

# SPT7810

## 10-BIT, 20 MSPS, ECL OUTPUT A/D CONVERTER

### **FEATURES**

- Monolithic 20 MSPS Converter
- On-Chip Track/Hold
- Bipolar ±2.0 V Analog Input
- 60 dB SNR @ 1 MHz Input
- Low Power (1.3 W Typical)
- 5 pF input Capacitance
- ECL Outputs

## APPLICATIONS

- Medical Imaging
- Professional Video
- Radar Receivers
- Instrumentation
- Electronic Warfare
- Digital Communications

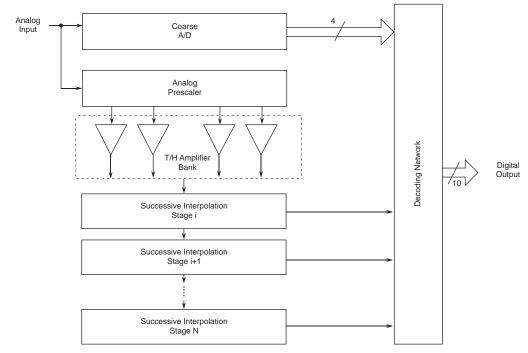
## **GENERAL DESCRIPTION**

The SPT7810 A/D converter is a 10-bit monolithic converter capable of word rates of a minimum of 20 MSPS. On board track/hold function assures excellent dynamic performance without the need for external components. Drive requirement problems are minimized with an input capacitance of only 5 pF.

Inputs and outputs are ECL to provide a higher level of noise immunity in high speed system applications. An overrange output signal is provided to indicate overflow conditions. Output data format is straight binary. Power dissipation is very low at only 1.3 watts with power supply voltages of +5.0 and -5.2 volts. The SPT7810 also provides a wide input voltage swing of  $\pm$ 2.0 volts.

The SPT7810 is available in a 28-lead ceramic sidebrazed DIP, PDIP, and die form. Commercial and industrial temperature ranges are currently offered. Contact the factory for availability of military temperature range and /883 processed units.

## **BLOCK DIAGRAM**



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## ABSOLUTE MAXIMUM RATINGS (Beyond which damage may occur)<sup>1</sup> 25 °C

Supply	Voltages	
\ /		

V <sub>CC</sub>	-0.3 to +6 V
V <sub>EE</sub>	+0.3 to -6 V

#### Input Voltages

Analog Input	V <sub>FB</sub> ≤V <sub>IN</sub> ≤V <sub>FT</sub>
V <sub>FT</sub> , V <sub>FB</sub>	+3.0 V, -3.0 V
Reference Ladder Current	

Out	put
Out	pui

Digital Outputs	+30 to -30 mA

#### Temperature

Operating Temperature25	to +85 °C
Junction Temperature (1)	175 °C
Lead Temperature, (soldering 10 seconds)	300 °C
Storage Temperature	o +150 °C

Note: 1. Operation at any Absolute Maximum Rating is not implied. See Electrical Specifications for proper nominal applied conditions in typical applications.

## **ELECTRICAL SPECIFICATIONS**

TA=Tmin - Tmax, VCC=+5.0 V, VEE=-5.2 V, VIN=±2.0 V, VSB=-2.0 V, VST=+2.0 V, fclock=20 MHz, 50% clock duty cycle, unless otherwise specified.

PARAMETERS	TEST CONDITIONS	TEST LEVEL	S MIN	SPT7810A TYP	MAX	MIN	SPT7810 TYP	)B MAX	UNITS
Resolution			10			10			Bits
DC Accuracy (+25 °C) Integral Nonlinearity Differential Nonlinearity No Missing Codes	± Full Scale 250 kHz Sample Rate	V V VI	G	±1.0 ±0.5 Guarantee	d	G	±1.5 ±0.75 Guarante	ed	LSB LSB
Analog Input Input Voltage Range Input Bias Current Input Resistance Input Capacitance Input Bandwidth +FS Error -FS Error	V <sub>IN</sub> =0 V 3 dB Small Signal	>  >  > > > > >	100	±2.0 30 300 5 120 ±2.0 ±2.0	60	100	±2.0 30 300 5 120 ±2.0 ±2.0	60	V μA pF MHz LSB LSB
Reference Input Reference Ladder Resistance Reference Ladder Tempco		VI V	500	800 0.8		500	800 0.8		Ω Ω/°C
Timing Characteristics Maximum Conversion Rate Overvoltage Recovery Time Pipeline Delay (Latency) Output Delay Aperture Delay Time Aperture Jitter Time	T <sub>A</sub> =+25 °C T <sub>A</sub> =+25 °C T <sub>A</sub> =+25 °C	VI V IV V V	20	20 5 1 5	1	20	20 5 1 5	1	MHz ns <sup>Clock Cyde</sup> ns ns ps-RMS
Dynamic Performance Effective Number of Bits f <sub>IN</sub> =1 MHz f <sub>IN</sub> =3.58 MHz f <sub>IN</sub> =10.3 MHz				9.2 8.8 7.5			8.7 8.3 7.0		Bits Bits Bits

