

# Mobile Low-Power DDR SDRAM

**MT46H8M16LF – 2 Meg x 16 x 4 Banks**
**MT46H4M32LF – 1 Meg x 32 x 4 Banks**

## Features

- $V_{DD}/V_{DDQ} = 1.70\text{--}1.95\text{V}$
- Bidirectional data strobe per byte of data (DQS)
- Internal, pipelined double data rate (DDR) architecture; two data accesses per clock cycle
- Differential clock inputs (CK and CK#)
- Commands entered on each positive CK edge
- DQS edge-aligned with data for READs; center-aligned with data for WRITEs
- 4 internal banks for concurrent operation
- Data masks (DM) for masking write data; one mask per byte
- Programmable burst lengths (BL): 2, 4, 8, or 16
- Concurrent auto precharge option is supported
- Auto refresh and self refresh modes
- 1.8V LVCMOS-compatible inputs
- Temperature-compensated self refresh (TCSR)
- Partial-array self refresh (PASR)
- Deep power-down (DPD)
- Status read register (SRR)
- Selectable output drive strength (DS)
- Clock stop capability
- 64ms refresh

## Options

- $V_{DD}/V_{DDQ}$ 
  - 1.8V/1.8V
- Configuration
  - 8 Meg x 16 (2 Meg x 16 x 4 banks)
  - 4 Meg x 32 (1 Meg x 32 x 4 banks)
- Row size option
  - JEDEC-standard option
- Plastic "green" package
  - 60-ball VFBGA (8mm x 9mm)<sup>1</sup>
  - 90-ball VFBGA (8mm x 13mm)<sup>2</sup>
- Timing – cycle time
  - 5ns @ CL = 3 (200 MHz)
  - 5.4ns @ CL = 3 (185 MHz)
  - 6ns @ CL = 3 (166 MHz)
  - 7.5ns @ CL = 3 (133 MHz)
- Operating temperature range
  - Commercial (0° to +70°C)
  - Industrial (–40°C to +85°C)
- Design revision

## Marking

H

8M16

4M32

LF

BF

B5

-5

-54

-6

-75

None

IT

:K

- Notes: 1. Only available for x16 configuration.  
 2. Only available for x32 configuration.

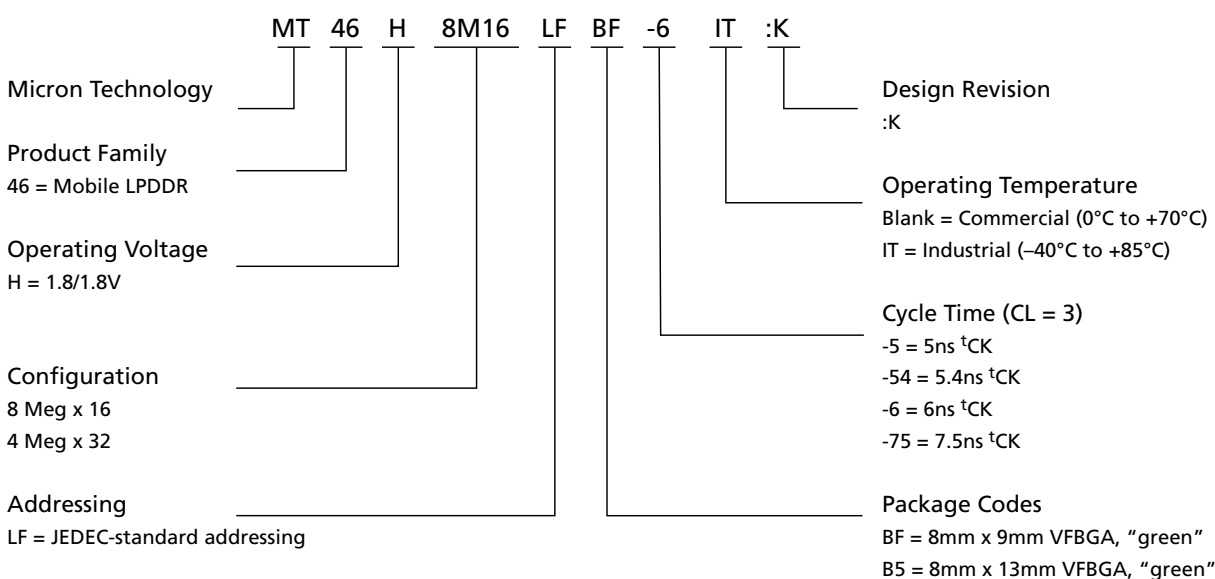
**Table 1: Key Timing Parameters (CL = 3)**

Speed Grade	Clock Rate	Access Time
-5	200 MHz	5.0ns
-54	185 MHz	5.0ns
-6	166 MHz	5.0ns
-75	133 MHz	6.0ns

**Table 2: Configuration Addressing**

Architecture	8 Meg x 16	4 Meg x 32
Configuration	2 Meg x 16 x 4 banks	1 Meg x 32 x 4 banks
Refresh count	4K	4K
Row addressing	4K A[11:0]	4K A[11:0]
Column addressing	512 A[8:0]	256 A[7:0]

**Figure 1: 128Mb Mobile LPDDR Part Numbering**



## FBGA Part Marking Decoder

Due to space limitations, FBGA-packaged components have an abbreviated part marking that is different from the part number. Micron's FBGA part marking decoder is available at [www.micron.com/decoder](http://www.micron.com/decoder).

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## General Description

The 128Mb Mobile DDR SDRAM is a high-speed CMOS, dynamic random-access memory containing 134,217,728 bits. It is internally configured as a quad-bank DRAM. Each of the x16's 33,554,432-bit banks is organized as 4096 rows by 512 columns by 16 bits. Each of the x32's 33,554,432-bit banks is organized as 4096 rows by 256 columns by 32 bits.

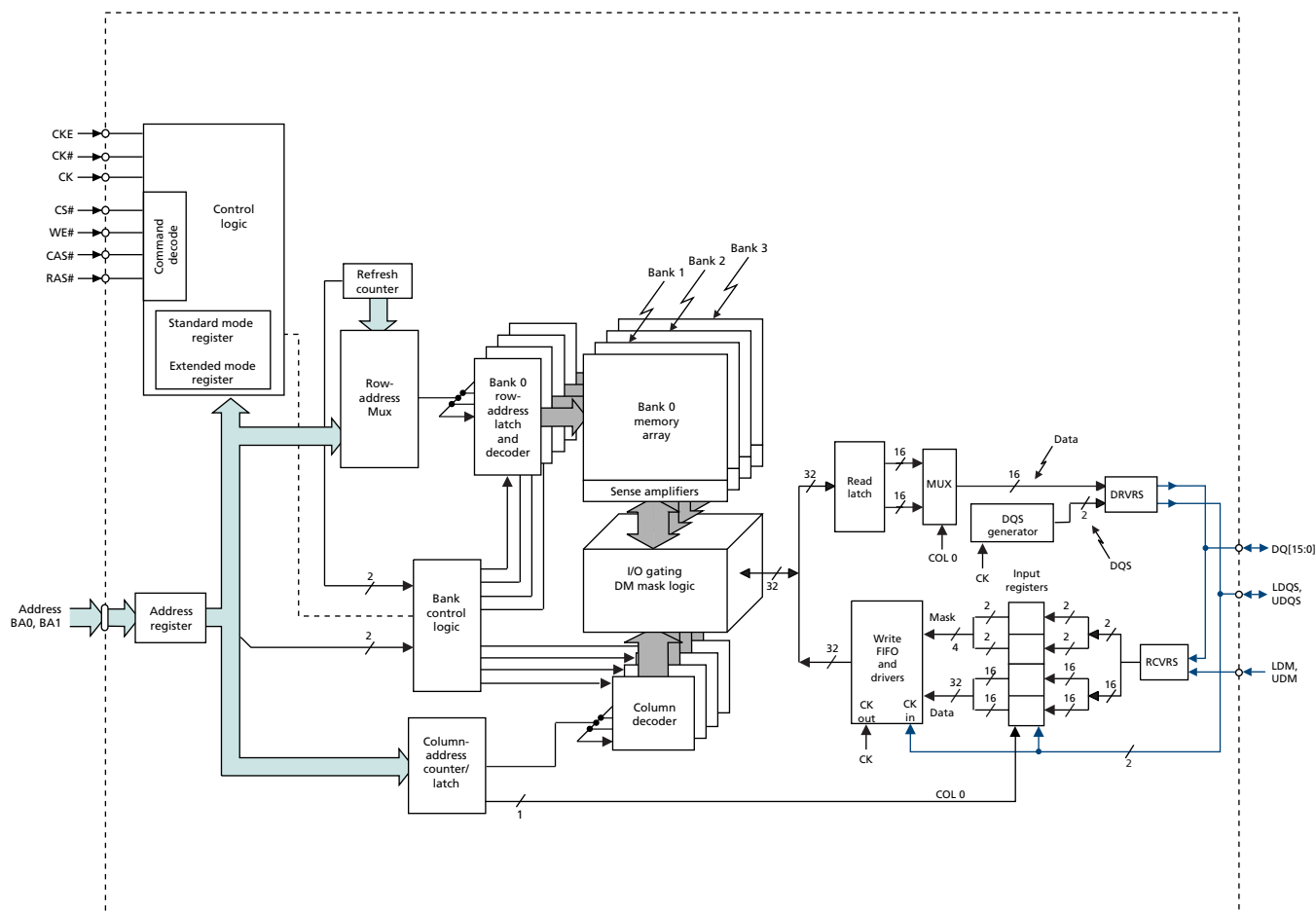
**Note:**

1. Throughout this data sheet, various figures and text refer to DQs as "DQ." DQ should be interpreted as any and all DQ collectively, unless specifically stated otherwise. Additionally, the x16 is divided into 2 bytes: the lower byte and the upper byte. For the lower byte (DQ[7:0]), DM refers to LDM and DQS refers to LDQS. For the upper byte (DQ[15:8]), DM refers to UDM and DQS refers to UDQS. The x32 is divided into 4 bytes. For DQ[7:0], DM refers to DM0 and DQS refers to DQS0. For DQ[15:8], DM refers to DM1 and DQS refers to DQS1. For DQ[23:16], DM refers to DM2 and DQS refers to DQS2. For DQ[31:24], DM refers to DM3 and DQS refers to DQS3.
2. Complete functionality is described throughout the document; any page or diagram may have been simplified to convey a topic and may not be inclusive of all requirements.
3. Any specific requirement takes precedence over a general statement.

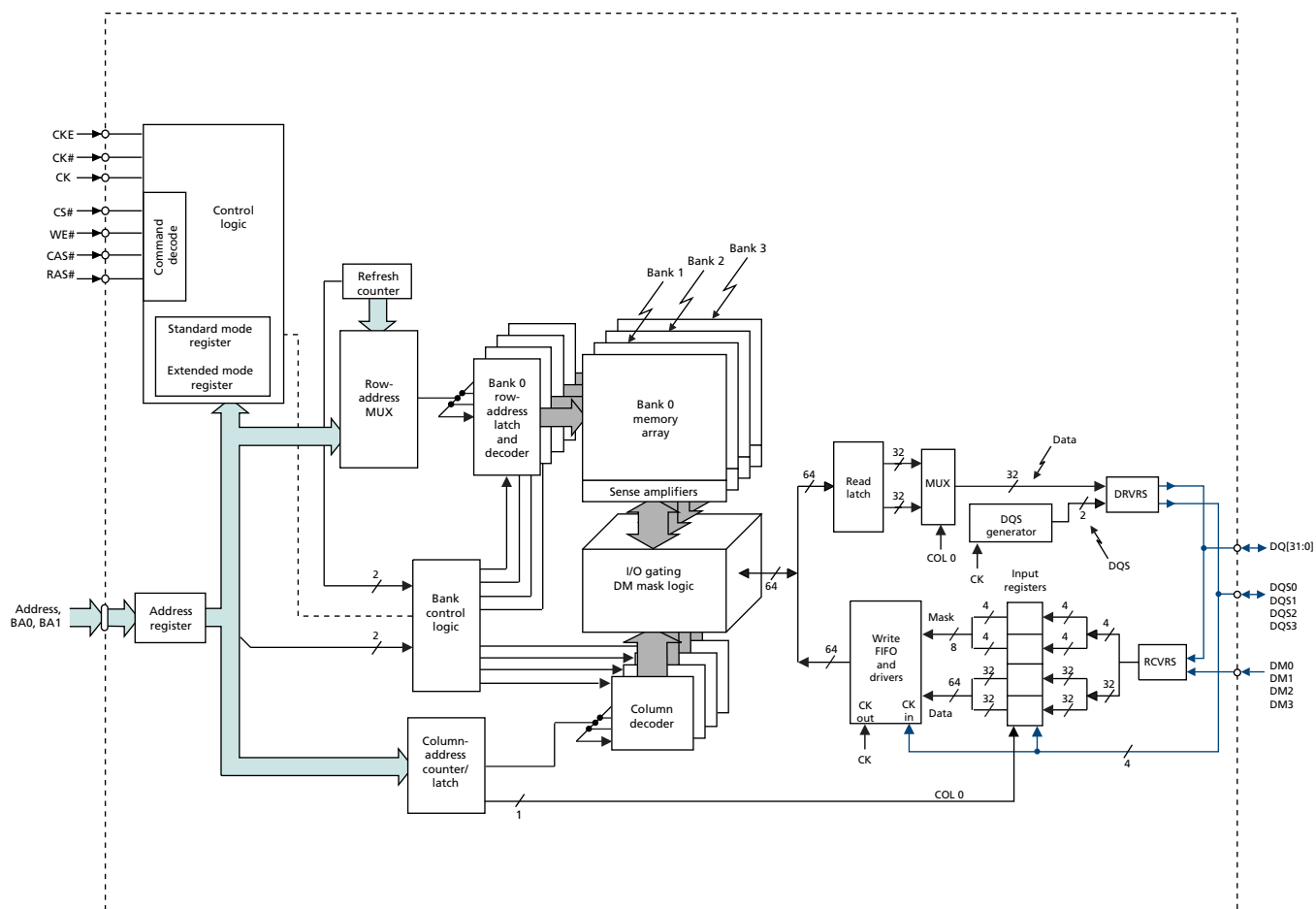


## Functional Block Diagrams

**Figure 2: Functional Block Diagram (x16)**

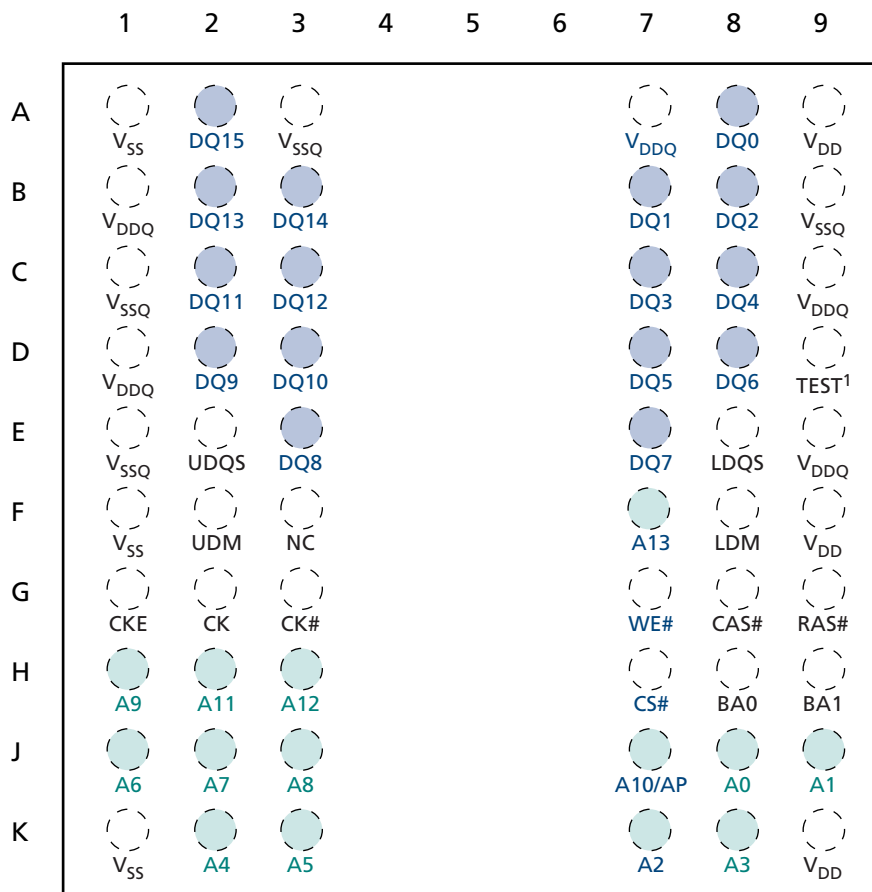


**Figure 3: Functional Block Diagram (x32)**



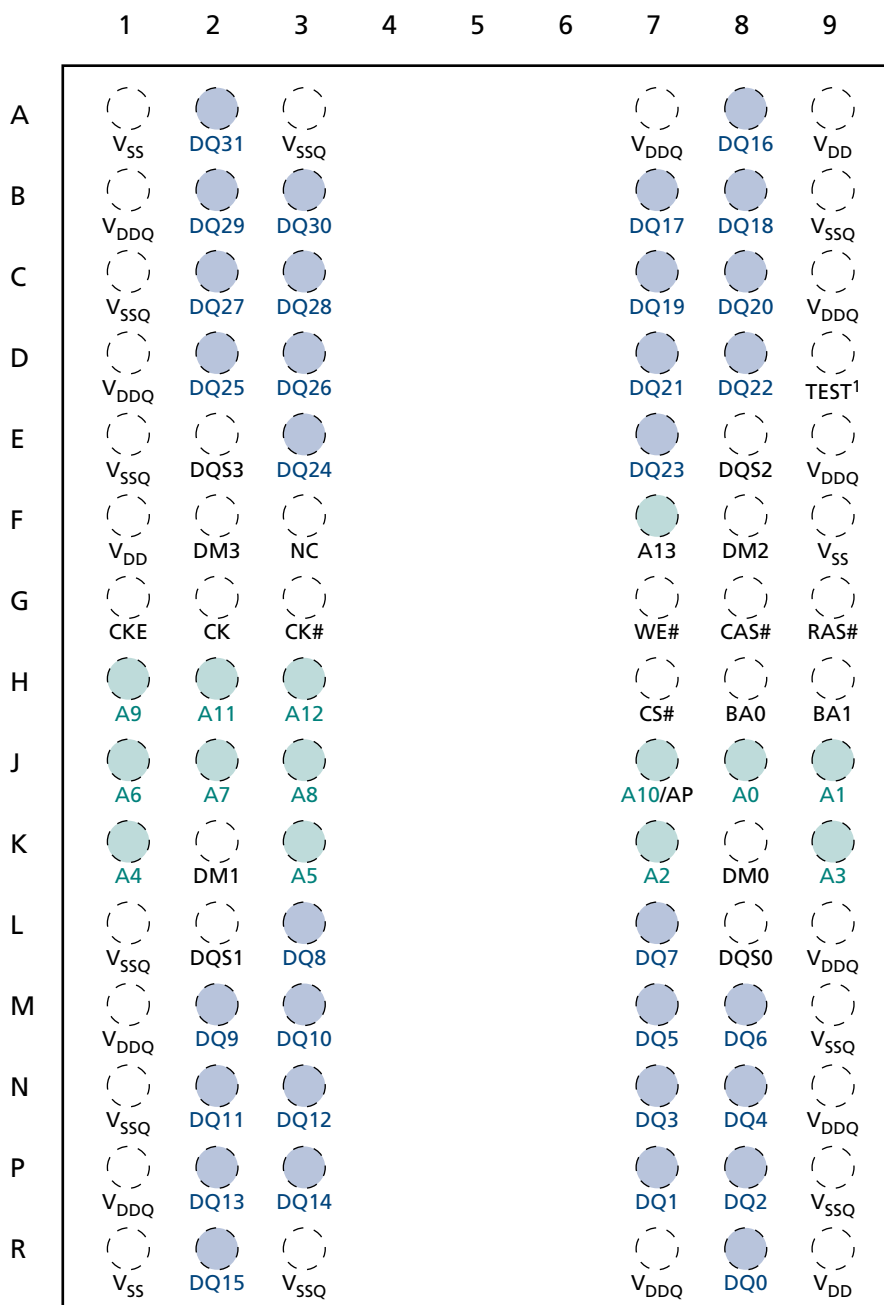
## Ball Assignments

Figure 4: 60-Ball VFBGA – Top View, x16 only



- Notes:
1. D9 is a test pin that must be tied to V<sub>SS</sub> or V<sub>SSQ</sub> in normal operations.
  2. Unused address pins become RFU.

**Figure 5: 90-Ball VFBGA – Top View, x32 only**



- Notes:
1. D9 is a test pin that must be tied to V<sub>SS</sub> or V<sub>SSQ</sub> in normal operations.
  2. Unused address pins become RFU.

## Ball Descriptions

The ball descriptions table is a comprehensive list of all possible balls for all supported packages. Not all balls listed are supported for a given package.

**Table 3: Ball Descriptions**

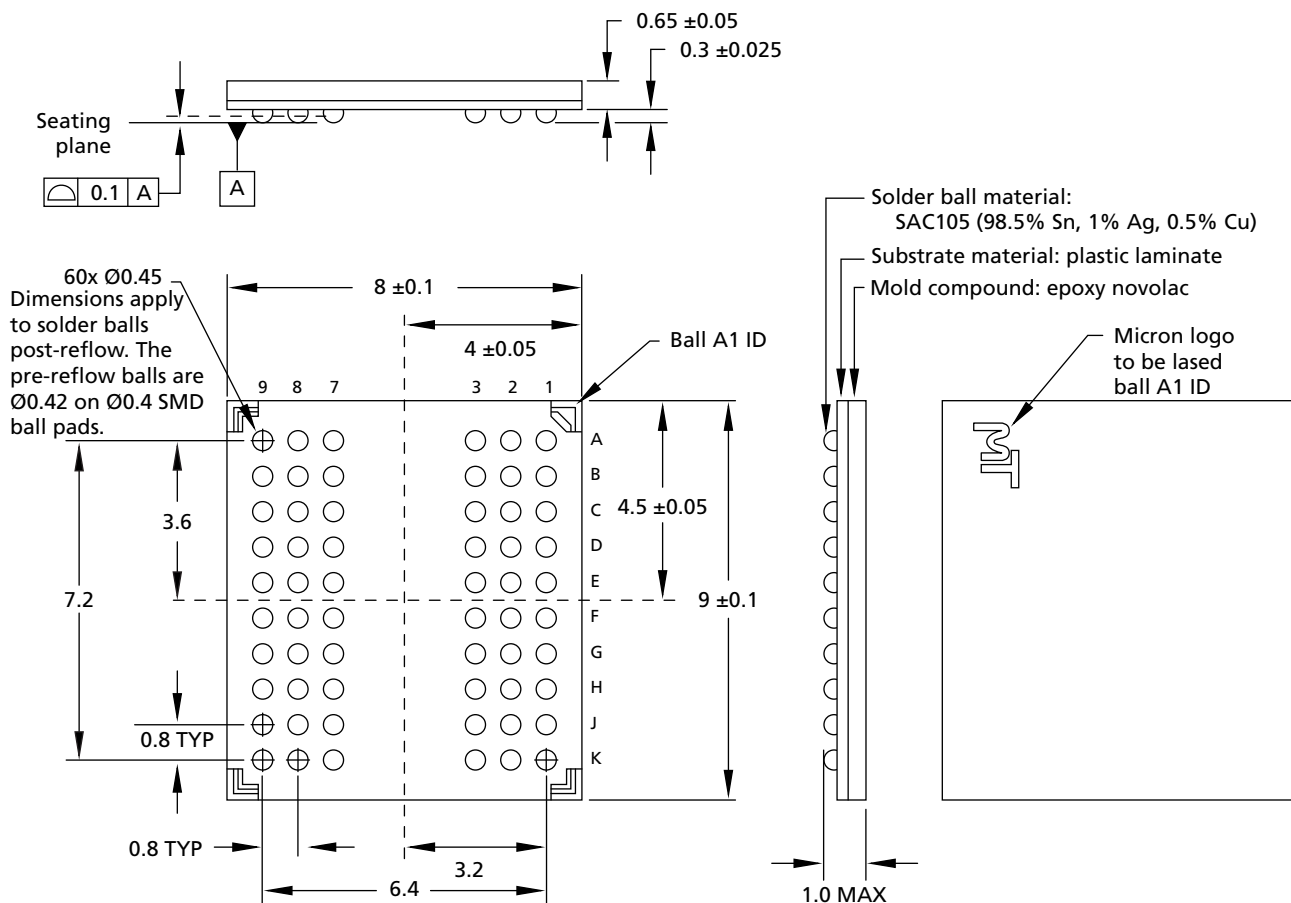
Symbol	Type	Description
CK, CK#	Input	Clock: CK is the system clock input. CK and CK# are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK and the negative edge of CK#. Input and output data is referenced to the crossing of CK and CK# (both directions of the crossing).
CKE CKE0, CKE1	Input	Clock enable: CKE HIGH activates, and CKE LOW deactivates, the internal clock signals, input buffers, and output drivers. Taking CKE LOW enables PRECHARGE power-down and SELF REFRESH operations (all banks idle), or ACTIVE power-down (row active in any bank). CKE is synchronous for all functions except SELF REFRESH exit. All input buffers (except CKE) are disabled during power-down and self refresh modes. CKE0 is used for a single LPDDR product. CKE1 is used for dual LPDDR products and is considered RFU for single LPDDR MCPs.
CS# CS0#, CS1#	Input	Chip select: CS# enables (registered LOW) and disables (registered HIGH) the command decoder. All commands are masked when CS# is registered HIGH. CS# provides for external bank selection on systems with multiple banks. CS# is considered part of the command code. CS0# is used for a single LPDDR product. CS1# is used for dual LPDDR products and is considered RFU for single LPDDR MCPs.
RAS#, CAS#, WE#	Input	Command inputs: RAS#, CAS#, and WE# (along with CS#) define the command being entered.
UDM, LDM (x16) DM[3:0] (x32)	Input	Input data mask: DM is an input mask signal for write data. Input data is masked when DM is sampled HIGH along with that input data during a WRITE access. DM is sampled on both edges of DQS. Although DM balls are input-only, the DM loading is designed to match that of DQ and DQS balls.
BA0, BA1	Input	Bank address inputs: BA0 and BA1 define to which bank an ACTIVE, READ, WRITE, or PRECHARGE command is being applied. BA0 and BA1 also determine which mode register is loaded during a LOAD MODE REGISTER command.
A[13:0]	Input	Address inputs: Provide the row address for ACTIVE commands, and the column address and auto precharge bit (A10) for READ or WRITE commands, to select one location out of the memory array in the respective bank. During a PRECHARGE command, A10 determines whether the PRECHARGE applies to one bank (A10 LOW, bank selected by BA0, BA1) or all banks (A10 HIGH). The address inputs also provide the op-code during a LOAD MODE REGISTER command. The maximum address range is dependent upon configuration. Unused address balls become RFU.
TEST	Input	Test pin: Must be tied to V <sub>SS</sub> or V <sub>SSQ</sub> in normal operations.
DQ[15:0] (x16) DQ[31:0] (x32)	Input/ output	Data input/output: Data bus for x16 and x32.
LDQS, UDQS (x16) DQS[3:0] (x32)	Input/ output	Data strobe: Output with read data, input with write data. DQS is edge-aligned with read data, center-aligned in write data. It is used to capture data.
TQ	Output	Temperature sensor output: TQ HIGH when LPDDR T <sub>J</sub> exceeds 85°C.
V <sub>DDQ</sub>	Supply	DQ power supply.

**Table 3: Ball Descriptions (Continued)**

Symbol	Type	Description
V <sub>SSQ</sub>	Supply	DQ ground.
V <sub>DD</sub>	Supply	Power supply.
V <sub>SS</sub>	Supply	Ground.
NC	–	No connect: May be left unconnected.
RFU	–	Reserved for future use. Balls marked RFU may or may not be connected internally. These balls should not be used. Contact factory for details.

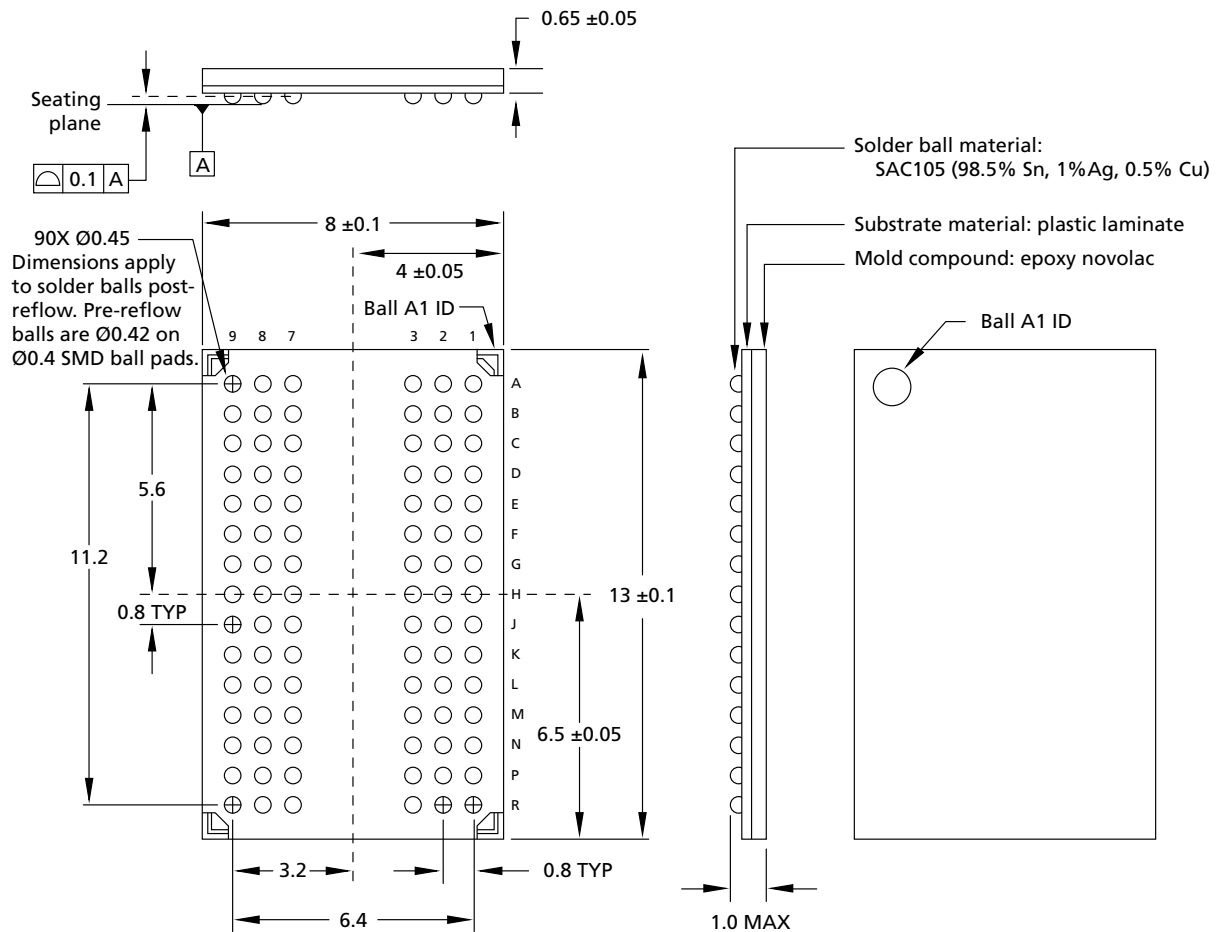
## Package Dimensions

**Figure 6: 60-Ball VFBGA (8mm x 9mm), Package code: BF**



Note: 1. All dimensions are in millimeters.

