

Ultra-low Power, 1-Channel 6th-Order SD Video Filter Driver **Description**

- Single SDTV Filter
- 6th-order Butterworth Low-Pass Filter:
 9MHz -3dB Bandwidth, 57.2dB Attenuation(27MHz)
- Support Multiple Input Biasing:
 - Provide 80-mV Level-Shift when DC-Coupled
 - Transparent Input Clamping when AC-Coupled
 - Support External DC Biasing when AC-Coupled
 - Very Low Quiescent Current: 3.85mA (Typical)
- Slew Rate: 38V/µs
- 6dB Gain (2V/V) with Rail-to-Rail Output
- AC- or DC-Coupled Output Driving Dual Video Loads (75Ω)
- Wide Operation Range: +2.85V to +5.5V Single Supply
- Robust ESD Protection:
 - HBM 8KV, CDM 2KV
- Lead-Free SC70 Package Available

Applications

- Video Signal Amplification
- Set-Top Box Video Driver
- PVR、 DVD Player Video Buffer
- Video Buffer for Portable or USB-Powered Video Devices
- HDTV

performance and very low power consumption, it is excellent choice for portable video applications. Integrated with a 6th-order Butterworth filter, TPF111 can be used as a DAC reconstruction filter or ADC anti-aliasing filter. With its -3dB frequency of 9MHz, it is an ideal choice for SD video applications including NTSC and PAL. TPF111 accepts both AC- and DC-coupled inputs and its flexible biasing options meet the requirement of the most demanding applications. The integrated Transparent Sync-tip Clamp circuit restores DC voltage level of an AC-coupled

TPF111 is a low-cost reconstruction video filter optimized for

consumer video devices. With its exceptional video

Video signal. It translates the sync tip of a CVBS, Y', or RGB signal to a fixed 40mV. External biasing resistors can be used to restore signals without sync tip such as Pb' or Pr'. Integrated level shifter then raises the clamped video signal by 80mV, assuring passing through of video signal without being distorted. When the signal is DC-coupled, the level shifter will raise the signal by 80mV.

TPF111 may be used for different kinds of video buffering with 6dB gain(2V/V) and rail-to-rail output. It supports AC-and DC-coupling at the output.

TPF111 is designed to have exceptional ESD rating. It is a great choice of protecting the main video processor chip from ESD or surge strikes in applications such as set-top-boxes.

TPF111 operates from a single supply with wide voltage range from +2.85V to +5.5V. Its very low quiescent current of 3.85mA makes it an perfect choice for battery-power or USB-power applications.

TPF111 is available in SC70-5 package (TPF111-C) and SOIC-8 package (TPF111-S). Its operation temperature range is from -40° C to $+85^{\circ}$ C.

Related Resources

AN-1201: Application notes of TPF1xx



Function Block

Order Information

Order Number	Marking Information	Operating Temperature Range	Package	Transport Media, Quantity
TPF111-CR	F1YW ⁽¹⁾	-40 to 85°C	5-Lead SC70	Tape and Reel, 3,000pcs
TPF111-SR	TPF111	-40 to 85°C	8-Lead SOIC	Tape and Reel, 4,000pcs
TPF111U-CR	F1UYW ⁽¹⁾	-40 to 85°C	5-Lead SC70	Tape and Reel, 3,000pcs
TPF111U-SR	TPF111U	-40 to 85°C	8-Lead SOIC	Tape and Reel, 4,000pcs

Note: (1). 'YW' is date coding scheme. 'Y' stands for calendar year, and 'W' stands for single workweek coding scheme.

Pin configuration (Top View)



Pin Functions

SC70	SOIC	Pin Name	Function
3 / 1	1 / 1	V _{IN}	Video Input
	2, 4, 5, 6 / 2, 3, 6, 7	N/C	No Connect
1, 2 / 2, 4	3 / 5	GND	Ground
5/3	7 / 4	+Vs	Positive Power Supply
4 / 5	8 / 8	V _{OUT}	Filtered Video Output

Absolute Maximum RatingsNote

	Parameters	Value	Unit
	Power Supply, V _{DD} to GND	6.0	V
PD	Power dissipation, $T_A = 25^{\circ}C$, 5-Lead SC70) 300 ⁽¹⁾ mW	
FD	Power dissipation , $T_A = 25^{\circ}C$, 8-Lead SOIC	800 ⁽¹⁾	11177
V _{IN}	Input Voltage	V _{DD} + 0.3∖	/ to GND - 0.3V
lo	Output Current	65	mA
TJ	Maximum Junction Temperature	150	°C
T _A	Operating Temperature Range	-45 to 85	°C
T _{STG}	Storage Temperature Range	-65 to 150	°C
TL	Lead Temperature (Soldering, 10 sec)	300	С°
θ _{JA}	5-Lead SC70	430 ⁽²⁾	°C/W
	8-Lead SOIC	130 ⁽²⁾	0/11

(1) This data was taken with the JEDEC low effective thermal conductivity test board.