Typical thermal impedances: 28L sidebrazed DIP.  $\theta_{ja}$  = 50 °C/W, 28L plastic DIP  $\theta_{ja} = 50 \text{ °C/W}$ .



## **ELECTRICAL SPECIFICATIONS**

 $T_{A}=T_{min} - T_{max}, V_{CC}=+5.0 \text{ V}, V_{EE}=-5.2 \text{ V}, V_{IN}=\pm 2.0 \text{ V}, V_{SB}=-2.0 \text{ V}, V_{ST}=+2.0 \text{ V}, f_{CLK}=20 \text{ MHz}, 50\% \text{ clock duty cycle, unless otherwise specified.}$ 

	TEST TEST SPT7810A SP			SPT7810	)B				
PARAMETERS	CONDITIONS	LEVEL	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
Dynamic Performance									
Signal-To-Noise Ratio									
(without Harmonics)									
f <sub>IN</sub> =1 MHz	+25 °C	I	57	60		54	57		dB
		IV	55	58		52	55		dB
f <sub>IN</sub> =3.58 MHz	+25 °C	I I	56	58		53	55		dB
		IV	54	56		51	53		dB
f <sub>IN</sub> =10.3 MHz	+25 °C	I I	50	53		47	49		dB
		IV	47	50		44	46		dB
Harmonic Distortion									
f <sub>IN</sub> =1 MHz	+25 °C	1	57	60		54	57		dB
	120 0	IV	54	57		51	54		dB
f <sub>IN</sub> =3.58 MHz	+25 °C		56	58		53	55		dB
11N-3.30 WH 12	+23 0	IV	53	55		50	52		dB
f <sub>IN</sub> =10.3 MHz	+25 °C		46	48		43	45		dB
1N=10:3 WI 12	+23 0	IV	40	40		43	43		dB
		IV	40	47		42			
Signal-to-Noise and Distortion									
f <sub>IN</sub> =1 MHz	+25 °C		55	57		52	54		dB
		IV	52			49			dB
f <sub>IN</sub> =3.58 MHz	+25 °C		54	55		51	52		dB
		IV	51			48			dB
f <sub>IN</sub> =10.3 MHz	+25 °C	I	44	47		41	44		dB
		IV	43			40			dB
Spurious Free Dynamic Range	+25 °C, f <sub>IN</sub> =1 MHz	V		67			67		dB
Differential Phase	+25 °C, f <sub>IN</sub> =3.58 & 4.35 MHz			0.2			0.2		Degree
Differential Gain	+25 °C, f <sub>IN</sub> =3.58 & 4.35 MHz	V		0.5			0.7		%
Digital Inputs									
Logic 1 Voltage		VI	-1.1			-1.1			V
		VI	-1.1		1 5	-1.1		15	1
Logic 0 Voltage		VI	500	1200	-1.5	500	1200	-1.5	1
Maximum Input Current Low			-500	±200	+750	-500	±200	+750	μA
Maximum Input Current High		VI	-500	±300	+750	-500	+300	+750	μA
Pulse Width Low (CLK)		IV	20		200	20		200	ns
Pulse Width High (CLK)		IV	20		300	20		300	ns
Digital Outputs									
Logic 1 Voltage	50 Ω to -2 V	VI	-1.1	-0.8		-1.1	-0.8		V
Logic 0 Voltage	50 Ω to -2 V	VI		-1.8	-1.5		-1.8	-1.5	V
Power Supply Requirements									
Voltages V <sub>CC</sub>		IV	+4.75	-5.0	+5.25	+4.75	+5.0	+5.25	
-V <sub>EE</sub>		IV	-4.95	-5.2	-5.45	-4.95	-5.2	-5.45	1
Currents I <sub>CC</sub>		VI		140	170		140	190	mA
- <sup>I</sup> EE		VI		115	140		115	160	mA
Power Dissipation	Outputs Open	VI		1.3	1.6		1.3	1.8	
Power Supply Rejection Ratio	(5 V ±0.25 V, -5.2 V ±2.0 V)	V		1.0			1.0		LSB

