

### **Dual Mode CDMA/AMPS Baseband Interface**

Advance Information

DS4722 - 1.8 July 1998

The PLUTO baseband interface circuit is designed for use in dual mode CDMA/AMPS digital cellular telephones. In the telephone, Pluto provides the interface between the radio (RF & IF) components and the baseband digital signal processor. Pluto is part of a complete chipset solution for CDMA phones entitled the Planet chipset.

The receive (RX) section converts the analog in-phase and quadrature (I & Q) signals into equivalent digital signals whilst the transmit (TX) circuits perform the complementary function of translating digital baseband information into the analog equivalent signals required for the modulator in the radio circuits. VHF PLLS are also included for second RXLO and TXIF generation.

PLUTO also contains a 4 channel general purpose ADC which is included for such purposes as environmental and signal strength monitoring.

### **FEATURES**

- Dual mode AMPS/CDMA compatible
- Low Power/Low Voltage operation
- Standard baseband I and Q interface
- 4 Input Auxiliary ADC
- Synthesisers

### **APPLICATIONS**

Dual Mode CDMA/AMPs digital cellular telephones

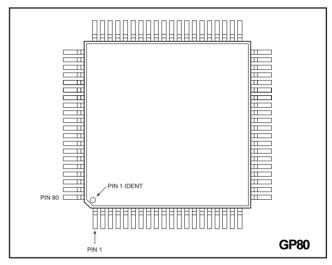


Figure 1 Pin connections - top view

### **ABSOLUTE MAXIMUM RATINGS**

Supply voltage -0.3 to 3.9V
Voltage applied to any other pin
Operating junction temperature 150°C
Storage temperature -55°C to 150°C
ESD (human body model) 2kV

