# 500mA Single-Cell Li-Ion Battery Charger with Power Path Management

#### **GENERAL DESCRIPTION**

The SGM41562 is highly integrated, I<sup>2</sup>C programmable, single-cell Li-lon or Li-polymer battery charger with system power path management. It is specifically designed for portable applications requiring minimum board space and small external components. The charging profile includes pre-charge, constant-current and constant-voltage phases. Several safety and protection features are included such as built-in safe charge timer to set maximum duration of charge and pre-charge, input voltage and current monitoring, internal (junction) and external (battery) temperature monitoring, input current limiting and load current limiting. SGM41562 can charge with an input voltage range up to 6V charging range.

The SGM41562 has 3 power ports: input power port (IN), battery port (BAT) and system or load port (SYS). The system is powered from the input whenever it is available. Input is typically a USB power source. If the input source is weak or removed, power source for the system will automatically switch to the battery. The voltage and currents from input and battery as power sources are continuously monitored to prevent battery damage due to excessive currents or over-discharge.

I<sup>2</sup>C serial interface is used to program the device functions and parameters or to read its status. 12 read/write or read only 8-bit registers (REG00 to REG0B) are accessible. A watchdog protection feature is also included. If this feature is enabled and there is no in time read/write activity or signal from the host, the device will reset the charging parameters to their defaults and recycles power to the system (turn off/on) that may reset the host.

The SGM41562 goes into voltage protection state if  $V_{\text{IN}} > 6V$ . The input changes are continuously monitored and a system power recycle (SYS) may occur if the system does not response to the input toggles.

The SGM41562 is available in a Green WLCSP-1.47×1.47-9B package. Device functionality and protection features are assured in the ambient temperature range from -40°C to +125°C. Charging parameters are guaranteed in 0°C to +55°C.

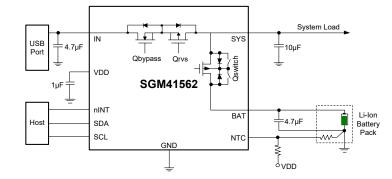
#### **FEATURES**

- Fully Autonomous Charger for Single-Cell Li-lon and Li-Polymer Battery
- 21V Maximum Input Voltage Rating with Over-Voltage Protection
- ±0.6% Charging Voltage Accuracy
- I<sup>2</sup>C Interface for Parameters Setting/Status Reporting
- Fully Integrated Power Switches
- No External Blocking Diode Required
- Built-in Robust Charge Protections Including Battery
   Temperature Monitor and Programmable Timer
- Battery or PCB Over-Temperature Protection
- System Reset Function
- Built-in Battery Disconnection Function
- Thermal Limit Regulation on Chip
- Available in a Green WLCSP-1.47×1.47-9B Package

#### **APPLICATIONS**

Wearable Devices IoT Gadgets

#### TYPICAL APPLICATION



**Figure 1. Typical Application Circuit** 

#### PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION	
SGM41562	WLCSP-1.47×1.47-9B	-40°C to +125°C	SGM41562XG/TR	XXXXX 41562	Tape and Reel, 3000	

#### MARKING INFORMATION

NOTE: XXXXX = Date Code, Trace Code and Vendor Code.



Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

#### **ABSOLUTE MAXIMUM RATINGS**

IN	0.3V to 21V
SYS	0.3V to 5.3V (5.5V for 500µs)
All Other Pins to GND	0.3V to 6V
I <sub>INCLAMP</sub>	5mA
Package Thermal Resistance	
WLCSP-1.47×1.47-9B, θ <sub>JA</sub>	95°C/W
Junction Temperature	+150°C
Storage Temperature Range.	65°C to +150°C
Lead Temperature (Soldering	, 10s)+260°C
ESD Susceptibility	
HBM	3000V
CDM	1000V

#### RECOMMENDED OPERATING CONDITIONS

Supply Voltage, V <sub>IN</sub>	4.35V to 5.5V (Charging)
I <sub>IN</sub>	Up to 500mA
I <sub>BAT</sub>	Up to 3.2A
Існ	Up to 456mA
V <sub>BAT_REG</sub>	Up to 4.545V
<b>Operating Junction Temperature</b>	Range40°C to +125°C

#### **OVERSTRESS CAUTION**

Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

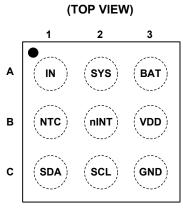
#### **ESD SENSITIVITY CAUTION**

This integrated circuit can be damaged if ESD protections are not considered carefully. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because even small parametric changes could cause the device not to meet the published specifications.

#### **DISCLAIMER**

SG Micro Corp reserves the right to make any change in circuit design, or specifications without prior notice.

## **PIN CONFIGURATION**



WLCSP-1.47×1.47-9B

## **PIN DESCRIPTION**

PIN	NAME	TYPE	FUNCTION
A1	IN	Р	Input Power Pin. Place a minimum 2.2μF ceramic capacitor between IN pin and GND pin as close as possible to these pins.
A2	SYS	Р	System Power Supply Output. Place a ceramic capacitor between SYS pin and GND pin as close as possible to these pins.
А3	BAT	Р	Battery Positive Terminal Connection Pin. Place a ceramic capacitor between BAT pin and GND pin as close as possible to the device. Connect the negative battery terminal to power GND.
B1	NTC	AIO	Battery Temperature Sense Input. Connect a negative temperature coefficient thermistor between this pin and GND pin. NTC is usually placed in touch with battery pack. Hot-cold temperature window can be programmed by a resistor divider network placed between VDD to NTC to GND pins. Charging will suspend if NTC function is enabled and NTC pin voltage goes out of the V <sub>HOT</sub> and V <sub>COLD</sub> range.
B2	nINT	DIO	Interrupt Output. The nINT pin can send a charging status and fault interrupt signal to the host. nINT is also used to disconnect the system from the battery. Pull nINT pin from high to low for > $t_{RST\_DGL}$ (16s default). The battery FET turns off and turns on again automatically after > $t_{RST\_DUR}$ (4s default) regardless of the nINT state. Both $t_{RST\_DGL}$ and $t_{RST\_DUR}$ can be programmed via the $l^2C$ interface.
В3	VDD	Р	Internal Power Supply Pin. Connect a minimum 0.1µF decoupling ceramic capacitor from this pin to GND. External load current on this pin should not exceed 1mA.
C1	SDA	DIO	$I^2C$ Bus Data. A $10k\Omega$ pull-up to the logic-high rail should be used on SDA line.
C2	SCL	DI	I <sup>2</sup> C Bus Clock. A 10kΩ pull-up to the logic-high rail should be used on SCL line.
C3	GND	_	Ground Pin of the Device.

NOTE:

AIO = Analog Input and Output; DI = Digital Input; DO = Digital Output; DIO = Digital Input and Output; P = Power.

# **ELECTRICAL CHARACTERISTICS**

 $(T_A = +25^{\circ}C, V_{IN} = 5V \text{ and } V_{BAT} = 3.5V, \text{ unless otherwise noted.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Source and Battery Protection	•					•
Input Under-Voltage Lockout Threshold	$V_{IN\_UVLO}$	Input falling	3.45	3.65	3.88	V
V <sub>IN_UVLO</sub> Threshold Hysteresis	V <sub>IN_UVLO_HYS</sub>	Input rising		105		mV
Input Over-Voltage Protection Threshold	$V_{IN\_OVLO}$	Input rising threshold	5.75	6	6.27	V
V <sub>IN_OVLO</sub> Threshold Hysteresis	V <sub>IN_OVLO_HYS</sub>			290		mV
Input Clamp Voltage	V <sub>IN_CLAMP</sub>	Test for having 1mA clamp current	19.5	21		V
Input vs. Battery Voltage Headroom Threshold	V <sub>HDRM</sub>	Input rising vs. battery		100		mV
V <sub>HDRM</sub> Threshold Hysteresis	V <sub>HDRM_HYS</sub>	Input vs. battery voltage headroom threshold hysteresis		150		mV
Input Power Detection Time	t <sub>PWD</sub>	Wait time before sending interrupt pulse for reporting input power new status	60	70	82	ms
nINT Output Pulse Duration	t <sub>INT_PULSE</sub>			250		μs
BAT Pin Input Voltage	$V_{BAT}$				4.5	V
		V <sub>BAT</sub> falling, VBAT_UVLO[2:0] = 000	2.30	2.40	2.66	V
Battery Under-Voltage Lockout Threshold	V <sub>BAT_UVLO</sub>	V <sub>BAT</sub> falling, VBAT_UVLO[2:0] = 100	2.69	2.76	2.86	
		V <sub>BAT</sub> falling, VBAT_UVLO[2:0] = 111	2.95	3.00	3.14	
Battery Under-Voltage Threshold Hysteresis	V <sub>BAT_UVLO_HYS</sub>	V <sub>BAT_UVLO</sub> = 2.76V		210		mV
Battery Over-Voltage Protection Threshold	$V_{BAT\_OVP}$	Rising, higher than V <sub>BAT_REG</sub>		100		mV
Power Path Management						
		$V_{IN}$ = 5.5V, $R_{SYS}$ = 100 $\Omega$ , $I_{CHG}$ = 0A, VSYS_REG[3:0] = 0000, $V_{SYS}$ REG = 4.2V	4.158	4.2	4.242	
Regulated System Output Voltage Accuracy	V <sub>SYS_REG_ACC</sub>	$V_{IN}$ = 5.5V, $R_{SYS}$ = 100 $\Omega$ , $I_{CHG}$ = 0A, VSYS_REG[3:0] = 1001, $V_{SYS}$ REG = 4.65V	4.603	4.65	4.696	V
		$V_{IN}$ = 5.5V, $R_{SYS}$ = 100 $\Omega$ , $I_{CHG}$ = 0A, VSYS_REG[3:0] = 1111, $V_{SYS_REG}$ = 4.95V	4.900	4.95	4.999	
		IIN_LIM[3:0] = 0000, I <sub>IN_LIM</sub> = 50mA		50		
Input Current Limit	I <sub>IN LIM</sub>	IIN_LIM[3:0] = 0011, I <sub>IN_LIM</sub> = 140mA		140		mA
mpat Ganoni Zimit	'IN_LIM	IIN_LIM[3:0] = 1001, I <sub>IN_LIM</sub> = 320mA		320		
		IIN_LIM[3:0] = 1111, I <sub>IN_LIM</sub> = 500mA		500		
		VIN_MIN[3:0] = 0000, V <sub>IN_MIN</sub> = 3.88V	3.710	3.88	4.060	
Input Minimum Voltage Regulation	$V_{IN\_MIN}$	VIN_MIN[3:0] = 1001, V <sub>IN_MIN</sub> = 4.60V	4.405	4.60	4.815	V
		VIN_MIN[3:0] = 1111, V <sub>IN_MIN</sub> = 5.08V	4.870	5.10	5.320	
IN to SYS Switch On-Resistance	R <sub>ON_Q1</sub>	V <sub>IN</sub> = 4.5V, I <sub>SYS</sub> = 100mA		235		mΩ
Input Quiescent Current	I <sub>IN Q</sub>	$V_{IN}$ = 5.5V, EN_HIZ = 0, CEB = 0, charge enable, $I_{CHG}$ = 0A, $I_{SYS}$ = 0A		220	265	μΑ
	IN_Q	V <sub>IN</sub> = 5.5V, EN_HIZ = 0, CEB = 1, charge disabled		180	220	F*, ,
Input Suspend Current	I <sub>IN_SUSP</sub>	$V_{IN}$ = 5.5V, EN_HIZ = 1, CEB = 0, charge enable		50	75	μA