**Figure 7: 90-Ball VFBGA (8mm x 13mm), Package code: B5**



Note: 1. All dimensions are in millimeters.



## Electrical Specifications

Stresses greater than those listed may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

**Table 4: Absolute Maximum Ratings**

Note 1 applies to all parameters in this table

Parameter	Symbol	Min	Max	Unit
$V_{DD}/V_{DDQ}$ supply voltage relative to $V_{SS}$	$V_{DD}/V_{DDQ}$	-1.0	+2.4	V
Voltage on any pin relative to $V_{SS}$	$V_{IN}$	-0.5	+2.4 or ( $V_{DDQ} + 0.3V$ ), whichever is less	V
Storage temperature (plastic)	$T_{STG}$	-55	+150	°C

Note: 1.  $V_{DD}$  and  $V_{DDQ}$  must be within 300mV of each other at all times.  $V_{DDQ}$  must not exceed  $V_{DD}$ .

**Table 5: AC/DC Electrical Characteristics and Operating Conditions**

Notes 1–5 apply to all parameters/conditions in this table;  $V_{DD}/V_{DDQ} = 1.70\text{--}1.95V$

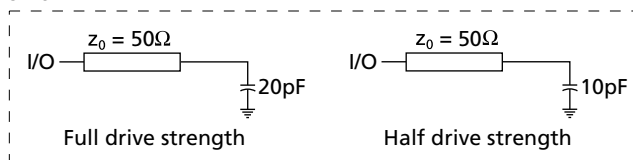
Parameter/Condition	Symbol	Min	Max	Unit	Notes
Supply voltage	$V_{DD}$	1.70	1.95	V	6, 7
I/O supply voltage	$V_{DDQ}$	1.70	1.95	V	6, 7
<b>Address and command inputs</b>					
Input voltage high	$V_{IH}$	$0.8 \times V_{DDQ}$	$V_{DDQ} + 0.3$	V	8, 9
Input voltage low	$V_{IL}$	-0.3	$0.2 \times V_{DDQ}$	V	8, 9
<b>Clock inputs (CK, CK#)</b>					
DC input voltage	$V_{IN}$	-0.3	$V_{DDQ} + 0.3$	V	10
DC input differential voltage	$V_{ID(DC)}$	$0.4 \times V_{DDQ}$	$V_{DDQ} + 0.6$	V	10, 11
AC input differential voltage	$V_{ID(AC)}$	$0.6 \times V_{DDQ}$	$V_{DDQ} + 0.6$	V	10, 11
AC differential crossing voltage	$V_{IX}$	$0.4 \times V_{DDQ}$	$0.6 \times V_{DDQ}$	V	10, 12
<b>Data inputs</b>					
DC input high voltage	$V_{IH(DC)}$	$0.7 \times V_{DDQ}$	$V_{DDQ} + 0.3$	V	8, 9, 13
DC input low voltage	$V_{IL(DC)}$	-0.3	$0.3 \times V_{DDQ}$	V	8, 9, 13
AC input high voltage	$V_{IH(AC)}$	$0.8 \times V_{DDQ}$	$V_{DDQ} + 0.3$	V	8, 9, 13
AC input low voltage	$V_{IL(AC)}$	-0.3	$0.2 \times V_{DDQ}$	V	8, 9, 13
<b>Data outputs</b>					
DC output high voltage: Logic 1 ( $I_{OH} = -0.1mA$ )	$V_{OH}$	$0.9 \times V_{DDQ}$	–	V	
DC output low voltage: Logic 0 ( $I_{OL} = 0.1mA$ )	$V_{OL}$	–	$0.1 \times V_{DDQ}$	V	
<b>Leakage current</b>					
Input leakage current Any input $0V \leq V_{IN} \leq V_{DD}$ (All other pins not under test = 0V)	$I_I$	-1	+1	$\mu A$	

**Table 5: AC/DC Electrical Characteristics and Operating Conditions (Continued)**

Notes 1–5 apply to all parameters/conditions in this table;  $V_{DD}/V_{DDQ} = 1.70\text{--}1.95V$

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Output leakage current (DQ are disabled; $0V \leq V_{OUT} \leq V_{DDQ}$ )	$I_{OZ}$	–5	+5	$\mu A$	
<b>Operating temperature</b>					
Commercial	$T_A$	0	+70	$^{\circ}C$	
Industrial	$T_A$	–40	+85	$^{\circ}C$	

- Notes:
1. All voltages referenced to  $V_{SS}$ .
  2. All parameters assume proper device initialization.
  3. Tests for AC timing,  $I_{DD}$ , and electrical AC and DC characteristics may be conducted at nominal supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.
  4. Outputs measured with equivalent load; transmission line delay is assumed to be very small:



5. Timing and  $I_{DD}$  tests may use a  $V_{IL}$ -to- $V_{IH}$  swing of up to 1.5V in the test environment, but input timing is still referenced to  $V_{DDQ/2}$  (or to the crossing point for CK/CK#). The output timing reference voltage level is  $V_{DDQ/2}$ .
6. Any positive glitch must be less than one-third of the clock cycle and not more than +200mV or 2.0V, whichever is less. Any negative glitch must be less than one-third of the clock cycle and not exceed either –150mV or +1.6V, whichever is more positive.
7.  $V_{DD}$  and  $V_{DDQ}$  must track each other and  $V_{DDQ}$  must be less than or equal to  $V_{DD}$ .
8. To maintain a valid level, the transitioning edge of the input must:
  - 8a. Sustain a constant slew rate from the current AC level through to the target AC level,  $V_{IL(AC)}$  Or  $V_{IH(AC)}$ .
  - 8b. Reach at least the target AC level.
  - 8c. After the AC target level is reached, continue to maintain at least the target DC level,  $V_{IL(DC)}$  or  $V_{IH(DC)}$ .
9.  $V_{IH}$  overshoot:  $V_{IHmax} = V_{DDQ} + 1.0V$  for a pulse width  $\leq 3ns$  and the pulse width cannot be greater than one-third of the cycle rate.  $V_{IL}$  undershoot:  $V_{ILmin} = -1.0V$  for a pulse width  $\leq 3ns$  and the pulse width cannot be greater than one-third of the cycle rate.
10. CK and CK# input slew rate must be  $\geq 1 V/ns$  (2 V/ns if measured differentially).
11.  $V_{ID}$  is the magnitude of the difference between the input level on CK and the input level on CK#.
12. The value of  $V_{IX}$  is expected to equal  $V_{DDQ/2}$  of the transmitting device and must track variations in the DC level of the same.
13. DQ and DM input slew rates must not deviate from DQS by more than 10%. 50ps must be added to  $t_{DS}$  and  $t_{DH}$  for each 100 mV/ns reduction in slew rate. If slew rate exceeds 4 V/ns, functionality is uncertain.

**Table 6: Capacitance (x16, x32)**

Note 1 applies to all the parameters in this table

Parameter	Symbol	Min	Max	Unit	Notes
Input capacitance: CK, CK#	$C_{CK}$	1.5	3.0	pF	
Delta input capacitance: CK, CK#	$C_{DCK}$	–	0.25	pF	2
Input capacitance: command and address	$C_I$	1.5	3.0	pF	
Delta input capacitance: command and address	$C_{DI}$	–	0.5	pF	2
Input/output capacitance: DQ, DQS, DM	$C_{IO}$	2.0	4.5	pF	
Delta input/output capacitance: DQ, DQS, DM	$C_{DIO}$	–	0.5	pF	3

- Notes:
1. This parameter is sampled.  $V_{DD}/V_{DDQ} = 1.70\text{--}1.95\text{V}$ ,  $f = 100\text{ MHz}$ ,  $T_A = 25^\circ\text{C}$ ,  $V_{OUT(DC)} = V_{DDQ}/2$ ,  $V_{OUT}$  (peak-to-peak) = 0.2V. DM input is grouped with I/O pins, reflecting the fact that they are matched in loading.
  2. The input capacitance per pin group will not differ by more than this maximum amount for any given device.
  3. The I/O capacitance per DQS and DQ byte/group will not differ by more than this maximum amount for any given device.

## Electrical Specifications – I<sub>DD</sub> Parameters

**Table 7: I<sub>DD</sub> Specifications and Conditions (x16)**

Notes 1–5 apply to all parameters/conditions in this table; V<sub>DD</sub>/V<sub>DDQ</sub> = 1.70–1.95V

Parameter/Condition	Symbol	Max				Unit	Notes	
		-5	-54	-6	-75			
Operating 1 bank active precharge current: $t_{RC} = t_{RC}(\text{MIN})$ ; $t_{CK} = t_{CK}(\text{MIN})$ ; CKE is HIGH; CS is HIGH between valid commands; Address inputs are switching every 2 clock cycles; Data bus inputs are stable	$I_{DD0}$	70	60	50	40	mA	6	
Precharge power-down standby current: All banks idle, CKE is LOW; CS is HIGH, $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2P}$	200	200	200	200	$\mu\text{A}$	7, 8	
Precharge power-down standby current: Clock stopped; All banks idle; CKE is LOW; CS is HIGH; CK = LOW, CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2PS}$	200	200	200	200	$\mu\text{A}$	7	
Precharge nonpower-down standby current: All banks idle CKE = HIGH; CS = HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2N}$	20	17	15	12	mA	9	
Precharge nonpower-down standby current: Clock stopped; All banks idle, CKE = HIGH; CS = HIGH; CK = LOW, CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2NS}$	15	12	10	8	mA	9	
Active power-down standby current: 1 bank active, CKE = LOW; CS = HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3P}$	3	3	3	3	mA	8	
Active power-down standby current: Clock stopped; 1 bank active, CKE = LOW; CS = HIGH; CK = LOW; CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3PS}$	3	3	3	3	mA		
Active nonpower-down standby: 1 bank active, CKE = HIGH; CS = HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3N}$	25	22	20	15	mA	6	
Active nonpower-down standby: Clock stopped; 1 bank active, CKE = HIGH; CS = HIGH; CK = LOW; CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3NS}$	20	17	15	10	mA	6	
Operating burst read: 1 bank active; BL = 4; $t_{CK} = t_{CK}(\text{MIN})$ ; Continuous READ bursts; $I_{OUT} = 0\text{mA}$ ; Address inputs are switching every 2 clock cycles; 50% data changing each burst	$I_{DD4R}$	90	85	80	70	mA	6	
Operating burst write: 1 bank active; BL = 4; $t_{CK} = t_{CK}(\text{MIN})$ ; Continuous WRITE bursts; Address inputs are switching; 50% data changing each burst	$I_{DD4W}$	90	85	80	70	mA	6	
Auto refresh: Burst refresh; CKE = HIGH; Address and control inputs are switching; Data bus inputs are stable	$t_{RFC} = 110\text{ns}$	$I_{DD5}$	95	92	90	85	mA	10
	$t_{RFC} = t_{REFI}$	$I_{DD5A}$	8	6	5	3	mA	10, 11
Deep power-down current: Address and control balls are stable; Data bus inputs are stable	$I_{DD8}$	10	10	10	10	$\mu\text{A}$	7, 13	

**Table 8: I<sub>DD</sub> Specifications and Conditions (x32)**

Notes 1–5 apply to all parameters/conditions in this table; V<sub>DD</sub>/V<sub>DDQ</sub> = 1.70–1.95V

Parameter/Condition	Symbol	Max				Unit	Notes	
		-5	-54	-6	-75			
Operating 1 bank active precharge current: $t_{RC} = t_{RC}(\text{MIN})$ ; $t_{CK} = t_{CK}(\text{MIN})$ ; CKE is HIGH; CS is HIGH between valid commands; Address inputs are switching every 2 clock cycles; Data bus inputs are stable	$I_{DD0}$	70	60	50	40	mA	6	
Precharge power-down standby current: All banks idle; CKE is LOW; CS is HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2P}$	200	200	200	200	$\mu\text{A}$	7, 8	
Precharge power-down standby current: Clock stopped; All banks idle; CKE is LOW; CS is HIGH, CK = LOW, CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2PS}$	200	200	200	200	$\mu\text{A}$	7	
Precharge nonpower-down standby current: All banks idle CKE = HIGH; CS = HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2N}$	20	17	15	12	mA	9	
Precharge nonpower-down standby current: Clock stopped; All banks idle, CKE = HIGH; CS = HIGH; CK = LOW, CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD2NS}$	15	12	10	8	mA	9	
Active power-down standby current: 1 bank active, CKE = LOW; CS = HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3P}$	3	3	3	3	mA	8	
Active power-down standby current: Clock stopped; 1 bank active, CKE = LOW; CS = HIGH; CK = LOW; CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3PS}$	3	3	3	3	mA		
Active nonpower-down standby: 1 bank active; CKE = HIGH; CS = HIGH; $t_{CK} = t_{CK}(\text{MIN})$ ; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3N}$	25	22	20	15	mA	6	
Active nonpower-down standby: Clock stopped; 1 bank active, CKE = HIGH; CS = HIGH; CK = LOW; CK# = HIGH; Address and control inputs are switching; Data bus inputs are stable	$I_{DD3NS}$	20	17	15	10	mA	6	
Operating burst read: 1 bank active; BL = 4; CL = 3; $t_{CK} = t_{CK}(\text{MIN})$ ; Continuous READ bursts; $I_{\text{OUT}} = 0\text{mA}$ ; Address inputs are switching every 2 clock cycles; 50% data changing each burst	$I_{DD4R}$	110	105	100	90	mA	6	
Operating burst write: One bank active; BL = 4; $t_{CK} = t_{CK}(\text{MIN})$ ; Continuous WRITE bursts; Address inputs are switching; 50% data changing each burst	$I_{DD4W}$	110	105	100	90	mA	6	
Auto refresh: Burst refresh; CKE = HIGH; Address and control inputs are switching; Data bus inputs are stable	$t_{RFC} = 110\text{ns}$	$I_{DD5}$	95	92	90	85	mA	10
	$t_{RFC} = t_{REFI}$	$I_{DD5A}$	8	6	5	3	mA	10, 11
Deep power-down current: Address and control pins are stable; Data bus inputs are stable	$I_{DD8}$	10	10	10	10	$\mu\text{A}$	7, 13	

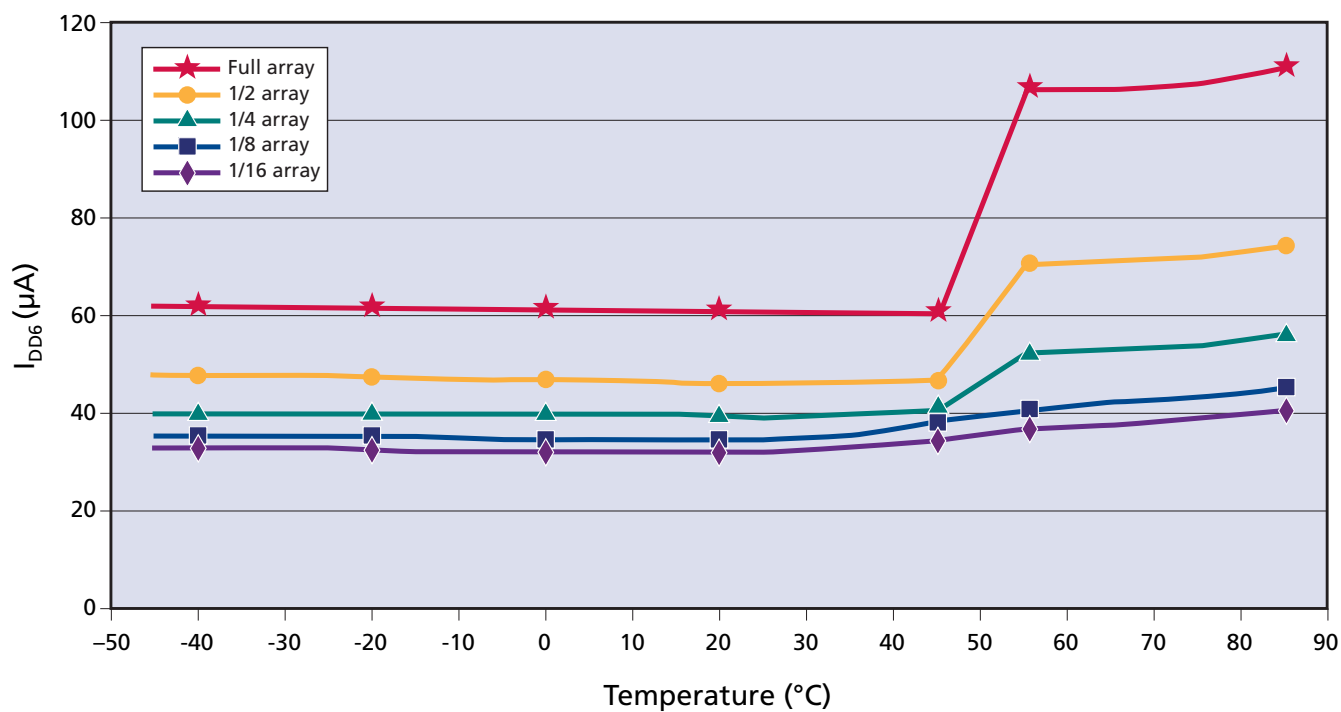
**Table 9: I<sub>DD6</sub> Specifications and Conditions**

Notes 1–5, 7, and 12 apply to all parameters/conditions in this table; V<sub>DD</sub>/V<sub>DDQ</sub> = 1.70–1.95V

Parameter/Condition		Symbol	I <sub>DD6</sub>	Units
Self refresh CKE = LOW; t <sub>CK</sub> = t <sub>CK</sub> (MIN); Address and control inputs are stable; Data bus inputs are stable	Full array, 85°C	I <sub>DD6A</sub>	200	μA
	Full array, 45°C	I <sub>DD6C</sub>	140	μA
	1/2 array, 85°C	I <sub>DD6A</sub>	160	μA
	1/2 array, 45°C	I <sub>DD6C</sub>	120	μA
	1/4 array, 85°C	I <sub>DD6A</sub>	140	μA
	1/4 array, 45°C	I <sub>DD6C</sub>	100	μA
	1/8 array, 85°C	I <sub>DD6A</sub>	120	μA
	1/8 array, 45°C	I <sub>DD6C</sub>	95	μA
	1/16 array, 85°C	I <sub>DD6A</sub>	100	μA
	1/16 array, 45°C	I <sub>DD6C</sub>	90	μA

- Notes:
1. All voltages referenced to V<sub>SS</sub>.
  2. Tests for I<sub>DD</sub> characteristics may be conducted at nominal supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.
  3. Timing and I<sub>DD</sub> tests may use a V<sub>IL</sub>-to-V<sub>IH</sub> swing of up to 1.5V in the test environment, but input timing is still referenced to V<sub>DDQ</sub>/2 (or to the crossing point for CK/CK#). The output timing reference voltage level is V<sub>DDQ</sub>/2.
  4. I<sub>DD</sub> is dependent on output loading and cycle rates. Specified values are obtained with minimum cycle time with the outputs open.
  5. I<sub>DD</sub> specifications are tested after the device is properly initialized and values are averaged at the defined cycle rate.
  6. MIN (t<sub>RC</sub> or t<sub>RFC</sub>) for I<sub>DD</sub> measurements is the smallest multiple of t<sub>CK</sub> that meets the minimum absolute value for the respective parameter. t<sub>RAS</sub> (MAX) for I<sub>DD</sub> measurements is the largest multiple of t<sub>CK</sub> that meets the maximum absolute value for t<sub>RAS</sub>.
  7. Measurement is taken 500ms after entering into this operating mode to allow settling time for the tester.
  8. V<sub>DD</sub> must not vary more than 4% if CKE is not active while any bank is active.
  9. I<sub>DD2N</sub> specifies DQ, DQS, and DM to be driven to a valid high or low logic level.
  10. CKE must be active (HIGH) during the entire time a REFRESH command is executed. From the time the AUTO REFRESH command is registered, CKE must be active at each rising clock edge until t<sub>RFC</sub> later.
  11. This limit is a nominal value and does not result in a fail. CKE is HIGH during REFRESH command period (t<sub>RFC</sub> [MIN]) else CKE is LOW (for example, during standby).
  12. Values for I<sub>DD6</sub> 85°C are guaranteed for the entire temperature range. All other I<sub>DD6</sub> values are estimated.
  13. Typical values at 25°C, not a maximum value.

**Figure 8: Typical Self Refresh Current vs. Temperature**





## Electrical Specifications – AC Operating Conditions

**Table 10: Electrical Characteristics and Recommended AC Operating Conditions**

Notes 1–9 apply to all the parameters in this table;  $V_{DD}/V_{DDQ} = 1.70\text{--}1.95V$

Parameter		Symbol	-5		-54		-6		-75		Unit	Notes
			Min	Max	Min	Max	Min	Max	Min	Max		
Access window of DQ from CK/CK#	CL = 3	$t_{AC}$	2.0	5.0	2.0	5.0	2.0	5.0	2.0	6.0	ns	
	CL = 2		2.0	6.5	2.0	6.5	2.0	6.5	2.0	6.5		
Clock cycle time	CL = 3	$t_{CK}$	5.0	–	5.4	–	6	–	7.5	–	ns	10
	CL = 2		12	–	12	–	12	–	12	–		
CK high-level width		$t_{CH}$	0.45	0.55	0.45	0.55	0.45	0.55	0.45	0.55	$t_{CK}$	
CK low-level width		$t_{CL}$	0.45	0.55	0.45	0.55	0.45	0.55	0.45	0.55	$t_{CK}$	
CKE minimum pulse width (high and low)		$t_{CKE}$	1	–	1	–	1	–	1	–	$t_{CK}$	11
Auto precharge write recovery + precharge time		$t_{DAL}$	–	–	–	–	–	–	–	–	–	12
DQ and DM input hold time relative to DQS (fast slew rate)		$t_{DH_f}$	0.48	–	0.54	–	0.6	–	0.8	–	ns	13, 14, 15
DQ and DM input hold time relative to DQS (slow slew rate)		$t_{DH_s}$	0.58	–	0.64	–	0.7	–	0.9	–	ns	
DQ and DM input setup time relative to DQS (fast slew rate)		$t_{DS_f}$	0.48	–	0.54	–	0.6	–	0.8	–	ns	13, 14, 15
DQ and DM input setup time relative to DQS (slow slew rate)		$t_{DS_s}$	0.58	–	0.64	–	0.7	–	0.9	–	ns	
DQ and DM input pulse width (for each input)		$t_{DIPW}$	1.8	–	1.9	–	2.1	–	1.8	–	ns	16
Access window of DQS from CK/CK#	CL = 3	$t_{DQSCK}$	2.0	5.0	2.0	5.0	2.0	5.0	2.0	6.0	ns	
	CL = 2		2.0	6.5	2.0	6.5	2.0	6.5	2.0	6.5	ns	
DQS input high pulse width		$t_{DQSH}$	0.4	0.6	0.4	0.6	0.4	0.6	0.4	0.6	$t_{CK}$	
DQS input low pulse width		$t_{DQSL}$	0.4	0.6	0.4	0.6	0.4	0.6	0.4	0.6	$t_{CK}$	
DQS–DQ skew, DQS to last DQ valid, per group, per access		$t_{DQSQ}$	–	0.4	–	0.45	–	0.5	–	0.6	ns	13, 17
WRITE command to first DQS latching transition		$t_{DQSS}$	0.75	1.25	0.75	1.25	0.75	1.25	0.75	1.25	$t_{CK}$	
DQS falling edge from CK rising – hold time		$t_{DSH}$	0.2	–	0.2	–	0.2	–	0.2	–	$t_{CK}$	
DQS falling edge to CK rising – setup time		$t_{DSS}$	0.2	–	0.2	–	0.2	–	0.2	–	$t_{CK}$	





# 128Mb: x16, x32 Mobile LPDDR SDRAM Electrical Specifications – AC Operating Conditions

**Table 10: Electrical Characteristics and Recommended AC Operating Conditions (Continued)**

Notes 1–9 apply to all the parameters in this table;  $V_{DD}/V_{DDQ} = 1.70\text{--}1.95V$

Parameter		Symbol	-5		-54		-6		-75		Unit	Notes
			Min	Max	Min	Max	Min	Max	Min	Max		
Data valid output window (DVW)		n/a	$t_{QH} - t_{DQSQ}$		$t_{QH} - t_{DQSQ}$		$t_{QH} - t_{DQSQ}$		$t_{QH} - t_{DQSQ}$		ns	17
Half-clock period		$t_{HP}$	$t_{CH}$ , $t_{CL}$	–	$t_{CH}$ , $t_{CL}$	–	$t_{CH}$ , $t_{CL}$	–	$t_{CH}$ , $t_{CL}$	–	ns	18
Data-out High-Z window from CK/CK#	CL = 3	$t_{HZ}$	–	5.0	–	5.0	–	5.0	–	6.0	ns	19, 20
	CL = 2		–	6.5	–	6.5	–	6.5	–	6.5	ns	
Data-out Low-Z window from CK/CK#		$t_{LZ}$	1.0	–	1.0	–	1.0	–	1.0	–	ns	19
Address and control input hold time (fast slew rate)		$t_{IH_F}$	0.9	–	1.0	–	1.1	–	1.3	–	ns	15, 21
Address and control input hold time (slow slew rate)		$t_{IH_S}$	1.1	–	1.2	–	1.3	–	1.5	–	ns	
Address and control input setup time (fast slew rate)		$t_{IS_F}$	0.9	–	1.0	–	1.1	–	1.3	–	ns	15, 21
Address and control input setup time (slow slew rate)		$t_{IS_S}$	1.1	–	1.2	–	1.3	–	1.5	–	ns	
Address and control input pulse width		$t_{IPW}$	2.3	–	2.5	–	$t_{IS} + t_{IH}$	–	$t_{IS} + t_{IH}$	–	ns	16
LOAD MODE REGISTER command cycle time		$t_{MRD}$	2	–	2	–	2	–	2	–	$t_{CK}$	
DQ–DQS hold, DQS to first DQ to go nonvalid, per access		$t_{QH}$	$t_{HP} - t_{QHS}$	–	$t_{HP} - t_{QHS}$	–	$t_{HP} - t_{QHS}$	–	$t_{HP} - t_{QHS}$	–	ns	13, 17
Data hold skew factor		$t_{QHS}$	–	0.5	–	0.5	–	0.65	–	0.75	ns	
ACTIVE-to-PRECHARGE command		$t_{RAS}$	40	70,000	42	70,000	42	70,000	45	70,000	ns	22
ACTIVE to ACTIVE/AUTO REFRESH command period		$t_{RC}$	55	–	58.2	–	60	–	67.5	–	ns	23
Active to read or write delay		$t_{RCD}$	15	–	16.2	–	18	–	22.5	–	ns	
Refresh period		$t_{REF}$	–	64	–	64	–	64	–	64	ms	
Average periodic refresh interval: 64Mb, 128Mb, and 256Mb (x32)		$t_{REFI}$	–	15.6	–	15.6	–	15.6	–	15.6	$\mu s$	
Average periodic refresh interval: 256Mb, 512Mb, 1Gb, 2Gb		$t_{REFI}$	–	7.8	–	7.8	–	7.8	–	7.8	$\mu s$	

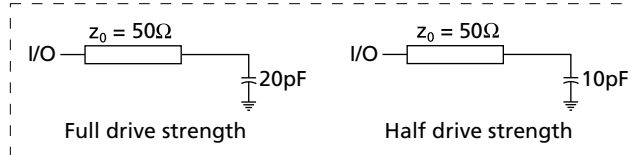
**Table 10: Electrical Characteristics and Recommended AC Operating Conditions (Continued)**

Notes 1–9 apply to all the parameters in this table;  $V_{DD}/V_{DDQ} = 1.70\text{--}1.95V$

Parameter	Symbol	-5		-54		-6		-75		Unit	Notes
		Min	Max	Min	Max	Min	Max	Min	Max		
AUTO REFRESH command period	$t_{RFC}$	80	–	80	–	80	–	80	–	ns	
PRECHARGE command period	$t_{RP}$	15	–	16.2	–	18	–	22.5	–	ns	
DQS read preamble	CL = 3 $t_{RPRE}$	0.9	1.1	0.9	1.1	0.9	1.1	0.9	1.1	$t_{CK}$	
	CL = 2 $t_{RPRE}$	0.5	1.1	0.5	1.1	0.5	1.1	0.5	1.1	$t_{CK}$	
DQS read postamble	$t_{RPST}$	0.4	0.6	0.4	0.6	0.4	0.6	0.4	0.6	$t_{CK}$	
Active bank <i>a</i> to active bank <i>b</i> command	$t_{RRD}$	10	–	10.8	–	12	–	15	–	ns	
Read of SRR to next valid command	$t_{SRC}$	CL + 1	–	CL + 1	–	CL + 1	–	CL + 1	–	$t_{CK}$	
SRR to read	$t_{SRR}$	2	–	2	–	2	–	2	–	$t_{CK}$	
DQS write preamble	$t_{WPRE}$	0.25	–	0.25	–	0.25	–	0.25	–	$t_{CK}$	
DQS write preamble setup time	$t_{WPRES}$	0	–	0	–	0	–	0	–	ns	24, 25
DQS write postamble	$t_{WPST}$	0.4	0.6	0.4	0.6	0.4	0.6	0.4	0.6	$t_{CK}$	26
Write recovery time	$t_{WR}$	15	–	15	–	15	–	15	–	ns	27
Internal WRITE-to-READ command delay	$t_{WTR}$	2	–	2	–	2	–	1	–	$t_{CK}$	
Exit power-down mode to first valid command	$t_{XP}$	1	–	1	–	1	–	1	–	$t_{CK}$	
Exit self refresh to first valid command	$t_{XSR}$	80	–	80	–	80	–	80	–	ns	28

- Notes:
1. All voltages referenced to  $V_{SS}$ .
  2. All parameters assume proper device initialization.
  3. Tests for AC timing and electrical AC and DC characteristics may be conducted at nominal supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage ranges specified.
  4. The circuit shown below represents the timing reference load used in defining the relevant timing parameters of the device. It is not intended to be either a precise representation of the typical system environment or a depiction of the actual load presented by a production tester. System designers will use IBIS or other simulation tools to correlate the timing reference load to system environment. Specifications are correlated to production test conditions (generally a coaxial transmission line terminated at the tester electronics). For the half-strength driver with a nominal 10pF load, parameters  $t_{AC}$  and  $t_{QH}$  are expected to be in the same range. However, these parameters are not subject to production test but are estimated by design/characterization. Use of IBIS or other simu-

Simulation tools for system design validation is suggested.