## ORDERING INFORMATION

PLUTO/KG/GP1R

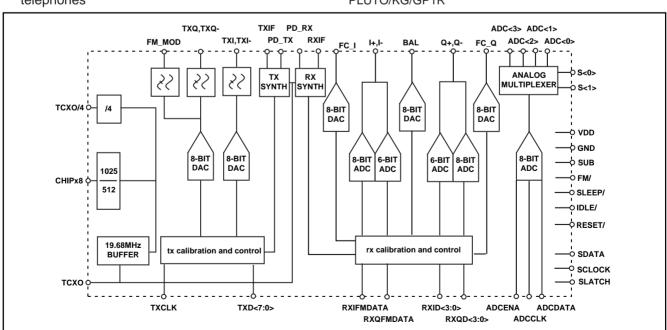


Figure 2 Block diagram

PIN DESCRIPTION

PIN DE	IN DESCRIPTION										
No	Pin Name	Type	A/D	Description							
1	VDD	Power		Power Supply							
2	RSET	Input	Analog	Bias current setting resistor - 40kΩ to ground							
3	GND	Ground		Ground							
4	TX_IF	Input	Analog	TX VCO output							
5	IDLEB	Digital		Idle mode control signal - active low - pulled low if left unconnected							
6	PD TX	Output		TX synthesiser charge pump output							
7	FMB	Input	Digital	FM mode control signal - active low - pulled low if left unconnected							
8	PD_RX	Output	Analog	RX synthesiser charge pump output							
9	SLEEPB	Input	Digital	Sleep mode control signal - active low - pulled low if left unconnected							
10	RX_IF	Input	Analog	RX VCO output							
11	TX_LOCK	Output	Digital	TX synthesiser lock detect open drain output - pulled high by ext. resistor							
12	RX_LOCK	Output	Digital	RX synthesiser lock detect open drain output - pulled high by ext. resistor							
13	TCXO/4	Output	Digital	TCXO divided by 4 output							
14	TXD<0>	Input	Digital	Transmit data bit 0 (lsb)							
15	TXD<1>	Input	Digital	Transmit data bit 1							
16	TXD<2>	Input	Digital	Transmit data bit 2							
17	TXD<3> TXD<4>	Input	Digital	Transmit data bit 3 Transmit data bit 4							
18 19	TXD<4>	Input	Digital Digital	Transmit data bit 5							
20	TXD<6>	Input	Digital	Transmit data bit 6							
21	TXD<0>	Input Input	Digital	Transmit data bit 6 Transmit data bit 7 (MSB)							
22	TXCLK	Input	Digital	Complimentary Transmit Clock (+ve)							
23	TXCLKB	Input	Digital	Complementary Transmit Clock (+ve)							
24	CHIPx8	Input	Digital	9.8304MHz synthesiser output							
25	VDD	Power	Digital	Power Supply							
26	TCXO	Input	Analog	TCXO 19.68MHz a.c. coupled sinewave input							
27	GND	Ground	7 tildlog	Ground							
28	SUB	Ground		Substrate-Ground							
29	RESET	Input	Digital	Chip master reset - pulled high if not connected							
30	SDATA	Input	Digital	Serial Interface Data Input							
31	SCLK	Input	Digital	Serial Interface Clock Input							
32	SLATCH	Input	Digital	Serial Interfce Latch Input							
33	S<0>	Input	Digital	Aux ADC mux channel select LSB							
34	n/c	·									
35	RXID<0>	Output	Digital	I-Channel RX CDMA output LSB - low when inactive							
36	RXID<1>	Output	Digital	I-Channel RX CDMA output bit 1 - low when inactive							
37	RXID<2>	Output	Digital	I-Channel RX CDMA output bit 2 - low when inactive							
38	RXID<3>	Output	Digital	I-Channel RX CDMA output bit 3 - low when inactive							
39	S<1>	Input	Digital	Aux ADC mux channel select MSB							
40	n/c										
41	RXQD<0>	Output	Digital	Q_Channel RX CDMA output LSB - low when inactive							
42	RXQD<1>	Output	Digital	Q_Channel RX CDMA output bit 1 - low when inactive							
43	RXQD<2>	Output	Digital	Q_Channel RX CDMA output bit 2 - low when inactive							
44	RXQD<3>	Output	Digital	Q_Channel RX CDMA output bit 3 - low when inactive							
45	GND	Ground		Ground							
46	VDD	Power		Power Supply							
47	RXFMSTB	input	Digital	Receive data FM strobe - pulled low if not connected							
48	FMCLK	Input	Digital	Receive data FM clock - pulled low if not connected							
49	RXQFMDATA	Output	Digital	Q-Channel RX FM data serial output - low when inactive							
50	RXIFMDATA	Output	Digital	I-Channel RX FM data serial output -low when inactive							
51	ADCLK	Output	Digital	Auxiliary ADC serial data clock. Low when inactive							
52	ADCDATA	Output	Digital	Auxiliary ADC serial data output. Low when inactive							
53	ADCENA	Input	Digital	Auxiliary ADC enable - pull down if not used							
54	SUB	Ground		Substrate - Ground							

PIN DESCRIPTION (continued)

No	Pin Name	Туре	A/D	Description
55	RXQP	Input	Analog	Receive Q Channel Input (+ve)
56	RXQM	Input	Analog	Receive Q Channel Input (-ve)
57	SUB	Ground	J	Substrate - Ground
58	RXIP	Input	Analog	Receive Q channel input (+ve)
59	RXIM	Input	Analog	Receive Q channel Input (-ve)
60	VDD	Power	· ·	Power Supply
61	GND	Ground		Ground
62	VREF<0>	Input/Output	Analog	CDMA Receive Circuit Voltage Reference De-Coupling
63	AD<0>	Input	Analog	AUX ADC Input
64	AD<1>	Input	Analog	AUX ADC Input
65	AD<2>	Input	Analog	AUX ADC Input
66	AD<3>	Input	Analog	AUX ADC Input
67	Vtest	Output	Digital	RX Filter tuning tone output - pulled low when inactive
68	EnTest	Output	Digital	RX Filter tuning mode control output - pulled low when inactive
69	FC_Q	Output	Analog	RX Filter Q channel FC control
70	FC_I	Output	Analog	RX Filter channel FC control
71	BAL	Output	Analog	RX Filter Gain Balance Control
72	VDD	Power		Power Supply
73	GND	Ground		Ground
74	TXIP	Output	Analog	Transmit Circuit channel Complementary Output (+ve)
75	TXIM	Output	Analog	Transmit Circuit I channel Complementary Output (-ve)
76	SUB	Ground		Substrate - Ground
77	TXQP	Output	Analog	Transmit Circuit Q channel Complementary Output (+ve)
78	TXQM	Output	Analog	Transmit Circuit Q channel Complementary Output (-ve)
79	FMTX	Output	Analog	Transmit Circuit FM output
80	VREF<1>	Input/Output	Analog	Transmit Circuit Voltage Reference De-coupling

#### **FUNCTIONAL DESCRIPTION**

Baseband TX interface circuit

The Pluto baseband transmit circuit acts as an interface between the baseband signal processor and the RF/IF sections in a CDMA/AMPS compatible mobile telephone.