# **ELECTRICAL CHARACTERISTICS (continued)**

 $(T_A = +25^{\circ}C, V_{IN} = 5V \text{ and } V_{BAT} = 3.5V, \text{ unless otherwise noted.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
		$V_{IN} = 5V$ , CEB = 0, $I_{SYS} = 0A$ , $V_{BAT} = 4.3V$ ,		15			
	I <sub>BAT_Q</sub>	charge complete $\begin{aligned} &V_{\text{IN}} = \text{GND, CEB} = 1, \text{VDD\_GATE} = 1, \\ &\text{FET\_DIS} = 0, \text{EN\_SHIP\_DGL}[1:0] \neq 11, \\ &I_{\text{SYS}} = 0\text{A, } V_{\text{BAT}} = 4.35\text{V, disable external NTC} \\ &\text{circuit driving} \end{aligned}$		10	40		
Battery Quiescent Current		$V_{\text{IN}}$ = GND, CEB = 1, $I_{\text{SYS}}$ = 0A, $V_{\text{BAT}}$ = 4.35V, enable PCB OTP function, excluding the external NTC bias.		11		μΑ	
		$V_{\text{IN}}$ = GND, CEB = 1, $I_{\text{SYS}}$ = 0A, $V_{\text{BAT}}$ = 4.35V, enable PCB OTP function and watchdog, excluding the NTC bias.		27			
		V <sub>BAT</sub> = 4.5V, IN is open or grounded, shipping mode		0.7	1.2		
Battery FET On-Resistance	R <sub>ON_Q2</sub>	V <sub>IN</sub> < 2V, V <sub>BAT</sub> = 3.5V, I <sub>SYS</sub> = 100mA		100		mΩ	
Datter FFT Dischause Comment Limit		IDSCHG[3:0] = 0001, I <sub>DSCHG</sub> = 400mA	295	400	545	Л	
Battery FET Discharge Current Limit	I <sub>DSCHG</sub>	IDSCHG[3:0] = 1001, I <sub>DSCHG</sub> = 2000mA		2000		mA	
Delay before Discharge Over Current Cut	t <sub>DSCHG_CUT</sub>	Delay after discharge OC detection and before turning switch off		64		μs	
Delay before Retry after Cut	t <sub>RETRY</sub>	Turn on retry delay after OC turn off		800		μs	
Ideal Diode Forward Voltage in Supplement Mode (BAT to SYS)	$V_{\text{FWD}}$	10mA discharge current		10		mV	
Shipping Mode							
Enter to Shipping Mode Deglitch Delay Time after Programming the Shipping Mode	t <sub>SMEN_DGL</sub>	FET_DIS is set from 0 to 1, EN_SHIP_DGL[1:0] = 00		1		s	
Exit Shipping Mode Delay (Initiated by nINT pin or V <sub>IN</sub> Plug-in)	t <sub>SMEX_DGL</sub>	nINT pin is pulled low		2		s	
Auto-Reset Mode							
Reset and Power Recycle	t <sub>RST DGI</sub>	tRST_DGL[1:0] = 00		8		s	
by nINT Pin is Pull Down		tRST_DGL[1:0] = 10		16			
Battery FET Off-Time Duration after Reset	t <sub>RST_DUR</sub>	tRST_DUR = 0		2		e	
Battery F E F On-Time Buration after Neset	4RSI_DUR	tRST_DUR = 1		4		s	
Battery Charger							
		VBAT_REG[5:0] = 000000, V <sub>BAT_REG</sub> = 3.6V	3.564	3.600	3.636		
Battery Charge Regulation Voltage	V	VBAT_REG[5:0] = 101000, V <sub>BAT_REG</sub> = 4.2V	4.175	4.200	4.225	V	
Dattery Charge Regulation Voltage	$V_{BAT\_REG}$	VBAT_REG[5:0] = 110100, V <sub>BAT_REG</sub> = 4.38V	4.354	4.380	4.406	V	
		VBAT_REG[5:0] = 1111111, V <sub>BAT_REG</sub> = 4.545V	4.518	4.545	4.572		
		ICC[5:0] = 000000, I <sub>CC</sub> = 8mA	4	8	10		
Chargo Current		ICC[5:0] = 001100, I <sub>CC</sub> = 96mA	84	96	106	mΛ	
Charge Current	Icc	ICC[5:0] = 100000, I <sub>CC</sub> = 264mA	242	264	300	mA	
		ICC[5:0] = 111000, I <sub>CC</sub> = 456mA	390	456	520	1	
Junction Temperature Population	т.	I <sup>2</sup> C programmable range	60		120	°C	
Junction Temperature Regulation	$T_{J_REG}$	TJ_REG[1:0] = 11, T <sub>J_REG</sub> = 120°C		120			
		ITERM[3:0] = 0000, I <sub>TERM</sub> = I <sub>PRE</sub> = 1mA	0.35	1	1.45		
Pre-Charge Current	1	ITERM[3:0] = 0001, I <sub>TERM</sub> = I <sub>PRE</sub> = 3mA	1.25	3	3.9	m۸	
r re-onarge ourrent	I <sub>PRE</sub>	ITERM[3:0] = 0101, I <sub>TERM</sub> = I <sub>PRE</sub> = 11mA	5.5	11	14.5	mA	
		ITERM[3:0] = 1111, I <sub>TERM</sub> = I <sub>PRE</sub> = 31mA	17	31	42	1	

# **ELECTRICAL CHARACTERISTICS (continued)**

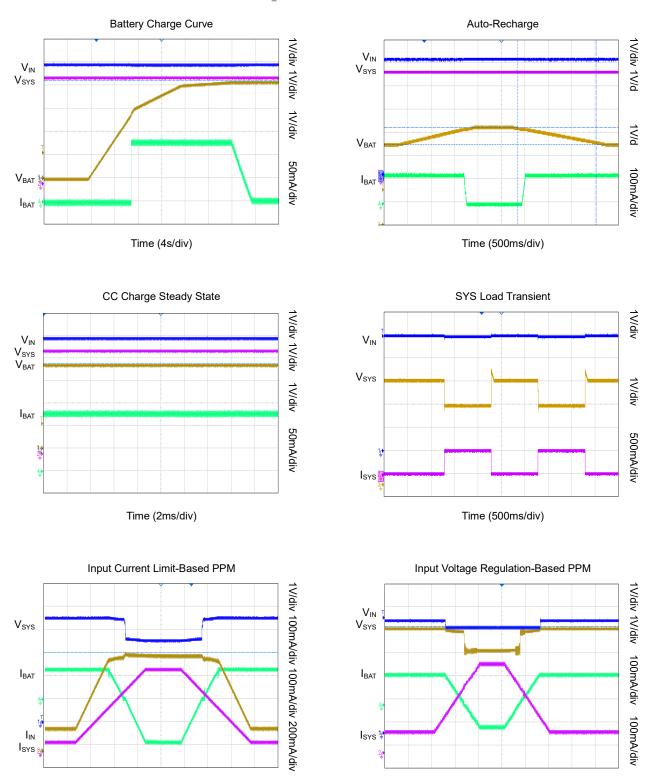
 $(T_A = +25^{\circ}C, V_{IN} = 5V \text{ and } V_{BAT} = 3.5V, \text{ unless otherwise noted.})$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		ITERM[3:0] = 0000, I <sub>TERM</sub> = 1mA	0.75	1	1.15	
Channa Tamain ation Commant Thursday	,	ITERM[3:0] = 0001, I <sub>TERM</sub> = 3mA	2	3	4	mA
Charge Termination Current Threshold	I <sub>TERM</sub>	ITERM[3:0] = 0101, I <sub>TERM</sub> = 11mA	7	11	15	
		ITERM[3:0] = 1111, I <sub>TERM</sub> = 31mA	22	31	42	
Termination Deglitch Time	t <sub>TERM_DGL</sub>			200		ms
Pre-Charge to Fast Charge Threshold	V <sub>BAT_PRE</sub>	V <sub>BAT</sub> Rising, VBAT_PRE = 1, V <sub>BAT_PRE</sub> = 3V	2.9	3	3.1	V
Pre-Charge to Fast Charge Threshold Hysteresis	V <sub>BAT_PRE_HYS</sub>			90		mV
Pottory Auto Bookergo Voltago Drop Throshold	V	Below $V_{BAT\_REG}$ , $VRECH = 0$	110	135	155	mV
Battery Auto-Recharge Voltage Drop Threshold	$V_{RECH}$	Below V <sub>BAT_REG</sub> , VRECH = 1	210	240	275	IIIV
Battery Auto-Recharge Deglitch Time	t <sub>RECH_DGL</sub>			200		ms
Thermal Protection						
Thermal Shutdown Threshold	$T_{J\_SHDN}$			150		°C
Thermal Shutdown Hysteresis				20		°C
NTC Pin Output Current	I <sub>NTC</sub>	CEB = 0, NTC = 3V	-150		150	nA
NTC Cold Temp Rising Threshold	V <sub>COLD</sub>	As percentage of V <sub>DD</sub>	63	65	67	%
NTC Cold Temp Rising Threshold Hysteresis				30		mV
NTC Hot Temp Falling Threshold	$V_{HOT}$	As percentage of V <sub>DD</sub>	30	33	35	%
NTC Hot Temp Falling Threshold Hysteresis				70		mV
NTC Hot Temp Falling Threshold for PCB OTP	V <sub>HOT_PCB</sub>	As percentage of V <sub>DD</sub>	30	32	34	%
NTC Hot Temp Falling Threshold Hysteresis for PCB OTP				90		mV
Logic IO Pin Characteristics						
Low Logic Voltage Threshold	$V_L$				0.4	V
High Logic Voltage Threshold	$V_{H}$		1.4			V
I <sup>2</sup> C Interface (SDA, SCL)						
Input Low Logic Voltage Threshold	V <sub>IL</sub>				0.4	V
Input High Logic Voltage Threshold	V <sub>IH</sub>		1.4			V
Output Low Threshold Level	V <sub>OL</sub>	I <sub>SINK</sub> = 5mA			0.2	V
I <sup>2</sup> C Clock Frequency	f <sub>SCL</sub>				400	kHz
Clock Frequency and Watchdog Timer						
Watchdog Timer	t <sub>wdt</sub>	WATCHDOG[1:0] = 11		160		s