#### **TEST LEVEL CODES**

#### TEST LEVEL TEST PROCEDURE

L

Ш

Ш

IV

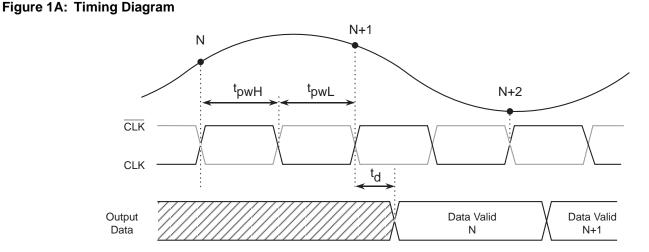
V

VI

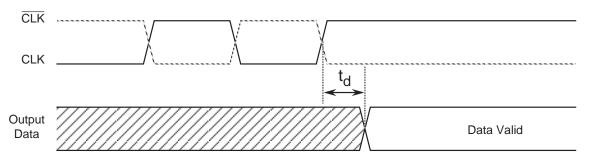
All electrical characteristics are subject to the following conditions:

All parameters having min/max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality Assurance inspection. Any blank section in the data column indicates that the specification is not tested at the specified condition.

- 100% production tested at the specified temperature.
- 100% production tested at  $T_A=25$  °C, and sample tested at the specified temperatures.
- QA sample tested only at the specified temperatures.
- Parameter is guaranteed (but not tested) by design and characterization data.
- Parameter is a typical value for information purposes only.
- 100% production tested at  $T_A = 25$  °C. Parameter is guaranteed over specified temperature range.



#### Figure 1B: Single Event Clock



#### **Table I - Timing Parameters**

PARAMETERS DESCRIPTION		MIN	ТҮР	МАХ	UNITS
t <sub>d</sub>	CLK to Data Valid Prop Delay	CLK to Data Valid Prop Delay - 5			ns
t <sub>pwH</sub>	CLK High Pulse Width 20		-	300	ns
t <sub>pwL</sub>	CLK Low Pulse Width	20	-	-	ns



## SPECIFICATION DEFINITIONS

#### APERTURE DELAY

Aperture delay represents the point in time, relative to the rising edge of the CLOCK input, that the analog input is sampled.

#### APERTURE JITTER

The variations in aperture delay for successive samples.

#### **DIFFERENTIAL GAIN (DG)**

A signal consisting of a sine wave superimposed on various DC levels is applied to the input. Differential gain is the maximum variation in the sampled sine wave amplitudes at these DC levels.

#### **DIFFERENTIAL PHASE (DP)**

A signal consisting of a sine wave superimposed on various DC levels that is applied to the input. Differential phase is the variation in the sampled sine wave phases at these DC levels.

#### **EFFECTIVE NUMBER OF BITS (ENOB)**

SINAD = 6.02N + 1.76, where N is equal to the effective number of bits.

$$N = \frac{SINAD - 1.76}{6.02}$$

#### $\pm$ FULL-SCALE ERROR (GAIN ERROR)

Difference between measured full scale response [(+Fs) - (-Fs)] and the theoretical response  $(+4 \vee -2 \text{ LSBs})$  where the +FS (full scale) input voltage is defined as the output transition between 1-10 and 1-11 and the -FS input voltage is defined as the output transition between 0-00 and 0-01.

#### INPUT BANDWIDTH

Small signal (50 mV) bandwidth (3 dB) of analog input stage.

#### DIFFERENTIAL NONLINEARITY (DNL)

Error in the width of each code from its theoretical value. (Theoretical =  $V_{FS}/2^{N}$ )

#### INTEGRAL NONLINEARITY (INL)

Linearity error refers to the deviation of each individual code (normalized) from a straight line drawn from -Fs through +Fs. The deviation is measured from the edge of each particular code to the true straight line.

#### OUTPUT DELAY

Time between the clock's triggering edge and output data valid.

#### OVERVOLTAGE RECOVERY TIME

The time required for the ADC to recover to full accuracy after an analog input signal 125% of full scale is reduced to 50% of the full-scale value.