The TX circuit has two modes of operation:

CDMA mode, transmit data that has previously been encoded by the baseband digital signal processor is converted to equivalent analog signals by matched digital-to-analog converters, these signals are then filtered to remove the image of the sample clock that would otherwise be present at the output before being output to the I and Q modulator as differential signals.

FM mode, transmit data is treated in much the same way as in CDMA mode except that only one DAC is used and (because of the much lower bandwidth of AMPS signals) a different reconstruction filter is used before the analog fm signal is output to the mixer as a single ended signal.

## CDMA Transmit Signal Path CDMA TX DACs

In CDMA mode two matched 8-bit DACs are used to generate the in-phase and quadrature signals, the input data for the DACs is obtained by multiplexing over an 8-bit parallel input port (TXD<7:0>). The transmit data rate is twice as fast as the differential transmit clock (TXCLK). Incoming data that is valid during the rising edge of the transmit clock is loaded into in In-Phase DAC & incoming data that is valid on the falling edge of the transmit clock is loaded into the Quadrature DAC - I and Q values must be modified in the digital baseband chip to account for the half-cycle delay between them.

### **CDMA Analog Reconstruction Filters**

The frequency spectrum at the output of the transmit DACs contains unwanted frequency components. Reconstruction filters are used to smooth the DAC output signals, providing continuous time output signals at the I and Q output pins thereby removing these undesirable signals.

The low pass filters used are 5th order Butterworth, continuous time filters with a nominal cut-off frequency of 1.2 MHz. These filters are designed to have a linear phase response in the pass band. On-chip reconstruction filters minimise the phase and amplitude mismatch between I and Q channels.

### **CDMA TX Section Analog Interface**

The ITx and QTx outputs can be d.c. or a.c. coupled to the external circuits and will differentially drive a minimum resistive load of 5 k $\Omega$  and a maximum capacitive load of 20 pF.

When the CDMA transmit path is in power-down mode the positive outputs goes high and the negative output goes low.

## FM Transmit Signal Path FM TX DAC

In FM mode, the Q-Channel DAC is used to generate an analog FM modulation signal from the data transmitted from the digital baseband processor. In this mode, all other CDMA TX circuits are powered down.

#### **FM Mode Analog Reconstruction Filters**

The frequency spectrum at the output of the transmit DAC contains unwanted frequency components. A reconstruction filter is used to smooth the DAC output signals.

Low-pass filters are used with a cut-off frequency of approximately 13 kHz. These filters are 3rd order Butterworth filters.

### **FM TX Section Analog Interface**

The FMTX output can be d.c. or a.c. coupled to the radio circuits and will drive a minimum resistive load of 5 k $\Omega$  and a maximum capacitive load of 20 pF.

When the FM mode is in power-down the output is in high impedance state.

# CDMA Receive Signal Path CDMA Receive ADC

In CDMA mode two high speed 4-bit ADCs are used to digitise the incoming signals before subsequent transmission to the baseband digital signal processor as two parallel 4 bit words (RXI<3:0> and RXQ<3:0>). The sample rate of 9.8304MHz is generated via an on chip synthesiser that requires no setting up or external components. On each falling edge of the synthesised clock (CHIPx8) a new digital sample is output on the digital bus.

#### **CDMA Receive Calibration Circuit**

On entering into CDMA mode from power down or from FM mode the calibration circuits are activated. These circuits measure the differences between the receive path gain in the pass band and in the transition band of both I and Q filters. Via a successive approximation process they tune the receive filters cut-off frequency and amplitude matching using the 8 bit DACs provided for this purpose (I\_FC, Q\_FC and BAL). Once both filters (I and Q) have been calibrated in this way the DAC outputs will not change until the chip is powered down or the calibration circuit is re-activated in some other way.

#### **FM Receive Signal Path**

In FM mode two low speed 8-bit ADCs are used to digitise the incoming signals before subsequent transmission to the baseband digital signal processor as two serial 8-bit words (FMRXI & FMRXQ). The sample rate is entirely determined by the digital baseband processor (up-to the maximum allowed) via the FMCLK input.

In FM mode the receive filters are assumed to track the filters used in CDMA mode i.e. there is no separate tuning mechanism.