## TYPICAL PERFORMANCE CHARACTERISTICS

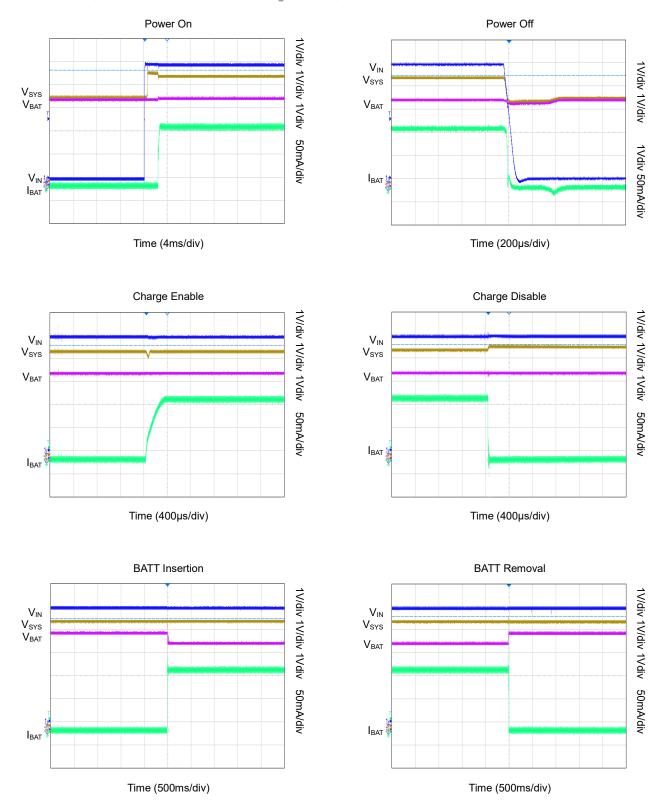
Time (4s/div)

 $T_A$  = +25°C,  $V_{IN}$  = 5V,  $I_{IN}$  = 500mA,  $I_{CC}$  = 128mA and  $V_{IN\_MIN}$  = 4.6V, unless otherwise noted.



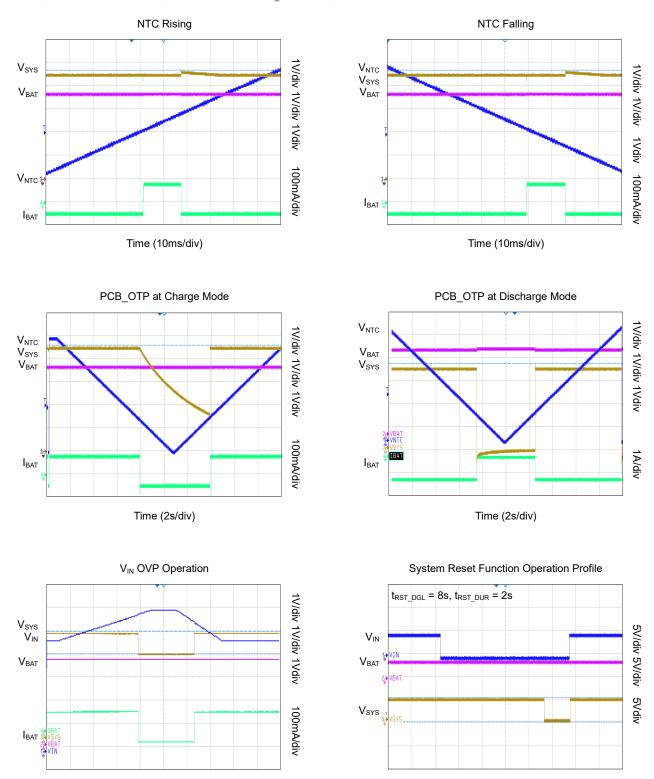
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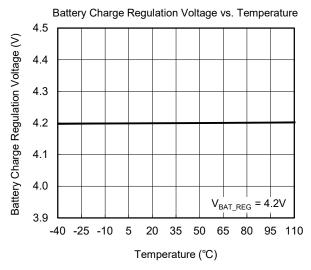
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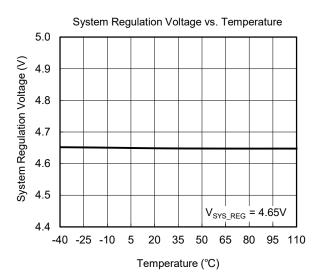
Time (2s/div)

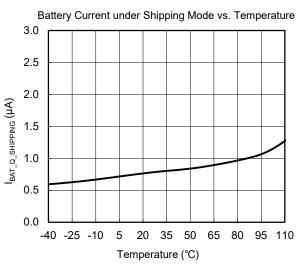


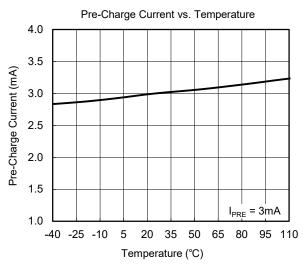
Time (2s/div)

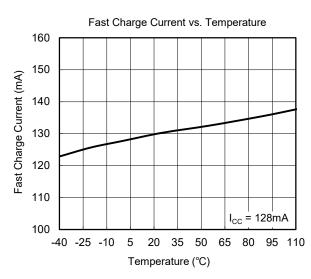
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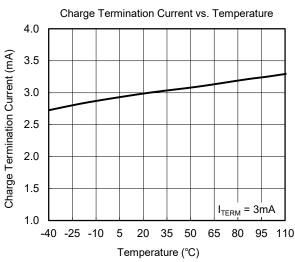




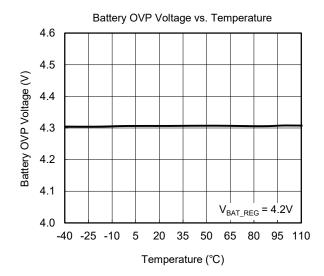


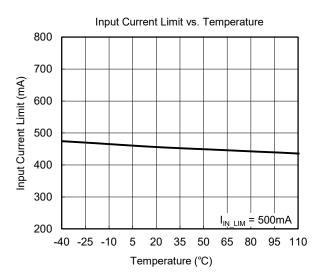


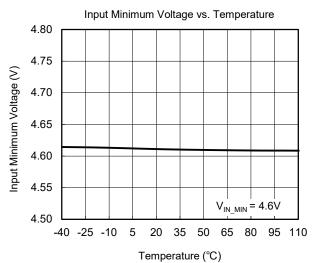




 $T_A$  = +25°C,  $V_{IN}$  = 5V,  $I_{IN}$  = 500mA,  $I_{CC}$  = 128mA and  $V_{IN\_MIN}$  = 4.6V, unless otherwise noted.







## **FUNCTIONAL BLOCK DIAGRAM**

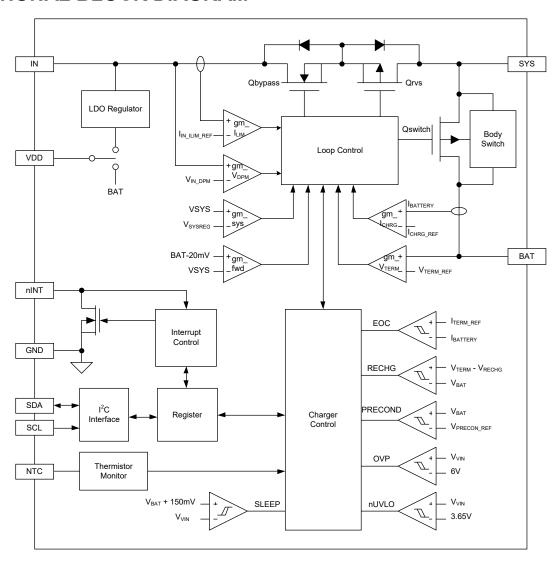


Figure 2. Functional Block Diagram

#### **DETAILED DESCRIPTION**

#### Introduction

The SGM41562 is a single-cell battery charger with power path management function for Li-lon and Li-polymer batteries. The charge features include pre-charge, fast charge including constant-current mode (CCM) and constant-voltage mode (CVM), end-of-charge termination, auto-recharge, and a built-in safe charge timer. The safe charge timer is used to prevent over-charging or other issues if the host runs out of control.

A bypass switch between IN and SYS pins, and a battery switch between SYS and BAT pins are integrated to provide complete power path management (PPM). The switches have low on-resistances to minimize loss and heat. System load is primarily powered from the input when it is available, and the remaining input power is used to charge the battery if needed. When the input source is weak, the load is powered partially from the battery. This mode in which the battery provides the power deficit, is called supplement mode. Battery will provide the full load power if input is removed or if V<sub>IN</sub> is out of range. For battery charging, the power to the battery is regulated by the battery switch. To prevent faulty charge conditions, input voltage, input current, system voltage, chip temperature and external temperature (sensed by NTC) are continuously monitored during charge.