5. The CK/CK# input reference voltage level (for timing referenced to CK/CK#) is the point at which CK and CK# cross; the input reference voltage level for signals other than CK/CK# is  $V_{DDQ/2}$ .
6. A CK and CK# input slew rate  $\geq 1$  V/ns (2 V/ns if measured differentially) is assumed for all parameters.
7. All AC timings assume an input slew rate of 1 V/ns.
8. CAS latency definition: with CL = 2, the first data element is valid at ( $t_{CK} + t_{AC}$ ) after the clock at which the READ command was registered; for CL = 3, the first data element is valid at ( $2 \times t_{CK} + t_{AC}$ ) after the first clock at which the READ command was registered.
9. Timing tests may use a  $V_{IL}$ -to- $V_{IH}$  swing of up to 1.5V in the test environment, but input timing is still referenced to  $V_{DDQ/2}$  or to the crossing point for CK/CK#. The output timing reference voltage level is  $V_{DDQ/2}$ .
10. Clock frequency change is only permitted during clock stop, power-down, or self refresh mode.
11. In cases where the device is in self refresh mode for  $t_{CKE}$ ,  $t_{CKE}$  starts at the rising edge of the clock and ends when CKE transitions HIGH.
12.  $t_{DAL} = (t_{WR}/t_{CK}) + (t_{RP}/t_{CK})$ : for each term, if not already an integer, round up to the next highest integer.
13. Referenced to each output group: for x16, LDQS with DQ[7:0]; and UDQS with DQ[15:8]. For x32, DQS0 with DQ[7:0]; DQS1 with DQ[15:8]; DQS2 with DQ[23:16]; and DQS3 with DQ[31:24].
14. DQ and DM input slew rates must not deviate from DQS by more than 10%. If the DQ/DM/DQS slew rate is less than 1.0 V/ns, timing must be derated: 50ps must be added to  $t_{DS}$  and  $t_{DH}$  for each 100 mV/ns reduction in slew rate. If the slew rate exceeds 4 V/ns, functionality is uncertain.
15. The transition time for input signals (CAS#, CKE, CS#, DM, DQ, DQS, RAS#, WE#, and addresses) are measured between  $V_{IL(DC)}$  to  $V_{IH(AC)}$  for rising input signals and  $V_{IH(DC)}$  to  $V_{IL(AC)}$  for falling input signals.
16. These parameters guarantee device timing but are not tested on each device.
17. The valid data window is derived by achieving other specifications:  $t_{HP}$  ( $t_{CK}/2$ ),  $t_{DQSQ}$ , and  $t_{QH}$  ( $t_{HP} - t_{QHS}$ ). The data valid window derates directly proportional with the clock duty cycle and a practical data valid window can be derived. The clock is provided a maximum duty cycle variation of 45/55. Functionality is uncertain when operating beyond a 45/55 ratio.
18.  $t_{HP}$  (MIN) is the lesser of  $t_{CL}$  (MIN) and  $t_{CH}$  (MIN) actually applied to the device CK and CK# inputs, collectively.
19.  $t_{HZ}$  and  $t_{LZ}$  transitions occur in the same access time windows as valid data transitions. These parameters are not referenced to a specific voltage level, but specify when the device output is no longer driving ( $t_{HZ}$ ) or begins driving ( $t_{LZ}$ ).
20.  $t_{HZ}$  (MAX) will prevail over  $t_{DQSCK}$  (MAX) +  $t_{RPST}$  (MAX) condition.
21. Fast command/address input slew rate  $\geq 1$  V/ns. Slow command/address input slew rate  $\geq 0.5$  V/ns. If the slew rate is less than 0.5 V/ns, timing must be derated:  $t_{IS}$  has an additional 50ps per each 100 mV/ns reduction in slew rate from the 0.5 V/ns.  $t_{IH}$  has 0ps added, therefore, it remains constant. If the slew rate exceeds 4.5 V/ns, functionality is uncertain.
22. READs and WRITEs with auto precharge must not be issued until  $t_{RAS}$  (MIN) can be satisfied prior to the internal PRECHARGE command being issued.

23. DRAM devices should be evenly addressed when being accessed. Disproportionate accesses to a particular row address may result in reduction of the product lifetime.
24. This is not a device limit. The device will operate with a negative value, but system performance could be degraded due to bus turnaround.
25. It is recommended that DQS be valid (HIGH or LOW) on or before the WRITE command. The case shown (DQS going from High-Z to logic low) applies when no WRITES were previously in progress on the bus. If a previous WRITE was in progress, DQS could be HIGH during this time, depending on  $t_{DQSS}$ .
26. The maximum limit for this parameter is not a device limit. The device will operate with a greater value for this parameter, but system performance (bus turnaround) will degrade accordingly.
27. At least 1 clock cycle is required during  $t_{WR}$  time when in auto precharge mode.
28. Clock must be toggled a minimum of two times during the  $t_{XSR}$  period.

## Output Drive Characteristics

**Table 11: Target Output Drive Characteristics (Full Strength)**

Notes 1–2 apply to all values; characteristics are specified under best and worst process variations/conditions

Voltage (V)	Pull-Down Current (mA)		Pull-Up Current (mA)	
	Min	Max	Min	Max
0.00	0.00	0.00	0.00	0.00
0.10	2.80	18.53	–2.80	–18.53
0.20	5.60	26.80	–5.60	–26.80
0.30	8.40	32.80	–8.40	–32.80
0.40	11.20	37.05	–11.20	–37.05
0.50	14.00	40.00	–14.00	–40.00
0.60	16.80	42.50	–16.80	–42.50
0.70	19.60	44.57	–19.60	–44.57
0.80	22.40	46.50	–22.40	–46.50
0.85	23.80	47.48	–23.80	–47.48
0.90	23.80	48.50	–23.80	–48.50
0.95	23.80	49.40	–23.80	–49.40
1.00	23.80	50.05	–23.80	–50.05
1.10	23.80	51.35	–23.80	–51.35
1.20	23.80	52.65	–23.80	–52.65
1.30	23.80	53.95	–23.80	–53.95
1.40	23.80	55.25	–23.80	–55.25
1.50	23.80	56.55	–23.80	–56.55
1.60	23.80	57.85	–23.80	–57.85
1.70	23.80	59.15	–23.80	–59.15
1.80	23.80	60.45	–23.80	–60.45
1.90	23.80	61.75	–23.80	–61.75

- Notes:
1. Based on nominal impedance of 25Ω (full strength) at  $V_{DDQ}/2$ .
  2. The full variation in driver current from minimum to maximum, due to process, voltage, and temperature, will lie within the outer bounding lines of the I-V curves.

**Table 12: Target Output Drive Characteristics (Three-Quarter Strength)**

Notes 1–2 apply to all values; characteristics are specified under best and worst process variations/conditions

Voltage (V)	Pull-Down Current (mA)		Pull-Up Current (mA)	
	Min	Max	Min	Max
0.00	0.00	0.00	0.00	0.00
0.10	1.96	12.97	–1.96	–12.97
0.20	3.92	18.76	–3.92	–18.76
0.30	5.88	22.96	–5.88	–22.96
0.40	7.84	25.94	–7.84	–25.94
0.50	9.80	28.00	–9.80	–28.00
0.60	11.76	29.75	–11.76	–29.75
0.70	13.72	31.20	–13.72	–31.20
0.80	15.68	32.55	–15.68	–32.55
0.85	16.66	33.24	–16.66	–33.24
0.90	16.66	33.95	–16.66	–33.95
0.95	16.66	34.58	–16.66	–34.58
1.00	16.66	35.04	–16.66	–35.04
1.10	16.66	35.95	–16.66	–35.95
1.20	16.66	36.86	–16.66	–36.86
1.30	16.66	37.77	–16.66	–37.77
1.40	16.66	38.68	–16.66	–38.68
1.50	16.66	39.59	–16.66	–39.59
1.60	16.66	40.50	–16.66	–40.50
1.70	16.66	41.41	–16.66	–41.41
1.80	16.66	42.32	–16.66	–42.32
1.90	16.66	43.23	–16.66	–43.23

- Notes:
1. Based on nominal impedance of  $37\Omega$  (three-quarter drive strength) at  $V_{DDQ}/2$ .
  2. The full variation in driver current from minimum to maximum, due to process, voltage, and temperature, will lie within the outer bounding lines of the I-V curves.

**Table 13: Target Output Drive Characteristics (One-Half Strength)**

Notes 1–3 apply to all values; characteristics are specified under best and worst process variations/conditions

Voltage (V)	Pull-Down Current (mA)		Pull-Up Current (mA)	
	Min	Max	Min	Max
0.00	0.00	0.00	0.00	0.00
0.10	1.27	8.42	–1.27	–8.42
0.20	2.55	12.30	–2.55	–12.30
0.30	3.82	14.95	–3.82	–14.95
0.40	5.09	16.84	–5.09	–16.84
0.50	6.36	18.20	–6.36	–18.20
0.60	7.64	19.30	–7.64	–19.30
0.70	8.91	20.30	–8.91	–20.30
0.80	10.16	21.20	–10.16	–21.20
0.85	10.80	21.60	–10.80	–21.60
0.90	10.80	22.00	–10.80	–22.00
0.95	10.80	22.45	–10.80	–22.45
1.00	10.80	22.73	–10.80	–22.73
1.10	10.80	23.21	–10.80	–23.21
1.20	10.80	23.67	–10.80	–23.67
1.30	10.80	24.14	–10.80	–24.14
1.40	10.80	24.61	–10.80	–24.61
1.50	10.80	25.08	–10.80	–25.08
1.60	10.80	25.54	–10.80	–25.54
1.70	10.80	26.01	–10.80	–26.01
1.80	10.80	26.48	–10.80	–26.48
1.90	10.80	26.95	–10.80	–26.95

- Notes:
1. Based on nominal impedance of 55Ω (one-half drive strength) at  $V_{DDQ}/2$ .
  2. The full variation in driver current from minimum to maximum, due to process, voltage, and temperature, will lie within the outer bounding lines of the I-V curves.
  3. The I-V curve for one-quarter drive strength is approximately 50% of one-half drive strength.

## Functional Description

The Mobile LPDDR SDRAM uses a double data rate architecture to achieve high-speed operation. The double data rate architecture is essentially a  $2n$ -prefetch architecture, with an interface designed to transfer two data words per clock cycle at the I/O. Single read or write access for the device consists of a single  $2n$ -bit-wide, one-clock-cycle data transfer at the internal DRAM core and two corresponding  $n$ -bit-wide, one-half-clock-cycle data transfers at the I/O.

A bidirectional data strobe (DQS) is transmitted externally, along with data, for use in data capture at the receiver. DQS is a strobe transmitted by the device during READs and by the memory controller during WRITEs. DQS is edge-aligned with data for READs and center-aligned with data for WRITEs. The x16 device has two data strobes, one for the lower byte and one for the upper byte; the x32 device has four data strobes, one per byte.

The LPDDR device operates from a differential clock (CK and CK#); the crossing of CK going HIGH and CK# going LOW will be referred to as the positive edge of CK. Commands (address and control signals) are registered at every positive edge of CK. Input data is registered on both edges of DQS, and output data is referenced to both edges of DQS, as well as to both edges of CK.

Read and write accesses to the device are burst-oriented; accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an ACTIVE command, followed by a READ or WRITE command. The address bits registered coincident with the ACTIVE command are used to select the bank and row to be accessed. The address bits registered coincident with the READ or WRITE command are used to select the starting column location for the burst access.

The device provides for programmable READ or WRITE burst lengths of 2, 4, 8, or 16. An auto precharge function can be enabled to provide a self-timed row precharge that is initiated at the end of the burst access.

As with standard DDR SDRAM, the pipelined, multibank architecture of LPDDR supports concurrent operation, thereby providing high effective bandwidth by hiding row precharge and activation time.

An auto refresh mode is provided, along with a power-saving power-down mode. Deep power-down mode is offered to achieve maximum power reduction by eliminating the power of the memory array. Data will not be retained after the device enters deep power-down mode.

Two self refresh features, temperature-compensated self refresh (TCSR) and partial-array self refresh (PASR), offer additional power savings. TCSR is controlled by the automatic on-chip temperature sensor. PASR can be customized using the extended mode register settings. The two features can be combined to achieve even greater power savings.

The DLL that is typically used on standard DDR devices is not necessary on LPDDR devices. It has been omitted to save power.



## Commands

A quick reference for available commands is provided in Table 14 and Table 15 (page 34), followed by a written description of each command. Three additional truth tables (Table 16 (page 40), Table 17 (page 42), and Table 18 (page 44)) provide CKE commands and current/next state information.

**Table 14: Truth Table – Commands**

CKE is HIGH for all commands shown except SELF REFRESH and DEEP POWER-DOWN; all states and sequences not shown are reserved and/or illegal

Name (Function)	CS#	RAS#	CAS#	WE#	Address	Notes
DESELECT (NOP)	H	X	X	X	X	1
NO OPERATION (NOP)	L	H	H	H	X	1
ACTIVE (select bank and activate row)	L	L	H	H	Bank/row	2
READ (select bank and column, and start READ burst)	L	H	L	H	Bank/column	3
WRITE (select bank and column, and start WRITE burst)	L	H	L	L	Bank/column	3
BURST TERMINATE or DEEP POWER-DOWN (enter deep power-down mode)	L	H	H	L	X	4, 5
PRECHARGE (deactivate row in bank or banks)	L	L	H	L	Code	6
AUTO REFRESH (refresh all or single bank) or SELF REFRESH (enter self refresh mode)	L	L	L	H	X	7, 8
LOAD MODE REGISTER	L	L	L	L	Op-code	9

- Notes:
1. Deselect and NOP are functionally interchangeable.
  2. BA0–BA1 provide bank address and A[0:I] provide row address (where I = the most significant address bit for each configuration).
  3. BA0–BA1 provide bank address; A[0:I] provide column address (where I = the most significant address bit for each configuration); A10 HIGH enables the auto precharge feature (nonpersistent); A10 LOW disables the auto precharge feature.
  4. Applies only to READ bursts with auto precharge disabled; this command is undefined and should not be used for READ bursts with auto precharge enabled and for WRITE bursts.
  5. This command is a BURST TERMINATE if CKE is HIGH and DEEP POWER-DOWN if CKE is LOW.
  6. A10 LOW: BA0–BA1 determine which bank is precharged.  
A10 HIGH: all banks are precharged and BA0–BA1 are “Don’t Care.”
  7. This command is AUTO REFRESH if CKE is HIGH, SELF REFRESH if CKE is LOW.
  8. Internal refresh counter controls row addressing; in self refresh mode all inputs and I/Os are “Don’t Care” except for CKE.
  9. BA0–BA1 select the standard mode register, extended mode register, or status register.

**Table 15: DM Operation Truth Table**

Name (Function)	DM	DQ	Notes
Write enable	L	Valid	1, 2
Write inhibit	H	X	1, 2

Notes: 1. Used to mask write data; provided coincident with the corresponding data.  
2. All states and sequences not shown are reserved and/or illegal.

## DESELECT

The Deselect function (CS# HIGH) prevents new commands from being executed by the device. Operations already in progress are not affected.

## NO OPERATION

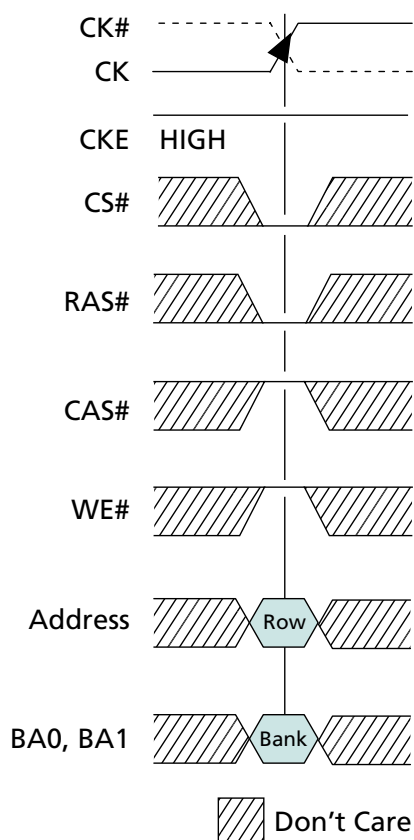
The NO OPERATION (NOP) command is used to instruct the selected device to perform a NOP. This prevents unwanted commands from being registered during idle or wait states. Operations already in progress are not affected.

## LOAD MODE REGISTER

The mode registers are loaded via inputs A[0:n]. See mode register descriptions in Standard Mode Register and Extended Mode Register. The LOAD MODE REGISTER command can only be issued when all banks are idle, and a subsequent executable command cannot be issued until  $t_{MRD}$  is met.

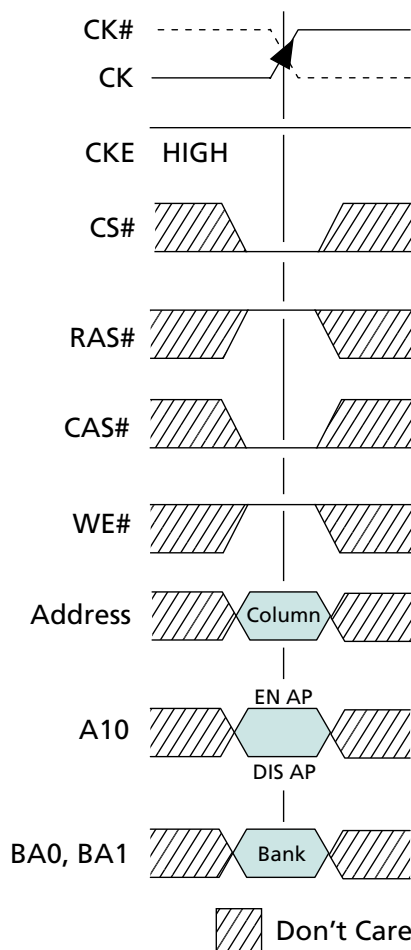
## ACTIVE

The ACTIVE command is used to activate a row in a particular bank for a subsequent access. The values on the BA0 and BA1 inputs select the bank, and the address provided on inputs A[0:n] selects the row. This row remains active for accesses until a PRECHARGE command is issued to that bank. A PRECHARGE command must be issued before opening a different row in the same bank.

**Figure 9: ACTIVE Command**


## READ

The READ command is used to initiate a burst read access to an active row. The values on the BA0 and BA1 inputs select the bank; the address provided on inputs A[*I*:0] (where *I* = the most significant column address bit for each configuration) selects the starting column location. The value on input A10 determines whether auto precharge is used. If auto precharge is selected, the row being accessed will be precharged at the end of the READ burst; if auto precharge is not selected, the row will remain open for subsequent accesses.

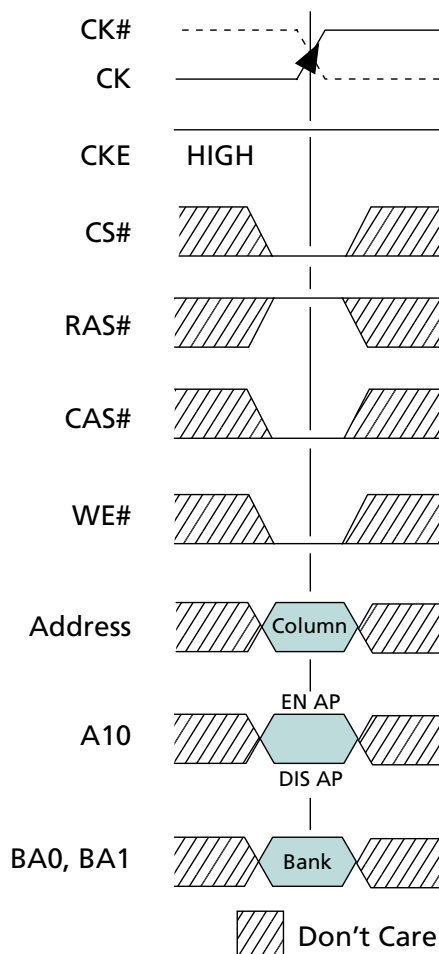
**Figure 10: READ Command**


Note: 1. EN AP = enable auto precharge; DIS AP = disable auto precharge.

## WRITE

The WRITE command is used to initiate a burst write access to an active row. The values on the BA0 and BA1 inputs select the bank; the address provided on inputs A[I:0] (where I = the most significant column address bit for each configuration) selects the starting column location. The value on input A10 determines whether auto precharge is used. If auto precharge is selected, the row being accessed will be precharged at the end of the WRITE burst; if auto precharge is not selected, the row will remain open for subsequent accesses. Input data appearing on the DQ is written to the memory array, subject to the DM input logic level appearing coincident with the data. If a given DM signal is registered LOW, the corresponding data will be written to memory; if the DM signal is registered HIGH, the corresponding data inputs will be ignored, and a WRITE will not be executed to that byte/column location.

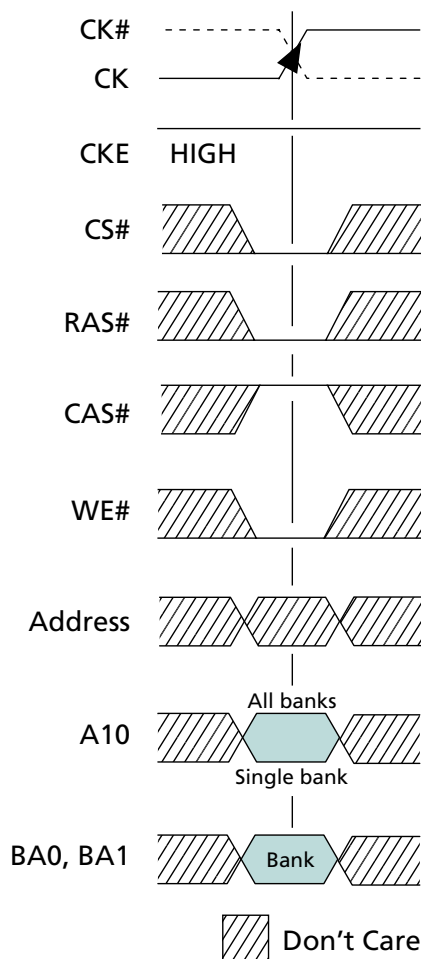
If a WRITE or a READ is in progress, the entire data burst must be complete prior to stopping the clock (see Clock Change Frequency (page 93)). A burst completion for WRITES is defined when the write postamble and <sup>t</sup>WR or <sup>t</sup>WTR are satisfied.

**Figure 11: WRITE Command**


Note: 1. EN AP = enable auto precharge; DIS AP = disable auto precharge.

## PRECHARGE

The PRECHARGE command is used to deactivate the open row in a particular bank or the open row in all banks. The bank(s) will be available for a subsequent row access a specified time (TRP) after the PRECHARGE command is issued. Input A10 determines whether one or all banks will be precharged, and in the case where only one bank is precharged, inputs BA0 and BA1 select the bank. Otherwise, BA0 and BA1 are treated as "Don't Care." After a bank has been precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued to that bank.

**Figure 12: PRECHARGE Command**


Note: 1. If A10 is HIGH, bank address becomes "Don't Care."

## BURST TERMINATE

The BURST TERMINATE command is used to truncate READ bursts with auto pre-charge disabled. The most recently registered READ command prior to the BURST TERMINATE command will be truncated, as described in READ Operation. The open page from which the READ was terminated remains open.

## AUTO REFRESH

AUTO REFRESH is used during normal operation of the device and is analogous to CAS#-BEFORE-RAS# (CBR) REFRESH in FPM/EDO DRAM. The AUTO REFRESH command is nonpersistent and must be issued each time a refresh is required.

Addressing is generated by the internal refresh controller. This makes the address bits a "Don't Care" during an AUTO REFRESH command.

For improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided. The auto refresh period begins when the AUTO REFRESH command is registered and ends <sup>t</sup>RFC later.

## SELF REFRESH

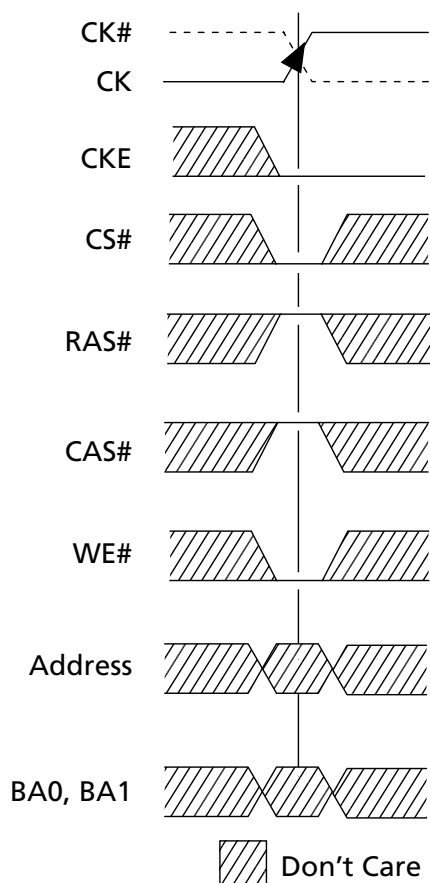
The SELF REFRESH command is used to place the device in self refresh mode; self refresh mode is used to retain data in the memory device while the rest of the system is powered down. When in self refresh mode, the device retains data without external clocking. The SELF REFRESH command is initiated like an AUTO REFRESH command, except that CKE is disabled (LOW). After the SELF REFRESH command is registered, all inputs to the device become "Don't Care" with the exception of CKE, which must remain LOW.

Micron recommends that, prior to self refresh entry and immediately upon self refresh exit, the user perform a burst auto refresh cycle for the number of refresh rows. Alternatively, if a distributed refresh pattern is used, this pattern should be immediately resumed upon self refresh exit.

## DEEP POWER-DOWN

The DEEP POWER-DOWN (DPD) command is used to enter DPD mode, which achieves maximum power reduction by eliminating the power to the memory array. Data will not be retained when the device enters DPD mode. The DPD command is the same as a BURST TERMINATE command with CKE LOW.

**Figure 13: DEEP POWER-DOWN Command**



## Truth Tables

**Table 16: Truth Table – Current State Bank  $n$  – Command to Bank  $n$** 

Notes 1–6 apply to all parameters in this table

Current State	CS#	RAS#	CAS#	WE#	Command/Action	Notes
Any	H	X	X	X	DESELECT (NOP/continue previous operation)	
	L	H	H	H	NO OPERATION (NOP/continue previous operation)	
Idle	L	L	H	H	ACTIVE (select and activate row)	
	L	L	L	H	AUTO REFRESH	7
	L	L	L	L	LOAD MODE REGISTER	7
Row active	L	H	L	H	READ (select column and start READ burst)	10
	L	H	L	L	WRITE (select column and start WRITE burst)	10
	L	L	H	L	PRECHARGE (deactivate row in bank or banks)	8
Read (auto pre-charge disabled)	L	H	L	H	READ (select column and start new READ burst)	10
	L	H	L	L	WRITE (select column and start WRITE burst)	10, 12
	L	L	H	L	PRECHARGE (truncate READ burst, start PRECHARGE)	8
	L	H	H	L	BURST TERMINATE	9
Write (auto pre-charge disabled)	L	H	L	H	READ (select column and start READ burst)	10, 11
	L	H	L	L	WRITE (select column and start new WRITE burst)	10
	L	L	H	L	PRECHARGE (truncate WRITE burst, start PRECHARGE)	8, 11

- Notes:
1. This table applies when  $\text{CKE}_{n-1}$  was HIGH,  $\text{CKE}_n$  is HIGH and after  $t_{\text{XSR}}$  has been met (if the previous state was self refresh), after  $t_{\text{XP}}$  has been met (if the previous state was power-down), or after a full initialization (if the previous state was deep power-down).
  2. This table is bank-specific, except where noted (for example, the current state is for a specific bank and the commands shown are supported for that bank when in that state). Exceptions are covered in the notes below.
  3. Current state definitions:
 

Idle: The bank has been precharged, and  $t_{\text{RP}}$  has been met.

Row active: A row in the bank has been activated, and  $t_{\text{RCD}}$  has been met. No data bursts/accesses and no register accesses are in progress.

Read: A READ burst has been initiated with auto precharge disabled and has not yet terminated or been terminated.

Write: A WRITE burst has been initiated with auto precharge disabled and has not yet terminated or been terminated.
  4. The states listed below must not be interrupted by a command issued to the same bank. COMMAND INHIBIT or NOP commands, or supported commands to the other bank, must be issued on any clock edge occurring during these states. Supported commands to any other bank are determined by that bank's current state.
 

Precharging: Starts with registration of a PRECHARGE command and ends when  $t_{\text{RP}}$  is met. After  $t_{\text{RP}}$  is met, the bank will be in the idle state.

Row activating: Starts with registration of an ACTIVE command and ends when  $t_{\text{RCD}}$  is met. After  $t_{\text{RCD}}$  is met, the bank will be in the row active state.



Read with auto-precharge enabled: Starts with registration of a READ command with auto precharge enabled and ends when  $t_{RP}$  has been met. After  $t_{RP}$  is met, the bank will be in the idle state.

Write with auto-precharge enabled: Starts with registration of a WRITE command with auto precharge enabled and ends when  $t_{RP}$  has been met. After  $t_{RP}$  is met, the bank will be in the idle state.

5. The states listed below must not be interrupted by any executable command; DESELECT or NOP commands must be applied on each positive clock edge during these states.

Refreshing: Starts with registration of an AUTO REFRESH command and ends when  $t_{RFC}$  is met. After  $t_{RFC}$  is met, the device will be in the all banks idle state.

Accessing mode register: Starts with registration of a LOAD MODE REGISTER command and ends when  $t_{MRD}$  has been met. After  $t_{MRD}$  is met, the device will be in the all banks idle state.

Precharging all: Starts with registration of a PRECHARGE ALL command and ends when  $t_{RP}$  is met. After  $t_{RP}$  is met, all banks will be in the idle state.