### **SYNTHESISERS**

The Synthesiser block comprises the input buffers, main dividers, phase comparator, charge pump and lock detect circuit for a TX and RX synthesiser. The loop filter components and the VCOs are external to the device. A common reference divider chain is also included together with bias and control circuitry. All blocks apart from reference divider, bias and control logic are duplicated exactly for RX and TX synthesisers.

The receive intermediate frequency (RX\_IF) is programmable and the transmit intermediate frequency (TX\_IF) is fixed at 130.38MHz.

### **AUX ADC**

The auxiliary converter section contains a single 8-bit successive approximation analog to digital converter, with serial output. In order to maximise the flexibility of Pluto, a 4 way analog multiplexer is provided, which enables the converter to encode any one of four selectable channels. The converter is intended for such applications as power supply and temperature monitoring. When not in use, the converter is powered down, and its outputs are held low.

### **TIMING INFORMATION**

	Parameter		Value		Units	Conditions
		Min	Тур	Max		
t1	TXCLOCK PERIOD (CDMA TX)		203.2		ns	CDMA TX Figure 3
t2	TXCLOCK HIGH TIME (CDMA TX)		101.6		ns	CDMA TX Figure 3
t3	TXCLOCK LOW TIME (CDMA TX)		101.6		ns	CDMA TX Figure 3
t4	TXCLOCK PHASE Delay (CDMA TX)			1.2	ns	CDMA TX Figure 3, FM TX Figure 4
t5	TXCLOCK RISE TIME (CDMA TX)			12	ns	CDMA TX Figure 3, FM TX Figure 4
t6	TXCLOCK FALL TIME (CDMA TX)			12	ns	CDMA TX Figure 3, FM TX Figure 4
t7	TXD-TXCLOCK SETUP TIME	20			ns	CDMA TX Figure 3, FM TX Figure 4
t8	TXCLOCK-TXD HOLD TIME	3			ns	CDMA TX Figure 3, FM TX Figure 4
t11	TXCLOCK PERIOD (FM TX)		2.78		μs	FM TX Figure 4
t12	TXCLOCK HIGH TIME (FM TX)		1.39		μs	FM TX Figure 4
t13	TXCLOCK LOW TIME (FM TX)		1.39		μs	FM TX Figure 4
t14	CHIPx8 PERIOD		101.6		ns	Figure 5
t15	CHIPx8 HIGH TIME		50.8		ns	Figure 5
t16	CHIPx8LOW TIME		50.8	152.4	ns	Figure 5
t17	CHIPx8 RISE TIME	3		12	ns	Figure 5
t18	CHIPx8 FALL TIME	3		12	ns	Figure 5
t19	RXD Hold Time After CHIPx8↓	10			ns	Figure 5
t20	RXD DELAY After CHIPx8↓			20	ns	Figure 5
t21	FMCLK PERIOD		2.78		μs	FM RX Figure 6
t22	FMCLK HIGH TIME		1.39		μs	FM RX Figure 6
t23	FMCLK LOW TIME		1.39		μs	FM RX Figure 6
t24	FMCLK RISE TIME			12	ns	FM RX Figure 6
t25	FMCLK FALL TIME			12	ns	FM RX Figure 6
t26	RXFMSTB HIGH TIME	1			μs	FM RX Figure 6
t27	RXFMSTB -FMCLK↓ SETUP TIME	50			ns	FM RX Figure 6
t28	FMCLK $\downarrow$ – RXFMSTB HOLD TIME	50			ns	FM RX Figure 6
t29	FMCLK $\downarrow$ – OUTPUT DATA DELAY			50	ns	FM RX Figure 6
t30	ADCENA HIGH _ CONVERSION			40	μs	General purpose ADC Figure 7
t31	ADCENA HIGH TIME	100			ns	General purpose ADC Figure 7
t32		100			ns	General purpose ADC Figure 7
t33	ADCCLK PERIOD		2.44		μs	General purpose ADC Figure 7
t34	ADCCLK HIGH TIME		0.81		μs	General purpose ADC Figure 7
t35	ADCCLK LOW TIME		1.62		μs	General purpose ADC Figure 7
t36	ADCCLK RISE TIME			12	ns	General purpose ADC Figure 7
t37	ADCCLK FALL TIME			12	ns	General purpose ADC Figure 7
t38	ADCDATA VALID BEFORE ADCCLK↑		1		μs	General purpose ADC Figure 7
t39	ADC DATA HOLD TIME	5			ns	General purpose ADC Figure 7
t40	SCLK-SDATA setup time	20			ns	Serial Interface Figure 8
t41	SCLK-SDATA hold time	20			ns	Serial Interface Figure 8
t42	SCLK pulse width	50			ns	Serial Interface Figure 8
t43	SLATCH-SCLK setup time	20			ns	Serial Interfsce Figure 8
t44	SLATCH pulse width	50			ns	Serial Interface Figure 8
t45	SCLK period	100			ns	Serial Interface Figure 8