(2) This data was taken with the JEDEC standard multilayer test boards.

* **Note:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

ESD, Electrostatic Discharge Protection

Symbol	Parameter	Condition	Minimum Level	Unit
HBM	Human Body Model ESD	MIL-STD-883H Method 3015.8	8	kV
CDM	Charged Device Model ESD	JEDEC-EIA/JESD22-C101E	2	kV

Electrical Characteristics All test condition is VDD = 3.3V, TA = +25°C, RL = 150 Ω to GND, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Input Electric	cal Specifications			1		
V _{DD}	Supply Voltage Range		2.85		5.5	V
1		V_{DD} = 3.3V, V_{IN} = 500mV, no load		3.85	4.85	mA
I _{DD}	Quiescent current (I _Q) ⁽¹⁾	V_{DD} = 5.0V, V_{IN} = 500mV, no load		5.00	6.25	mA
V _{CLAMP}	Input Voltage Clamp	I _{IN} = -100μA	-40	0	+40	mV
ICLAMP-CHG	Clamp Charge Current	V _{IN} = V _{CLAMP} - 200mV	-1.5	-1.7		mA
I _{CLAMP-DCHG}	Clamp Discharge Current	V _{IN} = 500mV	1.5	2.0	5.1	μA
R _{IN}	Input Impedance	0.5V < V _{IN} < 1.0V	0.5	3		MΩ
AV	Voltage Gain (1)	V _{IN} =0.5V, 1V and 2V R _L = 150Ω to GND	5.9	6.01	6.025	dB
Vols	Output Level Shift Voltage	V_{IN} = 0V, no load, input referred	53	80	124	mV
V _{OH}	Output Voltage High Swing	V_{IN} = 3V, R_L = 150 Ω to GND		3.18		V
V _{OL}	Output Voltage Low Swing	V _{IN} = -0.3V, R _L =75Ω		0.05		V
PSRR	Power Supply Rejection Ratio	ΔV_{DD} = 3.3V to 3.6V		61		dB
FORR		ΔV_{DD} = 5.0V to 5.5V, 50Hz, V _{IN} =0.7V		67		dB
l	Chart aircuit aurrant	V_{IN} =2V, output to GND through 10 Ω	65			mA
I _{SC}	Short-circuit current	V_{IN} =100mV, output short to V_{DD}	65			mA
AC Electrica	I Specifications					
f _{-1dB}	-1dB Bandwidth	R _L =150Ω	7.6	8.2	9.1	MHz
f _{-3dB}	-3dB Bandwidth	R _L =150Ω	7.8	9	10.5	MHz
Att _{27MHz}	Stop Band Attenuation	f = 27MHz	38.2	57.2	73.6	dB
dG	Differential Gain	Video input range 1V	-0.1	0.4	0.8	%
dP	Differential Phase	Video input range 1V	-1.1	0.7	1.1	0
THD	Output Distortion(All Channel)	f=1MHz, V _{OUT} =1.4V _{PP}	0.03	0.1	0.2	%
D/DT	Group Delay Variation	f = 100kHz, 5MHz		5.4		Ns
t _{PD}	Propagation Delay	Maximum delay from input to output: 100kHz to 4.43MHz	54	80	127	Ns
SNR	Signal-to-Noise Ration	f= 100kHz to 4.43MHz	65	69		dB
R _{OUT_AC}	Output Impedance	f = 4.2MHz		1.5		Ω
CLG	Chroma-Luma-Gain	400kHz to 3.58MHz and 4.43MHz		0.18	0.4	dB
CLD	Chroma-Luma-Delay	400kHz to 3.58MHz and 4.43MHz		5		ns

Note: (1). 100% tested at T_A=25°C.

Typical Performance Characteristics All test condition is VDD = 3.3V, TA = +25°C, RL = 150Ω to GND, unless otherwise noted.