6. All states and sequences not shown are illegal or reserved.
7. Not bank-specific; requires that all banks are idle, and bursts are not in progress.
8. May or may not be bank-specific; if multiple banks need to be precharged, each must be in a valid state for precharging.
9. Not bank-specific; BURST TERMINATE affects the most recent READ burst, regardless of bank.
10. READs or WRITEs listed in the Command/Action column include READs or WRITEs with auto precharge enabled and READs or WRITEs with auto precharge disabled.
11. Requires appropriate DM masking.
12. A WRITE command can be applied after the completion of the READ burst; otherwise, a BURST TERMINATE must be used to end the READ burst prior to asserting a WRITE command.

**Table 17: Truth Table – Current State Bank *n* – Command to Bank *m***

Notes 1–6 apply to all parameters in this table

Current State	CS#	RAS#	CAS#	WE#	Command/Action	Notes
Any	H	X	X	X	DESELECT (NOP/continue previous operation)	
	L	H	H	H	NO OPERATION (NOP/continue previous operation)	
Idle	X	X	X	X	Any command supported to bank <i>m</i>	
Row activating, active, or pre-charging	L	L	H	H	ACTIVE (select and activate row)	
	L	H	L	H	READ (select column and start READ burst)	
	L	H	L	L	WRITE (select column and start WRITE burst)	
	L	L	H	L	PRECHARGE	
Read (auto pre-charge disabled)	L	L	H	H	ACTIVE (select and activate row)	
	L	H	L	H	READ (select column and start new READ burst)	
	L	H	L	L	WRITE (select column and start WRITE burst)	7
	L	L	H	L	PRECHARGE	
Write (auto pre-charge disabled)	L	L	H	H	ACTIVE (select and activate row)	
	L	H	L	H	READ (select column and start READ burst)	
	L	H	L	L	WRITE (select column and start new WRITE burst)	
	L	L	H	L	PRECHARGE	
Read (with auto precharge)	L	L	H	H	ACTIVE (select and activate row)	
	L	H	L	H	READ (select column and start new READ burst)	
	L	H	L	L	WRITE (select column and start WRITE burst)	7
	L	L	H	L	PRECHARGE	
Write (with auto precharge)	L	L	H	H	ACTIVE (select and activate row)	
	L	H	L	H	READ (select column and start READ burst)	
	L	H	L	L	WRITE (select column and start new WRITE burst)	
	L	L	H	L	PRECHARGE	

- Notes:
1. This table applies when CKE<sub>*n* - 1</sub> was HIGH, CKE<sub>*n*</sub> is HIGH and after t<sub>XS</sub>R has been met (if the previous state was self refresh), after t<sub>XP</sub> has been met (if the previous state was power-down) or after a full initialization (if the previous state was deep power-down).
  2. This table describes alternate bank operation, except where noted (for example, the current state is for bank *n* and the commands shown are those supported for issue to bank *m*, assuming that bank *m* is in such a state that the given command is supported). Exceptions are covered in the notes below.
  3. Current state definitions:

Idle: The bank has been precharged, and t<sub>RP</sub> has been met.

Row active: A row in the bank has been activated, and t<sub>RCD</sub> has been met. No data bursts/accesses and no register accesses are in progress.

Read: A READ burst has been initiated and has not yet terminated or been terminated.

Write: A WRITE burst has been initiated and has not yet terminated or been terminated.

3a. Both the read with auto precharge enabled state or the write with auto precharge enabled state can be broken into two parts: the access period and the precharge period. For read with auto precharge, the precharge period is defined as if the same burst was

executed with auto precharge disabled and then followed with the earliest possible PRECHARGE command that still accesses all of the data in the burst. For write with auto precharge, the precharge period begins when  $t_{WR}$  ends, with  $t_{WR}$  measured as if auto precharge was disabled. The access period starts with registration of the command and ends when the precharge period (or  $t_{RP}$ ) begins. This device supports concurrent auto precharge such that when a read with auto precharge is enabled or a write with auto precharge is enabled, any command to other banks is supported, as long as that command does not interrupt the read or write data transfer already in process. In either case, all other related limitations apply (i.e., contention between read data and write data must be avoided).

3b. The minimum delay from a READ or WRITE command (with auto precharge enabled) to a command to a different bank is summarized below.

From Command	To Command	Minimum Delay (with Concurrent Auto Precharge)
WRITE with Auto Precharge	READ or READ with auto precharge WRITE or WRITE with auto pre-charge PRECHARGE ACTIVE	$[1 + (BL/2)] t_{CK} + t_{WTR}$ $(BL/2) t_{CK}$ $1 t_{CK}$ $1 t_{CK}$
READ with Auto Precharge	READ or READ with auto precharge WRITE or WRITE with auto pre-charge PRECHARGE ACTIVE	$(BL/2) \times t_{CK}$ $[CL + (BL/2)] t_{CK}$ $1 t_{CK}$ $1 t_{CK}$

4. AUTO REFRESH and LOAD MODE REGISTER commands can only be issued when all banks are idle.
5. All states and sequences not shown are illegal or reserved.
6. Requires appropriate DM masking.
7. A WRITE command can be applied after the completion of the READ burst; otherwise, a BURST TERMINATE must be used to end the READ burst prior to asserting a WRITE command.

**Table 18: Truth Table – CKE**

Notes 1–4 apply to all parameters in this table

Current State	CKE <sub>n-1</sub>	CKE <sub>n</sub>	COMMAND <sub>n</sub>	ACTION <sub>n</sub>	Notes
Active power-down	L	L	X	Maintain active power-down	
Deep power-down	L	L	X	Maintain deep power-down	
Precharge power-down	L	L	X	Maintain precharge power-down	
Self refresh	L	L	X	Maintain self refresh	
Active power-down	L	H	DESELECT or NOP	Exit active power-down	5
Deep power-down	L	H	DESELECT or NOP	Exit deep power-down	6
Precharge power-down	L	H	DESELECT or NOP	Exit precharge power-down	
Self refresh	L	H	DESELECT or NOP	Exit self refresh	5, 7
Bank(s) active	H	L	DESELECT or NOP	Active power-down entry	
All banks idle	H	L	BURST TERMINATE	Deep power-down entry	
All banks idle	H	L	DESELECT or NOP	Precharge power-down entry	
All banks idle	H	L	AUTO REFRESH	Self refresh entry	
	H	H	See Table 17 (page 42)		
	H	H	See Table 17 (page 42)		

- Notes:
1. CKE<sub>n</sub> is the logic state of CKE at clock edge *n*; CKE<sub>n-1</sub> was the state of CKE at the previous clock edge.
  2. Current state is the state of the DDR SDRAM immediately prior to clock edge *n*.
  3. COMMAND<sub>n</sub> is the command registered at clock edge *n*, and ACTION<sub>n</sub> is a result of COMMAND<sub>n</sub>.
  4. All states and sequences not shown are illegal or reserved.
  5. DESELECT or NOP commands should be issued on each clock edge occurring during the <sup>t</sup>XP or <sup>t</sup>XSR period.
  6. After exiting deep power-down mode, a full DRAM initialization sequence is required.
  7. The clock must toggle at least two times during the <sup>t</sup>XSR period.



## Initialization

Prior to normal operation, the device must be powered up and initialized in a predefined manner. Using initialization procedures other than those specified will result in undefined operation.

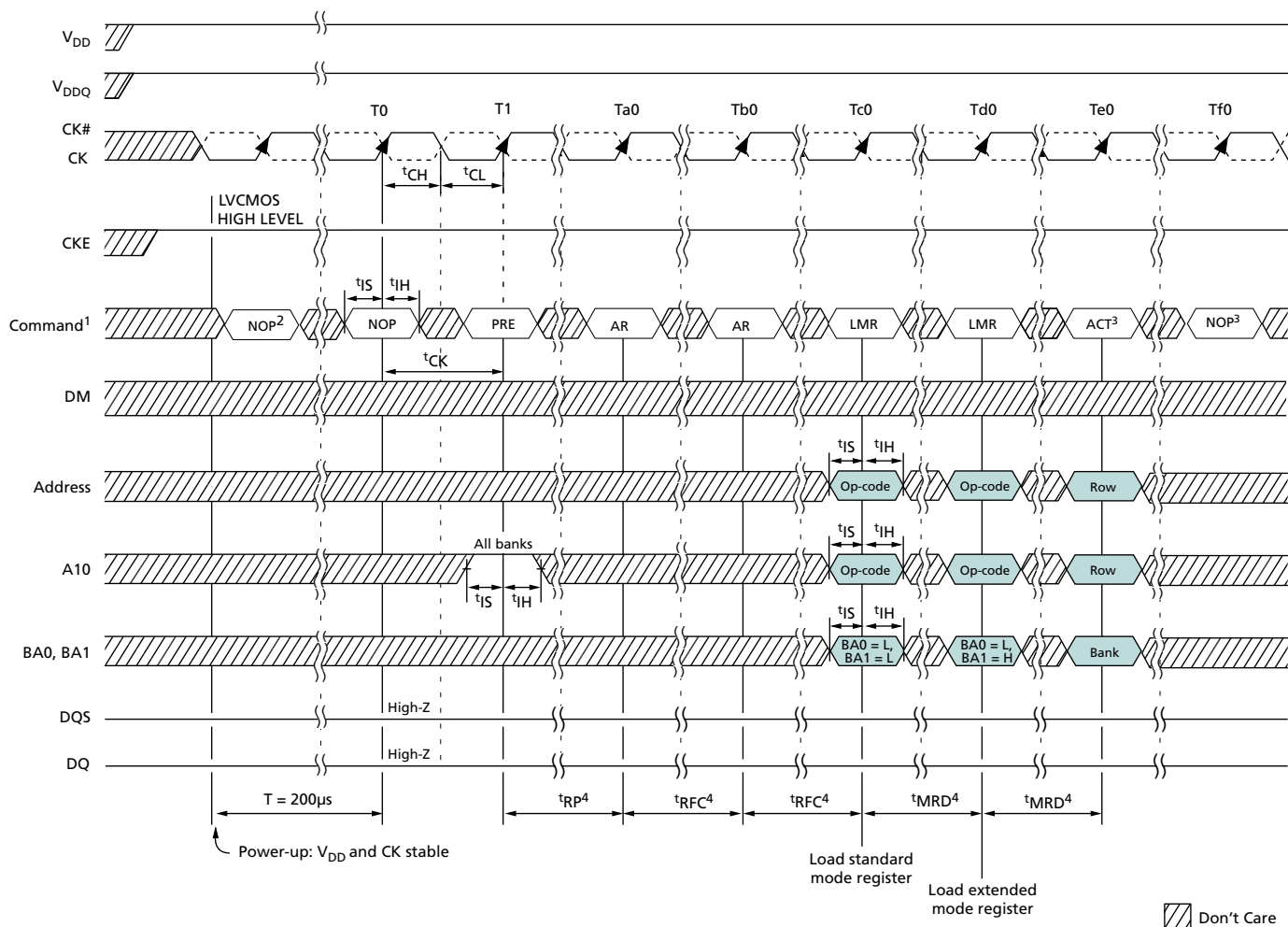
If there is an interruption to the device power, the device must be re-initialized using the initialization sequence described below to ensure proper functionality of the device.

To properly initialize the device, this sequence must be followed:

1. The core power ( $V_{DD}$ ) and I/O power ( $V_{DDQ}$ ) must be brought up simultaneously. It is recommended that  $V_{DD}$  and  $V_{DDQ}$  be from the same power source, or  $V_{DDQ}$  must never exceed  $V_{DD}$ . Standard initialization requires that CKE be asserted HIGH (see Figure 15 (page 47)). Alternatively, initialization can be completed with CKE LOW provided that CKE transitions HIGH  $t_{IS}$  prior to  $T_0$  (see Figure 16 (page 48)).
2. When power supply voltages are stable and the CKE has been driven HIGH, it is safe to apply the clock.
3. When the clock is stable, a 200 $\mu$ s minimum delay is required by the Mobile LPDDR prior to applying an executable command. During this time, NOP or DESELECT commands must be issued on the command bus.
4. Issue a PRECHARGE ALL command.
5. Issue NOP or DESELECT commands for at least  $t_{RP}$  time.
6. Issue an AUTO REFRESH command followed by NOP or DESELECT commands for at least  $t_{RFC}$  time. Issue a second AUTO REFRESH command followed by NOP or DESELECT commands for at least  $t_{RFC}$  time. Two AUTO REFRESH commands must be issued. Typically, both of these commands are issued at this stage as described above.
7. Using the LOAD MODE REGISTER command, load the standard mode register as desired.
8. Issue NOP or DESELECT commands for at least  $t_{MRD}$  time.
9. Using the LOAD MODE REGISTER command, load the extended mode register to the desired operating modes. Note that the sequence in which the standard and extended mode registers are programmed is not critical.
10. Issue NOP or DESELECT commands for at least  $t_{MRD}$  time.

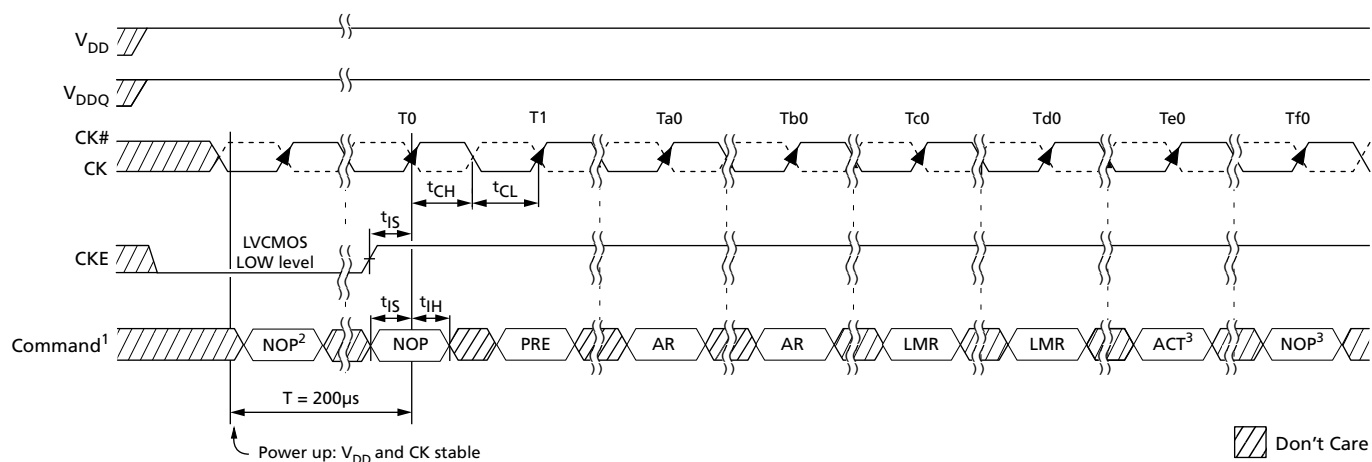
After steps 1–10 are completed, the device has been properly initialized and is ready to receive any valid command.

**Figure 15: Initialize and Load Mode Registers**



- Notes:
1. PRE = PRECHARGE command; LMR = LOAD MODE REGISTER command; AR = AUTO REFRESH command; ACT = ACTIVE command.
  2. NOP or DESELECT commands are required for at least 200µs.
  3. Other valid commands are possible.
  4. NOPs or DESELECTs are required during this time.

**Figure 16: Alternate Initialization with CKE LOW**



- Notes:
1. PRE = PRECHARGE command; LMR = LOAD MODE REGISTER command; AR = AUTO REFRESH command; ACT = ACTIVE command.
  2. NOP or DESELECT commands are required for at least 200µs.
  3. Other valid commands are possible.

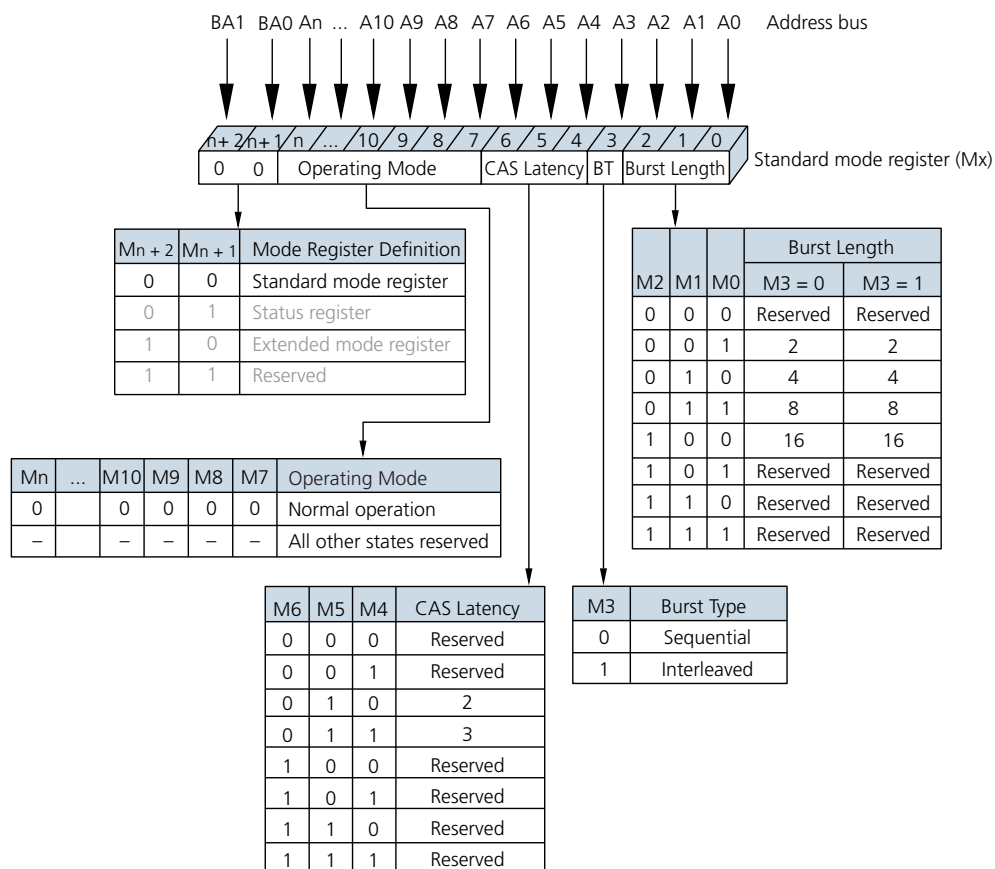


## Standard Mode Register

The standard mode register bit definition enables the selection of burst length, burst type, CAS latency (CL), and operating mode, as shown in Figure 17. Reserved states should not be used as this may result in setting the device into an unknown state or cause incompatibility with future versions of LPDDR devices. The standard mode register is programmed via the LOAD MODE REGISTER command (with BA0 = 0 and BA1 = 0) and will retain the stored information until it is programmed again, until the device goes into deep power-down mode, or until the device loses power.

Reprogramming the mode register will not alter the contents of the memory, provided it is performed correctly. The mode register must be loaded when all banks are idle and no bursts are in progress, and the controller must wait tMRD before initiating the subsequent operation. Violating any of these requirements will result in unspecified operation.

**Figure 17: Standard Mode Register Definition**



Note: 1. The integer  $n$  is equal to the most significant address bit.

## Burst Length

Read and write accesses to the device are burst-oriented, and the burst length (BL) is programmable. The burst length determines the maximum number of column locations that can be accessed for a given READ or WRITE command. Burst lengths of 2, 4, 8, or 16 locations are available for both sequential and interleaved burst types.

When a READ or WRITE command is issued, a block of columns equal to the burst length is effectively selected. All accesses for that burst take place within this block, meaning that the burst will wrap when a boundary is reached. The block is uniquely selected by  $A[i:1]$  when  $BL = 2$ , by  $A[i:2]$  when  $BL = 4$ , by  $A[i:3]$  when  $BL = 8$ , and by  $A[i:4]$  when  $BL = 16$ , where  $A_i$  is the most significant column address bit for a given configuration. The remaining (least significant) address bits are used to specify the starting location within the block. The programmed burst length applies to both READ and WRITE bursts.

## Burst Type

Accesses within a given burst can be programmed to be either sequential or interleaved via the standard mode register.

The ordering of accesses within a burst is determined by the burst length, the burst type, and the starting column address.

**Table 19: Burst Definition Table**

Burst Length	Starting Column Address			Order of Accesses Within a Burst	
				Type = Sequential	Type = Interleaved
2		A0			
			0	0-1	0-1
			1	1-0	1-0
4		A1 A0			
			0 0	0-1-2-3	0-1-2-3
			0 1	1-2-3-0	1-0-3-2
			1 0	2-3-0-1	2-3-0-1
			1 1	3-0-1-2	3-2-1-0
8		A2 A1 A0			
			0 0 0	0-1-2-3-4-5-6-7	0-1-2-3-4-5-6-7
			0 0 1	1-2-3-4-5-6-7-0	1-0-3-2-5-4-7-6
			0 1 0	2-3-4-5-6-7-0-1	2-3-0-1-6-7-4-5
			0 1 1	3-4-5-6-7-0-1-2	3-2-1-0-7-6-5-4
			1 0 0	4-5-6-7-0-1-2-3	4-5-6-7-0-1-2-3
			1 0 1	5-6-7-0-1-2-3-4	5-4-7-6-1-0-3-2
			1 1 0	6-7-0-1-2-3-4-5	6-7-4-5-2-3-0-1
			1 1 1	7-0-1-2-3-4-5-6	7-6-5-4-3-2-1-0
16	A3	A2 A1 A0			

**Table 19: Burst Definition Table (Continued)**

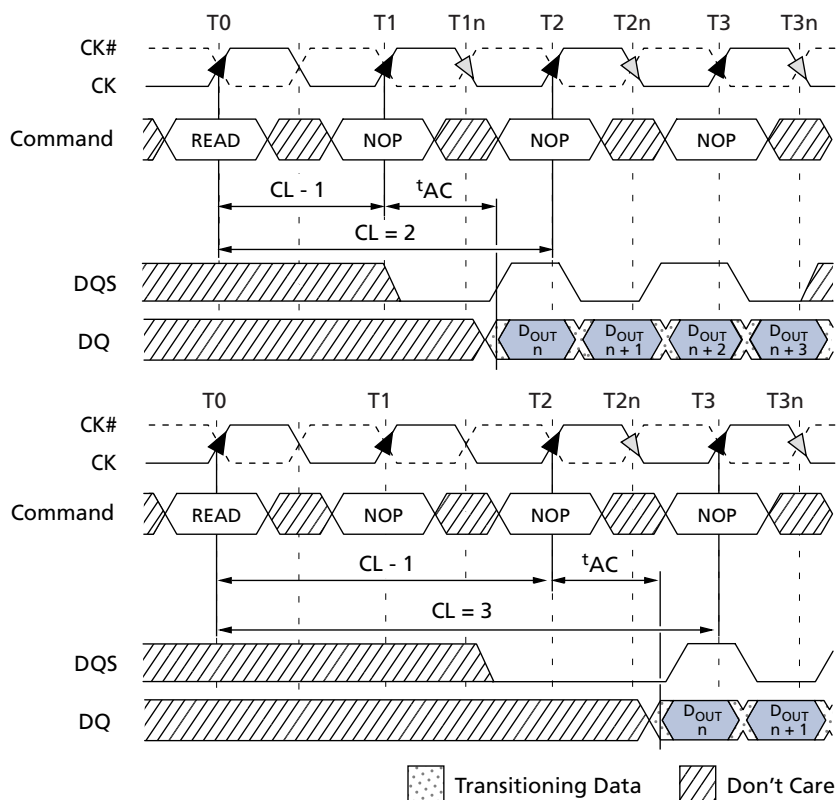
Burst Length	Starting Column Address				Order of Accesses Within a Burst	
					Type = Sequential	Type = Interleaved
	0	0	0	0	0-1-2-3-4-5-6-7-8-9-A-B-C-D-E-F	0-1-2-3-4-5-6-7-8-9-A-B-C-D-E-F
	0	0	0	1	1-2-3-4-5-6-7-8-9-A-B-C-D-E-F-0	1-0-3-2-5-4-7-6-9-8-B-A-D-C-F-E
	0	0	1	0	2-3-4-5-6-7-8-9-A-B-C-D-E-F-0-1	2-3-0-1-6-7-4-5-A-B-8-9-E-F-C-D
	0	0	1	1	3-4-5-6-7-8-9-A-B-C-D-E-F-0-1-2	3-2-1-0-7-6-5-4-B-A-9-8-F-E-D-C
	0	1	0	0	4-5-6-7-8-9-A-B-C-D-E-F-0-1-2-3	4-5-6-7-0-1-2-3-C-D-E-F-8-9-A-B
	0	1	0	1	5-6-7-8-9-A-B-C-D-E-F-0-1-2-3-4	5-4-7-6-1-0-3-2-D-C-F-E-9-8-B-A
	0	1	1	0	6-7-8-9-A-B-C-D-E-F-0-1-2-3-4-5	6-7-4-5-2-3-0-1-E-F-C-D-A-B-8-9
	0	1	1	1	7-8-9-A-B-C-D-E-F-0-1-2-3-4-5-6	7-6-5-4-3-2-1-0-F-E-D-C-B-A-9-8
	1	0	0	0	8-9-A-B-C-D-E-F-0-1-2-3-4-5-6-7	8-9-A-B-C-D-E-F-0-1-2-3-4-5-6-7
	1	0	0	1	9-A-B-C-D-E-F-0-1-2-3-4-5-6-7-8	9-8-B-A-D-C-F-E-1-0-3-2-5-4-7-6
	1	0	1	0	A-B-C-D-E-F-0-1-2-3-4-5-6-7-8-9	A-B-8-9-E-F-C-D-2-3-0-1-6-7-4-5
	1	0	1	1	B-C-D-E-F-0-1-2-3-4-5-6-7-8-9-A	B-A-9-8-F-E-D-C-3-2-1-0-7-6-5-4
	1	1	0	0	C-D-E-F-0-1-2-3-4-5-6-7-8-9-A-B	C-D-E-F-8-9-A-B-4-5-6-7-0-1-2-3
	1	1	0	1	D-E-F-0-1-2-3-4-5-6-7-8-9-A-B-C	D-C-F-E-9-8-B-A-5-4-7-6-1-0-3-2
	1	1	1	0	E-F-0-1-2-3-4-5-6-7-8-9-A-B-C-D	E-F-C-D-A-B-8-9-6-7-4-5-2-3-0-1
	1	1	1	1	F-0-1-2-3-4-5-6-7-8-9-A-B-C-D-E	F-E-D-C-B-A-9-8-7-6-5-4-3-2-1-0

## CAS Latency

The CAS latency (CL) is the delay, in clock cycles, between the registration of a READ command and the availability of the first output data. The latency can be set to 2 or 3 clocks, as shown in Figure 18 (page 52).

For CL = 3, if the READ command is registered at clock edge  $n$ , then the data will be nominally available at  $(n + 2 \text{ clocks} + {}^t\text{AC})$ . For CL = 2, if the READ command is registered at clock edge  $n$ , then the data will be nominally available at  $(n + 1 \text{ clock} + {}^t\text{AC})$ .

**Figure 18: CAS Latency**



## Operating Mode

The normal operating mode is selected by issuing a LOAD MODE REGISTER command with bits A[n:7] each set to zero, and bits A[6:0] set to the desired values.

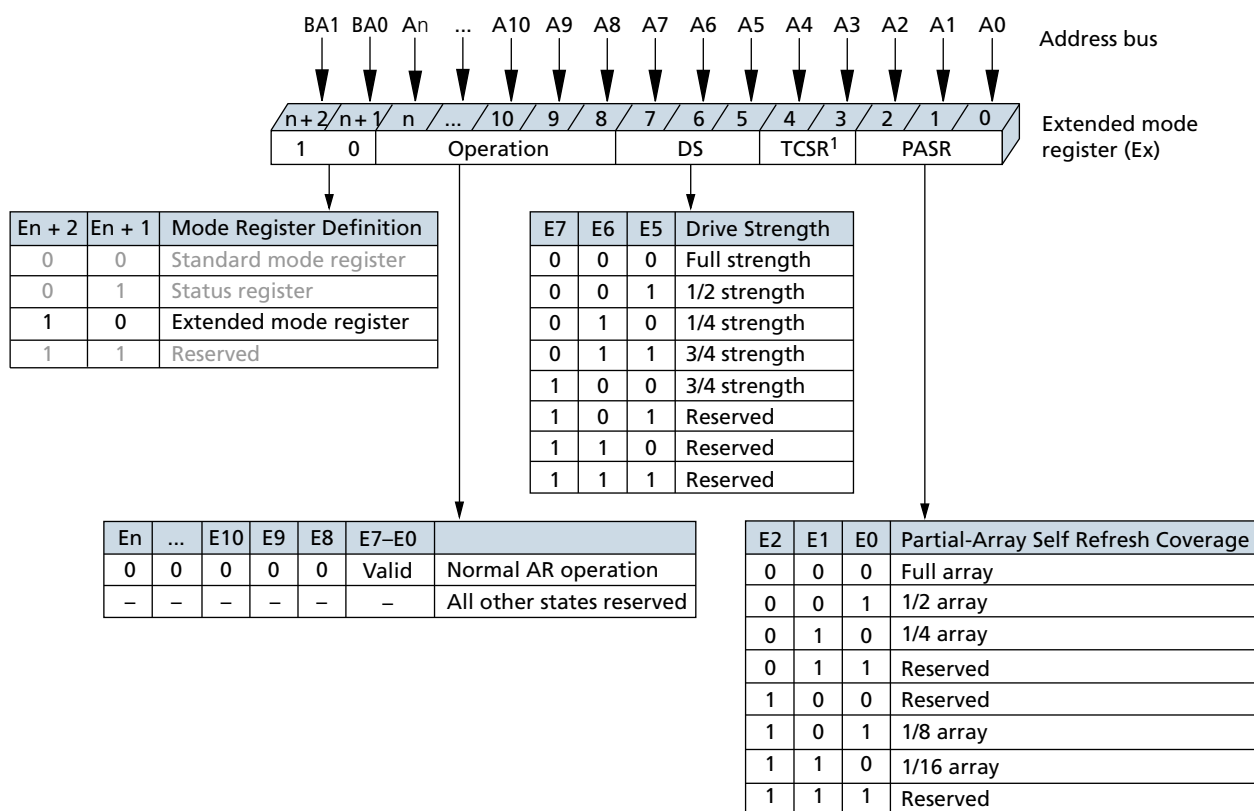
All other combinations of values for A[n:7] are reserved for future use. Reserved states should not be used because unknown operation or incompatibility with future versions may result.

## Extended Mode Register

The EMR controls additional functions beyond those set by the mode registers. These additional functions include drive strength, TCSR, and PASR.

The EMR is programmed via the LOAD MODE REGISTER command with BA0 = 0 and BA1 = 1. Information in the EMR will be retained until it is programmed again, the device goes into deep power-down mode, or the device loses power.