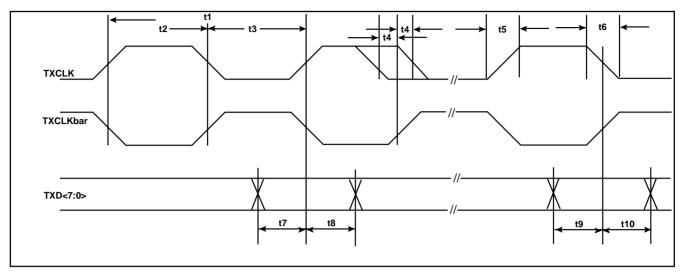


Figure 3 CDMA TX Mode

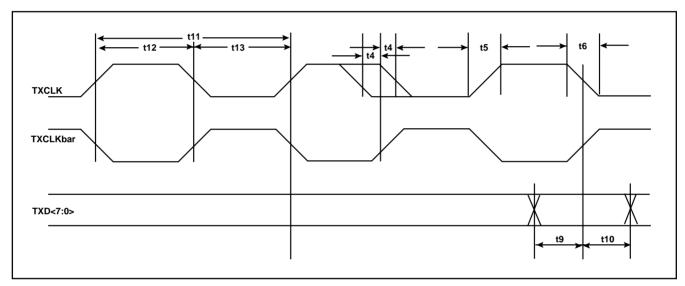


Figure 4. FM TX MODE

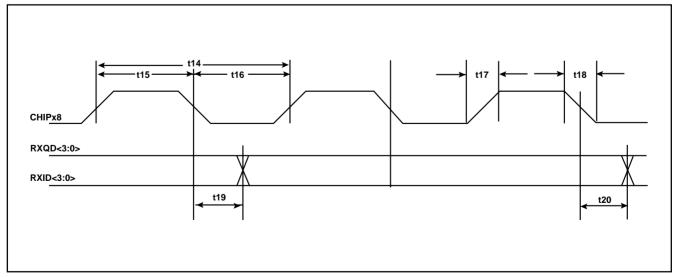


Figure 5. CDMA RX MODE

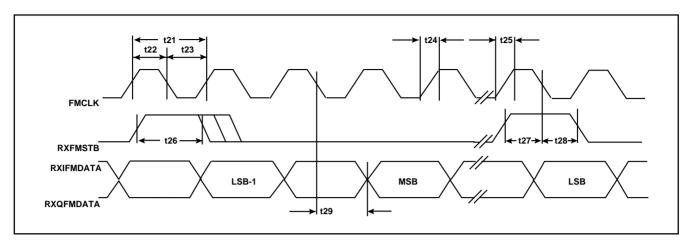


Figure 6. FM RX MODE

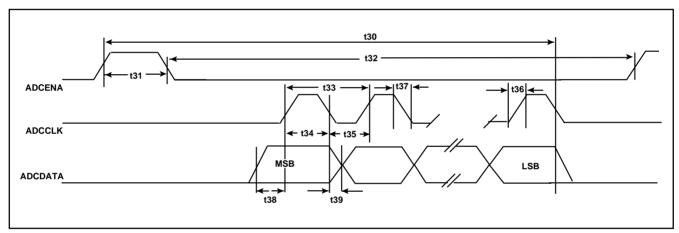


Figure 7. General purpose ADC

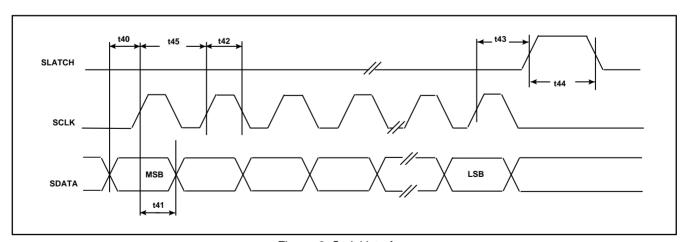


Figure 8. Serial interface

### **PROGRAMMING and CONTROL**

The control modes for Pluto can be set via external pins or via a 3 wire serial interface. On initialising Pluto control is from external pins but can then be set for programming from the serial interface by setting the appropriate bit in a serial input word. The Rx second LO synthesiser is programmed via the serial interface: the Tx IF synthesiser is fixed and requires no programming.