#### SIGNAL-TO-NOISE RATIO (SNR)

The ratio of the fundamental sinusoid power to the total noise power. Harmonics are excluded.

#### SIGNAL-TO-NOISE AND DISTORTION (SINAD)

The ratio of the fundamental sinusoid power to the total noise and distortion power.

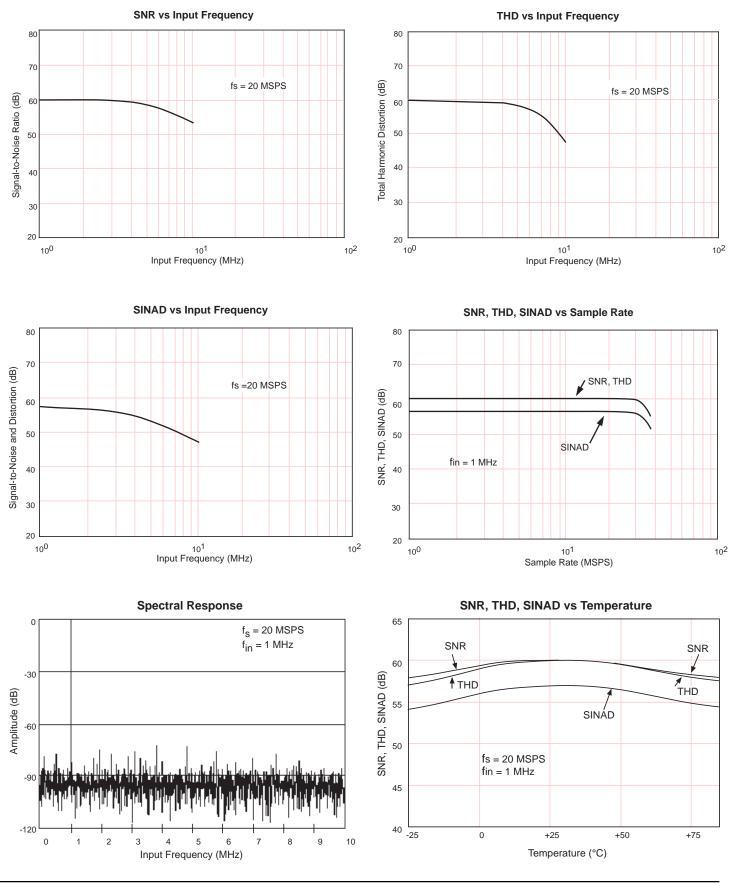
#### TOTAL HARMONIC DISTORTION (THD)

The ratio of the total power of the first 64 harmonics to the power of the measured sinusoidal signal.

#### SPURIOUS FREE DYNAMIC RANGE (SFDR)

The ratio of the fundamental sinusoidal amplitude to the single largest harmonic or spurious signal.

## **TYPICAL PERFORMANCE CHARACTERISTICS**



SPT

## TYPICAL INTERFACE CIRCUIT

The SPT7810 requires few external components to achieve the stated operation and performance. Figure 2 shows the typical interface requirements when using the SPT7810 in normal circuit operation.

The following section provides a description of the pin functions and outlines critical performance criteria to consider for achieving the optimal device performance.

#### POWER SUPPLIES AND GROUNDING

The SPT7810 requires the use of two supply voltages, V<sub>EE</sub> and V<sub>CC</sub>. Both supplies should be treated as analog supply sources. This means the V<sub>EE</sub> and V<sub>CC</sub> ground returns of the device should both be connected to the analog ground plane. All other -5.2 V requirements of the external digital logic circuit should be connected to the digital ground plane. Each power supply pin should be bypassed as closely as possible to the device with .01  $\mu$ F and 10  $\mu$ F capacitors as shown in figure 2.

The two grounds available on the SPT7810 are AGND and DGND. DGND is used only for ECL outputs and is to be referenced to the output pulldown voltage. These grounds are not tied together internal to the device. The use of ground planes is recommended to achieve the best performance of

the SPT7810. The AGND and the DGND ground planes should be separated from each other and only connected together at the device through an inductance. Doing this will minimize the ground noise pickup.