Figure 3 shows the power paths and key internal blocks of the device. The Qbypass switch regulates the voltage of the system and the internal charge circuit. The Qrvs switch acts as a near ideal blocking diode to prevent reverse power (or leakage) from the load (SYS pin) back to the input (IN pin). The Qswitch switch is responsible for battery charging regulation and connecting or disconnecting of the battery (BAT pin) to the system (SYS pin). The charge and discharge circuits in the Figure 3 that are connected to the IN and BAT pins have their own independent UVLO and power supply. The rest of the chip is powered by either IN or SYS pin, whichever has the higher voltage. The I/F interface (I<sup>2</sup>C communication and nINT) block is active whenever any of the power sources (IN or BAT pin) are available.

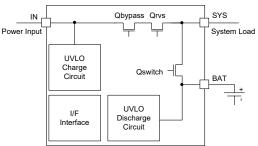


Figure 3. Power Path Management Structure

The chip has a watchdog timer as a protective feature against unexpected host malfunctions. When watchdog timer is enabled, it must be reset by host regularly to prevent watchdog timer overflow that results in a chip reset and power recycle. Watchdog reset is by writing into the watchdog register through I<sup>2</sup>C interface (I/F). If the watchdog is not reset on time, the power to the host will recycle.

The power fed to the SYS pin is recycled when watchdog times out, the host does not response to IN power input (when watchdog is forced on) or COLD\_RESET bit is set to 1, to clear the running environment before system program upgrade or release from locked situations.

#### **Input Detection**

Figure 4 shows how the input voltage status is detected and affects the device function along with the relevant timings and nINT output signal updates. The device continuously monitors the input voltage at the IN node. The SYS node and charge circuit is only started and connected to the input when for a duration of  $t_{\rm INI}$ ,  $V_{\rm IN}$  is within its normal range (above  $V_{\rm IN\_UVLO}$  and below  $V_{\rm IN\_OVLO}$ ). Qbypass and Qrvs switches will turn off as soon as an input UVLO or OVLO is detected.

As shown in Figure 4 any input state is considered stable if it continuously stays in the same condition for a duration of  $t_{PWD}$  after which the device sends out a negative pulse to the nINT pin with a pulse width of  $t_{INT\_PULSE}$  to inform the host about the input state change.

The watchdog timer WATCHDOG[1:0] register is set to 01 once the valid input is detected and when an INT pulse is asserted, which resumes its original setting when any writing to this device occurs. If the host does not clear the watchdog, power to the host is recycled for reset when watchdog runs time out.

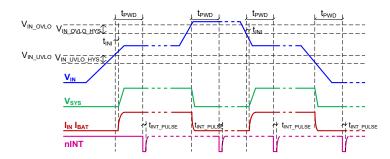


Figure 4. Input Power Detection and nINT Signaling Timings

#### **Power Path Management**

When the input voltage is normal and have enough headroom for powering the system ( $V_{IN} > V_{IN\_UVLO}$  and  $V_{IN} - V_{SYS} > V_{HDRM}$ ), the input power path will conduct and the device starts to power the system from input by setting the system voltage to  $V_{SYS\_REG}$ .  $V_{SYS\_REG}$  is selected by programming VSYS\_REG[3:0] register, the lower 4 bits of REG07 (also called system voltage register or VSYS\_REG[3:0] register). However, the actual system voltage ( $V_{SYS}$ ) can be affected by the input voltage level, input current limit and battery voltage.

I<sup>2</sup>C commands can directly control the power paths. Input path will be disconnected (high-impedance) by turning off Qbypass switch if the EN\_HIZ bit is set to 1. If the battery is getting charge and Qswitch switch is on, it can also be disconnected by setting charge enable bit, set the CEB bit to 1 (turn off Qswitch switch in charge direction). The power path control bits are explained in Table 1. When these bits are clear, they have no effect.

Table 1. Switch Control by I<sup>2</sup>C Interface

FETs	EN_HIZ = 1	CEB = 1
Qbypass	Off	X
Qswitch (Charging)	X	Off
Qswitch (Discharging)	X	X

NOTE: X = Don't Care.

#### **Battery Charge Profile**

Figure 5 shows the battery charge profile used in this device. The charge phases are explained below. Depending on the I<sup>2</sup>C settings and the battery state of charge (SOC), some or all of the phases may be skipped or used to finish a complete charge cycle as explained below:

**Pre-Charge**: If the battery voltage is less than the pre-charge threshold ( $V_{BAT\_PRE}$ ), the battery is charged with the small pre-charge current ( $I_{PRE}$ ). The pre-charge current value is the same as the termination current ( $I_{TERM}$ ) that is programmed via bit D[3:0] of the REG03, also called ITERM[3:0].

**Constant-Current Charge**: When battery voltage is higher than  $V_{BAT\_PRE}$ , and less than  $V_{BAT\_REG}$ , it will be charged with a constant current. The constant-current value is determined by bit D[5:0] of the REG02 that is called ICC[5:0] and a single scaling bit that if set, multiplies it by 1⁄4. This bit is used for finer CC adjustment (CC FINE bit in REG0A).

**Constant-Voltage Charge**: When the battery voltage reaches to the  $V_{BAT\_REG}$ , the voltage is kept constant and the charge current drown by battery will start to fall. The  $V_{BAT\_REG}$  value is determined by bit D[7:2] of the REG04 that is also called VBAT\_REG[5:0].

Charge Termination: A charge termination is recognized when the charge current drops to a small value represented by I<sub>TERM</sub>. If the termination detection is enabled by setting the EN\_TERM bit in REG05 D[4] to 1, then if the charge current (I<sub>CHG</sub>) stays equal or lower than I<sub>TERM</sub> for a period of t<sub>TERM\_DGL</sub> (termination deglitching time) the charge cycle is considered complete and charging current will be turned off and drop to zero. With no termination, the charge current will continue to drop. Note that a charge cycle is also considered complete and charging will be turned off, if the safe timer function runs out of time provided that the safe timer function is already enabled by setting EN\_TIMER bit in REG05 D[3] to 1.

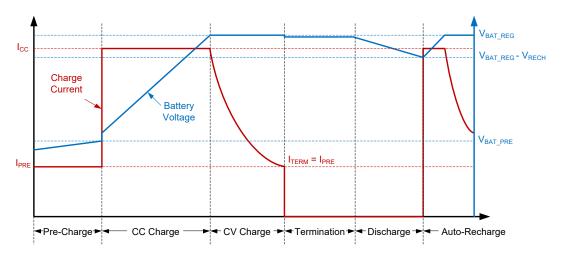


Figure 5. Battery Charge Profile

The charge status is updated to "charge complete" once the termination condition is detected. The charge current will be terminated when termination conditions are met and if the TERM\_TMR bit is set to 0 (REG05 D[0] = 0); The charge will not terminate and current keeps decreasing if TERM\_TMR bit is 1.

During the whole charging process, the actual charge current may fall below the set values due to the other regulations or controls such as dynamic power management (DPM) regulation caused by insufficient input voltage or current or due to thermal regulation. In thermal regulation the device reduces the power path currents to keep junction temperature below the programmed limit.

A new charge cycle starts when one of the following conditions occurs:

- The input power recycles (input on/off).
- Battery charging is enabled by I<sup>2</sup>C command.
- Auto-recharge kicks in due to battery charge state.

If all the following conditions are satisfied:

- · No NTC thermistor temperature fault.
- No safety (charge) timer fault.
- No battery over voltage event.
- The Qswitch switch is not forced to turn off (e.g. CEB = 1).

#### **Battery Over-Voltage Protection**

SGM41562 has a built-in battery over-voltage protection limit. A battery over-voltage event is detected when battery voltage is higher than  $V_{\text{BAT\_OVP}} + V_{\text{BAT\_REG}}$ . When this event occurs, the charging is immediately suspended and a fault is asserted. The discharging path will be turned on if battery over-voltage condition does not clear and continues.

# Input Current and Input Voltage Based Power Management

Usually the input source (typically USB) is not strong enough for all system power demands and a power management scheme is needed to keep the system voltage in desired level without over loading the source. Figure 6 shows the power management profile and explains how it is implemented in SGM41562 including the battery assist operation (supplement) when input source is not able to provide required power.

The input current is continuously monitored to make sure the input source maximum current limit specification is met. The total input current limit is programmable by I<sup>2</sup>C and is used to prevent over loading of the input source.

If the input source is weak and the programmed input current limit is higher than the effective capability of the source (like in a dynamic loading condition) the back-up power management will come in effect to prevent over loading of the input source. The back-up power management is based on limiting the input voltage drop to  $V_{\mathsf{IN\_MIN}}$  value (programmable). The voltage based dynamic power management (DPM) will regulate the input voltage to  $V_{\mathsf{IN\_MIN}}$  when the load is higher than the input current capacity. If input current and voltage limit are both reached, then the Qbypass switch (between IN and SYS pins) will regulate and limit the total power taken from the input. With the power limiting, if the system voltage drops to the minimum value of ( $V_{\mathsf{SYS\_REG}}$  - 90mV) or the input voltage falls below ( $V_{\mathsf{IN}}$  - 160mV), the device will finally reduce the charge current to prevent further voltage drops.

The programmed  $V_{\text{IN\_MIN}}$  must be at least 250mV higher than  $V_{\text{BAT\_REG}}$  to assure stable operation of the regulator.

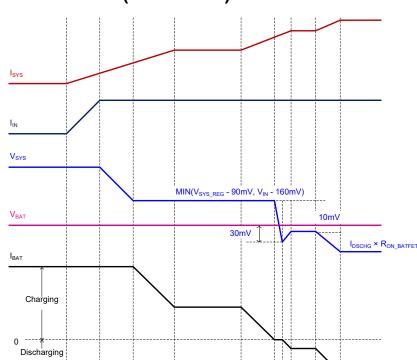


Figure 6. Dynamic Power Management and Battery Supplement Operation Profile

#### **Battery Supplement Mode**

As mentioned above, the DPM will reduce the charge current to keep the input current or voltage in regulation when source power is not sufficient for system demand. If the charge current is reduced to zero but still due to heavy system load the input source is overloaded and  $V_{SYS}$  continues to drop, then the battery will supply the deficit to assist the input source. This mode is called battery supplement mode in which the battery provides I<sub>DSCHG</sub> as supplement current to the load. This mode starts when the system drop reaches to 30mV below the battery voltage. In this mode the Qswitch switch acts as a near ideal diode from battery to the system. The Qswitch switch is controlled to regulate and maintain the  $V_{BAT}$  -  $V_{SYS}$  drop to a fixed 10mV value when  $I_{DSCHG}$  × Ron BATFET is less than 10mV. If IDSCHG × RON BATFET is larger than 10mV, the Qswitch switch is fully turned on to pass battery voltage to the system with minimum drop.

