













Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, TOLL, 750V, 5.4 mohm

Rev. B, January 2025

Description

The UJ4SC075005L8S is a 750V, $5.4m\Omega$ G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving MO-229 package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

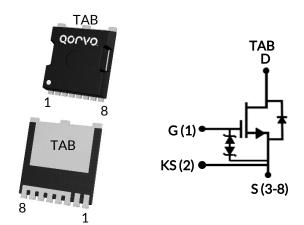
- On-resistance R_{DS(on)}: 5.4mΩ (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q_{rr} = 440nC
- Low body diode V_{FSD}: 1.03V
- ◆ Low gate charge: Q_G = 164nC
- Threshold voltage V_{G(th)}: 4.7V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected, HBM class 2
- MO-229 package for faster switching, clean gate waveforms

Typical applications

- Solid state relays and circuit-breakers
- Line rectification and active-bridge rectification circuits in AC/DC front-ends
- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating



UJ4SC075005L8S



| Part Number | Package | Marking |
|----------------|---------|-------------|
| UJ4SC075005L8S | MO-229 | UJ4SC075005 |





















Maximum Ratings

| Parameter | Symbol | Test Conditions | Value | Units |
|---|------------------|--|------------|-------|
| Drain-source voltage | V_{DS} | | 750 | V |
| Cata course voltage | V_{GS} | DC | -20 to +20 | V |
| Gate-source voltage | V GS | AC (f > 1Hz) | -25 to +25 | V |
| Continuous drain current ¹ | I _D | T _C < 144°C | 120 | Α |
| Pulsed drain current ² | I _{DM} | T _C = 25°C | 588 | Α |
| Single pulsed avalanche energy ³ | E _{AS} | $L=15mH$, $I_{AS}=6.5A$ | 316 | mJ |
| Short circuit withstand time ⁴ | t_SC | $V_{DS} = 400V, T_{J(START)} = 175^{\circ}C$ | 5 | μs |
| SiC FET dv/dt ruggedness | dv/dt | $V_{DS} \le 500V$ | 100 | V/ns |
| Power dissipation | P _{tot} | T _C = 25°C | 1153 | W |
| Maximum junction temperature | $T_{J,max}$ | | 175 | °C |
| Operating and storage temperature | T_J,T_STG | | -55 to 175 | °C |
| Reflow soldering temperature | T_{solder} | reflow MSL 1 | 260 | °C |

- 1. Limited by bondwires
- 2. Pulse width t_p limited by $T_{J,max}$
- 3. Starting $T_J = 25^{\circ}C$
- 4. Short circuit current is independent of the gate voltage $V_{\text{GS}} > 12V$

Thermal Characteristics

| Davameter | ameter Symbol Te | | Value | | | Linita |
|--------------------------------------|------------------------|-----------------|-------|------|------|--------|
| Parameter | Symbol Test Conditions | rest Conditions | Min | Тур | Max | Units |
| Thermal resistance, junction-to-case | $R_{\theta JC}$ | | | 0.10 | 0.13 | °C/W |













Electrical Characteristics (T_J = +25°C unless otherwise specified)

Typical Performance - Static

| Parameter | Cumbal | Test Conditions | | Value | | Units |
|--------------------------------|---------------------|--|-----|-------|-----|--------|
| Parameter | Symbol | rest Conditions | Min | Тур | Max | Offics |
| Drain-source breakdown voltage | BV_DS | V_{GS} =0V, I_D =1mA | 750 | | | V |
| Tabal dayla balana ayana t | | V _{DS} =750V, V _{GS} =0V, T _J =25°C | | 6 | 130 | Δ. |
| Total drain leakage current | I _{DSS} | V _{DS} =750V, V _{GS} =0V, T _J =175°C | | 45 | | μΑ |
| Total gate leakage current | I _{GSS} | V _{DS} =0V, T _J =25°C, V _{GS} =-20V / +20V | | 6 | 20 | μА |
| | | V_{GS} =12V, I_D =80A, T_J =25°C | | 5.4 | 7.2 | |
| Drain-source on-resistance | R _{DS(on)} | V _{GS} =12V, I _D =80A, T _J =125°C | | 9.3 | | mΩ |
| | | V _{GS} =12V, I _D =80A, T _J =175°C | | 12.2 | | |
| Gate threshold voltage | V _{G(th)} | V_{DS} =5V, I_{D} =10mA | 4 | 4.7 | 6 | V |
| Gate resistance | R_{G} | f=1MHz, open drain | | 0.8 | 1.5 | Ω |