**Figure 19: Extended Mode Register**



- Notes:
1. On-die temperature sensor is used in place of TCSR. Setting these bits will have no effect.
  2. The integer  $n$  is equal to the most significant address bit.

## Temperature-Compensated Self Refresh

This device includes a temperature sensor that is implemented for automatic control of the self refresh oscillator. Programming the temperature-compensated self refresh (TCSR) bits will have no effect on the device. The self refresh oscillator will continue to refresh at the optimal factory-programmed rate for the device temperature.

## Partial-Array Self Refresh

For further power savings during self refresh, the partial-array self refresh (PASR) feature enables the controller to select the amount of memory to be refreshed during self refresh. The refresh options include:

- Full array: banks 0, 1, 2, and 3
- One-half array: banks 0 and 1
- One-quarter array: bank 0
- One-eighth array: bank 0 with row address most significant bit (MSB) = 0
- One-sixteenth array: bank 0 with row address MSB = 0 and row address MSB - 1 = 0

READ and WRITE commands can still be issued to the full array during standard operation, but only the selected regions of the array will be refreshed during self refresh. Data in regions that are not selected will be lost.

## Output Drive Strength

Because the device is designed for use in smaller systems that are typically point-to-point connections, an option to control the drive strength of the output buffers is provided. Drive strength should be selected based on the expected loading of the memory bus. The output driver settings are 25Ω, 37Ω, and 55Ω internal impedance for full, three-quarter, and one-half drive strengths, respectively.

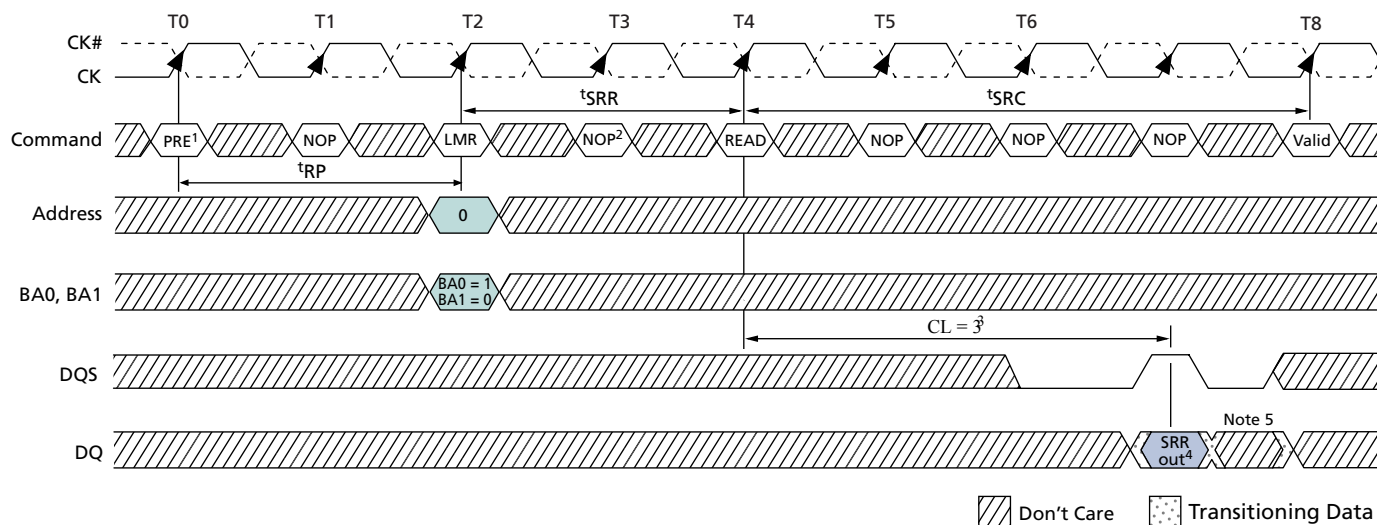
## Status Read Register

The status read register (SRR) is used to read the manufacturer ID, revision ID, refresh multiplier, width type, and density of the device, as shown in Figure 21 (page 56). The SRR is read via the LOAD MODE REGISTER command with BA0 = 1 and BA1 = 0. The sequence to perform an SRR command is as follows:

1. The device must be properly initialized and in the idle or all banks precharged state.
2. Issue a LOAD MODE REGISTER command with BA[1:0] = 01 and all address pins set to 0.
3. Wait  $t_{SRR}$ ; only NOP or DESELECT commands are supported during the  $t_{SRR}$  time.
4. Issue a READ command.
5. Subsequent commands to the device must be issued  $t_{SRC}$  after the SRR READ command is issued; only NOP or DESELECT commands are supported during  $t_{SRC}$ .

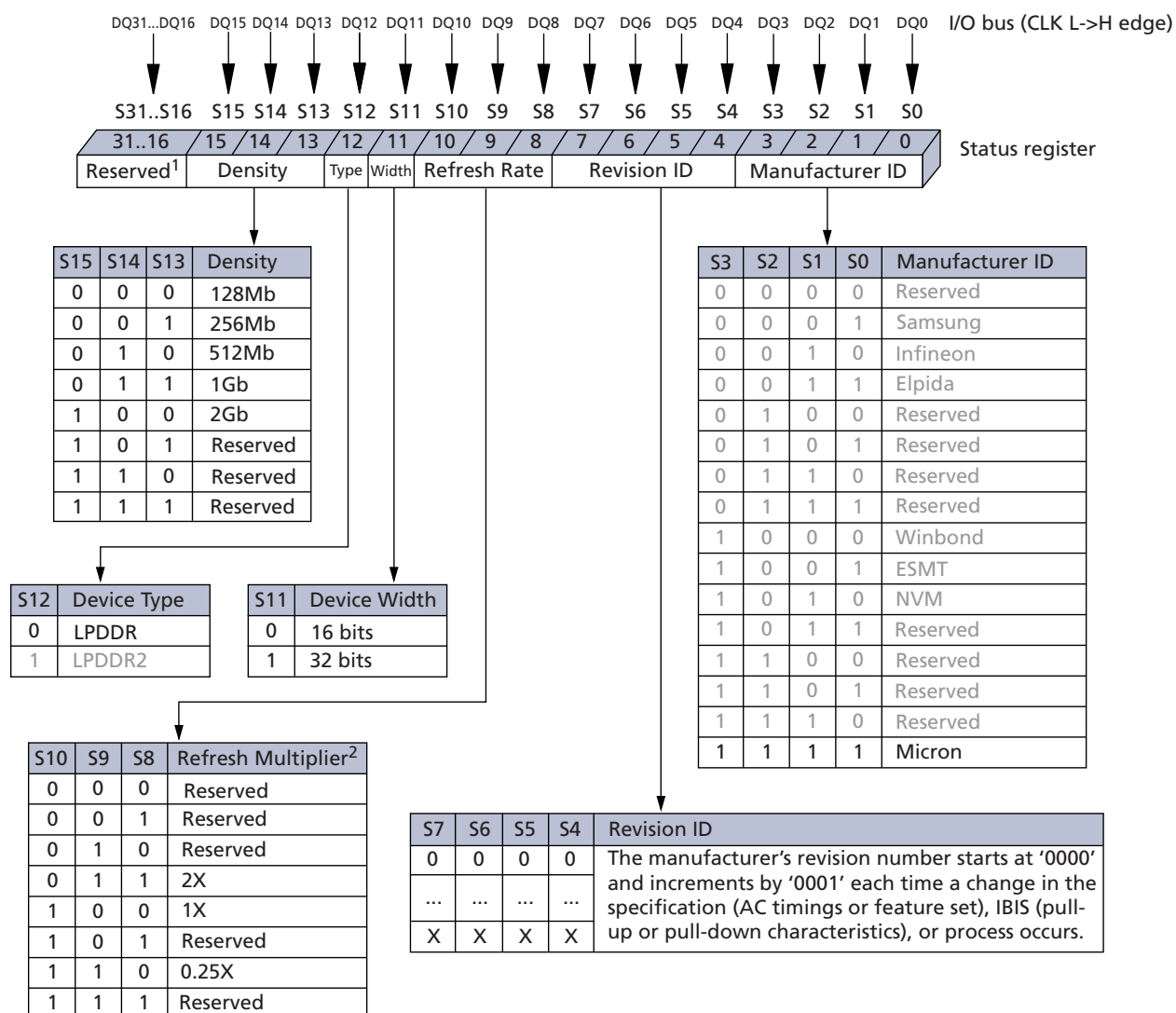
SRR output is read with a burst length of 2. SRR data is driven to the outputs on the first bit of the burst, with the output being “Don’t Care” on the second bit of the burst.

**Figure 20: Status Read Register Timing**



- Notes:
1. All banks must be idle prior to status register read.
  2. NOP or DESELECT commands are required between the LMR and READ commands ( $t_{SRR}$ ), and between the READ and the next VALID command ( $t_{SRC}$ ).
  3. CAS latency is predetermined by the programming of the mode register. CL = 3 is shown as an example only.
  4. Burst length is fixed to 2 for SRR regardless of the value programmed by the mode register.
  5. The second bit of the data-out burst is a “Don’t Care.”

**Figure 21: Status Register Definition**



- Notes:
1. Reserved bits should be set to 0 for future compatibility.
  2. Refresh multiplier is based on the memory device on-board temperature sensor. Required average periodic refresh interval =  $t_{REFI} \times \text{multiplier}$ .



## Bank/Row Activation

Before any READ or WRITE commands can be issued to a bank within the device, a row in that bank must be opened. This is accomplished via the ACTIVE command, which selects both the bank and the row to be activated (see the ACTIVE Command figure). After a row is opened with the ACTIVE command, a READ or WRITE command can be issued to that row, subject to the  $t_{RCD}$  specification.

A subsequent ACTIVE command to a different row in the same bank can only be issued after the previous active row has been precharged. The minimum time interval between successive ACTIVE commands to the same bank is defined by  $t_{RC}$ .

A subsequent ACTIVE command to another bank can be issued while the first bank is being accessed, which results in a reduction of total row access overhead. The minimum time interval between successive ACTIVE commands to different banks is defined by  $t_{RRD}$ .

## READ Operation

READ burst operations are initiated with a READ command, as shown in Figure 10 (page 36). The starting column and bank addresses are provided with the READ command, and auto precharge is either enabled or disabled for that burst access. If auto precharge is enabled, the row being accessed is precharged at the completion of the burst. For the READ commands used in the following illustrations, auto precharge is disabled.

During READ bursts, the valid data-out element from the starting column address will be available following the CL after the READ command. Each subsequent data-out element will be valid nominally at the next positive or negative clock edge. Figure 22 (page 59) shows general timing for each possible CL setting.

DQS is driven by the device along with output data. The initial LOW state on DQS is known as the read preamble; the LOW state coincident with the last data-out element is known as the read postamble. The READ burst is considered complete when the read postamble is satisfied.

Upon completion of a burst, assuming no other commands have been initiated, the DQ will go to High-Z. A detailed explanation of  $t_{DQSQ}$  (valid data-out skew),  $t_{QH}$  (data-out window hold), and the valid data window is depicted in Figure 29 (page 66) and Figure 30 (page 67). A detailed explanation of  $t_{DQSCK}$  (DQS transition skew to CK) and  $t_{AC}$  (data-out transition skew to CK) is depicted in Figure 31 (page 68).

Data from any READ burst can be truncated by a READ or WRITE command to the same or alternate bank, by a BURST TERMINATE command, or by a PRECHARGE command to the same bank, provided that the auto precharge mode was not activated.

Data from any READ burst can be concatenated with or truncated with data from a subsequent READ command. In either case, a continuous flow of data can be maintained. The first data element from the new burst either follows the last element of a completed burst or the last desired data element of a longer burst that is being truncated. The new READ command should be issued  $x$  cycles after the first READ command, where  $x$  equals the number of desired data element pairs (pairs are required by the  $2n$ -prefetch architecture). This is shown in Figure 23 (page 60).

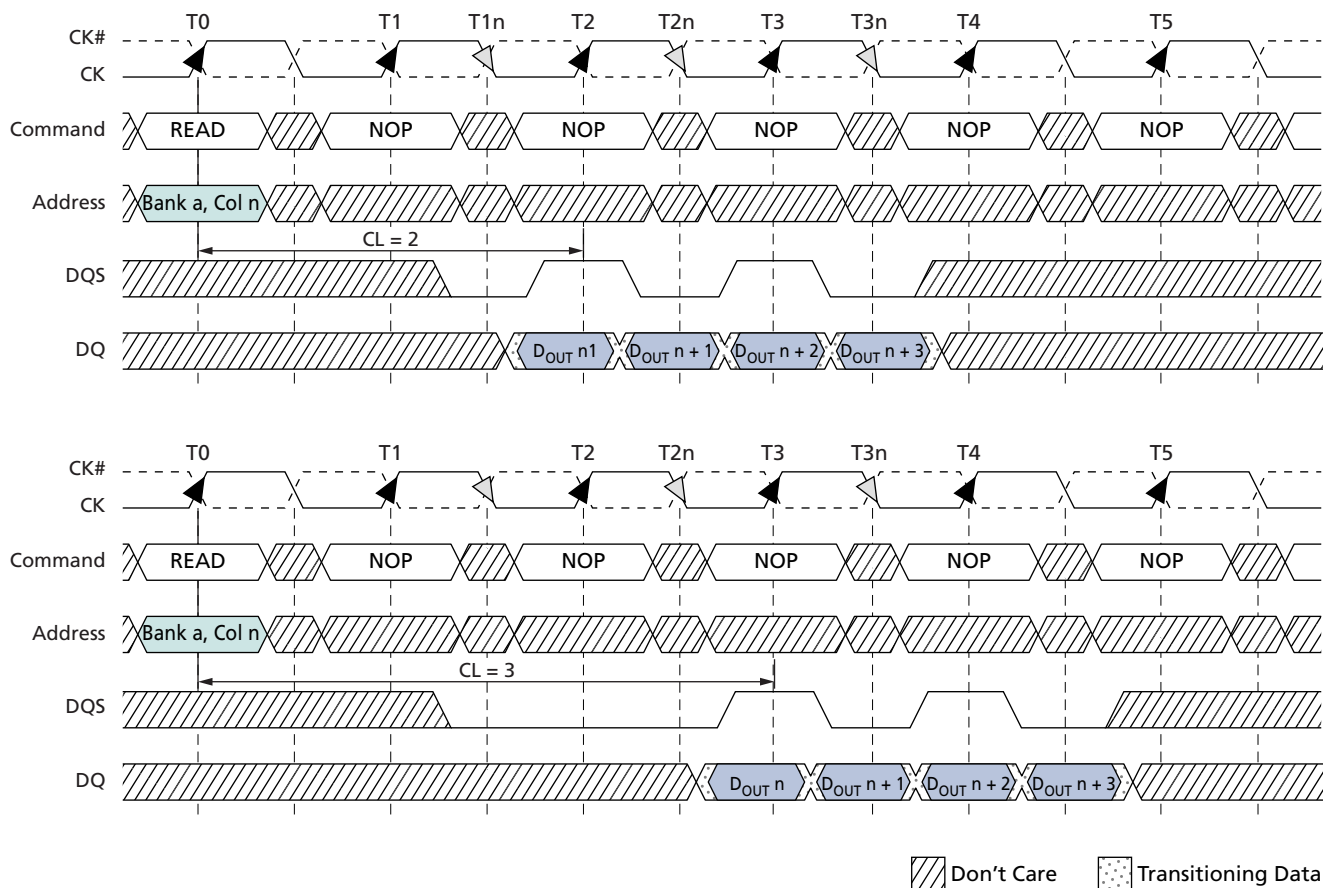
A READ command can be initiated on any clock cycle following a previous READ command. Nonconsecutive read data is shown in Figure 24 (page 61). Full-speed random read accesses within a page (or pages) can be performed as shown in Figure 25 (page 62).

Data from any READ burst can be truncated with a BURST TERMINATE command, as shown in Figure 26 (page 63). The BURST TERMINATE latency is equal to the READ (CAS) latency; for example, the BURST TERMINATE command should be issued  $x$  cycles after the READ command, where  $x$  equals the number of desired data element pairs (pairs are required by the  $2n$ -prefetch architecture).

Data from any READ burst must be completed or truncated before a subsequent WRITE command can be issued. If truncation is necessary, the BURST TERMINATE command must be used, as shown in Figure 27 (page 64). A READ burst can be followed by, or truncated with, a PRECHARGE command to the same bank, provided that auto precharge was not activated. The PRECHARGE command should be issued  $x$  cycles after the READ command, where  $x$  equals the number of desired data element pairs. This is shown in Figure 28 (page 65). Following the PRECHARGE command, a subsequent

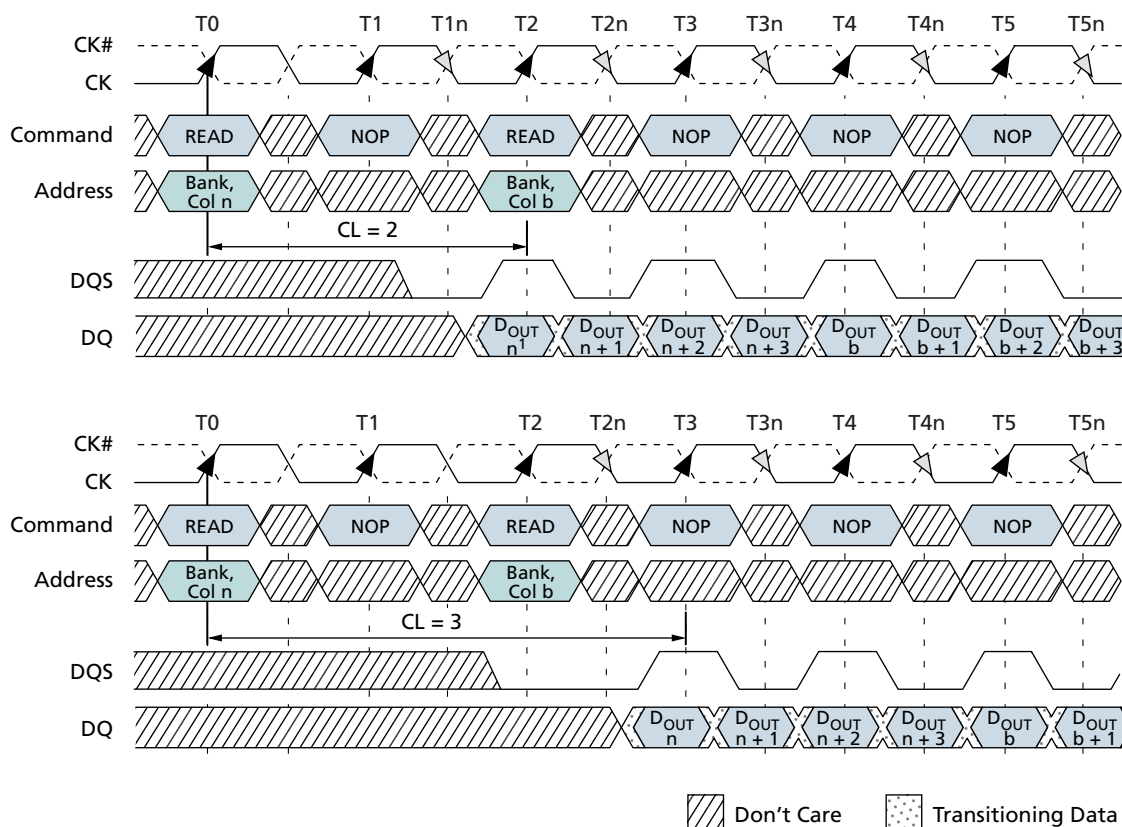
command to the same bank cannot be issued until  $t_{RP}$  is met. Part of the row precharge time is hidden during the access of the last data elements.

**Figure 22: READ Burst**



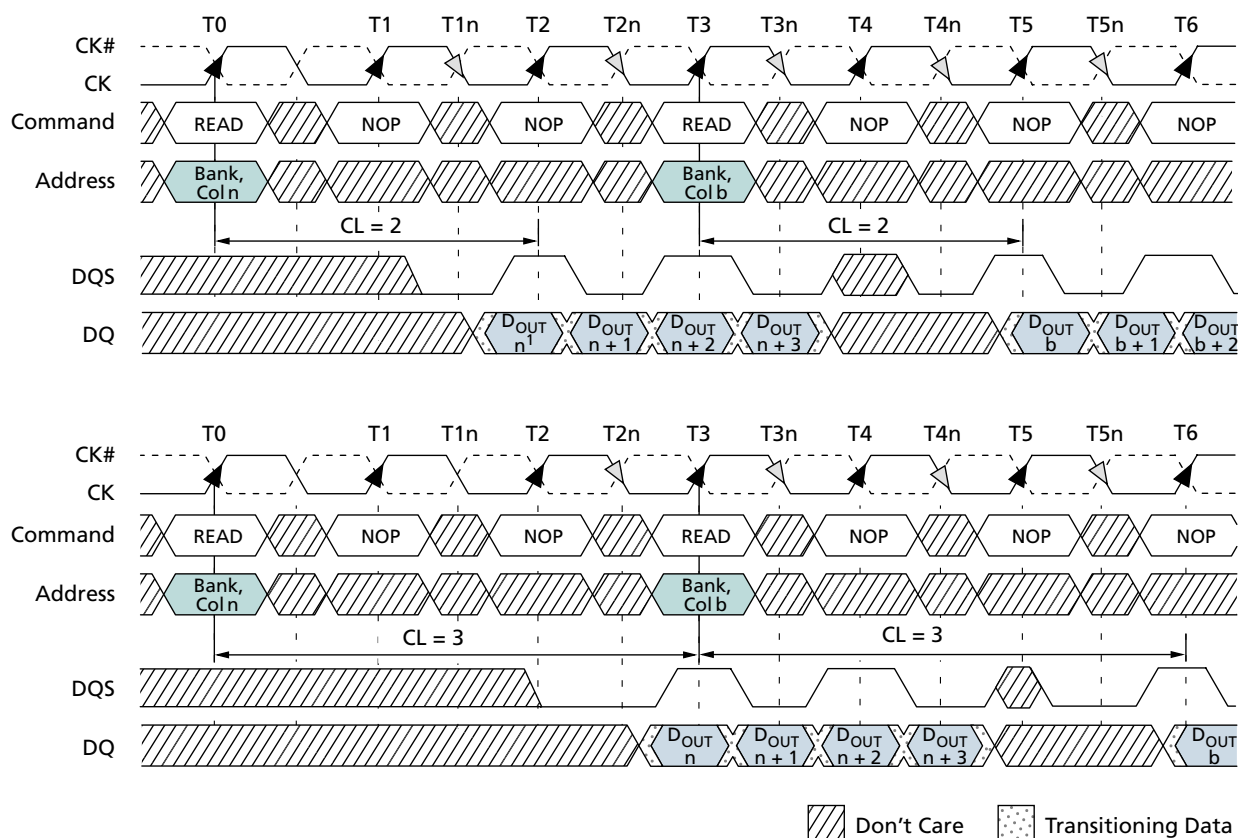
- Notes:
1.  $D_{OUT} n$  = data-out from column  $n$ .
  2.  $BL = 4$ .
  3. Shown with nominal  $t_{AC}$ ,  $t_{DQSCK}$ , and  $t_{DQSQ}$ .

**Figure 23: Consecutive READ Bursts**



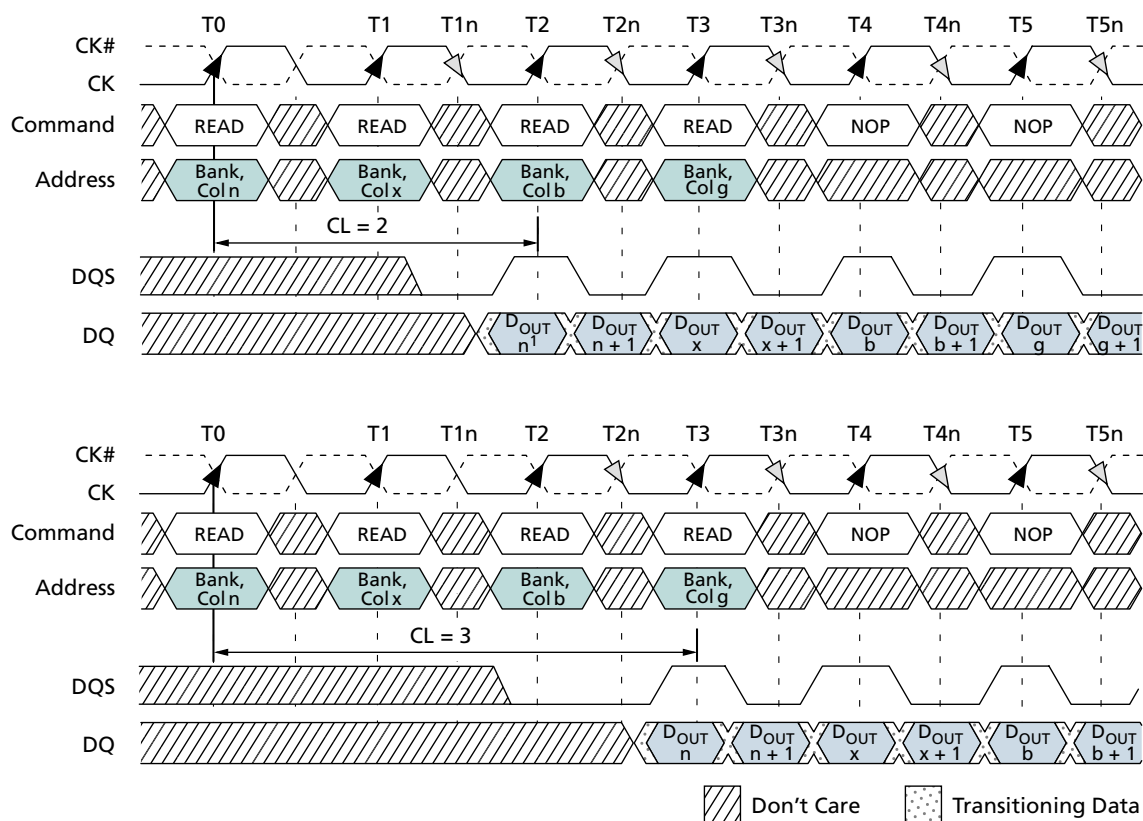
- Notes:
1.  $D_{OUT}n$  (or  $b$ ) = data-out from column  $n$  (or column  $b$ ).
  2. BL = 4, 8, or 16 (if 4, the bursts are concatenated; if 8 or 16, the second burst interrupts the first).
  3. Shown with nominal  $t_{AC}$ ,  $t_{DQSK}$ , and  $t_{DQSQ}$ .
  4. Example applies only when READ commands are issued to same device.

**Figure 24: Nonconsecutive READ Bursts**



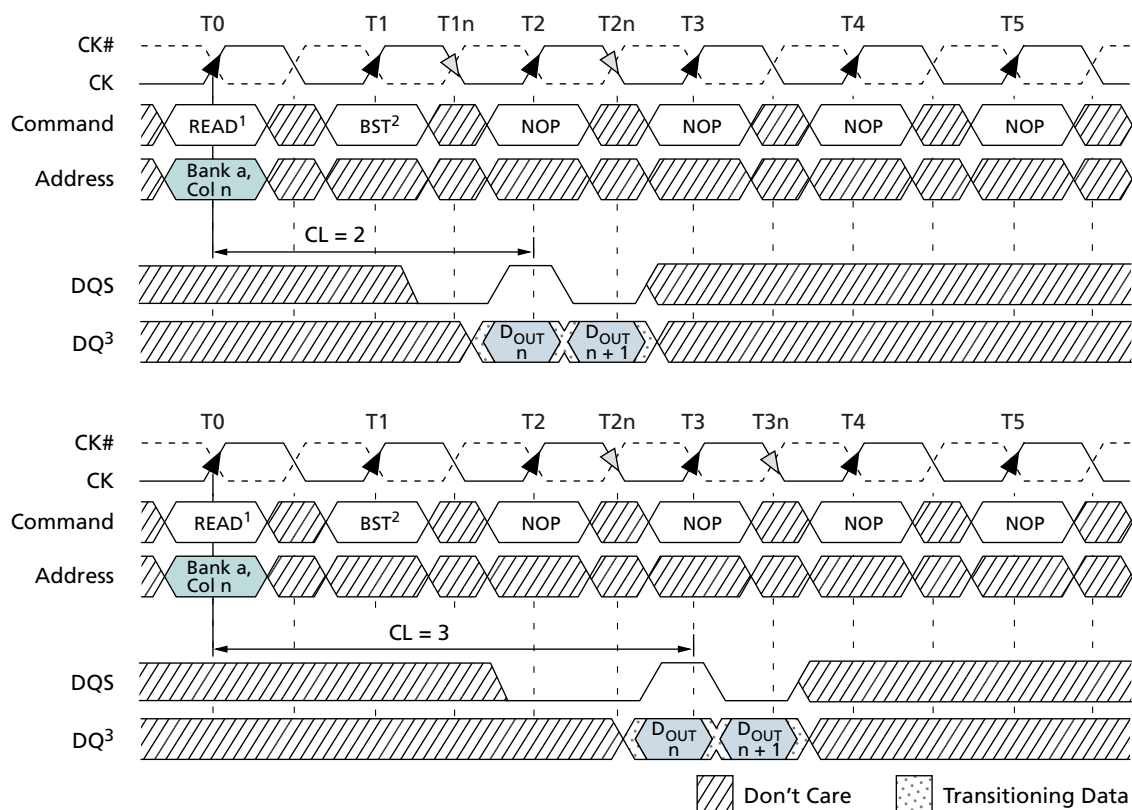
- Notes:
1.  $D_{OUT}n$  (or  $b$ ) = data-out from column  $n$  (or column  $b$ ).
  2.  $BL = 4, 8$ , or  $16$  (if burst is 8 or 16, the second burst interrupts the first).
  3. Shown with nominal  $t_{AC}$ ,  $t_{DQSQ}$ , and  $t_{DQSQ}$ .
  4. Example applies when READ commands are issued to different devices or nonconsecutive READs.