In the battery supplement mode the ideal diode mode will be disabled as soon as the system load decreases and  $V_{\text{SYS}}$  exceeds the  $V_{\text{BAT}}$  + 20mV value.

When  $V_{\text{IN}}$  source is not available, the device operates in discharge mode (battery power) in which the Qswitch switch is always fully on to reduce the losses.

#### **Battery Regulation Voltage**

The battery voltage for the constant-voltage regulation phase (CV) is represented by  $V_{\text{BAT}}$  REG.

#### Thermal Regulation and Shutdown

SGM41562 continuously monitors its internal junction temperature to avoid junction overheating while keeping the power delivery at its maximum. When the internal junction temperature reaches its programmable limit (T<sub>J\_REG</sub>), the device starts to reduce the charge current to prevent higher power dissipation. The thermal regulation limit is programmable to help adjusting the design for the thermal requirements in different applications. 4 different junction temperature regulation thresholds can be chosen by programming the TJ\_REG[1:0] register.

The device fixed thermal shutdown limit ( $T_{J\_SHDN}$ ) is slightly higher than the highest programmable  $T_{J\_REG}$ . If  $T_{J}$  rises above this limit, both Qbypass and Qswitch switches will turn off.

#### **NTC Function and VDD Gating**

The NTC pin is provided to sense the battery temperature using an NTC thermistor. Thermistors are usually included in the rechargeable battery packs to ensure safe operation by monitoring the battery temperature and making sure it is between hot and cold limits. To adjust the temperature limits for the device, two resistors (R<sub>T1</sub> and R<sub>T2</sub> in Figure 13) should be connected to NTC pin as a divider between VDD and GND pins. The thermistor itself is connected between NTC pin and GND. The voltage on the NTC pin is determined by all three resistors. This resistor divider along with the hot and cold limit voltages defined in the EC table determines the hot-cold operating window. Note that due to the negative temperature coefficient of NTC, when its voltage drops below V<sub>HOT</sub>, it means the battery temperature is exceeding the hot limit. The NTC protection function can be disabled by clearing the EN\_NTC bit to 0. The default settings for NTC function are the PCB OTP levels specified in EC table that can be change by I<sup>2</sup>C as explained in Table 2.

**Table 2. NTC Function Selection** 

I <sup>2</sup> C Ce	I <sup>2</sup> C Control				
EN_NTC	EN_PCB OTP	Function			
0	don't care	Disable			
1	1	NTC			
1	0	PCB OTP			

NTC function only works in charge mode. When NTC pin voltage falls out of the hot-cold window it means that the temperature is outside the safe operating range and results in a pause in charging and sets the fault bits. Charging will resume when the temperature falls back into the safe range.

If DIS\_VDD bit is disabled and  $V_{IN}$  is removed,  $V_{DD}$  power turns off and becomes high-impedance leaving only  $R_{T2}$  in parallel with the NTC thermistor. If DIS\_VDD bit is enabled,  $V_{DD}$  remains active.  $V_{DD}$  uses battery power if  $V_{IN}$  is removed.

With PCB OTP selected, if the NTC pin voltage is lower than the NTC hot threshold, Qbypass and Qswitch switches will turn off. The PCB OTP fault also will set the NTC\_FAULT status bit to 1. The operation will resume when the NTC pin voltage goes back above the NTC hot threshold.

#### **Safety Timer**

Using an internal safety timer, SGM41562 is capable to limit the maximum duration of the pre-charge and charge periods to avoid extended charging cycles that may happen due to abnormal battery conditions. This protection can be disabled by I<sup>2</sup>C. The safety timer starts counting if one of the following occurs:

- A new charge cycle is started.
- Write in REG01 D[3] bit: from 1 to 0 (charge enable)
- Write in REG05 D[3] bit: from 0 to 1 (safety timer enable)
- Write in REG02 D[7] bit: from 0 to 1 (software reset)
- ullet Write in REG0A D[4] bit: from 0 to 1 (software power recycle) The safety time limit is 1 hour for pre-charge condition in which the battery voltage stays lower than  $V_{BAT\_PRE}$  and cannot go higher. For the charge phase the time limit is programmable through  $I^2C$  and the safety timer starts counting when the battery enters in constant-current charge mode or constant-voltage charge mode.

#### **Host Mode and Default Mode**

SGM41562 can operate in either default mode (with default parameters) or host mode (parameters programmed by host). It will go to the default mode if one of the following occurs:

- · Input refresh with no battery connected.
- Re-insert battery with no input source connected.
- Device registers reset by writing 1 to REG RST bit.
- Watchdog timer expiry.

Upon a power on reset, the device starts in default mode and in the same state as if watchdog timer expiration has occurred. In this mode all registers take their default values, including EN\_HIZ = 0 and CEB = 1, that means the input power path is enabled and device is set to battery discharge mode. Note that by default the battery will not be charged after a reset.

When the device is in the host mode, watchdog function can be activated and works in both charge and discharge modes (Watchdog timer is independent of the charge safety timer). Watchdog timer can be enabled by programming a non-zero expiry time in its register, that is WATCHDOG[1:0]  $\neq$  00. If watchdog timer is enabled, it must be reset regularly before it runs out of time by writing 1 to WD\_RST bit in REG02. Otherwise the watchdog timer will expire and results in a power recycle to the system. Therefore, resetting the watchdog timer by host must happen in the intervals shorter than watchdog time limit. The power recycle is performed by turning off Qswitch and Qbypass for a duration of  $t_{RST\_DUR}$  and then turning them on again. After watchdog timer expiration, all registers will reset to their default values and the device goes to the default mode.

To reduce the quiescent current during discharge mode, the watchdog timer can be turned off by setting the EN\_WD\_DISCHG bit to 0. If the WATCHDOG[1:0] is set to 00, the watchdog timer is disabled under charge and discharge modes independent of the EN\_WD\_DISCHG bit value.

#### **Battery Discharge Function**

If the battery is connected ( $V_{BAT}$  is above the  $V_{BAT\_UVLO}$  threshold) and the input source is missing, the Qswitch turns fully on. The low on resistance of the Qswitch minimizes the conduction loss during discharge. The quiescent current of the device is as low as 11µA in this mode. By setting REG0A D[3] bit to 1, the Qswitch will stay on even if the rest of the internal blocks are turned off, to reduce the device quiescent current to less than 1.2µA. The low on-resistance and low quiescent current of the device extend the run time.

#### **Over-Discharge Current Protection**

The over-discharge current protection is effective in discharge mode and supplement mode. If the  $I_{BAT}$  exceeds discharge current limit value programmed in the REG03 D[7:4], the Qswitch turns off after a wait delay ( $t_{DSCHG\_CUT}$ ) and then resumes conducting after a retry delay time ( $t_{RETRY}$ ). Qswitch turns off instantly if the discharge current exceeds 3.7A value.

When the battery voltage falls below the  $V_{BAT\_UVLO}$  limit that is programmed in the REG01 D[2:0], the Qswitch turns off to prevent over discharging the battery.

If SWITCH\_MODE bit (REG0A D[3]) is set to 1, the Qswitch is forced to remain on like a simple switch and the over-discharge is ignored during battery discharge. This bit will reset if power is re-applied to the input. It will also reset if the battery is connected or disconnected while power is applied to the input.

#### **System Short Circuit Protection**

If a short circuit (to GND) occurs on the load connected to SYS pin, the Qswitch disconnects the BAT to SYS path and the Qbypass limits the current flowing in the IN to SYS path. If the short circuit persists, the die temperature goes high and causes a thermal shutdown.

#### Interrupt to Host (nINT Pin)

The nINT output signal is provided to alert the host on power events. SGM41562 sends out a negative pulse (width =  $t_{\text{INT PULSE}}$ ) to nINT if any of the following events occurs:

- A good input source is detected (UVLO < V<sub>IN</sub> < OVLO).</li>
- UVLO or OVLO is detected (input).
- Charge completed.
- A charging status change.
- A fault record in REG09 occurs (input fault, thermal fault, safety timer fault, battery OVP fault or NTC fault).
- Watchdog expiration (WTD\_FAULT in REG08 D[7]).

When one of the mentioned faults occurs, the relevant fault bit will latch in the register except for NTC fault bit that always reports the current status of the thermistor. A fault status bit is unlatched if the device quits that fault state. It will reset to 0 after the host reads the register if the bit is unlatched.

The assertion of nINT signal pulse can be masked for some of the events listed above when the corresponding mask control bits are set in REG06 D[4:0]. If a mask bit is set, and the event occurs, the nINT signals stays high.

The nINT pin is also used as an input to initiate a power recycle on the SYS output for example when a turn off/turn on is needed on the system when battery is not removable. This input is also used to exit the shipping mode that keeps the battery disconnected. The nINT pin is weakly pulled up internally to an unregulated low voltage that is not high enough for most logic devices. The weak and low voltage pull-up is used to avoid unexpected battery drain or leakage when the product is in shipping or is stocked for prolonged times (called shipping mode). The pull-up can be overdriven externally to higher appropriate logic voltages when it is in operation.

#### **Battery Disconnection Function**

When the battery is not removable, it's essential to disconnect the battery from the system to allow system power recycling or to put that in the shipping mode. It is performed by forcing the Qswitch to remain off by setting FET\_DIS bit to 1. Table 3 explains how the SGM41562 can be programmed in shipping mode (or to do a power recycle on SYS) and how to exit the shipping mode. To exit shipping mode either the input power should be applied to IN port, or a low voltage (ground) should be applied to nINT pin for a short time (for example by holding a push bottom).

**Table 3. Shipping Mode Control** 

	Enter Shipping Mode	Exit Shipping Mode			
Items	FET_DIS = 1	nINT Pin H to L for 2s	V <sub>IN</sub> Plug-in		
Qbypass	don't care	don't care	On		
Qswitch (Charging)	Off	On	On (64ms Later)		
Qswitch (Discharging)	Off	On	On (64ms Later)		

The FET\_DIS bit is used for battery disconnection control. If this bit is set to 1, the device enters shipping mode after a delay time, which can be programmed by EN\_SHIP\_DGL[1:0]. After the delay the Qswitch turns off and the FET\_DIS bit resets to 0. The device wakes up from shipping mode by pulling down nINT pin or detecting an acceptable voltage on the IN pin. The device exits from shipping mode 2 seconds after pulling nINT pin down or 64ms after detecting an acceptable  $V_{\text{IN}}$ . For the application of nINT pulled down to a low voltage in the shipping mode, EN\_SHIP\_DGL[1:0] must keep default value.