Typical Performance - Reverse Diode

| Parameter | Cymahal | Test Conditions | Value | | | 11.20 |
|---|----------------------|--|-------|------|------|-------|
| | Symbol | rest Conditions | Min | Тур | Max | Units |
| Diode continuous forward current ¹ | I _S | T _C < 144°C | | | 120 | Α |
| Diode pulse current ² | I _{S,pulse} | T _C =25°C | | | 588 | Α |
| Forward voltage | V _{FSD} | V _{GS} =0V, I _S =50A, T _J =25°C | | 1.03 | 1.16 | |
| | | V _{GS} =0V, I _S =50A, T _J =175°C | | 1.06 | | V |
| Reverse recovery charge | Q_{rr} | V_{DS} =400V, I_{S} =80A, V_{GS} =0V, R_{G} =20 Ω | | 440 | | nC |
| Reverse recovery time | t _{rr} | di/dt=2800A/μs, Τ _J =25°C | | 31 | | ns |
| Reverse recovery charge | Q _{rr} | V_{DS} =400V, I_{S} =80A, V_{GS} =0V, R_{G} =20 Ω | | 525 | | nC |
| Reverse recovery time | t _{rr} | di/dt=2800A/μs, Τ _J =150°C | | 37 | | ns |















Typical Performance - Dynamic

| D t | 6 | Total Constitution | | Value | | 11.20 |
|---|----------------------|--|-----|-------|-----|-------|
| Parameter | Symbol | Test Conditions – | Min | Тур | Max | Units |
| Input capacitance | C _{iss} | \/ -400\/ \/ -0\/ | | 8374 | | |
| Output capacitance | C _{oss} | V _{DS} =400V, V _{GS} =0V f=100kHz | | 362 | | pF |
| Reverse transfer capacitance | C _{rss} | | | 4 | | |
| Effective output capacitance, energy related | C _{oss(er)} | V _{DS} =0V to 400V, V _{GS} =0V | | 475 | | pF |
| Effective output capacitance, time related | C _{oss(tr)} | V _{DS} =0V to 400V, V _{GS} =0V | | 950 | | pF |
| C _{OSS} stored energy | E _{oss} | V _{DS} =400V, V _{GS} =0V | | 38 | | μJ |
| Total gate charge | Q_G | V _{DS} =400V, I _D =80A, | | 164 | | |
| Gate-drain charge | Q_{GD} | $V_{GS} = 0V \text{ to } 15V$ | | 24 | | nC |
| Gate-source charge | Q_{GS} | V _{GS} – 0 V to 13 V | | 46 | | |
| Turn-on delay time | t _{d(on)} | Notes 5 and 6, V_{DS} =400V, I_{D} =80A, Gate Driver =0V to +15V, Turn-on $R_{G,EXT}$ =1.5 Ω , | | 35 | | |
| Rise time | t _r | | | 39 | | ns |
| Turn-off delay time | t _{d(off)} | | | 109 | | |
| Fall time | t _f | | | 13 | | |
| Turn-on energy including R_S energy | E _{ON} | Turn-off $R_{G,EXT}$ =5 Ω , | | 766 | | μJ |
| Turn-off energy including R_S energy | E _{OFF} | inductive Load, FWD: | | 162 | | |
| Total switching energy | E _{TOTAL} | same device with $V_{GS} = 0V$ and $R_G = 5\Omega$, RC snubber: | | 928 | | |
| Snubber R_S energy during turn-on | E _{RS_ON} | $R_S=5\Omega$ and $C_S=680pF$, | | 17.6 | | |
| Snubber R _S energy during turn-off | E _{RS_OFF} | T _J =25°C | | 7.2 | | |
| Turn-on delay time | t _{d(on)} | | | 37 | | |
| Rise time | t _r | Notes 5 and 6, V _{DS} =400V, I _D =80A, Gate | | 41 | | |
| Turn-off delay time | t _{d(off)} | Driver = $0V$ to +15V, | | 114 | | - ns |
| Fall time | t _f | Turn-on $R_{G,EXT}$ =1.5 Ω , | | 13 | | |
| Turn-on energy including R _S energy | E _{ON} | Turn-off $R_{G,EXT}$ =5 Ω , inductive Load, FWD: same device with V_{GS} = 0V and R_{G} = 5 Ω , RC snubber: R_{S} =5 Ω and C_{S} =680pF, T_{J} =150°C | | 808 | | |
| Turn-off energy including R _S energy | E _{OFF} | | | 187 | | |
| Total switching energy | E _{TOTAL} | | | 995 | | μЈ |
| Snubber R _S energy during turn-on | E _{RS_ON} | | | 18.3 | | 1 |
| Snubber R _S energy during turn-off | E _{RS_OFF} | | | 10.3 | | |

^{5.} Measured with the switching test circuit in Figure 26.

