470uF$

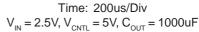


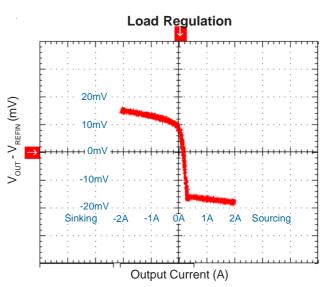
Time: 100us/Div $V_{IN} = 2.5V, V_{CNTL} = 5V, C_{OUT} = 1000uF$



Time: 100us/Div $V_{IN} = 2.5V, V_{REFIN} = 1.25V, C_{OUT} = 470uF$

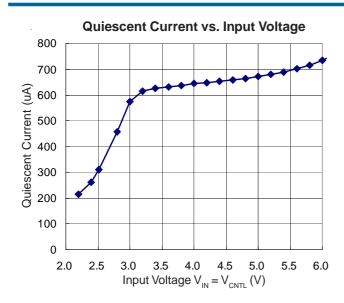


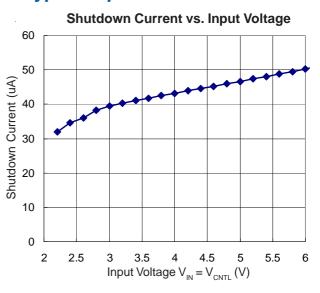


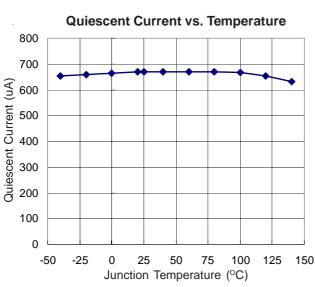


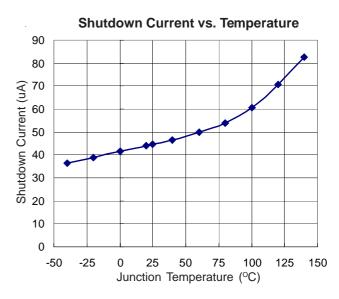


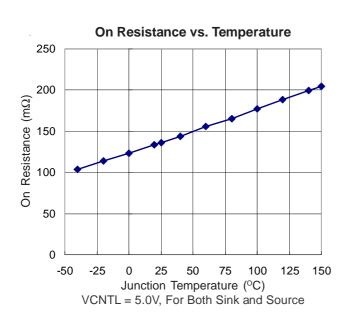
Typical Operation Characteristics













Application Information

The uP8809 is an ultra low dropout linear regulator specifically designed to provide termination voltage for DDR memory system. Designed with low on-resistance NMOSFETs, this device is capable of sinking/sourcing up to 2A output current. The output voltage is tightly regulated to track reference voltage input within 20mV variation with fast to line/load transient.

Supply Voltage for Control Circuit \mathbf{V}_{CNTL}

This uP8809 works with dual supplies, a control input for the control circuitry and a power input as low as 1.0V for providing current to output.

The control input provides bias current for control circuit and gate voltage for turning on and off the NMOSFETs. It is highly recommended to keep the control input 2.0V higher than the output voltage for optimal performance. Connect VCNTL pin to a 5V voltage source when available. The control voltage should be locally bypassed by a minimum 1uF ceramic capacitor plus a 10Ω resistor.

Power Input V_{IN} and Input Capacitor C_{CNTL} , C_{IN}

The VIN pin supplies current to output when the upper MOSFET turns on. The uP8809 is designed to work with minimum 10uF ceramic input capacitor. However, a bulk capacitor is still recommended in parallel with the ceramic capacitor to stabalize the input voltage during output soft start and load transient.

Since both power and control inputs are independently monitored for power on reset, special power sequence concern is not required for control input and power input.

When work with large output capacitor, the uP8809 may demand large input current during soft start. Make sure the power input is capable of delevering up to 3A.

Component	Recommended Value
	Minimum 1uF/10uF Ceramic Capacitor

Note: A low-ESR 1uF/10uF capacitor with minimal susceptibility to temperature is recommended, and stability is highly dependent on the input power supply characteristics and the substrate wiring pattern. The $\rm C_{IN}$ total capacitance of input capacitors value including ceramic and aluminum electrolytic capacitors should be larger than 10uF. A 1uF ceramic capacitor ($\rm C_{CNTL}$) or higher is recommended for noise decoupling.