System power can be recycled by turning off the Qswitch and Qbypass if nINT pin is pulled low for a duration of more than  $t_{RST\_DGL}$ . It is the time delay to avoid noise and glitches or to hold a push bottom. The  $t_{RST\_DGL}$  time is programmed by  $t_{RST\_DGL}$ [1:0] in REG01. The off state lasts for a duration of  $t_{RST\_DUR}$  which can be programmed via  $t_{RST\_DUR}$  in REG01. After this time the Qswitch and/or Qbypass will be automatically turned on and the system is powered again. During the off period, the nINT pin is biased to a lower voltage.

The waveforms of power recycling are shown in Figure 7.

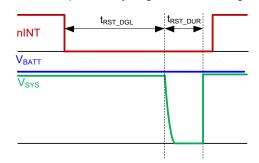


Figure 7. Power Recycling Waveforms

## **REGISTER MAP**

All registers are 8-bit and individual bits are named from D[0] (LSB) to D[7] (MSB).

#### I<sup>2</sup>C Slave Address is: 03H

R/W: Read/Write bit(s).
R: Read only bit(s).
PORV: Power-On-Reset value.

n: Parameter code formed by the bits as an unsigned binary number.

#### REG00

Register address: 0x00; R/W

PORV = 10011111

#### Table 4. REG00 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
		VIN_MIN[3] 1 = 640mV	Minimum Input Voltage Limit (n: 4 bits): = 3.88 + 0.08n (V)	1	R/W	REG_RST
D[7:4]	VIN MIN[3:0]	VIN_MIN[2] 1 = 320mV	Offset: 3.88V	0	R/W	REG_RST
D[7.4]	VIN_MIN[1]   Range:3.88V (0000) - 5.08\   Default: 4.60V (1001)	Range:3.88V (0000) - 5.08V (1111) Default: 4.60V (1001)	0	R/W	REG_RST	
				1	R/W	REG_RST
	IIN_LIM[3:0]	IIN_LIM[3] 1 = 240mA	Input Current Limit (n: 4 bits): = 50 + 30n (mA) Offset: 50mA Range: 50mA (0000) - 500mA (1111) Default: 500mA (1111)	1	R/W	REG_RST
D[3:0]		IIN_LIM[2] 1 = 120mA		1	R/W	REG_RST
D[0.0]		IIN_LIM[1] 1= 60mA		1	R/W	REG_RST
		IIN_LIM[0] 1 = 30mA		1	R/W	REG_RST

#### REG01

Register address: 0x01; R/W

PORV = 10101100

#### Table 5. REG01 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
D[7:6]	+DCT_DC1[1:0]	00 = 8s 01 = 12s	nINT Pull-Down Period to Disconnect the Battery (n: 2 bits):	1	R/W	REG_RST or Watchdog
D[7:6]	tRST_DGL[1:0]	10 = 16s (default) 11 = 20s	= 8s + 4n (seconds)	0	R/W	REG_RST or Watchdog
D[5]	tRST_DUR	0 = 2s 1 = 4s (default)	Battery FET off-time duration after reset. The Qbypass and Qswitch off-time before auto turn-on.	1	R/W	REG_RST or Watchdog
D[4]	EN_HIZ	HIZ Mode Enable 0 = Disable (default) 1 = Enable	Control Qbypass switch. Default: disable (0) or switch on Note: The EN_HIZ bit only controls the on and off of the Qbypass.	0	R/W	REG_RST or Watchdog
D[3]	CEB	Setting Charge Enable 0 = Charge Enable 1 = Charge Disabled (default)	Charge enable/disable Qswitch configuration. Default: charge disabled (1) or Qswitch off	1	R/W	REG_RST or Watchdog
		VBAT_UVLO[2] 1 = 360mV	Battery UVLO Threshold Value (n: 3 bits): = 2.4V + 0.09n (V)	1	R/W	REG_RST or Watchdog
D[2:0]	VBAT_UVLO[2:0]	VBAT_UVLO[1] 1 = 180mV	Offset: 2.4V	0	R/W	REG_RST or Watchdog
		VBAT_UVLO[0] 1 = 90mV	Range: 2.4V (000) - 3.03V (111) Default: 2.76V (100)	0	R/W	REG_RST or Watchdog

#### REG02

Register address: 0x02; R/W

PORV = 00001111

#### Table 6. REG02 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
D[7]	REG_RST	Software Reset 0 = Keep Current Setting (default) 1 = Reset	If set, will reset most parameters to default. (as explained in the last column of register map tables)	0	R/W	REG_RST
D[6]	WD_RST	I <sup>2</sup> C Watchdog Timer Reset 0 = Normal (default) 1 = Reset	If set, will reset watchdog timer.	0	R/W	REG_RST or Watchdog
		ICC[5] 1= 256mA	Fast Charge Current Value (CC Mode) (n: 5 bits): = 8mA + 8n (mA) (n ≤ 56)	0	R/W	REG_RST or Watchdog
		ICC[4] 1 = 128mA	Offset: 8mA	0	R/W	REG_RST or Watchdog
DIE:01	ICCIE:01	ICC[3] 1 = 64mA	Range: 8mA (000000) - 456mA (111000) Default: 128mA (001111)	1	R/W	REG_RST or Watchdog
D[5:0]	100[5.0]	ICC[5:0] ICC[2] 1 = 32mA	Note: Values above 56D = 111000 (456mA) are clamped to 56D = 111000 (456mA).	1	R/W	REG_RST or Watchdog
		ICC[1] 1 = 16mA		1	R/W	REG_RST or Watchdog
		ICC[0] 1 = 8mA		1	R/W	REG_RST or Watchdog

#### REG03

Register address: 0x03; R/W

Table 7. REG03 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
		IDSCHG[3] 1 = 1600mA	BAT to SYS Discharge Current Limit Value (n: 4 bits): = 200mA + 200n (mA), n ≠ 0	1	R/W	REG_RST or Watchdog
D[7:4]	IDSCHG[3:0]	IDSCHG[2] 1 = 800mA	Offset: 200mA	0	R/W	REG_RST or Watchdog
D[7.4]	iD3CHG[3.0]	IDSCHG[1] 1 = 400mA	Valid Range: 400mA (0001) - 3.2A (1111) Default: 2000mA (1001)	0	R/W	REG_RST or Watchdog
		IDSCHG[0] 1 = 200mA		1	R/W	REG_RST or Watchdog
		ITERM[3] 1 = 16mA	Charge Termination Current Value (n: 4 bits): = 1mA + 2n (mA)	0	R/W	REG_RST or Watchdog
D[3:0]	ITERM[3:0]	ITERM[2] 1 = 8mA	Offset: 1mA	0	R/W	REG_RST or Watchdog
D[3.0]	TTEIXIN[3.0]	ITERM[1] 1 = 4mA	Range: 1mA (0000) - 31mA (1111) Default: 3mA (0001)	0 R/	R/W	REG_RST or Watchdog
		ITERM[0] 1 = 2mA		1	R/W	REG_RST or Watchdog

#### REG04

Register address: 0x04; R/W

PORV = 10100011

#### Table 8. REG04 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
		VBAT_REG[5]	Battery Charge Regulation Voltage Value	1	R/W	REG_RST
		1 = 480mV	(CV Mode) (n: 6 bits):			or Watchdog
		VBAT_REG[4]	= 3.6V + 0.015n (V)	0	R/W	REG_RST
		1 = 240mV				or Watchdog
		VBAT_REG[3]	Offset: 3.60V	1	R/W	REG_RST
D[7:2]	VBAT REG[5:0]	1 = 120mV	Range: 3.60V (000000) - 4.545V (111111)	,	17/77	or Watchdog
D[1.2]	VDAT_INEO[0.0]	VBAT_REG[2]	Default: 4.2V (101000)	0	R/W	REG_RST
		1 = 60mV		U	17/77	or Watchdog
		VBAT_REG[1]		0	R/W	REG_RST
		1 = 30mV		U	FK/VV	or Watchdog
		VBAT_REG[0]		0	DAM	REG_RST
		1 = 15mV		0	R/W	or Watchdog
		Pre-Charge to Fast Charge				
D[4]	VDAT DDE	Threshold		4	R/W	REG_RST
D[1]	VBAT_PRE	0 = 2.8V		I	FK/VV	or Watchdog
		1 = 3.0V (default)				
		Battery Recharge Threshold	Offset below VBAT_REG.			REG RST
D[0]	VRECH	0 = 100mV		1	R/W	or Watchdog
		1 = 200mV (default)				or wateridog

#### REG05

Register address: 0x05; R/W

Table 9. REG05 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
D[7]	EN_WD_DISCHG	Watchdog Control 0 = Disable (default) 1 = Enable	Watchdog control in discharge mode.	0	R/W	REG_RST
		Watchdog Timer	If WATCHDOG[1:0] = 00, then watchdog timer	1	R/W	REG_RST
D[6:5]	WATCHDOG[1:0]	00 = Disable timer 01 = 40s 10 = 80s 11 = 160s (default)	is disabled no matter EN_WD_DISCHG is set or not.	1	R/W	REG_RST
D[4]	EN_TERM	Termination Control 0 = Disable 1 = Enable (default)	Use termination or not.	1	R/W	REG_RST or Watchdog
D[3]	EN_TIMER	Safety Timer Control 0 = Disable 1 = Enable (default)	Charge safety timer enable/disable setting.	1	R/W	REG_RST or Watchdog
		Charge Timer 00 = 3hrs		0	R/W	REG_RST or Watchdog
D[2:1]	CHG_TMR[1:0]	01 = 5hrs (default) 10 = 8hrs 11 = 12hr		1	R/W	REG_RST or Watchdog
D[0]	TERM_TMR	Termination Timer Control 0 = Disable (default) 1 = Enable	When TERM_TMR is enabled, the device will not suspend the charge current after charge termination.	0	R/W	REG_RST or Watchdog