Figure1. Small-Scale Frequency Response



Figure3. Gain Vs. Frequency







Figure2. Large-Scale Frequency Response



Figure4. Gain Vs. Frequency With CLOAD



Figure6. Group Delay vs Frequency



Figure 7. PSRR Vs. Frequency



Figure9. Large-Signal Pulse Response Vs. Time



Figure11. Differential Gain(dG)



Figure8. Output Impedance Vs. Frequency



Figure10. Large-Signal Pulse Response Vs. Time



Figure12. Differential Phase(dP)

Application Information

The TPF111 is a single supply rail-to-rail output amplifier achieving a -3dB bandwidth of around 9MHz and slew rate of about 38V/µs while demanding only 3.85mA of supply current. This part is ideally suited for applications with specific micropower consumption and high bandwidth demands. As the performance characteristics above and the features described below, the TPF111 is designed to be very attractive for portable composite video applications.

Internal Sync Clamp

The typical embedded video DAC operates from a ground referenced single supply. This becomes an issue because the lower level of the sync pulse output may be at a 0V reference level to some positive level. The problem is presenting a 0V input to most single supply driven amplifiers will saturate the output stage of the amplifier resulting in a clipped sync tip and degrading the video image. A larger positive reference may offset the input above its positive range.

The TPF111 features an internal sync clamp and offset function to level shift the entire video signal to the best level before it reaches the input of the amplifier stage. These features are also helpful to avoid saturation of the output stage of the amplifier by setting the signal closer to the best voltage range.

The simplified block diagram of the TPF111 in Page-1.The AC coupled video sync signal is pulled negative by a current source at the input of the comparator amplifier. When the sync tip goes below the comparator threshold the output comparator is driven negative, The PMOS device turns on clamping sync tip to near ground level. The network triggers on the sync tip of video signal.

Droop Voltage and DC Restoration

Selection of the input AC-coupling capacitance is based on the system requirements. A typical sync tip width of a 64 μ s NTSC line is 4 μ s during which clamp circuit restores its DC level. In the remaining 60 μ s period, the voltage droops because of a small constant 2.0 μ A sinking current. If the AC-coupling capacitance is 0.1 μ F, the maximum droop voltage is about 1mV which is restored by the clamp circuit. The maximum pull-up current of the clamp circuit is 1.7mA. For a 4µs sync tip width and 0.1µF capacitor, the maximum restoration voltage is about 80mV.

The line droop voltage will increase if a smaller AC-coupling capacitance is used. For the same reason, if larger capacitance is used the line droop voltage will decrease. Table 1 is droop voltage and maximum restoration voltage of the clamp for typical capacitance.

 Table 1.
 Maximum restoration voltage and droop voltage of Y and CVBS signals for different capacitance

CAP VALUE (nF)	DROOP IN 60µs (mV)	CHARGE IN 4µs (mV)
100	1.2	68
1,000	0.12	6.8

Low Pass Filter--Sallen Key

The Sallen Key is a classic low pass configuration. This provides a very stable low pass function, and in the case of the TPF111, a six-pole roll-off at around 9MHz. The six-pole function is accomplished with an RC low pass network placed in series with and before the Sallen Key.

Output Couple

TPF111 output could support both "AC Couple" and "DC Couple", if use "AC Couple", this capacitor is typically between 220- μ F and 1000- μ F, although 470- μ F is common. This value of this capacitor must be this large to minimize the line tilt (droop) and/or field tilt associated with ac-coupling as described previously in this document.

The TPF111 internal sync clamp makes it possible to DC couple the output to a video load, eliminating the need for any AC coupling capacitors, thereby saving board space and additional expense for capacitors. This makes the TPF111 extremely attractive for portable video applications. Additionally, this solution completely eliminates the issue of field tilt in the lower frequency. The trade off is greater demand of supply current. Typical load current for AC coupled is around 1mA, compared to typical 6.6mA used when DC coupling.