^{6.} In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.







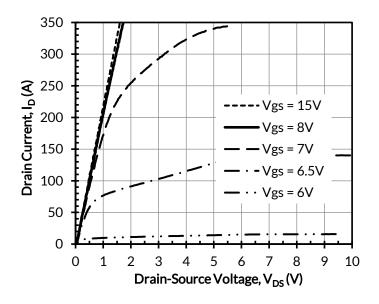


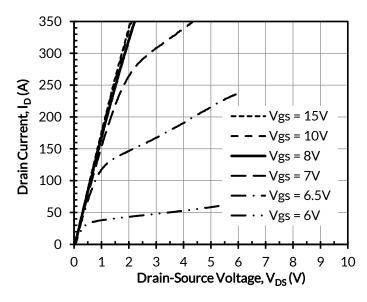






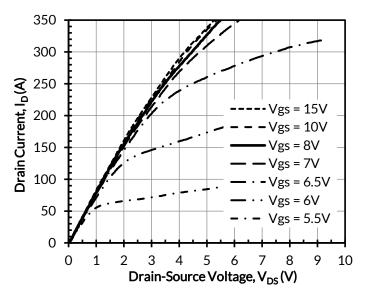
Typical Performance Diagrams





< 250µs

Figure 1. Typical output characteristics at $T_J = -55$ °C, tp Figure 2. Typical output characteristics at $T_J = 25$ °C, tp < 250µs



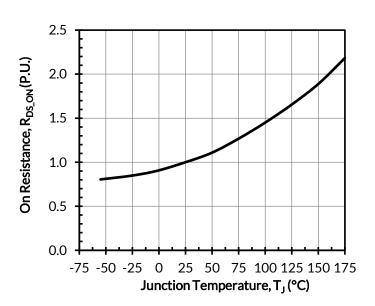


Figure 3. Typical output characteristics at $T_J = 175$ °C, tp < 250µs

Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_D = 80A





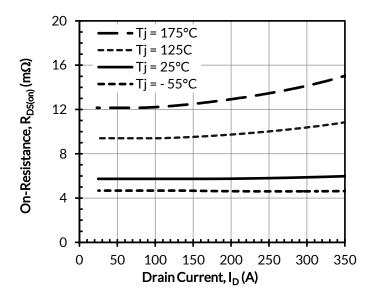








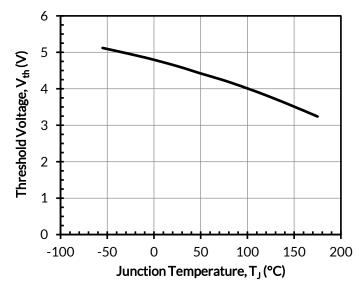




300 Tj = -55°C 250 Tj = 25°C Tj = 175°C Drain Current, I_D (A) 200 150 100 50 0 5 6 8 2 3 0 1 Gate-Source Voltage, V_{GS} (V)

Figure 5. Typical drain-source on-resistances at V_{GS} = 12V

Figure 6. Typical transfer characteristics at $V_{DS} = 5V$



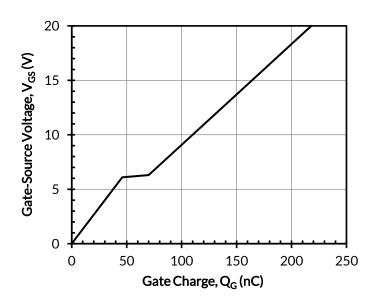


Figure 7. Threshold voltage vs. junction temperature at V_{DS} = 5V and I_{D} = 10mA

Figure 8. Typical gate charge at V_{DS} = 400V and I_D = 80A













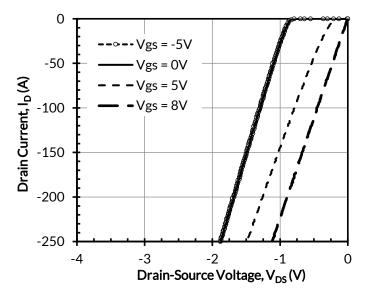
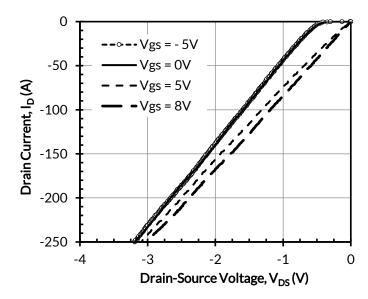