Reference Input

The output voltage is regulated to track the reference input at REFIN pin. The reference input can be obtained from power input by voltage divider or from an independent voltage reference. A ceramic capacitor physically near the IC is required to filter the reference voltage.

Output Voltage and Output Capacitor, Cout

The uP8809 is designed to work with low ESR ceramic capacitors. No special concern is required on ESR for stable operation. A minimum bulk capacitance of 10uF, along with a 0.1uF ceramic decoupling capacitor is recommended. Increasing the bulk capacitance will improve the overall transient response. The use of multiple lower value ceramic capacitors in parallel to achieve the desired bulk capacitance will not cause stability issues. Although designed for use with ceramic output capacitors, the uP8809 is extremely tolerant of output capacitor ESR values and thus will also work comfortably with tantalum output capacitors.

Component	Recommended Value
C _{OUT}	Minimum 10uF Ceramic Capacitor

Note: A 10uF ceramic capacitor is recommended, and actual stability is highly dependent on temperature and load conditions. Insufficient capacitance may cause oscillation, while high equivalent series resistance (ESR) exacerbates output voltage fluctuation under rapid load change conditions. Total output capacitors value including ceramic and aluminum electrolytic capacitors should be larger than 10uF.

Thermal Consideration

The uP8809 integrates internal thermal limiting function to protect the device from damage during fault conditions. However, continuously keeping the junction near the thermal shutdown temperature may remain possibility to affect device reliability. It is highly recommended to keep the junction temperature below the recommended operation condition 125°C for maximum reliability.

Power dissipation in the device is calculated as:

$$\mathsf{P}_{\scriptscriptstyle \mathsf{D}} = (\mathsf{V}_{\scriptscriptstyle \mathsf{IN}} - \mathsf{V}_{\scriptscriptstyle \mathsf{OUT}}) \; \mathsf{X} \; \mathsf{I}_{\scriptscriptstyle \mathsf{OUT}} + \mathsf{V}_{\scriptscriptstyle \mathsf{CNTL}} \; \mathsf{X} \; \mathsf{I}_{\scriptscriptstyle \mathsf{CNTL}}$$

It is adequate to neglect power loss with respective to control circuit V_{CNTL} x I_{CNTL} when considering thermal management in uP8809. Take the following moderate operation condition as an example: $V_{IN} = 2.5 \text{V}$, $V_{OUT} = 1.5 \text{V}$, $I_{OUT} = 1 \text{A}$, the power dissipation is:

$$P_D = (2.5V-1.5V) \times 1A = 1.0W$$

This power dissipation is conducted through the package into the ambient environment, and, in the process, the temperature of the die (T_J) rises above ambient. Large power dissipation may cause considerable temperature raise in the regulator in large dropout applications. The geometry of the package and of the printed circuit board (PCB) greatly influence how quickly the heat is transferred to the PCB and away from the chip. The most commonly used thermal metrics for IC packages are thermal resistance from the chip junction to the ambient air surrounding the package (θ_{ID}):

$$\theta_{JA} = (T_J - T_A) / P_D$$



Application Information

 θ_{JA} specified in the *Thermal Information* section is measured in the natural convection at $T_A = 25^{\circ}C$ on a high effective thermal conductivity test board (4 Layers, 2S2P) of JEDEC 51-7 thermal measurement standard. The case point of θ_{JC} is on the exposed pad for PSOP-8 package.