Output Drive Capability and Power Dissipation

With the high output drive capability of the TPF111, it is possible to exceed the +125°C absolute maximum junction temperature under certain load current conditions. Therefore, it is important to calculate the maximum junction temperature for an application to determine if load conditions or package types need to be modified to assure operation of the amplifier in a safe operating area. The maximum power dissipation allowed in a package is determined according to Equation:

$$PD_{MAX} = \frac{T_{JMAX} - T_{AMAX}}{\theta_{JA}}$$

 $\begin{array}{l} Where: \\ T_{JMAX} = Maximum junction temperature \\ T_{AMAX} = Maximum ambient temperature \\ \textcircled{m}{\ensuremath{\Theta}}_{JA} = Thermal resistance of the package \end{array}$

The maximum power dissipation actually produced by an IC is the total quiescent supply current times the total power supply voltage, plus the power in the IC due to the load, or: for sourcing:

$$PD_{MAX} = V_{s} \times I_{SMAX} + (V_{s} - V_{OUT}) \times \frac{V_{OUT}}{R_{t}}$$

Where:

$$\label{eq:VS} \begin{split} V_S &= Supply \ voltage \\ I_{SMAX} &= Maximum \ quiescent \ supply \ current \\ V_{OUT} &= Maximum \ output \ voltage \ of \ the \ application \\ R_{LOAD} &= Load \ resistance \ tied \ to \ ground \end{split}$$

By setting the two PDMAX equations equal to each other, we can solve the output current and RLOAD to avoid the device overheat.

Power Supply Bypassing Printed Circuit Board Layout

As with any modern operational amplifier, a good printed circuit board layout is necessary for optimum performance. Lead lengths should be as short as possible. The power supply pin must be well bypassed to reduce the risk of oscillation. For normal single supply operation, a single 4.7μ F tantalum capacitor in parallel with a 0.1μ F ceramic capacitor from VS+ to GND will suffice.

VIDEO FILTER DRIVER SELECTION GUIDE

P/N	Product Description	Channel	-3dB Bandwidth	Package
TPF110	Low power, enable function and	1-SD	9MHz	SC70-5
/TPF110L	SAG correction, 1 channel 6 th order			SOT23-6
	9MHz			
TPF113	Low power 3 channel, 6th-order	3-SD	9MHz	SO-8
	9MHz SD video filter			
TPF114	Low power 4 channel, 6th-order	4-SD	9MHz	MSOP-10
	9MHz SD video filter			TSSOP-14
TPF116	Low power 4 channel, 6th-order	6-SD	9MHz	TSSOP-14
	9MHz SD video filter for CVBS,			
	SVIDEO			
TPF123	3 channel 6th-order 13.5MHz,	3-ED	13.5MHz	SO-8
	960H/720H-CVBS video filter or			
	Y'Pb'Pr 480P/576P video filter			
TPF133	Low power 3 channel, 6th-order	3-HD	36MHz	SO-8
	36MHz HD video filter			
TPF134	Low power 3 channel, 6th-order	1-SD&	9MHz	MSOP-10
	36MHz HD video filter and 1 channel	3-SD	36MHz	TSSOP-14
	SD video filter			
TPF136	Low power 3 channel, 6th-order	3-SD&	9MHz	TSSOP-20
	36MHz HD video filter and 3 channel	3-HD	36MHz	
	•	-		

	SD video filter			
TPF143	Low power 3 channel, 6th-order	3-FHD	72MHz	SO-8
	72MHz Full HD video filter			
TPF144	Low power 3 channel, 6th-order	1-SD&	9MHz	MSOP-10
	72MHz Full HD video filter and 1	3-FHD	72MHz	TSSOP-14
	channel SD video filter			
TPF146	Low power 3 channel, 6th-order	3-SD&	9MHz	TSSOP-20
	72MHz Full HD video filter and3	3-FHD	72MHz	
	channel SD video filter			
TPF153	Low power 3 channel, 6th-order	3-CH	220MHz	SO-8
	220MHz Full HD video filter			

Package Outline Dimensions



	Dimensi	ons	Dimensions In			
Symbol	In Millim	eters	Inches	Inches		
	Min	Max	Min	Max		
А	0.900	1.100	0.035	0.043		
A1	0.000	0.100	0.000	0.004		
A2	0.900	1.000	0.035	0.039		
b	0.150	0.350	0.006	0.014		
С	0.080	0.150	0.003	0.006		
D	2.000	2.200	0.079	0.087		
E	1.150	1.350	0.045	0.053		
E1	2.150	2.450	0.085	0.096		
е	0.650TYF)	0.026TYP			
e1	1.200	1.400	0.047	0.055		
L	0.525REF		0.021REF			
L1	0.260	0.460	0.010	0.018		
θ	0°	8°	0°	8°		



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