Figure 9. 3rd quadrant characteristics at $T_J = -55$ °C

Figure 10. 3rd quadrant characteristics at $T_J = 25^{\circ}C$



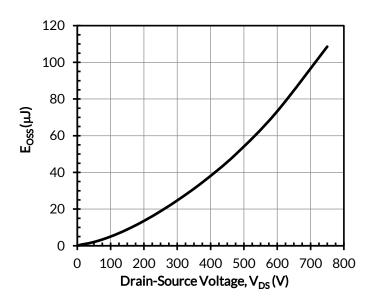


Figure 11. 3rd quadrant characteristics at $T_J = 175$ °C

Figure 12. Typical stored energy in C_{OSS} at V_{GS} = 0V



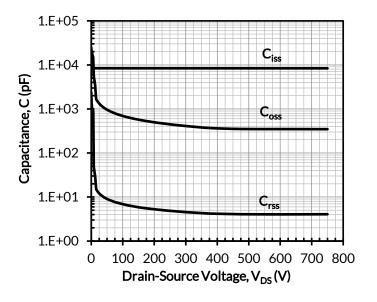












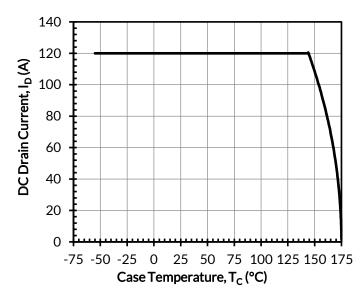


Figure 13. Typical capacitances at f = 100kHz and $V_{GS} = Figure 14$. DC drain current derating 0V

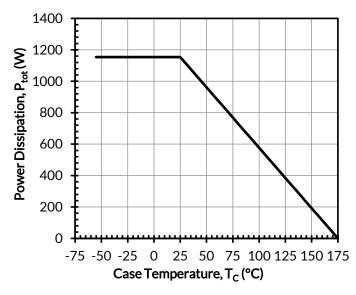


Figure 15. Total power dissipation

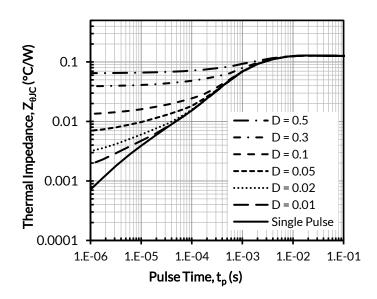


Figure 16. Maximum transient thermal impedance



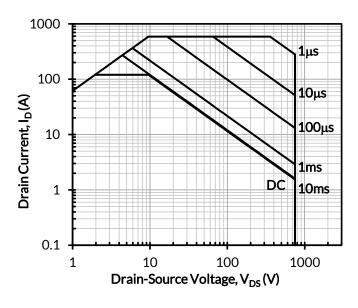








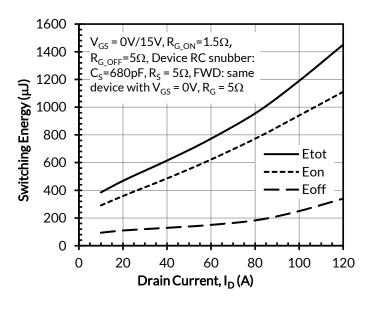


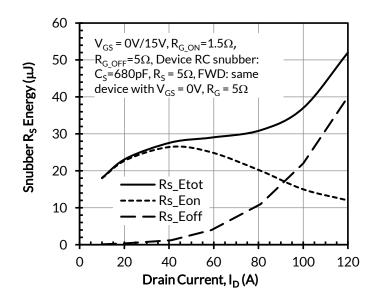


600 500 400 Qrr (nC) 300 200 $I_{s} = 80A$, $di/dt = 2800A/\mu s$, 100 $V_{GS} = 0V, R_G = 20\Omega$ 0 0 25 50 75 100 125 150 175 Junction Temperature, T_J (°C)

Figure 17. Safe operation area at $T_C = 25$ °C, D = 0, Parameter t_n

Figure 18. Reverse recovery charge Qrr vs. junction temperature at V_{DS} = 400V





current at V_{DS} = 400V and T_J = 25°C

Figure 19. Clamped inductive switching energy vs. drain Figure 20. RC snubber energy loss vs. drain current at $V_{DS} = 400V$ and $T_{J} = 25$ °C