**Figure 25: Random Read Accesses**



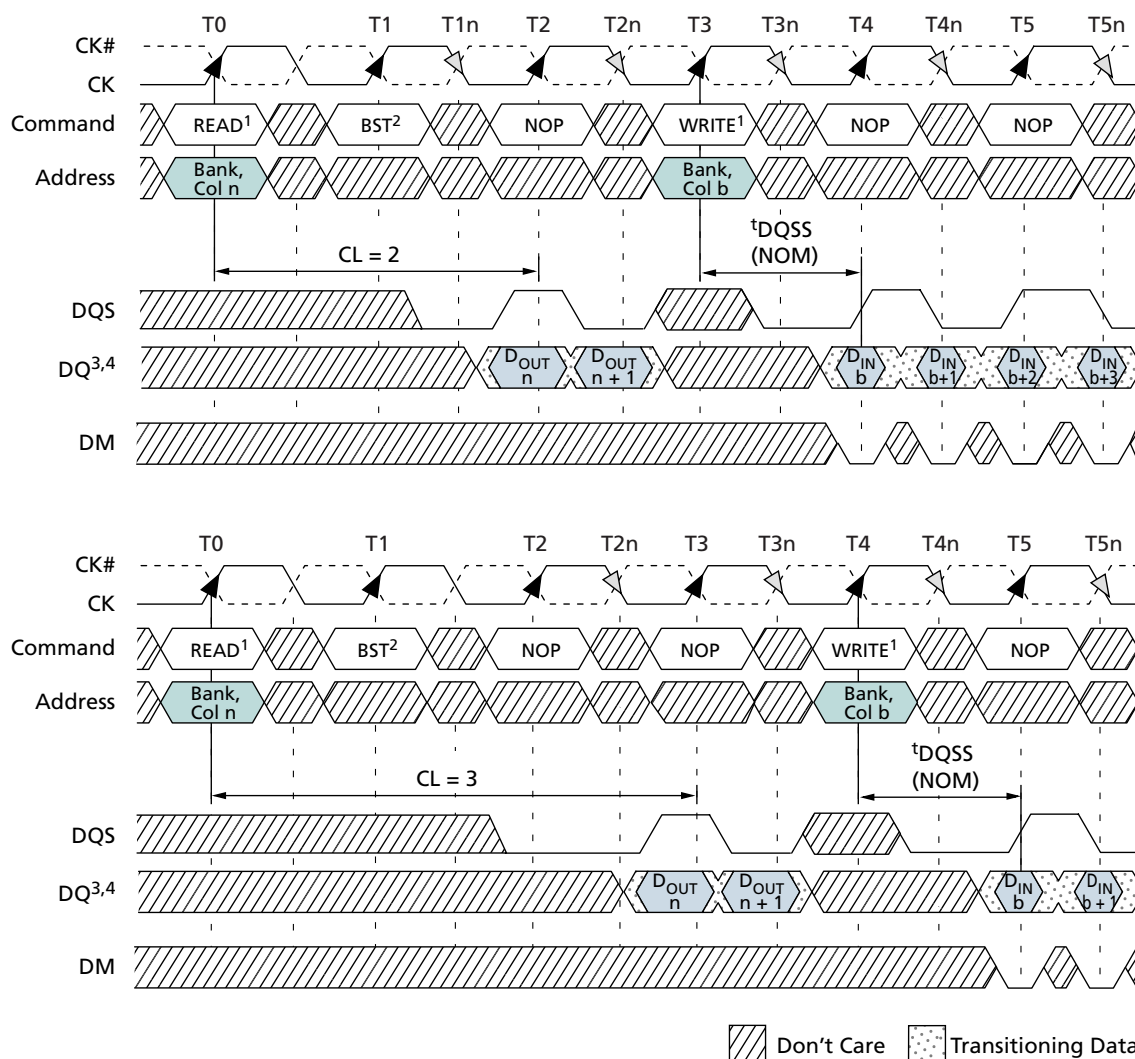
- Notes:
1.  $D_{OUTn}$  (or  $x$ ,  $b$ ,  $g$ ) = data-out from column  $n$  (or column  $x$ , column  $b$ , column  $g$ ).
  2.  $BL = 2, 4, 8$ , or  $16$  (if  $4, 8$ , or  $16$ , the following burst interrupts the previous).
  3. READs are to an active row in any bank.
  4. Shown with nominal  $t_{AC}$ ,  $t_{DQSCK}$ , and  $t_{DQSQ}$ .

**Figure 26: Terminating a READ Burst**



- Notes:
1. BL = 4, 8, or 16.
  2. BST = BURST TERMINATE command; page remains open.
  3. DOUT<sub>n</sub> = data-out from column *n*.
  4. Shown with nominal *t*<sub>AC</sub>, *t*<sub>DQSC</sub>, and *t*<sub>DQSQ</sub>.
  5. CKE = HIGH.

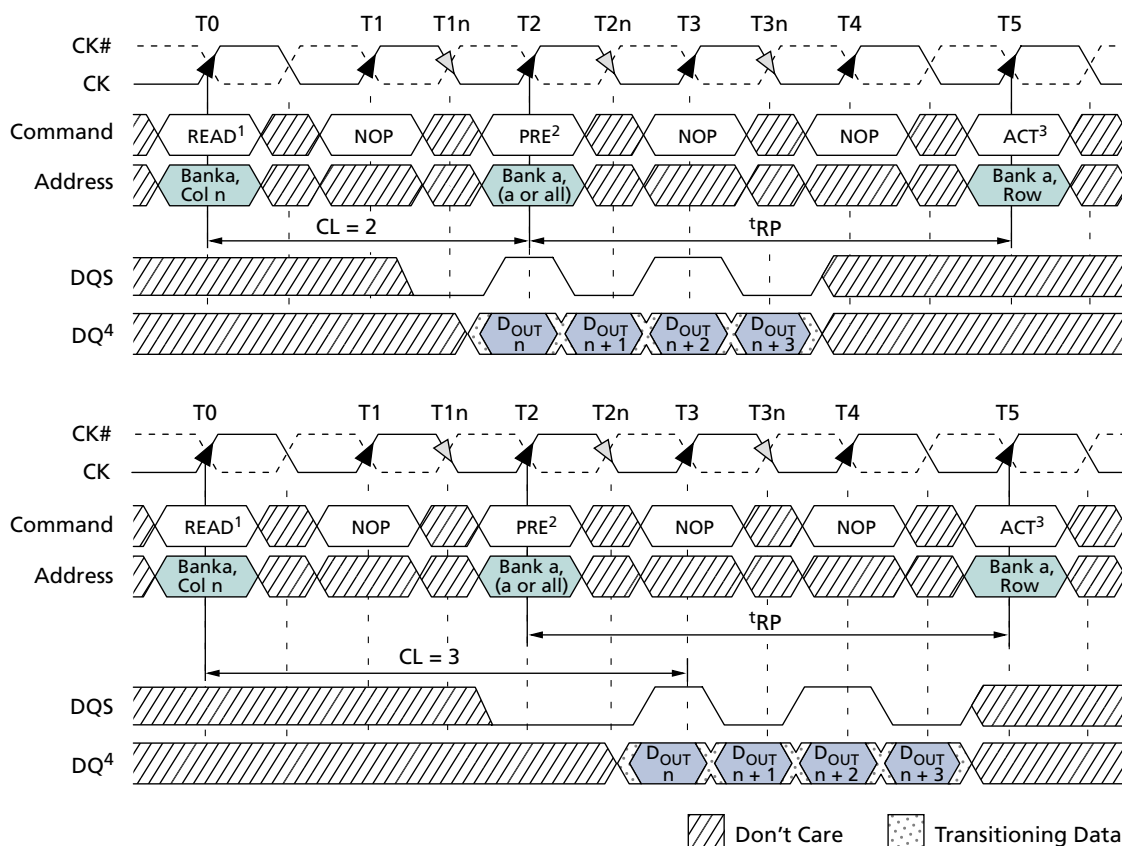
**Figure 27: READ-to-WRITE**



- Notes:
1. BL = 4 in the cases shown (applies for bursts of 8 and 16 as well; if BL = 2, the BST command shown can be NOP).
  2. BST = BURST TERMINATE command; page remains open.
  3.  $D_{OUTn}$  = data-out from column  $n$ .
  4.  $D_{INb}$  = data-in from column  $b$ .
  5. Shown with nominal  $t_{AC}$ ,  $t_{DQSQ}$ , and  $t_{DQSQ}$ .
  6. CKE = HIGH.

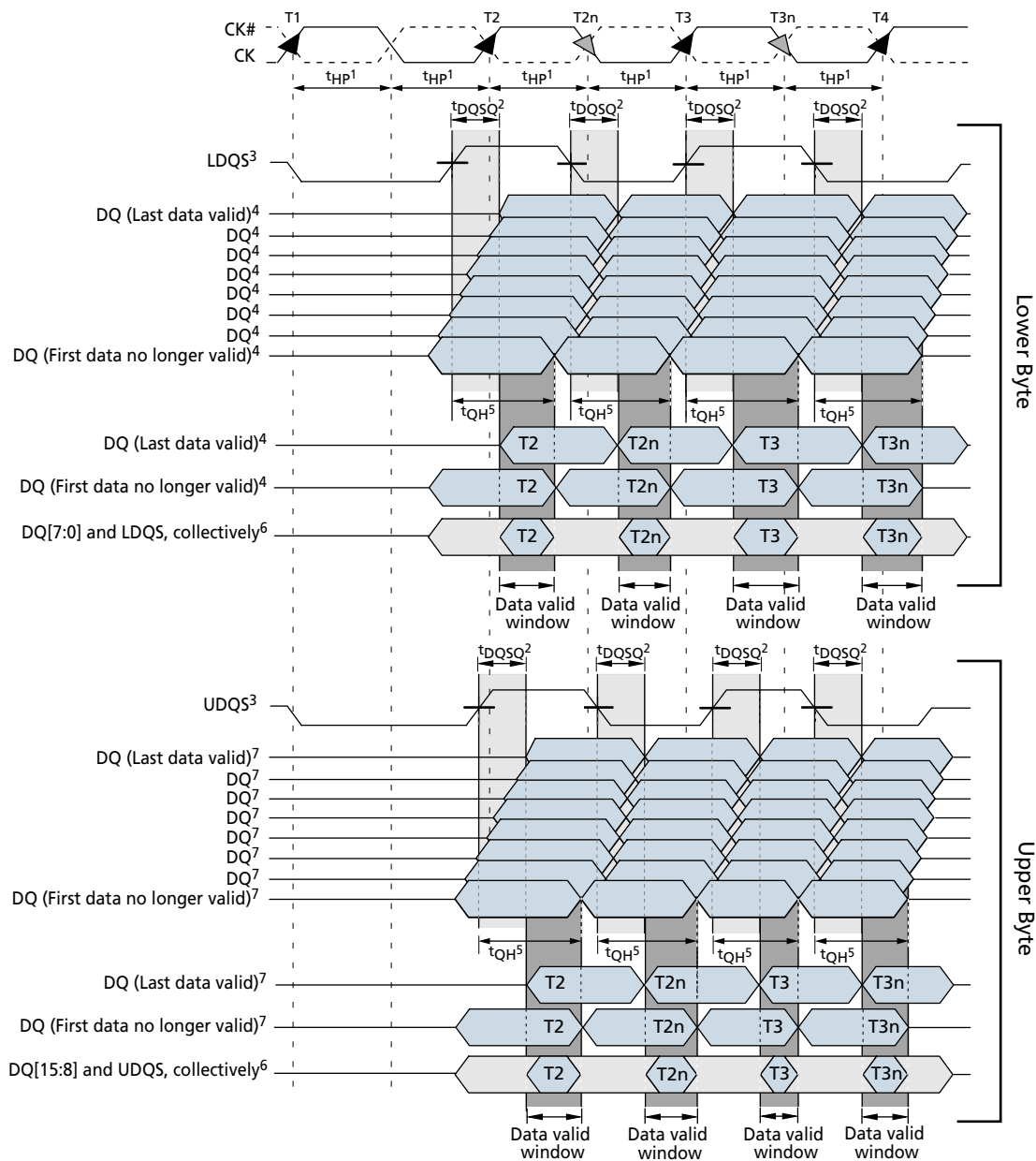


**Figure 28: READ-to-PRECHARGE**



- Notes:
1. BL = 4, or an interrupted burst of 8 or 16.
  2. PRE = PRECHARGE command.
  3. ACT = ACTIVE command.
  4. DOUT $n$  = data-out from column  $n$ .
  5. Shown with nominal  $t_{AC}$ ,  $t_{DQSCK}$ , and  $t_{DQSQ}$ .
  6. READ-to-PRECHARGE equals 2 clocks, which enables 2 data pairs of data-out.
  7. A READ command with auto precharge enabled, provided  $t_{RAS}$  (MIN) is met, would cause a precharge to be performed at  $x$  number of clock cycles after the READ command, where  $x = BL/2$ .

**Figure 29: Data Output Timing –  $t_{DQSQ}$ ,  $t_{QH}$ , and Data Valid Window (x16)**

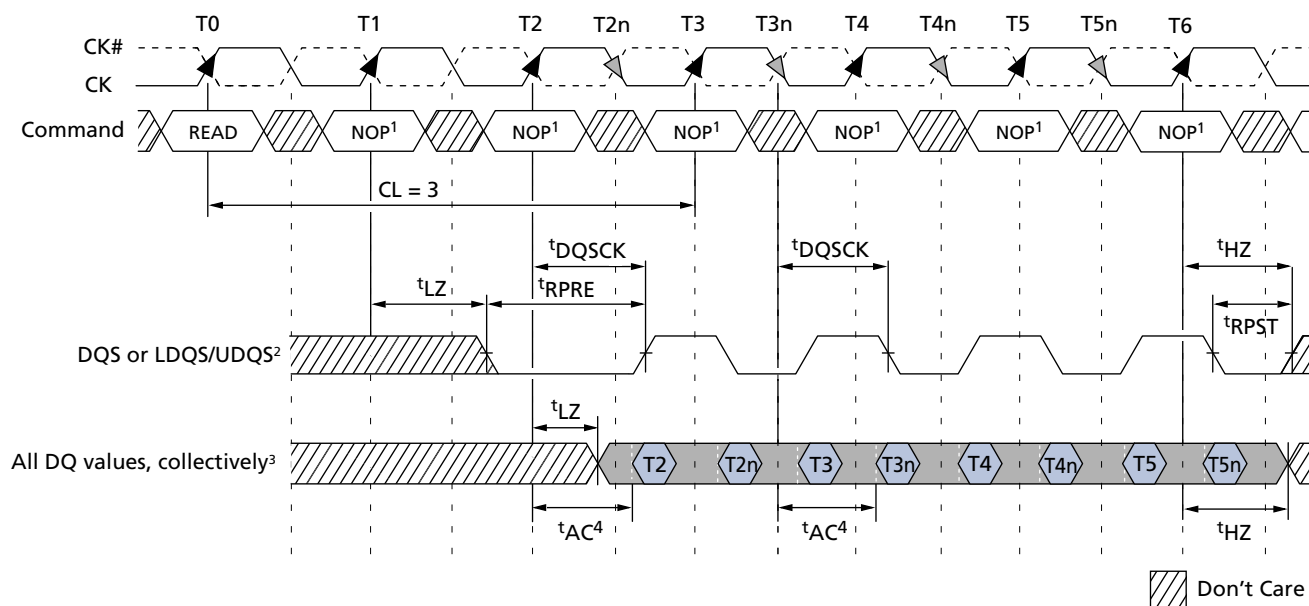


- Notes:
- $t_{HP}$  is the lesser of  $t_{CL}$  or  $t_{CH}$  clock transition collectively when a bank is active.
  - $t_{DQSQ}$  is derived at each DQS clock edge and is not cumulative over time and begins with DQS transition and ends with the last valid DQ transition.
  - DQ transitioning after DQS transitions define the  $t_{DQSQ}$  window. LDQS defines the lower byte and UDQS defines the upper byte.
  - DQ0, DQ1, DQ2, DQ3, DQ4, DQ5, DQ6, or DQ7.
  - $t_{QH}$  is derived from  $t_{HP}$ :  $t_{QH} = t_{HP} - t_{QHS}$ .
  - The data valid window is derived for each DQS transitions and is defined as  $t_{QH} - t_{DQSQ}$ .
  - DQ8, DQ9, DQ10, DQ11, DQ12, DQ13, DQ14, or DQ15.

The diagram illustrates the timing of a 4-lane DRAM array. At the top, a clock signal (CK) is shown with transitions T1, T2, T2n, T3, T3n, and T4. The period between consecutive clock transitions is labeled  $t_{HP1}$ . Below the clock, the data strobe signal (DQS0/DQS1/DQS2/DQS3) is shown as a series of pulses. The duration of each pulse is labeled  $t_{DQSQ^{2,3}}$ . The data bus (DQ) is shown as a series of data words (hexagons) for each lane. The data is valid during the data valid windows, which are indicated by dashed vertical lines. The data is no longer valid after the data valid windows. The diagram also shows the data hold time ( $t_{QH^5}$ ) and the data valid time ( $t_{DV}$ ) for each lane. The data valid windows are labeled "Data valid window" at the bottom. The data is valid for 4 lanes (Byte 0, Byte 1, Byte 2, Byte 3) and is no longer valid after the data valid windows.

- Notes:
1.  $t_{HP}$  is the lesser of  $t_{CL}$  or  $t_{CH}$  clock transition collectively when a bank is active.
  2. DQ transitioning after DQS transitions define the  $t_{DQSQ}$  window.
  3.  $t_{DQSQ}$  is derived at each DQS clock edge and is not cumulative over time; it begins with DQS transition and ends with the last valid DQ transition.
  4. Byte 0 is DQ[7:0], byte 1 is DQ[15:8], byte 2 is DQ[23:16], byte 3 is DQ[31:24].
  5.  $t_{QH}$  is derived from  $t_{HP}$ :  $t_{QH} = t_{HP} - t_{QHS}$ .
  6. The data valid window is derived for each DQS transition and is  $t_{QH} - t_{DQSQ}$ .
  7. DQ[7:0] and DQS0 for byte 0; DQ[15:8] and DQS1 for byte 1; DQ[23:16] and DQS2 for byte 2; DQ[31:23] and DQS3 for byte 3.

### Figure 31: Data Output Timing – $t_{AC}$ and $t_{DQSCK}$



- Notes:
1. Commands other than NOP can be valid during this cycle.
  2. DQ transitioning after DQS transitions define  ${}^t\text{DQSQ}$  window.
  3. All DQ must transition by  ${}^t\text{DQSQ}$  after DQS transitions, regardless of  ${}^t\text{AC}$ .
  4.  ${}^t\text{AC}$  is the DQ output window relative to CK and is the long-term component of DQ skew.

## WRITE Operation

WRITE bursts are initiated with a WRITE command, as shown in Figure 11 (page 37). The starting column and bank addresses are provided with the WRITE command, and auto precharge is either enabled or disabled for that access. If auto precharge is enabled, the row being accessed is precharged at the completion of the burst. For the WRITE commands used in the following illustrations, auto precharge is disabled. Basic data input timing is shown in Figure 32 (page 70) (this timing applies to all WRITE operations).

Input data appearing on the data bus is written to the memory array subject to the state of data mask (DM) inputs coincident with the data. If DM is registered LOW, the corresponding data will be written; if DM is registered HIGH, the corresponding data will be ignored, and the write will not be executed to that byte/column location. DM operation is illustrated in Figure 33 (page 71).

During WRITE bursts, the first valid data-in element will be registered on the first rising edge of DQS following the WRITE command, and subsequent data elements will be registered on successive edges of DQS. The LOW state of DQS between the WRITE command and the first rising edge is known as the write preamble; the LOW state of DQS following the last data-in element is known as the write postamble. The WRITE burst is complete when the write postamble and  $t_{WR}$  or  $t_{WTR}$  are satisfied.

The time between the WRITE command and the first corresponding rising edge of DQS ( $t_{DQSS}$ ) is specified with a relatively wide range (75%–125% of one clock cycle). All WRITE diagrams show the nominal case. Where the two extreme cases (that is,  $t_{DQSS}$  [MIN] and  $t_{DQSS}$  [MAX]) might not be obvious, they have also been included. Figure 34 (page 72) shows the nominal case and the extremes of  $t_{DQSS}$  for a burst of 4. Upon completion of a burst, assuming no other commands have been initiated, the DQ will remain High-Z and any additional input data will be ignored.

Data for any WRITE burst can be concatenated with or truncated by a subsequent WRITE command. In either case, a continuous flow of input data can be maintained. The new WRITE command can be issued on any positive edge of clock following the previous WRITE command. The first data element from the new burst is applied after either the last element of a completed burst or the last desired data element of a longer burst that is being truncated. The new WRITE command should be issued  $x$  cycles after the first WRITE command, where  $x$  equals the number of desired data element pairs (pairs are required by the  $2n$ -prefetch architecture).

Figure 35 (page 73) shows concatenated bursts of 4. An example of nonconsecutive WRITES is shown in Figure 36 (page 73). Full-speed random write accesses within a page or pages can be performed, as shown in Figure 37 (page 74).

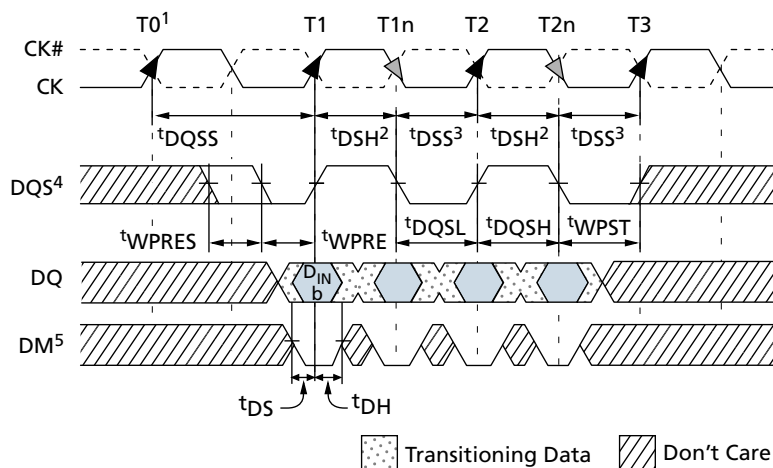
Data for any WRITE burst can be followed by a subsequent READ command. To follow a WRITE without truncating the WRITE burst,  $t_{WTR}$  should be met, as shown in Figure 38 (page 75).

Data for any WRITE burst can be truncated by a subsequent READ command, as shown in Figure 39 (page 76). Note that only the data-in pairs that are registered prior to the  $t_{WTR}$  period are written to the internal array, and any subsequent data-in should be masked with DM, as shown in Figure 40 (page 77).

Data for any WRITE burst can be followed by a subsequent PRECHARGE command. To follow a WRITE without truncating the WRITE burst,  $t_{WR}$  should be met, as shown in Figure 41 (page 78).

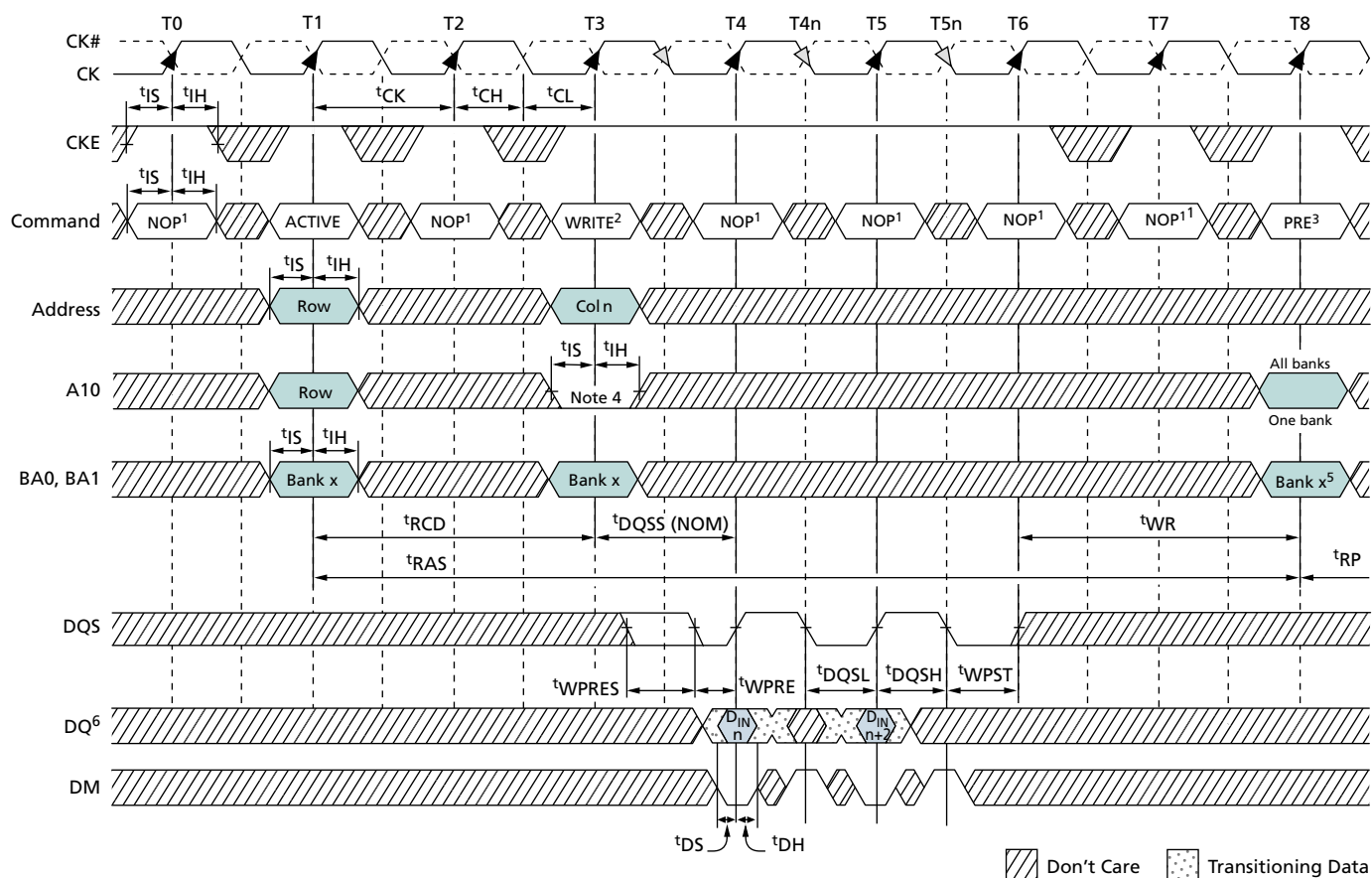
Data for any WRITE burst can be truncated by a subsequent PRECHARGE command, as shown in Figure 42 (page 79) and Figure 43 (page 80). Note that only the data-in pairs that are registered prior to the  $t_{WR}$  period are written to the internal array, and any subsequent data-in should be masked with DM, as shown in Figure 42 (page 79) and Figure 43 (page 80). After the PRECHARGE command, a subsequent command to the same bank cannot be issued until  $t_{RP}$  is met.

**Figure 32: Data Input Timing**



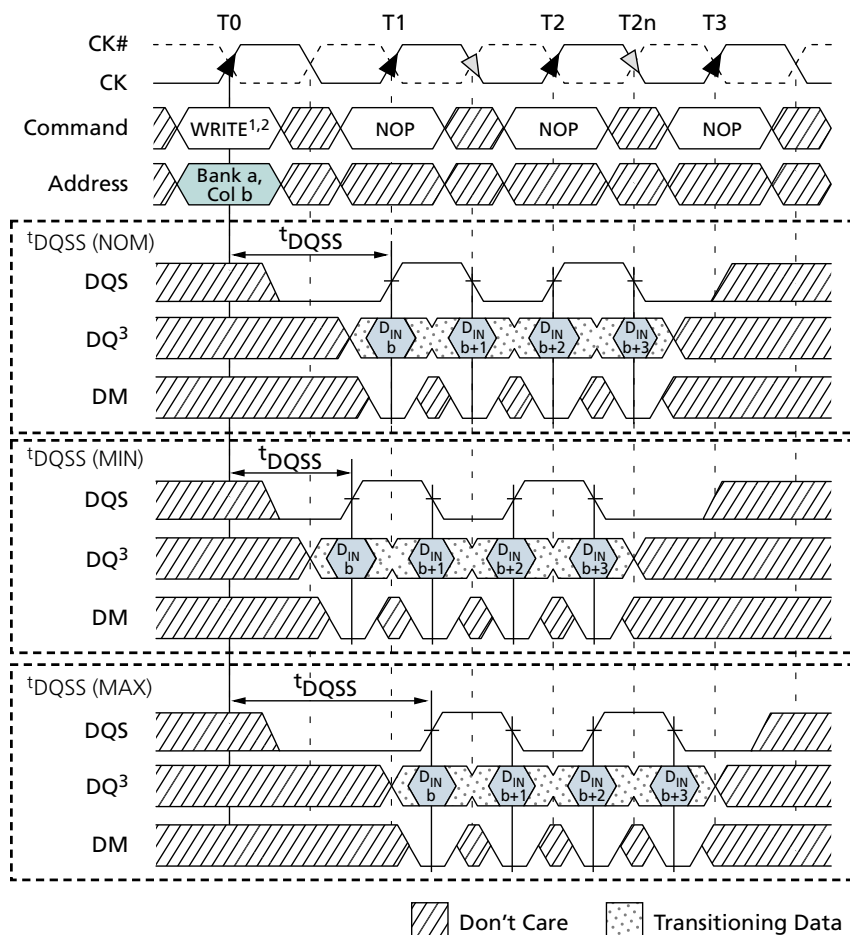
- Notes:
1. WRITE command issued at T0.
  2.  $t_{DSH}$  (MIN) generally occurs during  $t_{DQSS}$  (MIN).
  3.  $t_{DSS}$  (MIN) generally occurs during  $t_{DQSS}$  (MAX).
  4. For x16, LDQS controls the lower byte; UDQS controls the upper byte. For x32, DQS0 controls DQ[7:0], DQS1 controls DQ[15:8], DQS2 controls DQ[23:16], and DQS3 controls DQ[31:24].
  5. For x16, LDM controls the lower byte; UDM controls the upper byte. For x32, DM0 controls DQ[7:0], DM1 controls DQ[15:8], DM2 controls DQ[23:16], and DM3 controls DQ[31:24].

**Figure 33: Write – DM Operation**



- Notes:
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
  2. BL = 4 in the case shown.
  3. PRE = PRECHARGE.
  4. Disable auto precharge.
  5. Bank x at T8 is "Don't Care" if A10 is HIGH at T8.
  6.  $D_{INn}$  = data-in from column n.