#### **Mode Control - External**

The control modes are set by the pins SLEEPB, FMB and IDLEB as shown in the table below:

SLEEPB (Pin 9)	FMB (Pin 7)	IDLEB (Pin 5)	Mode				
0	X	Х	Sleep Mode				
1	0	0	FM Receive only				
1	0	1	FM Receive and Transmit				
1	1	0	CDMA Receive only				
1	1	1	CDMA Receive and Transmit				

#### **Aux ADC Selection - External**

The auxiliary analog to digital converters can be selected via pins S0 and S1 as shown in the table below:

S1 (Pin 39)	S0 (Pin 33)	ADC selected
0	0	ADC<0>
0	0	ADC<1>
1	0	ADC<2>
1	1	ADC<3>

ADC selection can also be programmed to be via serial interface if required

#### **Serial Interface**

The 3 wire serial interface (SDATA, SCLK and SLATCH) is programmed using 24 bit words as shown below. Timing details are shown in Figure 8.

	MSB																							LSB
	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
WORD1	0	0	0	0	0	Х	Χ	RXC						RX	DIV<	:13:0>							0	0
WORD2	SOP	FMB	IDB	SLB	0	0	0	0	0	0	AD1	AD0	TST	0	0	0	0	0	MXS	СТВ	0	0	0	1

X unused

RXC RX Synth comparison frequency :0 = 30kHz, 1 = 5kHZ

RXDIV<13:0> RX Synth divider ratio

SOP External/ Serial mode selection: 0 = Ext, 1 = Serial SLB, IDB, FMB Sleep, Idle and FM mode control bits (serial mode)

TST Test Mode Control - This is for test purposes only and should be set to 0

AD1, AD0 Aux ADC select bits (serial mode)

MXS External / Serial Aux ADC select: 0 = Ext

CTB Enable Rx Calibration: 0 = calibration mode

If SOP is high mode control is via serial bits FMB, IDB and SLB, instead of external pins FMB, IDLEB and SLEEPB. If MXS is high then ADC selection is via AD1, AD0 instead of S<1>, S<0>

### INITIALISATION

#### **Transmit**

On power-up or reset (RESETB) the Tx reconstruction filters are tuned to give the specified cut-off frequency. This calibration is internal and requires no external input. The calibration time is 1ms.

#### Receive

On power-up or reset (RESETB) an autocalibration algorithm is started which can be used to tune the programmable filters in Jupiter. (Jupiter is a programmable active filter designed for use in dual mode CDMA/AMPS system -further details of which can be found in the Jupiter Datasheet). The autocalibration is also initiated when Pluto is switched into CDMA mode via FMB control.

When Pluto enters calibration mode the En Test (pin 68) goes high. A test signal at 364kHz is then generated at the Vtest output (pin 67). This signal is input to Jupiter which provides a response which is digitised by the I and Q Rx ADCs. An output DAC - BAL - (Pin 71) then tunes the Q channel to match the I and Q channel amplitude via a successive approximation routine. The test signal is then switched to 728kHz which is above the required cut off of the filter. DAC outputs, FC\_I and FC\_Q are then adjusted to tune the I and Q filters to the correct amplitude with reference to the in band test signal. The filter cut off is tuned to 690kHz. Oversampling in the Rx ADC's ensures sufficient accuracy for the calibration. This calibration routine takes 26ms and after completion En Test goes low and the test signal Vtest is disabled. Only the CDMA filter is tuned, the matching within Jupiter ensures that the FM (AMPS) filter performance meets specification.