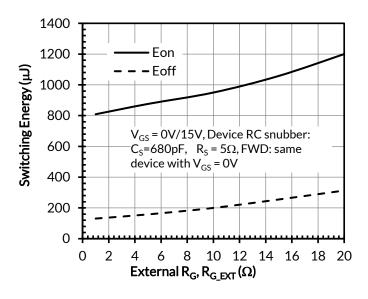








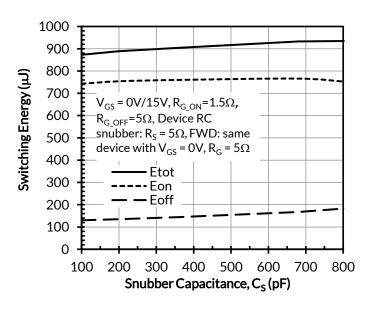




25 $V_{GS} = 0V/15V$, Device RC snubber: C_s =680pF, R_s = 5 Ω , FWD: same 20 Snubber R_S Energy (µJ) device with $V_{GS} = 0V$ Rs_Eon 15 Rs_Eoff 10 5 0 0 2 10 12 14 16 18 20 6 External R_G , $R_{G, EXT}(\Omega)$

Figure 21. Clamped inductive switching energies vs. $R_{G,EXT}$ at V_{DS} = 400V, I_{D} = 80A, and T_{J} = 25°C

Figure 22. RC snubber energy losses vs. $R_{G.EXT}$ at V_{DS} = 400V, $I_D = 80A$, and $T_J = 25$ °C



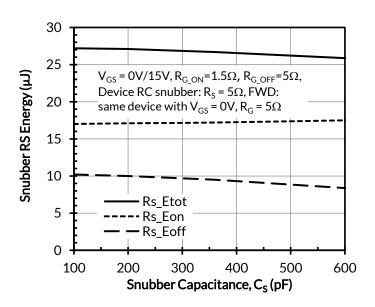


Figure 23. Clamped inductive switching energies vs. snubber capacitance C_S at V_{DS} = 400V, I_D = 80A, and T_J = capacitance C_S at V_{DS} = 400V, I_D = 80A, and T_J = 25°C 25°C

Figure 24. RC snubber energy losses vs. snubber





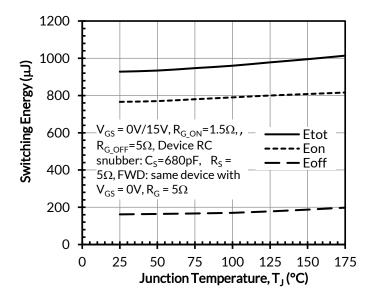












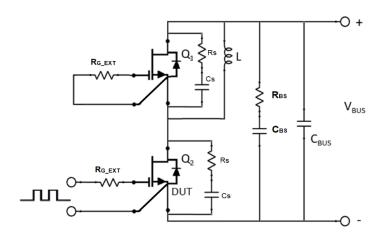


Figure 25. Clamped inductive switching energy vs. junction temperature at V_{DS} =400V and I_D = 80A

Figure 26. Schematic of the half-bridge mode switching test circuit. Note, a device snubber (Rs = 5Ω , Cs = 680pF) and bus RC snubber ($R_{BS} = 1\Omega$, $C_{BS} = 100$ nF) is used to reduce the power loop high frequency oscillations.













Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com

Important notice

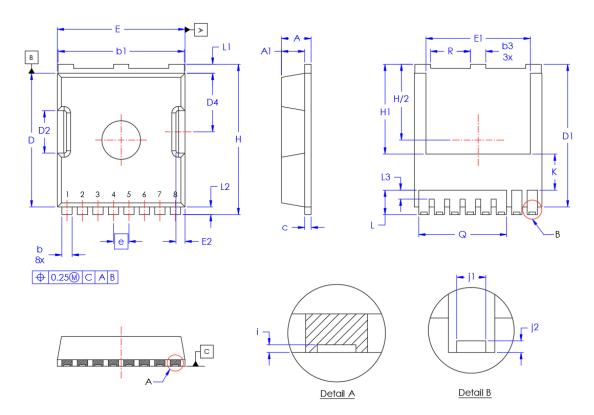
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Datasheet: UJ4SC075005L8S Rev. B, January 2025



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|---|---------------------------|
| DS_TOLL | Rev B |