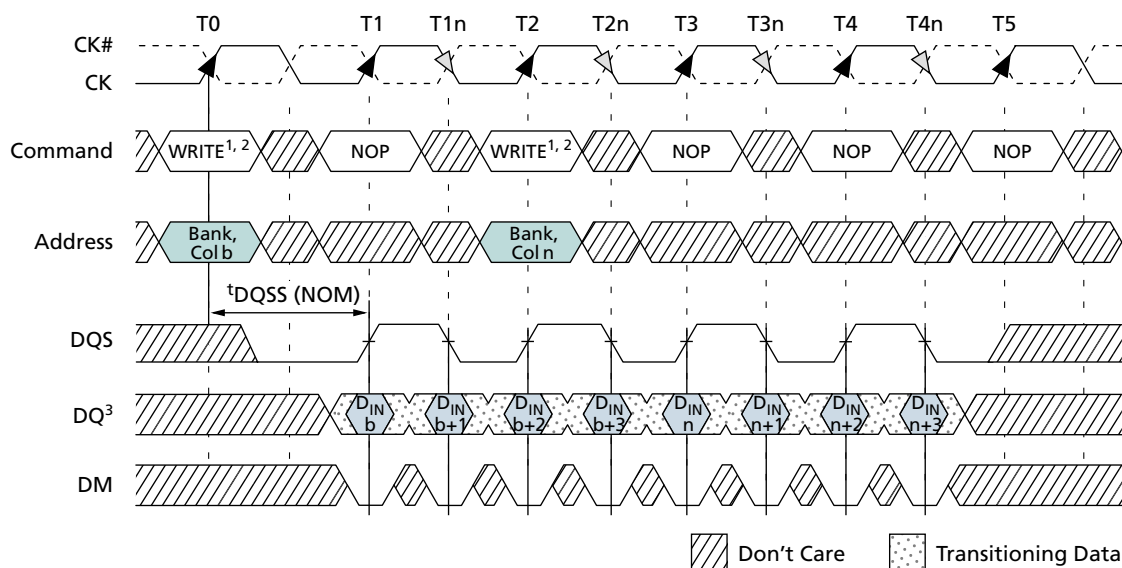
**Figure 34: WRITE Burst**



- Notes:
1. An uninterrupted burst of 4 is shown.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3.  $D_{IN}b$  = data-in for column  $b$ .

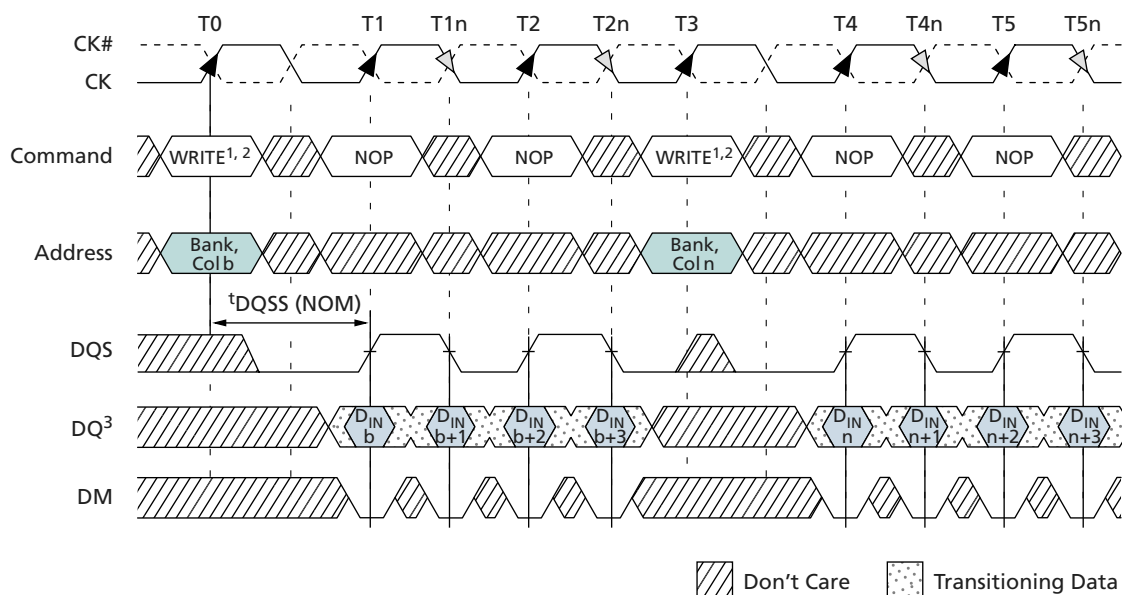


**Figure 35: Consecutive WRITE-to-WRITE**



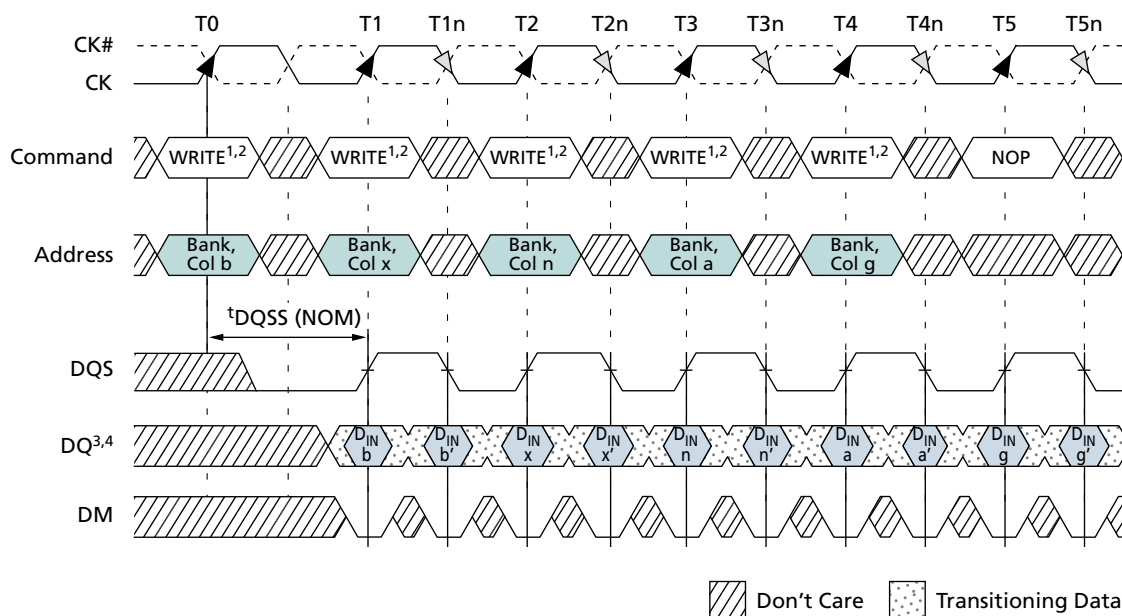
- Notes:
1. Each WRITE command can be to any bank.
  2. An uninterrupted burst of 4 is shown.
  3.  $D_{IN}b(n)$  = data-in for column  $b(n)$ .

**Figure 36: Nonconsecutive WRITE-to-WRITE**



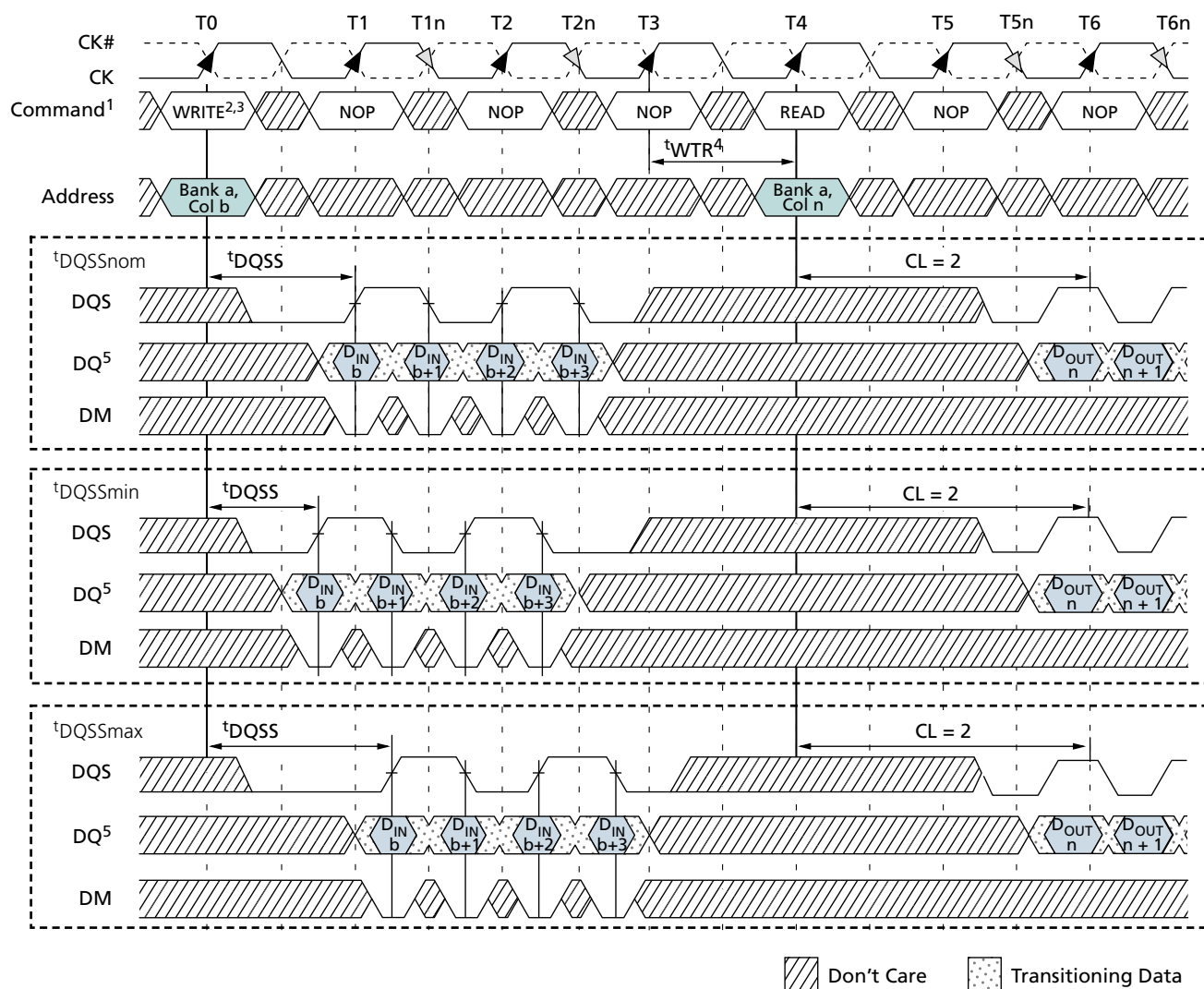
- Notes:
1. Each WRITE command can be to any bank.
  2. An uninterrupted burst of 4 is shown.
  3.  $D_{IN}b(n)$  = data-in for column  $b(n)$ .

**Figure 37: Random WRITE Cycles**



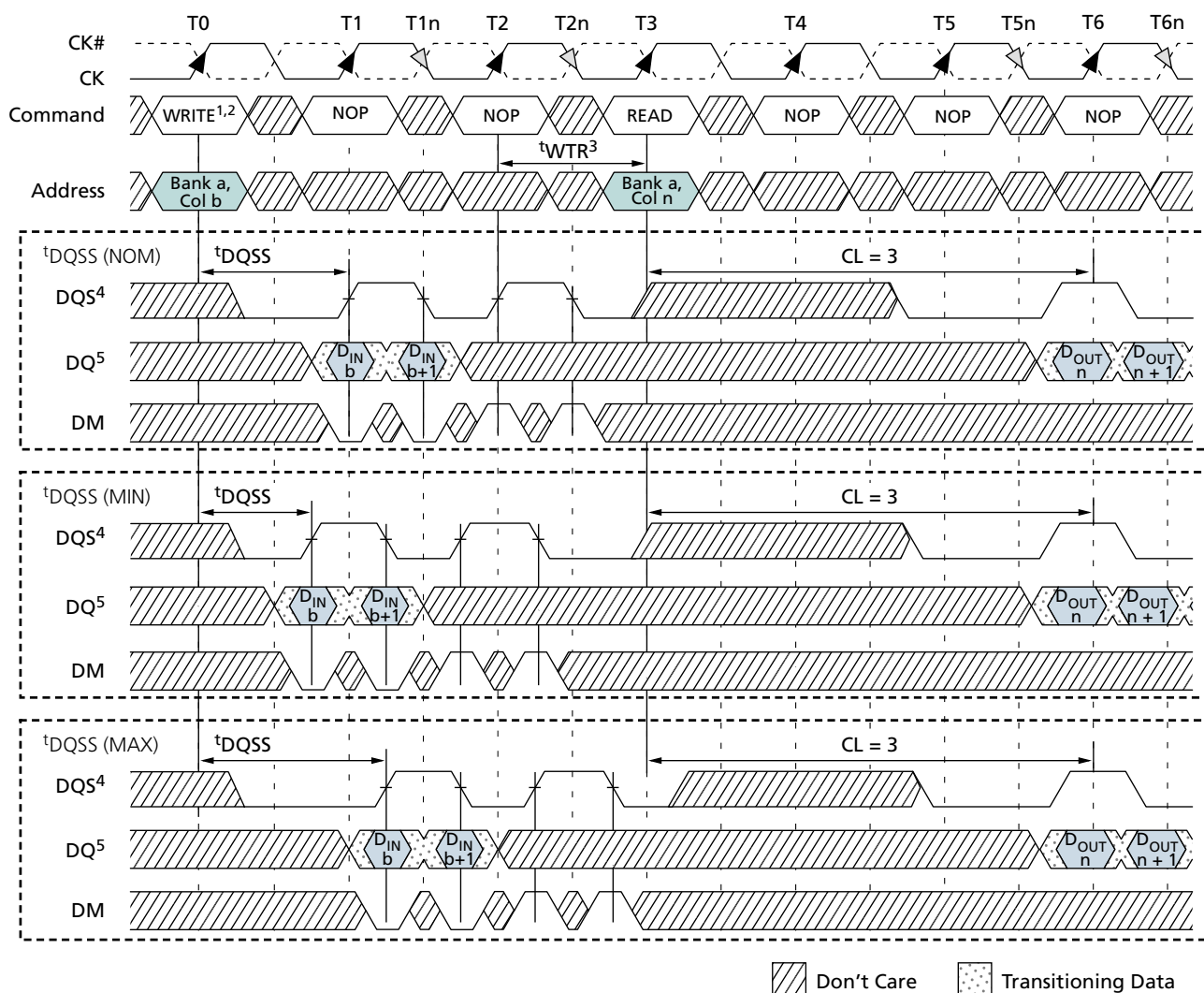
- Notes:
1. Each WRITE command can be to any bank.
  2. Programmed BL = 2, 4, 8, or 16 in cases shown.
  3.  $D_{IN}b$  (or  $x, n, a, g$ ) = data-in for column  $b$  (or  $x, n, a, g$ ).
  4.  $b'$  (or  $x, n, a, g$ ) = the next data-in following  $D_{IN}b$  ( $x, n, a, g$ ) according to the programmed burst order.

**Figure 38: WRITE-to-READ – Uninterrupting**



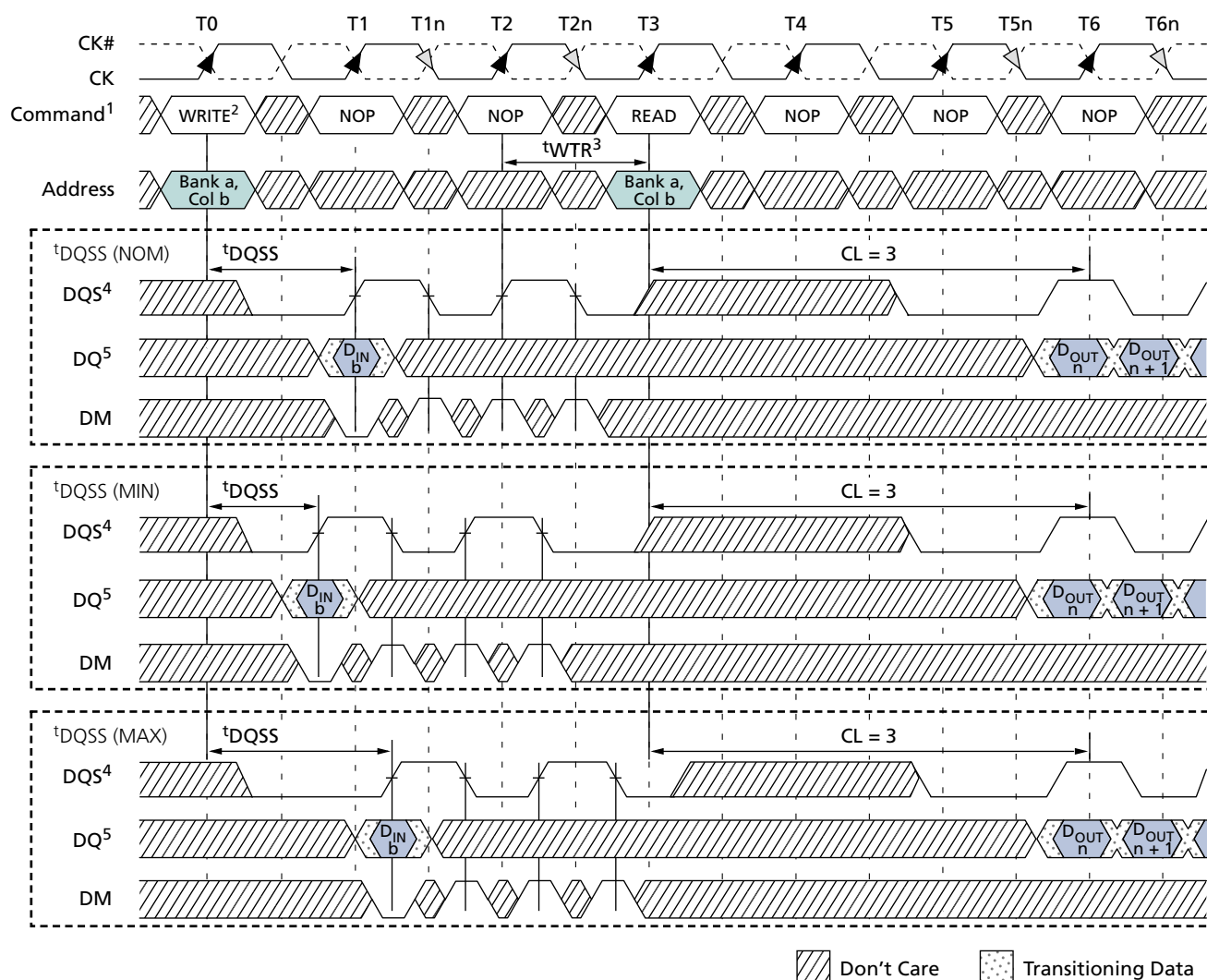
- Notes:
1. The READ and WRITE commands are to the same device. However, the READ and WRITE commands may be to different devices, in which case  $t_{WTR}$  is not required and the READ command could be applied earlier.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3. An uninterrupted burst of 4 is shown.
  4.  $t_{WTR}$  is referenced from the first positive CK edge after the last data-in pair.
  5.  $D_{IN}b$  = data-in for column  $b$ ;  $D_{OUT}n$  = data-out for column  $n$ .

**Figure 39: WRITE-to-READ – Interrupting**



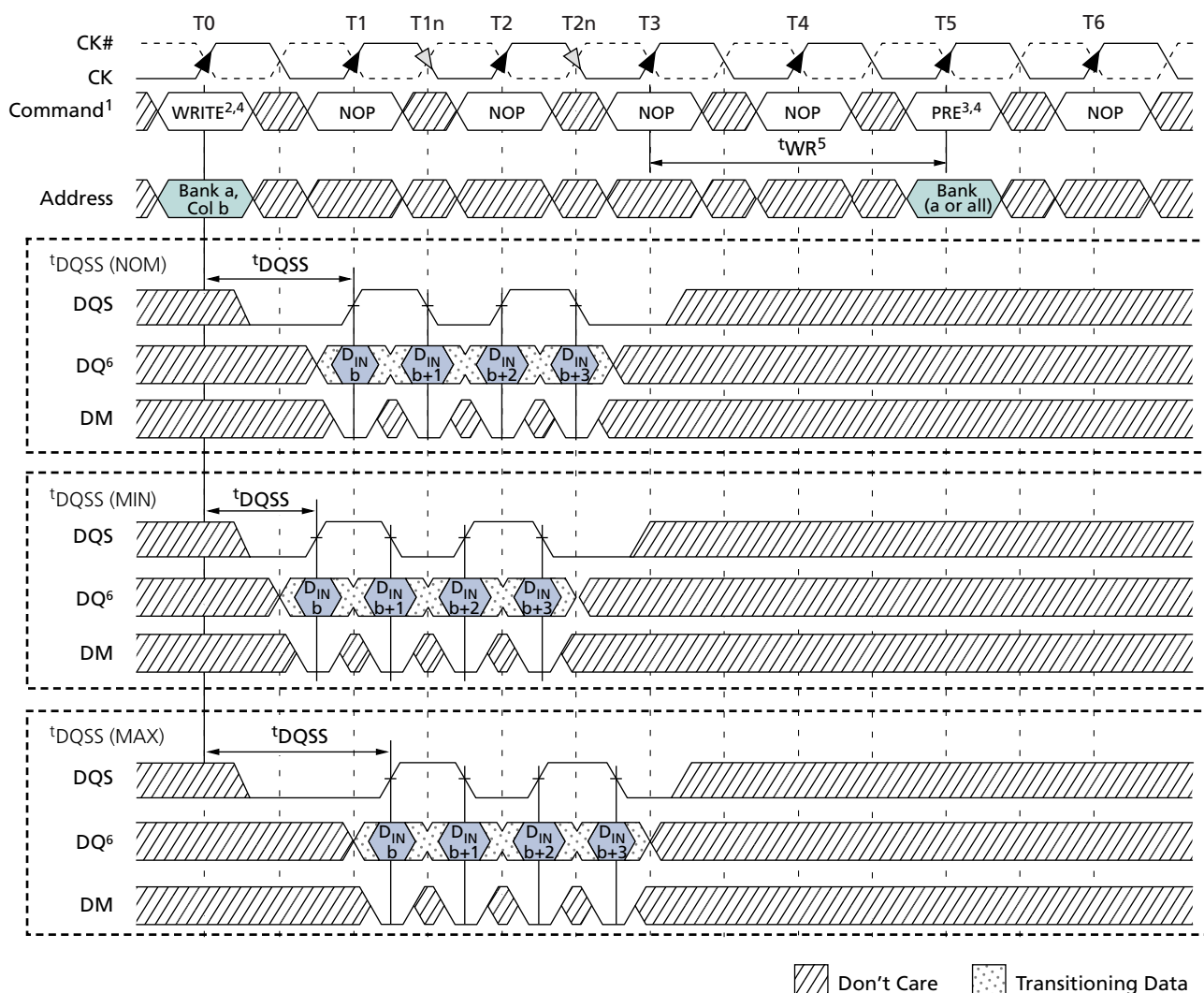
- Notes:
1. An interrupted burst of 4 is shown; 2 data elements are written.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3.  $t_{WTR}$  is referenced from the first positive CK edge after the last data-in pair.
  4. DQS is required at T2 and T2n (nominal case) to register DM.
  5.  $D_{IN}b$  = data-in for column b;  $D_{OUT}n$  = data-out for column n.

**Figure 40: WRITE-to-READ – Odd Number of Data, Interrupting**



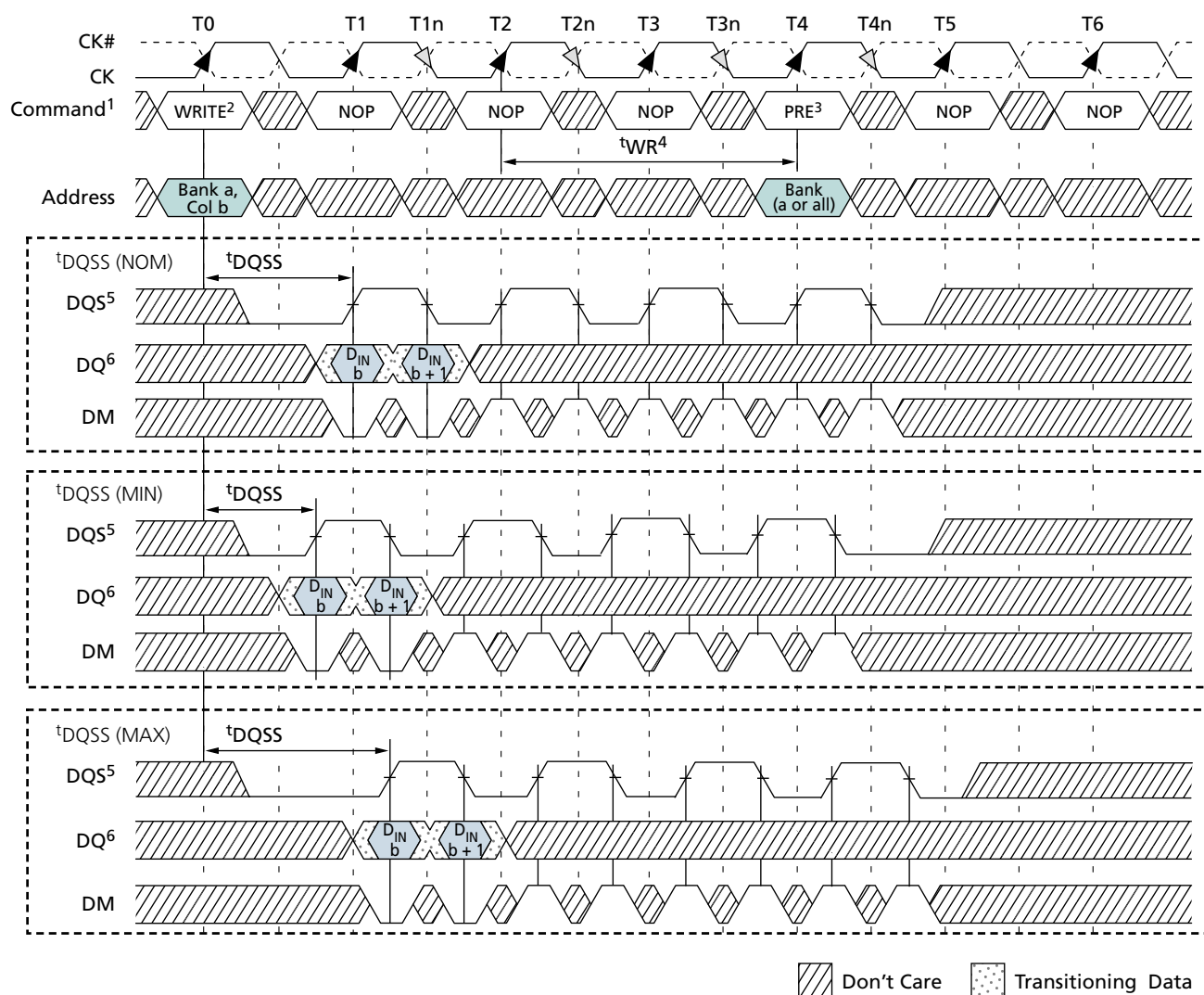
- Notes:
1. An interrupted burst of 4 is shown; 1 data element is written, 3 are masked.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3.  $t_{WTR}$  is referenced from the first positive CK edge after the last data-in pair.
  4. DQS is required at T2 and T2n (nominal case) to register DM.
  5.  $D_{INb}$  = data-in for column  $b$ ;  $D_{OUTn}$  = data-out for column  $n$ .

**Figure 41: WRITE-to-PRECHARGE – Uninterrupting**



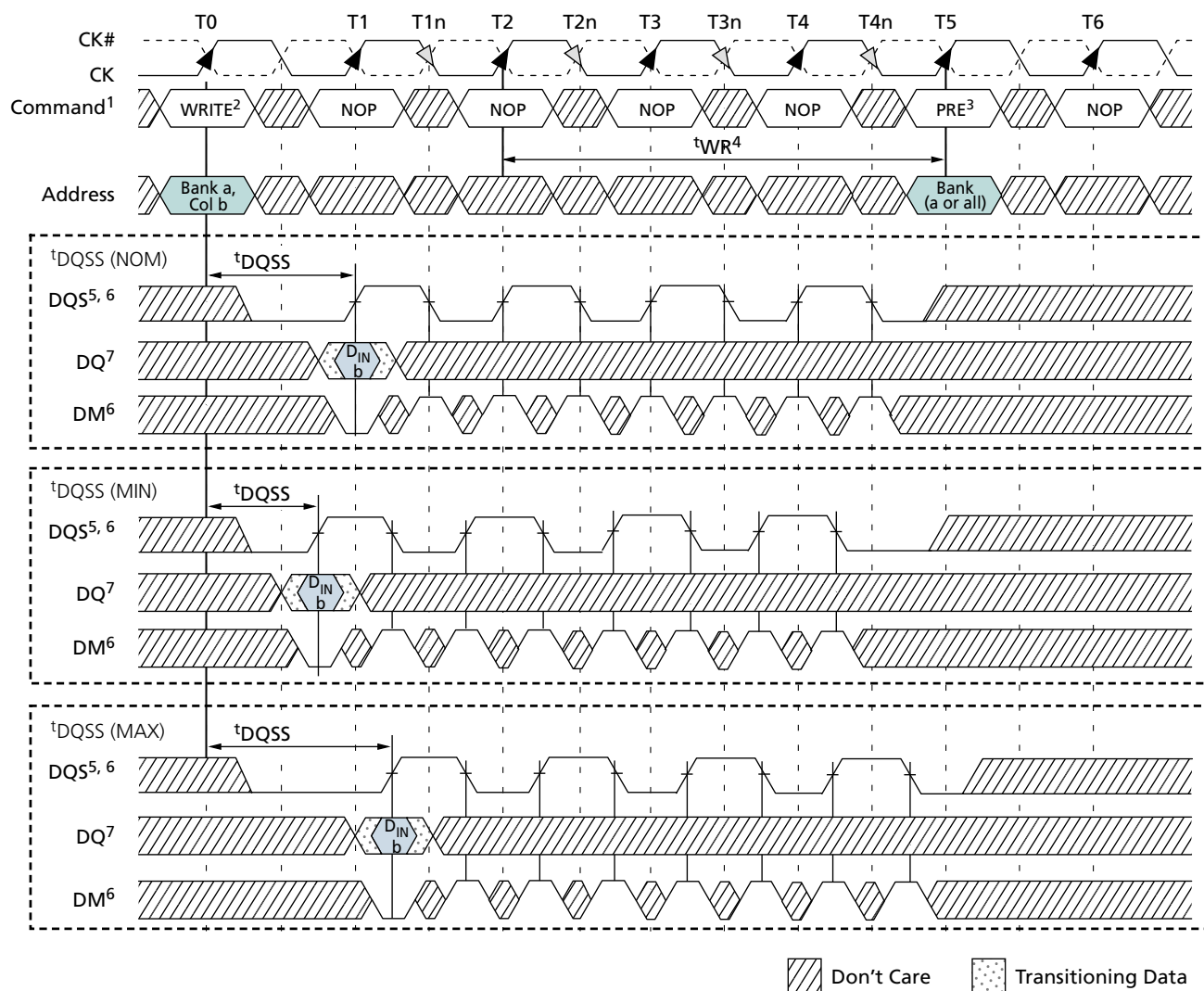
- Notes:
1. An uninterrupted burst 4 of is shown.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3. PRE = PRECHARGE.
  4. The PRECHARGE and WRITE commands are to the same device. However, the PRECHARGE and WRITE commands can be to different devices; in this case,  $t_{WR}$  is not required and the PRECHARGE command can be applied earlier.
  5.  $t_{WR}$  is referenced from the first positive CK edge after the last data-in pair.
  6.  $D_{IN}b$  = data-in for column  $b$ .