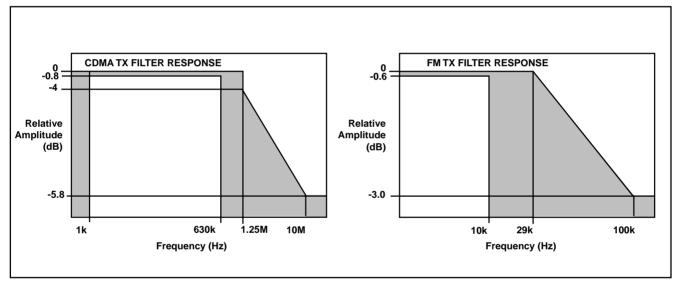


Figure 9 Baseband RX interface circuit

### **RECOMMENDED OPERATING CONDITIONS**

Characteristic		Value		Units	Conditions
	Min	Тур	Max		
Operating voltage range	2.7		3.6	V	
Operating temperature range	-40		+85	°C	
Input high voltage, VIH	VDD-0.8			V	
Input low voltage, VIL			0.8	V	
Master clock amplitude	800			mV pk-pk	AC coupled 19.68MHz sinusoidal
					signal
Input current, IIH			0.1	μΑ	
Input capacitance, CIN			5	pF	
Output high voltage, VOH	VDD-0.4			V	IOUT = 100μA
Output low voltage, VOL			0.4	V	IOUT = 100μA
Tri-state leakage current			10	μΑ	

### **ELECTRICAL CHARACTERISTICS**

 $T_{AMB} = -30$ °C to +70°C,  $V_{CC} = +2.7$  to +3.6V. These characteristics are guaranteed by either production test or design. They apply within the specified ambient temperature and supply voltage ranges unless otherwise stated.

Characteristic		Value		Units	Conditions
	Min	Тур	Max		
CDMA TX Interface					
Resolution	8			Bits	
Integral non-linearity			0.5	LSB	
Differential non-linearity			0.5	LSB	
Full scale output Voltage	1.6	2.05	2.4	V	Vpp differential
Output common mode Voltage	1.1	1.2	1.3	V	
+Ve output Voltage in Power-down mode	Vdd-0.16	Vdd-0.1	Vdd-0.04	V	
-Ve output Voltage in Power-down mode	0.04	0.1	0.16	V	
I,Q gain mismatch			0.15	dB	
I,Q phase imbalance			1	degrees	
Differential offset			TBD	mV	
PSRR		50		dB	VDD to differential I & Q outputs,
					100mV pk-pk at 100kHz
Load resistance	5			$k\Omega$	
Load capacitance			20	pF	
Filter type					Butterwoth low pass
Filter order		5th			
Filter cut off					See Figure 9
Filter Pass Band ripple					See Figure 9
Filter stop band attenuation					See Figure 9
Sample rate			6		Msample/s

**ELECTRICAL CHARACTERISTICS** (continued)  $T_{AMB} = -30^{\circ}\text{C to } +70^{\circ}\text{C}, \ V_{CC} = +2.7 \text{ to } +3.6 \text{V}. \ \text{These characteristics are guaranteed by either production test or design. They apply within the specified ambient temperature and supply voltage ranges unless otherwise stated.}$ 

Characteristic		Value		Units	Conditions
	Min	Тур	Max		
FM TX Interface					
Resolution	8			Bits	
Integral non linearity			0.5	LSB	
Differential non linearity			0.5	LSB	
Output Voltage range		550		mVpp	
Output Voltage mid scale		0		V	Differential
PSRR		50		dB	Vdd to output, 100mV
					pk-pk at 100kHz
Load resistance	5			kΩ	
Load capacitance			20	pF	
Filter type					Butterworth low pass
Filter order		3rd			
Filter Pass Band ripple					See Figure 9
Filter stop band attenuation					See Figure 9
Filter cut-off					See Figure 9
CDMA RX Interface					
Resolution	4			Bits	
Full scale input voltage		1		V pk-pk	Measured differentially
Input common mode range		Vdd -1.4		V	
Input sample rate		9.8304		Ms/s	
Input resistance (dc)	20			kΩ	
Input capacitance		10	20	pF	
Integral non linearity			0.15	LSB	
Differential non linearity			0.175	LSB	
IRX and QRX gain matching			0.25	dB	