#### REG06

Register address: 0x06; R/W

PORV = 11000000

#### Table 10. REG06 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	POR	TYPE	RESET BY
D[7]	EN_NTC	NTC Control 0 = Disable 1 = Enable (default)		1	R/W	REG_RST or Watchdog
D[6]	TMR2X_EN	Enable Half Clock Rate Safety Timer 0 = Disable. 1 = Enable 2× extended safety timer during PPM. (default)		1	R/W	REG_RST or Watchdog
D[5]	FET_DIS	0 = Enable (default) 1 = Disable	Qswitch control for shipping mode and system power recycle.  Note: The FET_DIS bit controls the on and off of the Qswitch in both charging and discharging.	0	R/W	REG_RST
D[4]	PG_INT_CTL	0 = On (default) 1 = Off		0	R/W	REG_RST or Watchdog
D[3]	EOC_INT_CTL	Charge Completed INT Mask Control 0 = On (default) 1 = Off		0	R/W	REG_RST or Watchdog
D[2]	CHG_STATUS_ INT_CTL	Charging Status Change INT Mask Control 0 = On (default) 1 = Off	Charging statuses are: not charging, pre-charge and charge.	0	R/W	REG_RST or Watchdog
D[1]	NTC_INT_CTL	0 = On (default) 1 = Off		0	R/W	REG_RST or Watchdog
D[0]	BATOVP_INT_CTL	0 = On (default) 1 = Off		0	R/W	REG_RST or Watchdog

#### REG07

Register address: 0x07; R/W

Table 11. REG07 Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
D[7]	EN_PCB OTP	PCB OTP Enable 0 = Enable (default) 1 = Disable		0	R/W	REG_RST or Watchdog
D[6]	EN_VINLOOP	0 = Enable (default) 1 = Disable		0	R/W	REG_RST or Watchdog
		Thermal Regulation Threshold 00 = 60°C		1	R/W	REG_RST or Watchdog
D[5:4]	TJ_REG[1:0]	01 = 80°C 10 = 100°C 11 = 120°C (default)		1	R/W	REG_RST or Watchdog
		VSYS_REG[3] 1 = 400mV	System Regulation Voltage Value: = 4.2V + 0.05n (V) (n: 4 bits)	0	R/W	REG_RST
D[3:0]	VSYS REG[3:0]	VSYS_REG[2] 1 = 200mV	Offset: 4.2V	1	R/W	REG_RST
[٥.0]	V313_REG[3.0]	VSYS_REG[1] 1 = 100mV	Range: 4.2V (0000) - 4.95V (1111) Default: 4.55V (0111)	1	R/W	REG_RST
		VSYS_REG[0] 1 = 50mV		1	R/W	REG_RST

#### REG08

Register address: 0x08; R and RW

PORV = 00000000

#### Table 12. REG08 Register Details

BITS	BIT NAME	DESCRIPTION	PORV	TYPE	RESET BY
D[7]	WTD_FAULT	Watchdog Expiration 0 = Normal (default) 1 = Watchdog Timer Expiration		R	NA
D[6]	IIN_LIM_REL	Input Current Limit Release 0 = Disable (default) 1 = Enable	0	R/W	REG_RST or Watchdog
D[5]	IIN_LIM_ADD200	add 200mA to Input Current Limit = Disable (default) = Enable		R/W	REG_RST or Watchdog
		Charging Status	0	R	NA
D[4:3]	CHG_STAT[1:0]	00 = Not Charging (default) 01 = Pre-Charge 10 = Charge 11 = Charge Done	0	R	NA
D[2]	PPM_STAT	Device in Power Path Management Mode (PPM) 0 = No PPM (default) 1 = IN PPM	0	R	NA
D[1]	PG_STAT	Input Power (IN) Status 0 = Power Fail (default) 1 = Power Good	0	R	NA
D[0]	THERM_STAT	Thermal Regulation Status 0 = No Thermal Regulation (default) 1 = In Thermal Regulation	0	R	NA

#### REG09

Register address: 0x09; R and R/W

Table 13. REG09 Register Details

BITS	BIT NAME	DESCRIPTION	PORV	TYPE	RESET BY
		Enter Shipping Mode Deglitch Time 00 = 1s (default)	0	R/W	REG_RST
D[7:6]	EN_SHIP_DGL [1:0]	01 = 2s 10 = 4s 11 = 8s	0	R/W	REG_RST
D[5]	VIN_FAULT	Input VIN Fault Status 0 = Normal (default) 1 = Input Fault (OVP or bad source)	0	R	NA
D[4]	THEM_SD	Thermal Shutdown Fault Status 0 = Normal (default) 1 = Thermal Shutdown	0	R	NA
D[3]	BAT_FAULT	Battery Over-Voltage Fault Status 0 = Normal (default) 1 = Battery OVP	0	R	NA
D[2]	STMR_FAULT	Safety Timer Expiration Fault Status 0 = Normal (default) 1 = Safety Timer Expiration	0	R	NA
D[1]	NTC_FAULT[1]	NTC Exceeding Hot Level 0 = Normal (default) 1 = NTC Hot	0	R	NA
D[0]	NTC_FAULT[0]	NTC Exceeding Cold Level 0 = Normal (default) 1 = NTC Cold	0	R	NA

#### **REG0A**

Register address: 0x0A; R and R/W

PORV = 01100000

#### Table 14. REG0A Register Details

BITS	BIT NAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
		Slave Address		0	R	NA
		001 = 01H 010 = 02H		1	R	NA
D[7:5]	ADDR[2:0]	011 = 03H (default) 100 = 04H 101 = 05H 110 = 06H 111 = 07H		1	R	NA
D[4]	COLD_RESET	Software Power Recycle 0 = No Action (default) 1 = Power Recycle Reset	Causes a system power recycles if set to 1. Automatically clears after power recycle.	0	R/W	NA
D[3]	SWITCH_MODE	0 = Normal Power Path (default) 1 = Qswitch Forced On	Effective in battery discharge mode only. When Qswitch is forced on, there is no current and voltage limit because the internal circuits are shut down for lower consumption.	0	R/W	NA
D[2]	DIS_VDD	0 = Enable Battery Power (default) 1 = Disable Battery Power	If set to 1, $V_{\text{DD}}$ becomes high-impedance when $V_{\text{IN}}$ is removed.  If the PCB OTP of NTC function is be selected, the DIS_VDD bit setting is invalid.	0	R/W	NA
D[1]	DIS_VINOVP	0 = Enable (default) 1 = Disable	Disables over-voltage lockout detection of $V_{\text{IN}}$ if set to 1.	0	R/W	NA
D[0]	CC_FINE	0 = Normal Scale (default) 1 = Fine Scale	If set to 1, the programmed charge current in ICC[5:0] is weighted to 1/4.	0	R/W	NA

#### **REG0B**

Register address: 0x0B; R PORV = 00000100 (SGM41562)

#### Table 15. REG0B Register Details

BITS	BITNAME	DESCRIPTION	COMMENT	PORV	TYPE	RESET BY
D[7:0]	ID[7:0]	Device ID	SGM41562 = 00000100		R	NA

## **OTP MAP**

The following table shows the one time programmable (OTP) regions of the register map. The OTP bits can be read only.

ADDRESS	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0x0A	ОТ	P BITS: ADDR[2	2:0]	COLD_ RESET	SWITCH_ MODE	DIS_VDD	DIS_VINOVP	CC_FINE
0x0B				OTP BIT	S: ID[7:0]			

## **OTP DEFAULT**

OTP ITEMS	DEFAULT
ICC	128mA
ITERM	3mA
VBAT_REG	4.2V
WATCHDOG	160s
EN_VINLOOP	Enable
Address	03H
Device ID	SGM41562: 00000100. V <sub>IN_OVLO</sub> = 6V

## STATE CONVERSION CHART

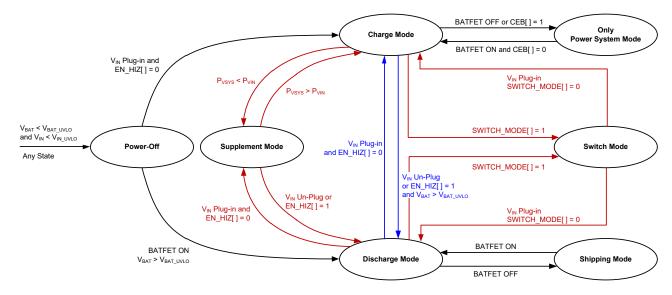


Figure 8. State Machine Conversion

## **CONTROL FLOW CHART**

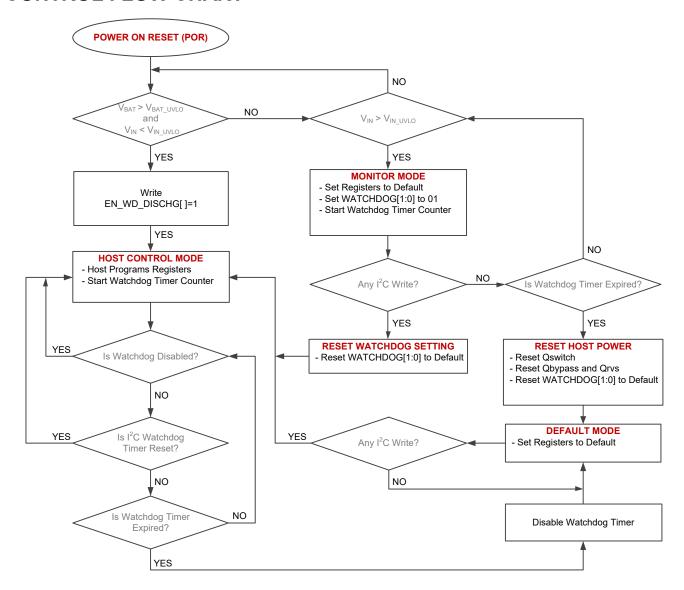


Figure 9. Startup, Host Mode, Default Mode and Host Power Reset

# **CONTROL FLOW CHART (continued)**

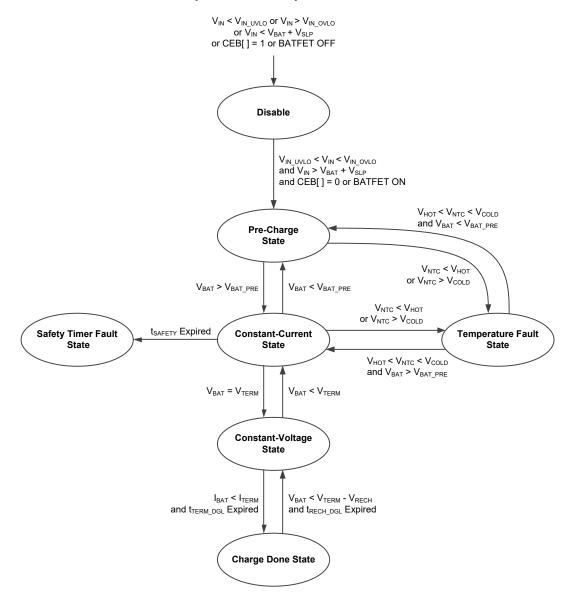
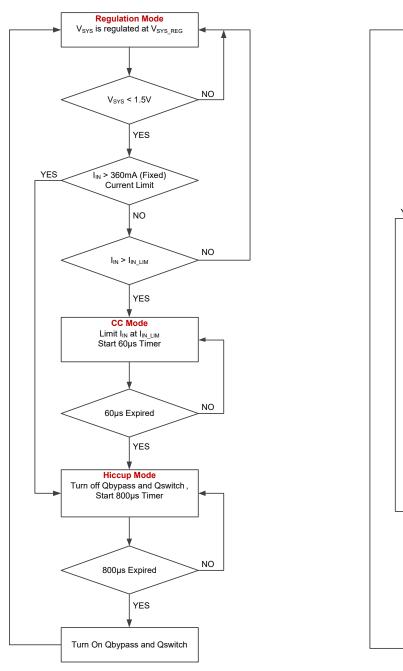