PACKAGE OUTLINE



| TO-LL | | | | | |
|--------|-----------|-------|--|--|--|
| SYMBOL | Va | | | | |
| | Min | Max | | | |
| A | 2.15 | 2.45 | | | |
| Al | 1.80 REF | | | | |
| b | 0.65 | 0.90 | | | |
| b1 | 9.65 | 9.95 | | | |
| b3 | 1.10 | 1.30 | | | |
| С | 0.40 | 0.60 | | | |
| D | 10.18 | 10.58 | | | |
| DI | 10.88 | 11.28 | | | |
| D2 | 3.15 | 3.45 | | | |
| D4 | 4.40 | 4.70 | | | |
| Е | 9.70 | 10.10 | | | |
| El | 7.95 | 8.25 | | | |
| E2 | 0.60 0.80 | | | | |
| е | 1.20 BSC | | | | |
| Н | 11.48 | 11.88 | | | |
| HI | 6.80 | 7.10 | | | |
| i | 0.10 | REF | | | |
| jl | 0.46 | REF | | | |
| j2 | 0.20 | REF | | | |
| K | 2.80 | REF | | | |
| L | 1.40 | 2.10 | | | |
| L1 | 0.50 | 0.90 | | | |
| L2 | 0.48 | 0.72 | | | |
| L3 | 0.30 | 0.80 | | | |
| Q | 6.80 REF | | | | |
| R | 3.00 | 3 20 | | | |

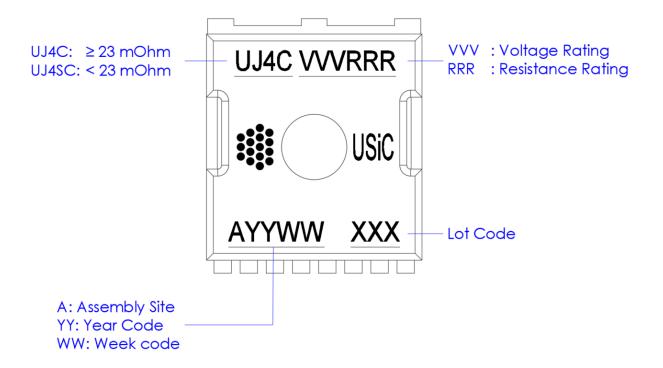
Note:

- 1. All dimensions in millimeters
- 2. Dimensions does not include Burrs and Mold Flashes



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|---|---------------------------|
| DS_TOLL | Rev B |

PART MARKING



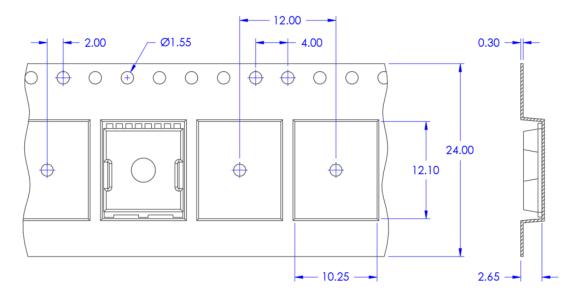
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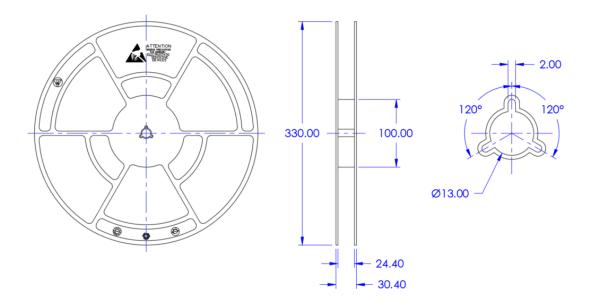
| TOLL PACKAGE OUTLINE, PART MARKING, TAPE AND REEL SPECIFICATION | Page 3 of 4 |
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| DS_TOLL | Rev B |

PACKING TYPE

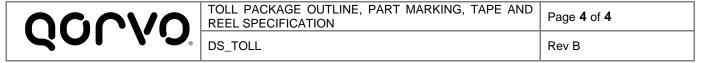
Carrier Tape



Reel



All dimensions in millimeters Quantity per Reel: 2000 units



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REVISION HISTORY

| Revision | Create Date (mm/dd/yyyy) | Description of Change | Initiator of Change |
|----------|--------------------------|--|---------------------|
| Α | 10/13/2023 | Initial Production Release | Glenn Galang |
| В | 01/31/2024 | Corrected device orientation inside carrier tape pocket (Page 3) | Glenn Galang |
| | | | |
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