#### **VOLTAGE REFERENCE**

The SPT7810 requires the use of two voltage references: VFT and VFB. VFT is the force for the top of the voltage reference ladder (+2.5 V typ), VFB (-2.5 V typ) is the force for the bottom of the voltage reference ladder. Both voltages are applied across an internal reference ladder resistance of 800 ohms. In addition, there are 3 reference ladder taps (VST, VRM and VSB). VST is the sense for the top of the reference ladder (+2.0 V), V<sub>RM</sub> is the midpoint of the ladder (0.0 V typ) and V<sub>SB</sub> is the sense for the bottom of the reference ladder (-2.0 V). The voltages seen at VST and VSB are the true full scale input voltages of the device when VFT and VFB are driven to the recommended voltages (+2.5 V and -2.5 V typical respectively). These points should be used to monitor the actual full scale input voltage of the device and should not be driven to the expected ideal values as is commonly done with standard flash converters. When not being used, a decoupling capacitor of .01 uF connected to AGND from each tap is recommended to minimize high frequency noise injection.

An example of a reference driver circuit recommended is shown in figure 2. IC1 is REF-03, the +2.5 V reference with a

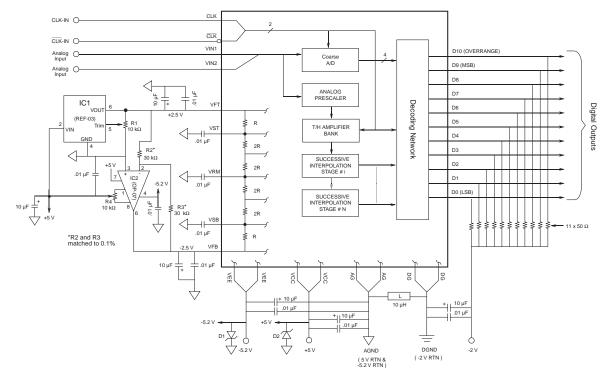


Figure 2 - Typical Interface Circuit

NOTE: D1=D2=1N5817 or equivalent. (Used to prevent damage caused by power sequencing.)

tolerance of 0.6% or  $\pm$  0.015 V. The potentiometer R1 is 10 k $\Omega$  and supports a minimum adjustable range of up to 150 mV. IC2 is recommended to be an OP-07 or equivalent device. R2 and R3 must be matched to within 0.1% with good TC tracking to maintain a 0.3 LSB matching between VFT and VFB. If 0.1% matching is not met, then potentiometer R4 can be used to adjust the VFB voltage to the desired level. R1 and R4 should be adjusted such that VST and VSB are exactly +2.0 V and -2.0V respectively.

The analog input range will scale proportionally with respect to the reference voltage if a different input range is required. The maximum scaling factor for device operation is  $\pm$  20% of the recommended reference voltages of V<sub>FT</sub> and V<sub>FB</sub>. However, because the device is laser trimmed to optimize performance with  $\pm$  2.5 V references, the accuracy of the device will degrade if operated beyond a  $\pm$  2% range.

The following errors are defined:

+FS error = top of ladder offset voltage =  $\Delta$ (+FS -V<sub>ST</sub>+1 LSB) -FS error = bottom of ladder offset voltage =  $\Delta$ (-FS -V<sub>SB</sub>-1 LSB)

where the +FS (full scale) input voltage is defined as the output transition between 1-10 and 1-11 and the -FS input voltage is defined as the output transition between 0-00 and 0-01.

#### ANALOG INPUT

 $V_{IN1}$  and  $V_{IN2}$  are the analog inputs. Both inputs are tied to the same point internally. Either one may be used as an analog input "sense" and the other for an input "force." The inputs can also be tied together and driven from the same source. The full scale input range will be 80% of the reference voltage or  $\pm 2$  volts with V<sub>FB</sub>=-2.5 V and V<sub>FT</sub>=+2.5 V.

The drive requirements for the analog inputs are minimal when compared to conventional Flash converters due the SPT7810's extremely low input capacitance of only 5 pF and very high input resistance of  $300 \text{ k}\Omega$ . For example, for an input signal of  $\pm 2 \text{ V}$  p-p with an input frequency of 10 MHz, the peak output current required for the driving circuit is only 628  $\mu$ A.

#### **CLOCK INPUT**

The clock inputs (CLK,  $\overline{\text{CLK}}$ ) are designed to be driven differentially with ECL levels. The clock may be driven single ended since  $\overline{\text{CLK}}$  is internally biased to -1.3 V.  $\overline{\text{CLK}}$  may be left open, but a .01  $\mu$ F bypass capacitor to AGND is recommended. As with all high speed circuits, proper terminations are required to avoid signal reflections and possible ringing that can cause the device to trigger at an unwanted time.