Given power dissipation P_D , ambient temperature and thermal resistance θ_{JA} , the junction temperature is calculated as:

$$\mathsf{T}_{\mathsf{J}} = \mathsf{T}_{\mathsf{A}} + \Delta \mathsf{T}_{\mathsf{J}\mathsf{A}} = \mathsf{T}_{\mathsf{A}} + \mathsf{P}_{\mathsf{D}} \times \theta_{\mathsf{J}\mathsf{A}}$$

To limit the junction temperature within its maximum rating, the allowable maximum power dissipation is calculated as:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where T $_{J(MAX)}$ is the maximum operation junction temperature 125°C, T $_{A}$ is the ambient temperature and the θ_{JA} is the junction to ambient thermal resistance. θ_{JA} of PSOP-8 packages is 75°C/W on JEDEC 51-7 (4 layers, 2S2P) thermal test board with minimum copper area. The maximum power dissipation at T $_{A}$ = 25°C can be calculated as:

$$P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / 75^{\circ}C/W = 1.33W$$

The thermal resistance θ_{JA} highly depends on the PCB design. Copper plane under the exposed pad is an effective heatsink and is useful for improving thermal conductivity. Figure 3 shows the relationship between thermal resistance θ_{JA} vs. copper area on a standard JEDEC 51-7 (4 layers, 2S2P) thermal test board at $T_A = 25^{\circ}$ C. A 50mm² copper plane reduces θ_{JA} from 75°C/W to 50°C/W and increases maximum power dissipation from 1.33W to 2W.

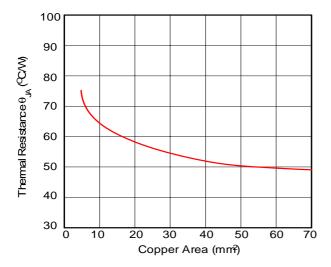


Figure 3. Thermal Resistance θ_{JA} vs. Copper Area

Figure 4 illustrated the recommended PCB layout for best thermal performance.

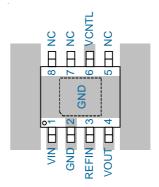


Figure 4. Recommended PCB Layout.

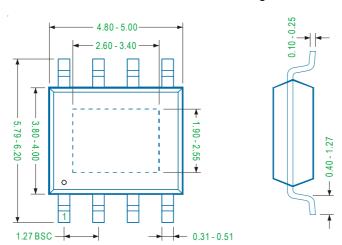
Layout Consideration

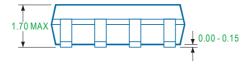
- 1. Place a local bypass capacitor as closed as possible to the VIN pin. Use short and wide traces to minimize parasitic resistance and inductance.
- 2. The exposed pad should be soldered on GND plane with maximum area and with multiple vias to inner layer of ground place for improved thermal performance.
- Connect voltage divider directly to the point where regulation is required. Place voltage divider close to the device.



Package Information

PSOP-8 Package





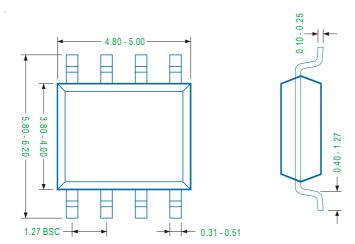
Note

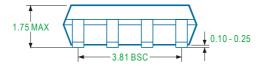
- 1. Package Outline Unit Description:
 - BSC: Basic. Represents theoretical exact dimension or dimension target
 - MIN: Minimum dimension specified.
 - MAX: Maximum dimension specified.
 - REF: Reference. Represents dimension for reference use only. This value is not a device specification.
 - TYP. Typical. Provided as a general value. This value is not a device specification.
- 2. Dimensions in Millimeters.
- 3. Drawing not to scale.
- 4. These dimensions do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm.



. Package Information

FSOP-8 Package





Note

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MAX: Maximum dimension specified.

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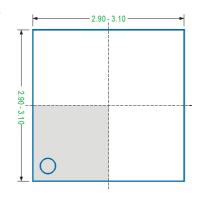
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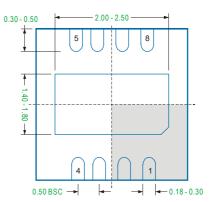
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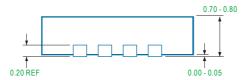


Package Information

WDFN3x3-8L







Note

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BSC: Basic. Represents theoretical exact dimension or dimension target

MIN: Minimum dimension specified.

MAX: Maximum dimension specified.

REF: Reference. Represents dimension for reference use only. This value is not a device specification.

TYP. Typical. Provided as a general value. This value is not a device specification.

- 2. Dimensions in Millimeters.
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- 4. These dimensions do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm.



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