Figure 10. Charging Process

# **CONTROL FLOW CHART (continued)**



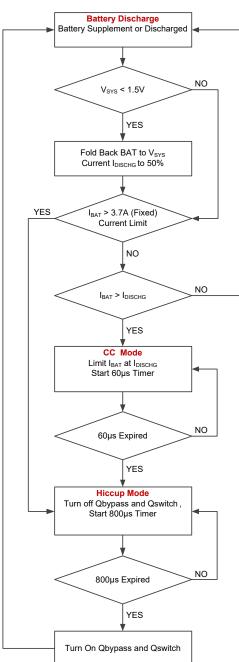


Figure 11. System Short Circuit Protection

#### APPLICATION INFORMATION

#### **Resistor Divider for NTC Sensor**

A resistor divider between VDD and GND pins can be used to adjust the battery temperature limits sensed by the NTC sensor. The  $R_{T1}$  and  $R_{T2}$  resistors (see Figure 12) allow independent programming of the high and low temperature limits for any type of NTC temperature characteristics.

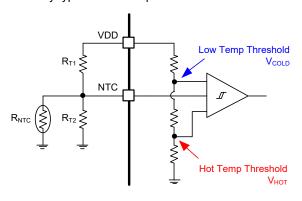


Figure 12. NTC Function Block

For a given NTC thermistor, if the NTC resistances at the desired high and low temperatures are  $R_{\text{NTCH}}$  and  $R_{\text{NTCL}}$  respectively,  $R_{\text{T1}}$  and  $R_{\text{T2}}$  values can be calculated by:

$$\begin{split} R_{\text{T2}} = & \frac{\left(V_{\text{COLD}} - V_{\text{HOT}}\right) \times R_{\text{NTCH}} \times R_{\text{NTCL}}}{\left(V_{\text{HOT}} - V_{\text{COLD}}V_{\text{HOT}}\right) \times R_{\text{NTCL}} - \left(V_{\text{COLD}} - V_{\text{COLD}}V_{\text{HOT}}\right) \times R_{\text{NTCH}}} \\ R_{\text{T1}} = & \frac{1 - V_{\text{COLD}}}{V_{\text{COLD}}} \times \left(R_{\text{T2}} \, / \, / R_{\text{NTCL}}\right) \end{split}$$

where  $V_{\text{COLD}}$  and  $V_{\text{HOT}}$  thresholds values are voltage levels on the NTC pin given in the EC table for hot and cold detection.

For example, for a thermistor with  $R_{25^{\circ}C}=10k\Omega$  and  $\beta=3260$ ,  $R_{NTCL}$  is  $27.2k\Omega$  at  $T_{COLD}=0^{\circ}C$ , and  $R_{NTCH}$  is  $4.29k\Omega$  at  $T_{HOT}=50^{\circ}C$ . Using Equation 1 and Equation 2 to calculate  $R_{T1}=7.6k\Omega$  and  $R_{T2}=29.33k\Omega$  (to be recalculated when the EC table mean values are characterized), assuming that the NTC window is between  $0^{\circ}C$  and  $50^{\circ}C$  and using the  $V_{COLD}$  and  $V_{HOT}$  values from the EC table.

#### **External Capacitor Selection**

Like many low-dropout regulators, SGM41562 requires external capacitors on its power ports for stability and noise or spike voltage immunity. These capacitors must be properly selected and placed near the device.

#### Input Capacitor (IN to GND)

A minimum 2.2µF input capacitor must be connected between IN and GND pins for stable operation over full load range. In general an output capacitance larger than the input capacitor is acceptable if the input capacitor is at least 2.2µF.

#### **Output Capacitor (SYS to GND)**

SGM41562 is designed specifically to operate with small ceramic output capacitance. A ceramic capacitor (X5R or X7R) larger than  $2.2\mu F$  is suitable for the SGM41562 applications. The output capacitor should be connected close to the device between SYS and GND pins with thick traces and small loop area.

#### **BAT to GND Capacitor**

A capacitor is needed between BAT and GND pins. Use a ceramic capacitor (X5R or X7R) that is at least 2.2µF.

#### **VDD to GND Capacitor**

VDD voltage powers the internal control and logic circuit. It is critical to use a 0.1µF decoupling ceramic capacitor between VDD pin and GND close to the device with thick PCB traces to decouple noise and stabilize VDD voltage.

#### **PCB Layout Guide**

- Place external capacitors as close as possible to the device to minimize stray inductances and connection impedance.
- The GND for the I<sup>2</sup>C signals should be clean and directly connected to GND pin, without sharing its route with GND returns that carry high current or switching currents.
- 3. Due to relatively slow rise/fall times, it is ok to route the I<sup>2</sup>C wires in parallel on the same PCB layer.

# **TYPICAL APPLICATION CIRCUIT**

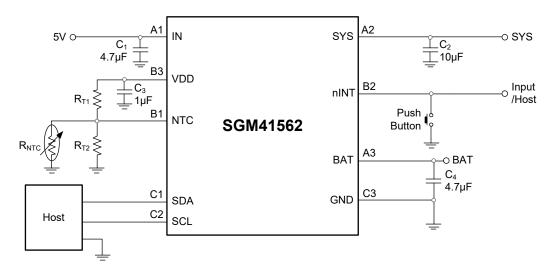
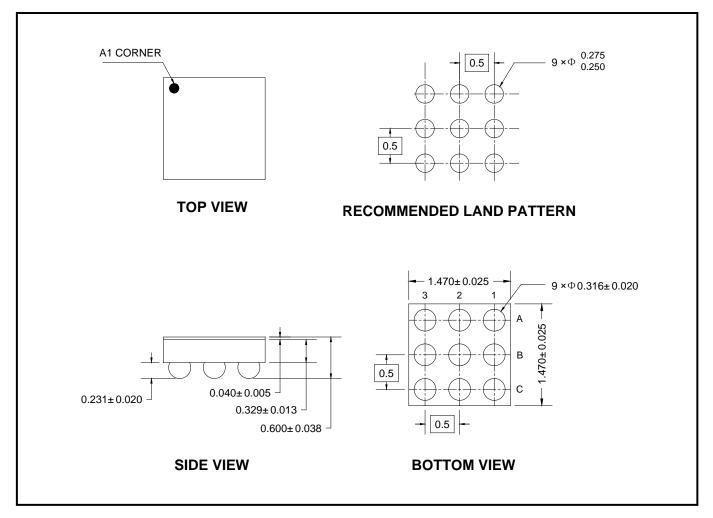


Figure 13. SGM41562 Typical Application Circuit with 5V Input

Table 16. The Key BOM of Figure 13

QTY	REF	VALUE	DESCRIPTION	PACKAGE	
1	C1, C4	4.7µF	Ceramic Capacitor; 16V; X5R or X7R	0603	
2	C2	10µF	Ceramic Capacitor; 16V; X5R or X7R	0603	
1	C3	1µF	Ceramic Capacitor; 16V; X5R or X7R	0603	

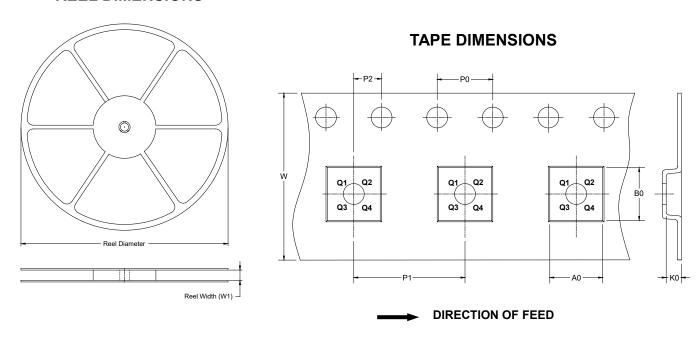
# PACKAGE OUTLINE DIMENSIONS WLCSP-1.47×1.47-9B



NOTE: All linear dimensions are in millimeters.

## TAPE AND REEL INFORMATION

#### **REEL DIMENSIONS**



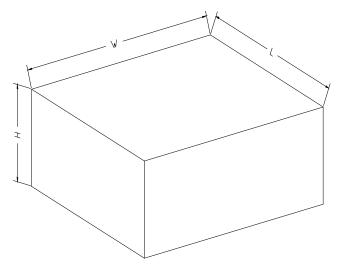
NOTE: The picture is only for reference. Please make the object as the standard.

#### **KEY PARAMETER LIST OF TAPE AND REEL**

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
WLCSP-1.47×1.47-9B	7"	9.2	1.61	1.61	0.7	4.0	4.0	2.0	8.0	Q1

TX10000.000

## **CARTON BOX DIMENSIONS**



NOTE: The picture is only for reference. Please make the object as the standard.

### **KEY PARAMETER LIST OF CARTON BOX**

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton
7" (Option)	368	227	224	8
7"	442	410	224	18