The CLK pulse width (tpwH) must be kept between 10 ns and 300 ns to ensure proper operation of the internal track-and-hold amplifier. (See timing diagram.) When operating the SPT7810 at sampling rates above 3 MSPS, it is recommended that the clock input duty cycle be kept at 50% to optimize performance. The analog input signal is latched on the rising edge of the CLK.

#### **DIGITAL OUTPUTS**

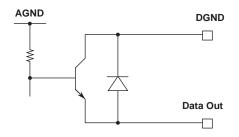
The format of the output data (D0-D9) is straight binary. These outputs are ECL with the output circuit shown in figure 4. The outputs are latched on the rising edge of CLK with a propagation delay of 4 ns. There is a one clock cycle latency between CLK and the valid output data (see timing diagram). These digital outputs can drive 50 ohms to ECL levels when pulled down to -2 V. The total specified power dissipation of the device does not include the power used by these loads. The additional power used by these loads can vary between 10 and 300 mW typically (including the overrange load) depending on the output codes. If lower power levels are desired, the output loads can be reduced, but careful consideration to the capacitive loads in relation to the operating frequency must be considered.

#### Table II - Output Data Information

ANALOG INPUT	OVERRANGE D1O	OUTPUT CODE D9-DO	
>+2.0 V + 1/2 LSB	1	11 1111 1111	
+2.0 V -1 LSB	0	11 1111 111Ø	
0.0 V	0	ØØ ØØØØ ØØØØ	
-2.0 V +1 LSB	0	00 0000 000Ø	
<-2.0 V	0	00 0000 0000	

(Ø indicates the flickering bit between logic 0 and 1).

#### Figure 3 - Output Circuit



#### **OVERRANGE OUTPUT**

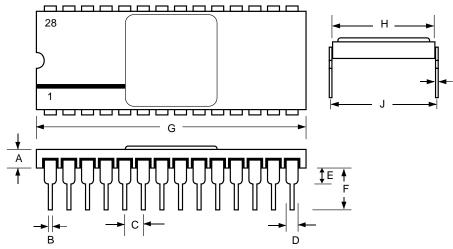
The OVERRANGE OUTPUT (D10) is an indication that the analog input signal has exceeded the positive full scale input voltage by 1 LSB. When this condition occurs, D10 will switch to logic 1. All other data outputs (D0 to D9) will remain at logic 1 as long as D10 remains at logic 1. This feature makes it possible to include the SPT7810 into higher resolution systems.

#### **EVALUATION BOARD**

The EB7810 evaluation board is available to aid designers in demonstrating the full performance of the SPT7810. This board includes a reference circuit, clock driver circuit, output data latches and an on-board reconstruction of the digital data. An application note describing the operation of this board as well as information on the testing of the SPT7810 is also available. Contact the factory for price and availability.

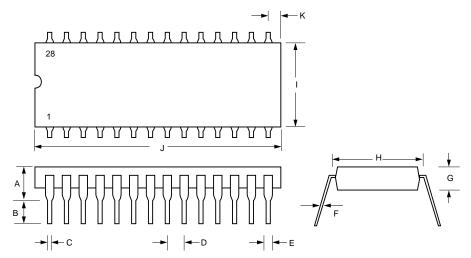
## PACKAGE OUTLINE

28-Lead Sidebrazed



		INCHE	ES	MILLIME	TERS
	SYMBOL	MIN	MAX	MIN	MAX
	Α	0.077	0.093	1.96	2.36
	В	0.016	0.020	0.41	0.51
	С	0.095	0.105	2.41	2.67
1	D		.050 typ	0.00	1.27
I	E	0.040	0.060	1.02	1.52
	F	0.215	0.235	5.46	5.97
	G	1.388	1.412	35.26	35.86
	Н	0.585	0.605	14.86	15.37
	1	0.009	0.012	0.23	0.30
	J	0.600	0.620	15.24	15.75

28-Lead Plastic DIP



	INCHES		MILLIME	TERS
SYMBOL	MIN	MAX	MIN	MAX
А		0.200		5.08
В	0.120	0.135	3.05	3.43
С		0.020		0.51
D		0.100		2.54
E		0.067		1.70
F		0.013		0.33
G	0.170	0.180	4.32	4.57
Н		0.622		15.80
I		0.555		14.10
J		1.460		37.08
К		0.085		2.16

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