### **ELECTRICAL CHARACTERISTICS** (CDMA BASEBAND RX INTERFACE CIRCUIT) continued

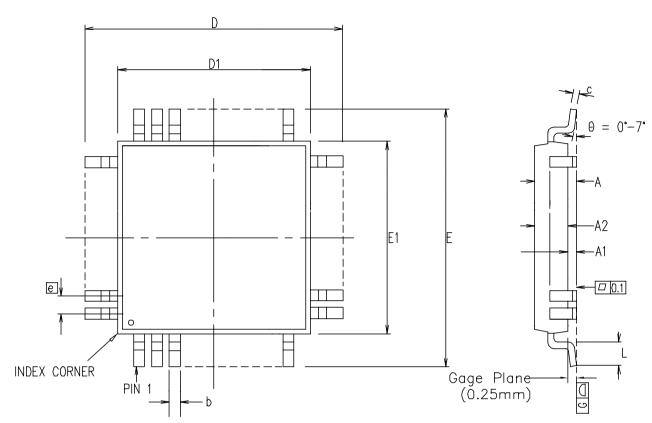
 $T_{AMB}$  = -30°C to +70°C,  $V_{CC}$  = +2.7 to +3.6V. These characteristics are guaranteed by either production test or design. They apply within the specified ambient temperature and supply voltage ranges unless otherwise stated.

Characteristic	Value			Units	Conditions
	Min	Тур	Max		
FM RX Interface					
Resolution	8			Bits	
Full scale input voltage		1		V pk-pk	Measured differentially
Input dc level		Vdd-1.4		V	
Input sample rate		30	50	ks/s	
Input resistance (dc)	100			kΩ	
Input capacitance		10	20	pF	
Integral non linearity			±1.5	LSB	
Differential non linearity			±0.75	LSB	
AUXILIARY CONVERTER SECTION					
Resolution	8			Bits	
ADC full scale range		2.5		V	
ADC zero scale range		0.5		V	
Integral non linearity			±1.25	LSB	
Differential non linearity			±0.75	LSB	
Conversion time		20		ks/s	
ADCCLK		410		kHz	

### **ELECTRICAL CHARACTERISTICS** (Continued)

 $T_{AMB}$  = -30°C to +70°C,  $V_{CC}$  = +2.7 to +3.6V. These characteristics are guaranteed by either production test or design. They apply within the specified ambient temperature and supply voltage ranges unless otherwise stated.

Characteristic		Value		Units	Conditions
	Min	Тур	Max		
TRANSMIT SYNTHESISER					
Input Frequency		65.19		MHz	
Lock mode output current current		16		μΑ	Rset = $40k\Omega$
Acquisition mode output current		176		μΑ	Rset = $40k\Omega$
PD output compliance	0.5		Vdd-0.5	V	
RECEIVER SYNTHESISER					
Input frequency	42.69		52.595	MHz	
Lock mode output current		16		μΑ	Rset = $40k\Omega$
Acquisition mode output current		176		V	Rset = $40k\Omega$
PD output compliance	0.5		Vdd-0.5	V	
POWER SUPPLY CURRENTS					
Sleep			1	mA	
CDMA_IDLE		11	15	mA	
FM_ILDE		5	7	mA	
CDMA_RTTX		18	32	mA	
FM_RXTX		11	15	mA	



	Control D	imensions		Altern. D	imensions					
Symbol		metres			ches					
	MIN	MAX		MIN	MAX					
Α		1.60			0.063					
A1	0.05	0.15		0.002	0.006					
A2	1.35	1.45		0.053	0.057					
D	14.00	O BSC		0.551 BSC						
D1	12.00	O BSC		0.472 BSC						
E	14.00	O BSC		0.55	1 BSC					
E1	12.00	O BSC		0.472	2 BSC					
L	0.45	0.75		0.018	0.030					
е	0.50	BSC		0.020 BSC						
b	0.17	0.27		0.007	0.011					
С	0.09	0.20		0.004	0.008					
		Pin	featı	ıres						
N			08							
ND	20									
NE	20									
NOTE		S	QUAR	!E						

Conforms to JEDEC MS-026 BDD Iss. C

### Notes:

- Pin 1 indicator may be a corner chamfer, dot or both.
   Controlling dimensions are in millimeters.
   The top package body size may be smaller than the bottom package body size by a max. of 0.15 mm.
   Dimension D1 and E1 do not include mould protusion.
   Dimension b does not include dambar protusion.
   Coplanarity, measured at seating plane G, to be 0.08 mm max.

This drawing supersedes 418/ED/51210/032 (Swindon)

© Mitel							ORIGINATING SITE: SWINDON
ISSUE	1	2			MITEL	SEMICONDUCTOR	Title: Package Outline Drawing for 80 Ids LQFP (GP) (12x12x1.4) mm, Body+2.0 mm  Drawing Number
ACN	201375	207088					
DATE	290CT96	7JUL99					
APPROVED							GPD00255



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