**Figure 42: WRITE-to-PRECHARGE – Interrupting**



- Notes:
1. An interrupted burst of 8 is shown; two data elements are written.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3. PRE = PRECHARGE.
  4.  $t_{WR}$  is referenced from the first positive CK edge after the last data-in pair.
  5. DQS is required at T4 and T4n to register DM.
  6.  $D_{IN}b$  = data-in for column  $b$ .

**Figure 43: WRITE-to-PRECHARGE – Odd Number of Data, Interrupting**



- Notes:
1. An interrupted burst of 8 is shown; one data element is written.
  2. A10 is LOW with the WRITE command (auto precharge is disabled).
  3. PRE = PRECHARGE.
  4. t<sub>WR</sub> is referenced from the first positive CK edge after the last data-in pair.
  5. DQS is required at T4 and T4n to register DM.
  6. If a burst of 4 is used, DQS and DM are not required at T3, T3n, T4, and T4n.
  7. D<sub>IN</sub>b = data-in for column b.



## PRECHARGE Operation

The PRECHARGE command is used to deactivate the open row in a particular bank or the open row in all banks. The bank(s) will be available for a subsequent row access some specified time ( $t_{RP}$ ) after the PRECHARGE command is issued. Input A10 determines whether one or all banks will be precharged, and in the case where only one bank is precharged (A10 = LOW), inputs BA0 and BA1 select the bank. When all banks are precharged (A10 = HIGH), inputs BA0 and BA1 are treated as “Don’t Care.” After a bank has been precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued to that bank. A PRECHARGE command will be treated as a NOP if there is no open row in that bank (idle state), or if the previously open row is already in the process of precharging.

## Auto Precharge

Auto precharge is a feature that performs the same individual bank PRECHARGE function described previously, without requiring an explicit command. This is accomplished by using A10 to enable auto precharge in conjunction with a specific READ or WRITE command. A precharge of the bank/row that is addressed with the READ or WRITE command is automatically performed upon completion of the READ or WRITE burst. Auto precharge is nonpersistent; it is either enabled or disabled for each individual READ or WRITE command.

Auto precharge ensures that the precharge is initiated at the earliest valid stage within a burst. This earliest valid stage is determined as if an explicit PRECHARGE command was issued at the earliest possible time without violating  $t_{RAS}$  (MIN), as described for each burst type in Table 17 (page 42). The READ with auto precharge enabled state or the WRITE with auto precharge enabled state can each be broken into two parts: the access period and the precharge period. The access period starts with registration of the command and ends where  $t_{RP}$  (the precharge period) begins. For READ with auto precharge, the precharge period is defined as if the same burst was executed with auto precharge disabled, followed by the earliest possible PRECHARGE command that still accesses all the data in the burst. For WRITE with auto precharge, the precharge period begins when  $t_{WR}$  ends, with  $t_{WR}$  measured as if auto precharge was disabled. In addition, during a WRITE with auto precharge, at least one clock is required during  $t_{WR}$  time. During the precharge period, the user must not issue another command to the same bank until  $t_{RP}$  is satisfied.

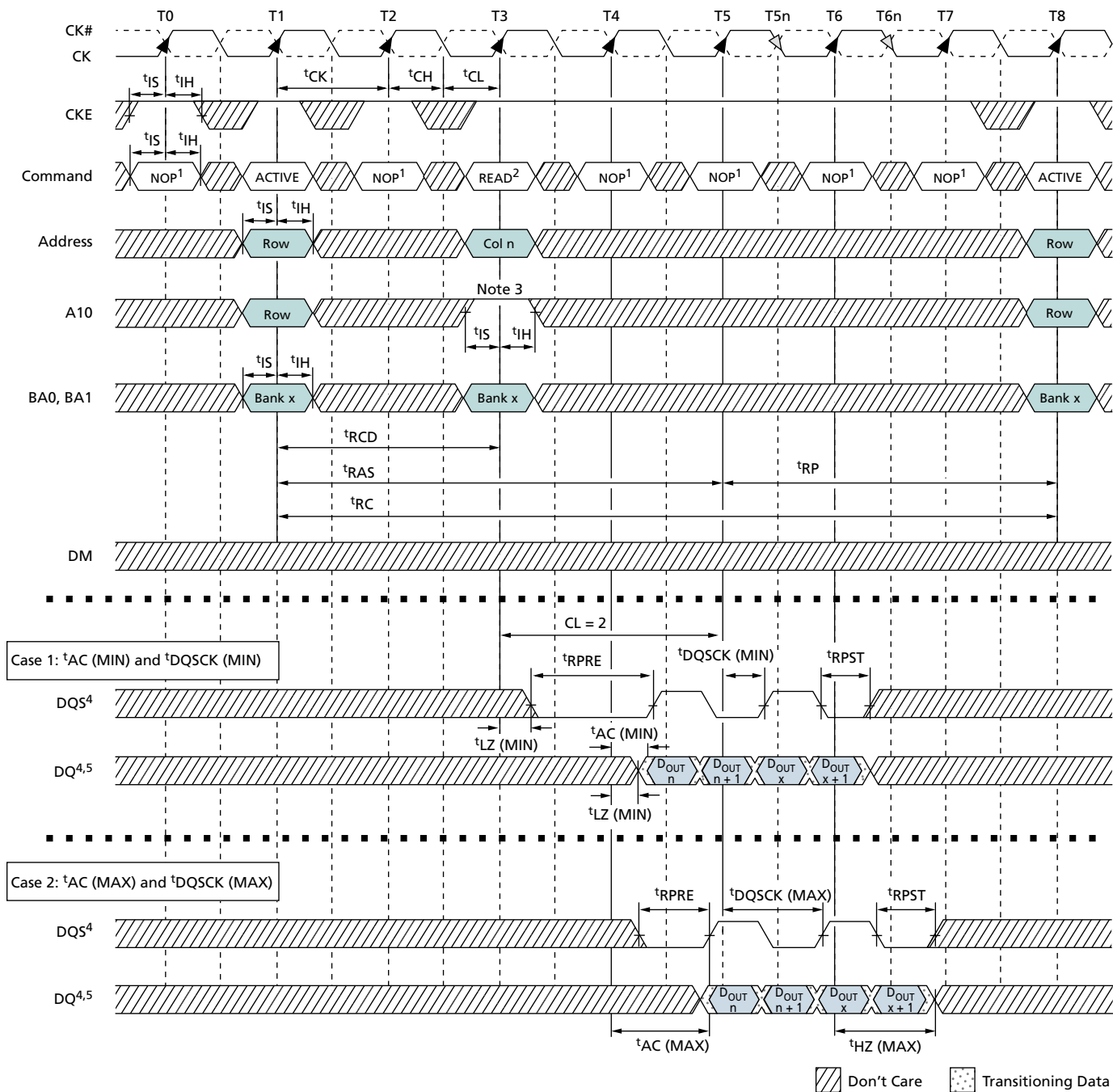
This device supports  $t_{RAS}$  lock-out. In the case of a single READ with auto precharge or single WRITE with auto precharge issued at  $t_{RCD}$  (MIN), the internal precharge will be delayed until  $t_{RAS}$  (MIN) has been satisfied.

Bank READ operations with and without auto precharge are shown in Figure 44 (page 83) and Figure 45 (page 84). Bank WRITE operations with and without auto precharge are shown in Figure 46 (page 85) and Figure 47 (page 86).

## **Concurrent Auto Precharge**

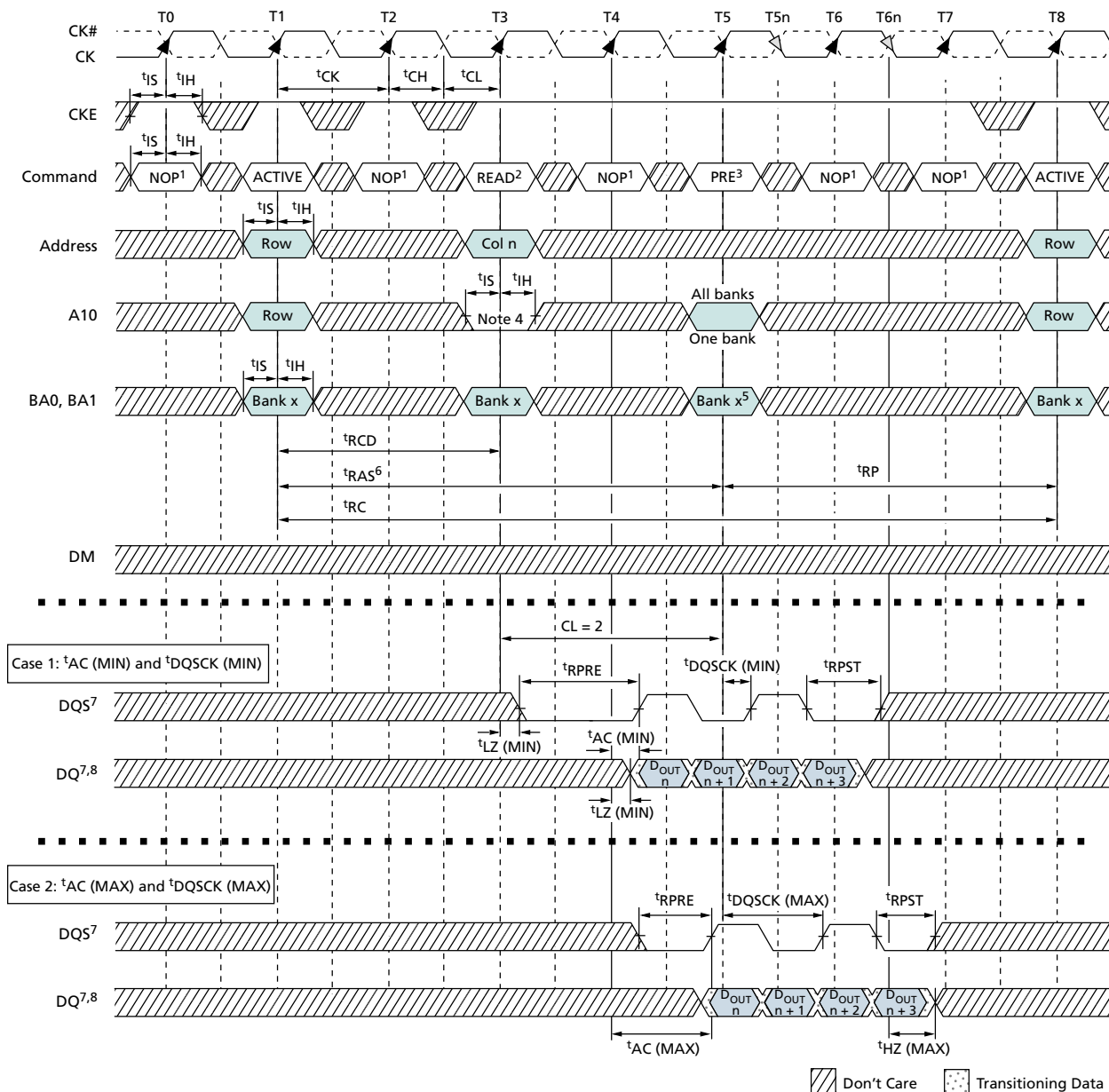
This device supports concurrent auto precharge such that when a READ with auto precharge is enabled or a WRITE with auto precharge is enabled, any command to another bank is supported, as long as that command does not interrupt the read or write data transfer already in process. This feature enables the precharge to complete in the bank in which the READ or WRITE with auto precharge was executed, without requiring an explicit PRECHARGE command, thus freeing the command bus for operations in other banks.

**Figure 44: Bank Read – With Auto Precharge**



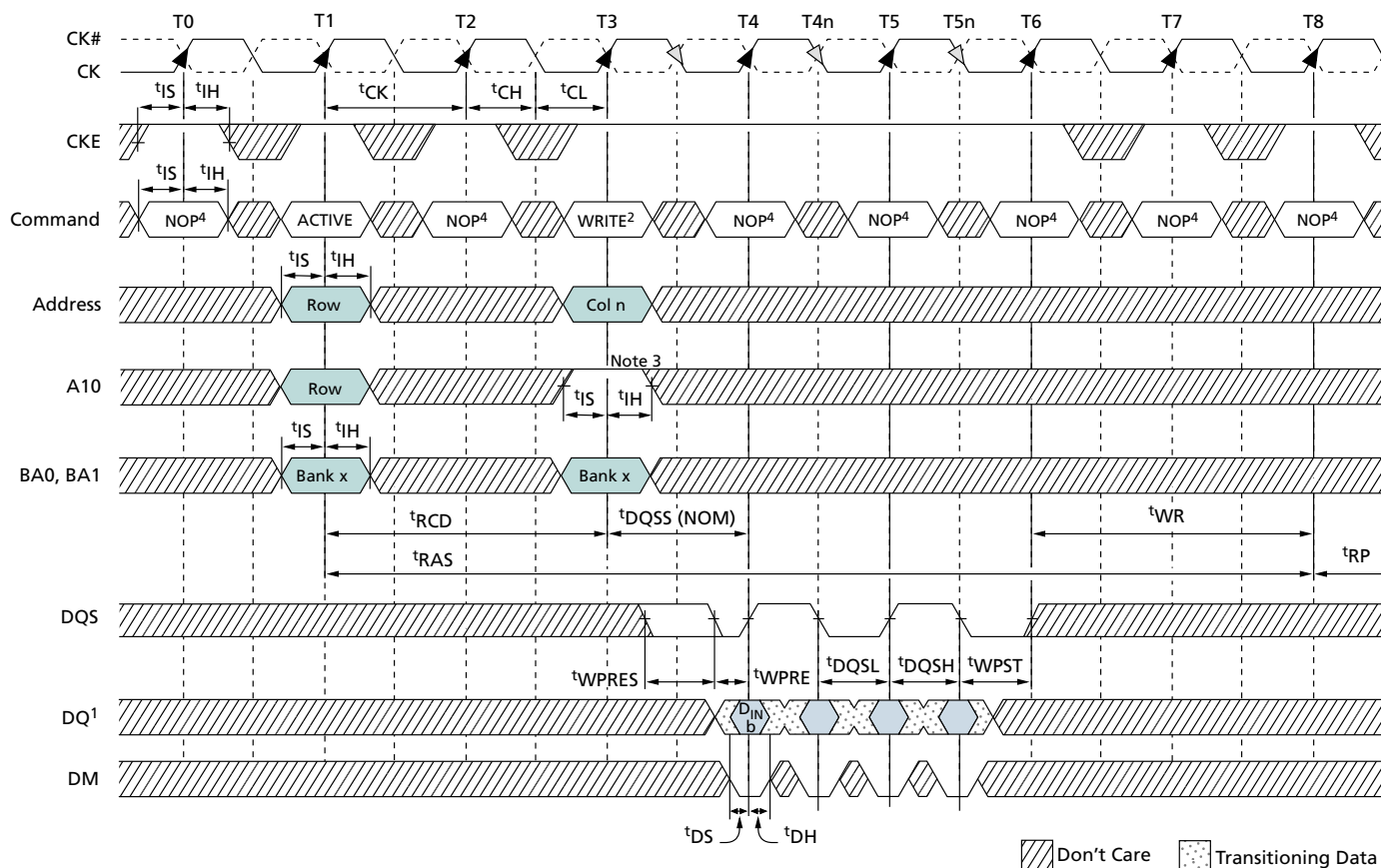
- Notes:
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
  2. BL = 4 in the case shown.
  3. Enable auto precharge.
  4. Refer to Figure 29 (page 66) and Figure 30 (page 67) for detailed DQS and DQ timing.
  5.  $D_{OUT\ n}$  = data-out from column  $n$ .

**Figure 45: Bank Read – Without Auto Precharge**



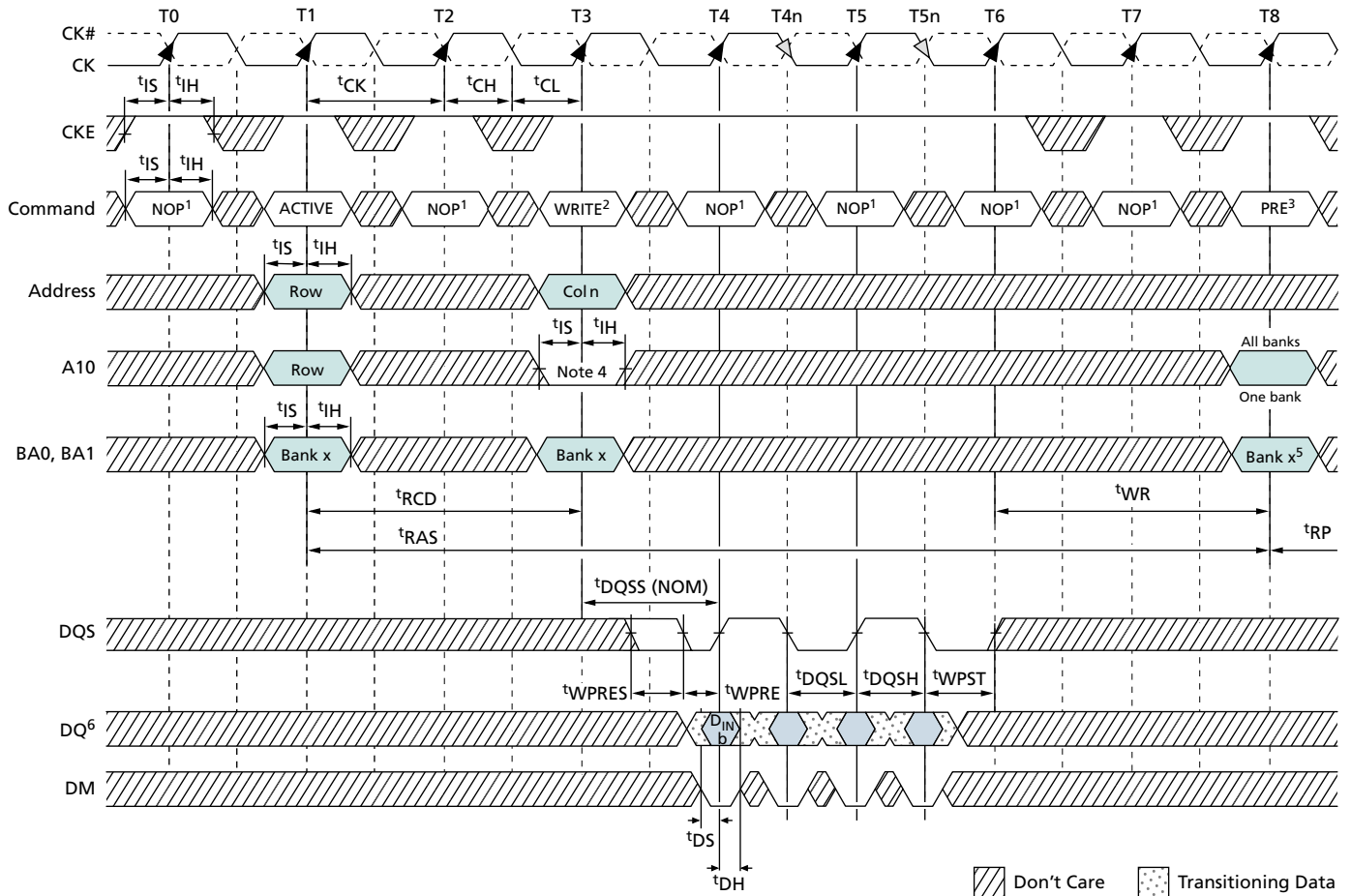
- Notes:
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
  2. BL = 4 in the case shown.
  3. PRE = PRECHARGE.
  4. Disable auto precharge.
  5. Bank x at T5 is "Don't Care" if A10 is HIGH at T5.
  6. The PRECHARGE command can only be applied at T5 if  $t_{RAS}(\text{MIN})$  is met.
  7. Refer to Figure 29 (page 66) and Figure 30 (page 67) for DQS and DQ timing details.
  8.  $D_{OUTn}$  = data out from column  $n$ .

**Figure 46: Bank Write – With Auto Precharge**



- Notes:
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
  2. BL = 4 in the case shown.
  3. Enable auto precharge.
  4.  $D_{INn}$  = data-out from column  $n$ .

**Figure 47: Bank Write – Without Auto Precharge**



- Notes:
1. NOP commands are shown for ease of illustration; other commands may be valid at these times.
  2. BL = 4 in the case shown.
  3. PRE = PRECHARGE.
  4. Disable auto precharge.
  5. Bank x at T8 is "Don't Care" if A10 is HIGH at T8.
  6.  $D_{OUTn}$  = data-out from column n.

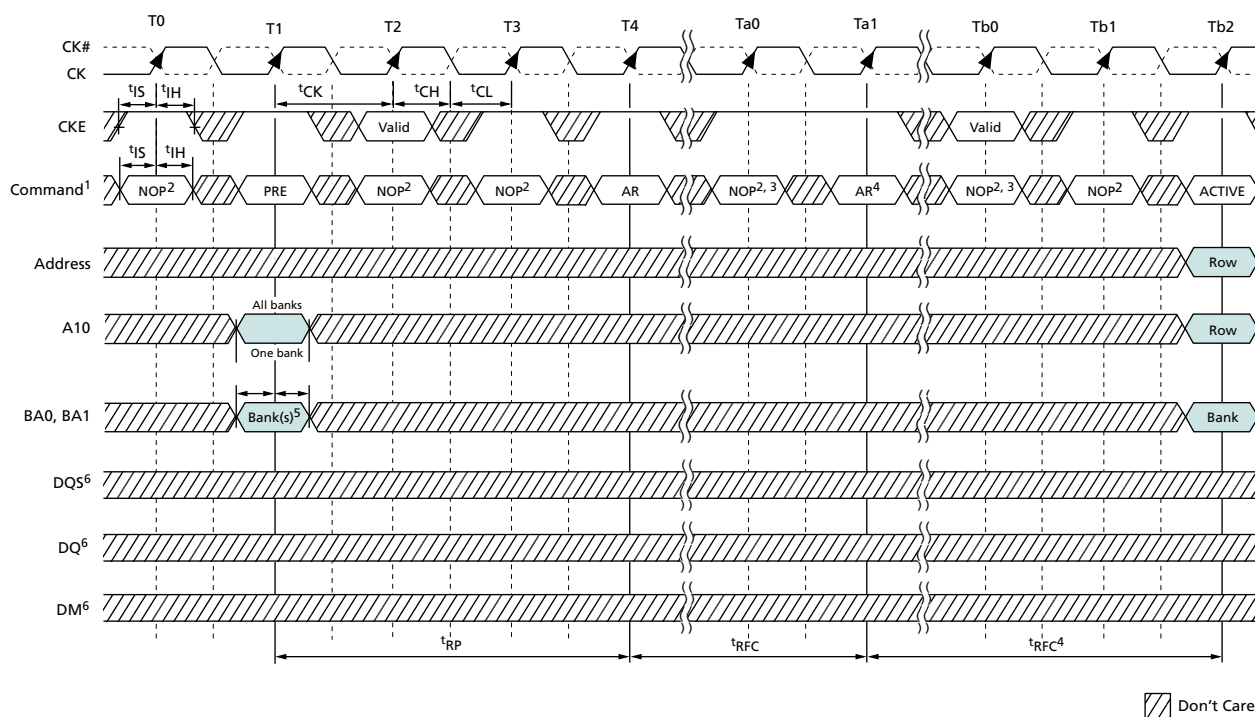
## AUTO REFRESH Operation

Auto refresh mode is used during normal operation of the device and is analogous to CAS#-BEFORE-RAS# (CBR) REFRESH in FPM/EDO DRAM. The AUTO REFRESH command is nonpersistent and must be issued each time a refresh is required.

The addressing is generated by the internal refresh controller. This makes the address bits a “Don’t Care” during an AUTO REFRESH command.

For improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided. The auto refresh period begins when the AUTO REFRESH command is registered and ends  $t_{RFC}$  later.

**Figure 48: Auto Refresh Mode**



- Notes:
1. PRE = PRECHARGE; AR = AUTO REFRESH.
  2. NOP commands are shown for ease of illustration; other commands may be valid during this time. CKE must be active during clock positive transitions.
  3. NOP or COMMAND INHIBIT are the only commands supported until after  $t_{RFC}$  time; CKE must be active during clock positive transitions.
  4. The second AUTO REFRESH is not required and is only shown as an example of two back-to-back AUTO REFRESH commands.
  5. Bank x at T1 is “Don’t Care” if A10 is HIGH at this point; A10 must be HIGH if more than one bank is active (for example, must precharge all active banks).
  6. DM, DQ, and DQS signals are all “Don’t Care”/High-Z for operations shown.

Although it is not a JEDEC requirement, CKE must be active (HIGH) during the auto refresh period to provide support for future functional features. The auto refresh period begins when the AUTO REFRESH command is registered and ends  $t_{RFC}$  later.

## **SELF REFRESH Operation**

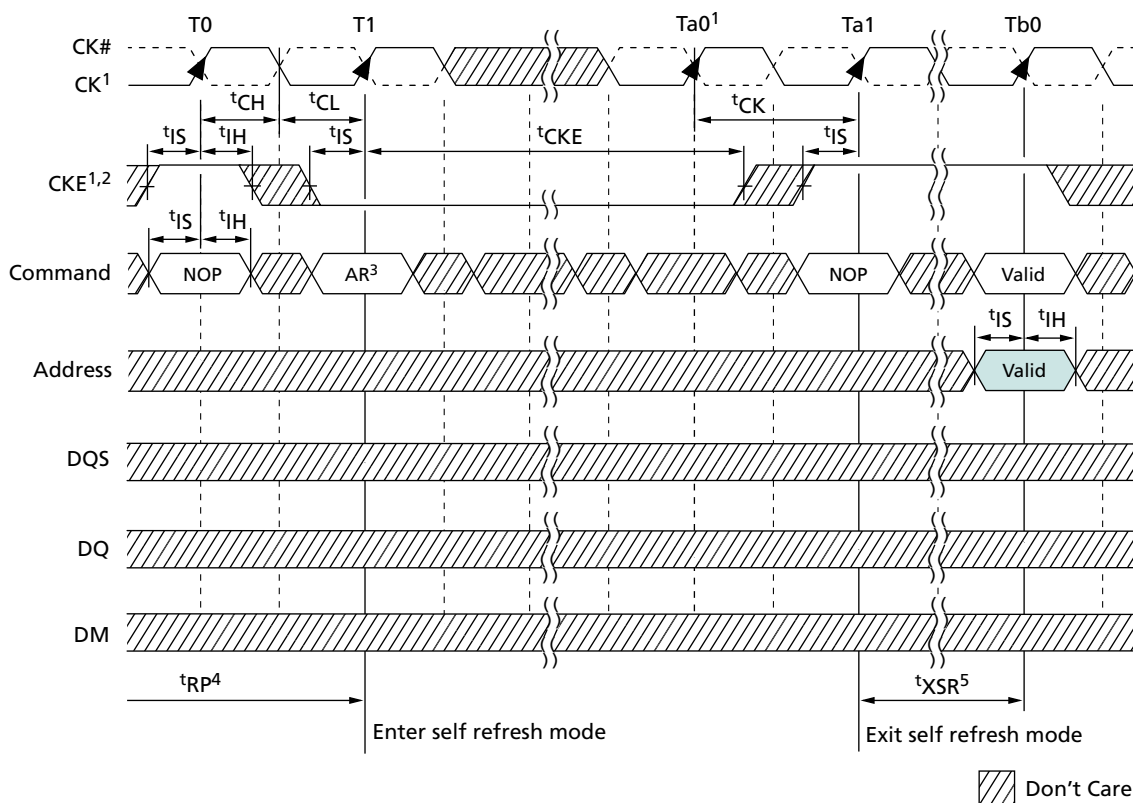
The SELF REFRESH command can be used to retain data in the device while the rest of the system is powered down. When in self refresh mode, the device retains data without external clocking. The SELF REFRESH command is initiated like an AUTO REFRESH command, except that CKE is disabled (LOW). All command and address input signals except CKE are “Don’t Care” during self refresh.

During self refresh, the device is refreshed as defined in the extended mode register. (see Partial-Array Self Refresh (page 54).) An internal temperature sensor adjusts the refresh rate to optimize device power consumption while ensuring data integrity. (See Temperature-Compensated Self Refresh (page 53).)

The procedure for exiting self refresh requires a sequence of commands. First, CK must be stable prior to CKE going HIGH. When CKE is HIGH, the device must have NOP commands issued for <sup>t</sup>XSR to complete any internal refresh already in progress.

During SELF REFRESH operation, refresh intervals are scheduled internally and may vary. These refresh intervals may differ from the specified <sup>t</sup>REFI time. For this reason, the SELF REFRESH command must not be used as a substitute for the AUTO REFRESH command.



**Figure 49: Self Refresh Mode**


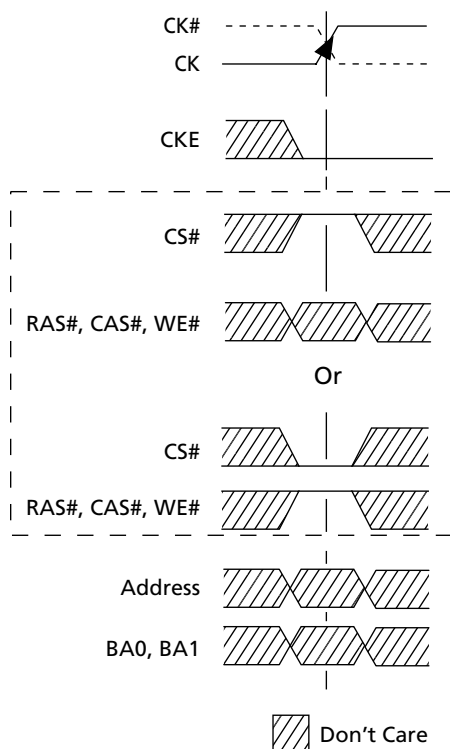
- Notes:
1. Clock must be stable, cycling within specifications by Ta0, before exiting self refresh mode.
  2. CKE must remain LOW to remain in self refresh.
  3. AR = AUTO REFRESH.
  4. Device must be in the all banks idle state prior to entering self refresh mode.
  5. Either a NOP or DESELECT command is required for t<sub>XSR</sub> time with at least two clock pulses.

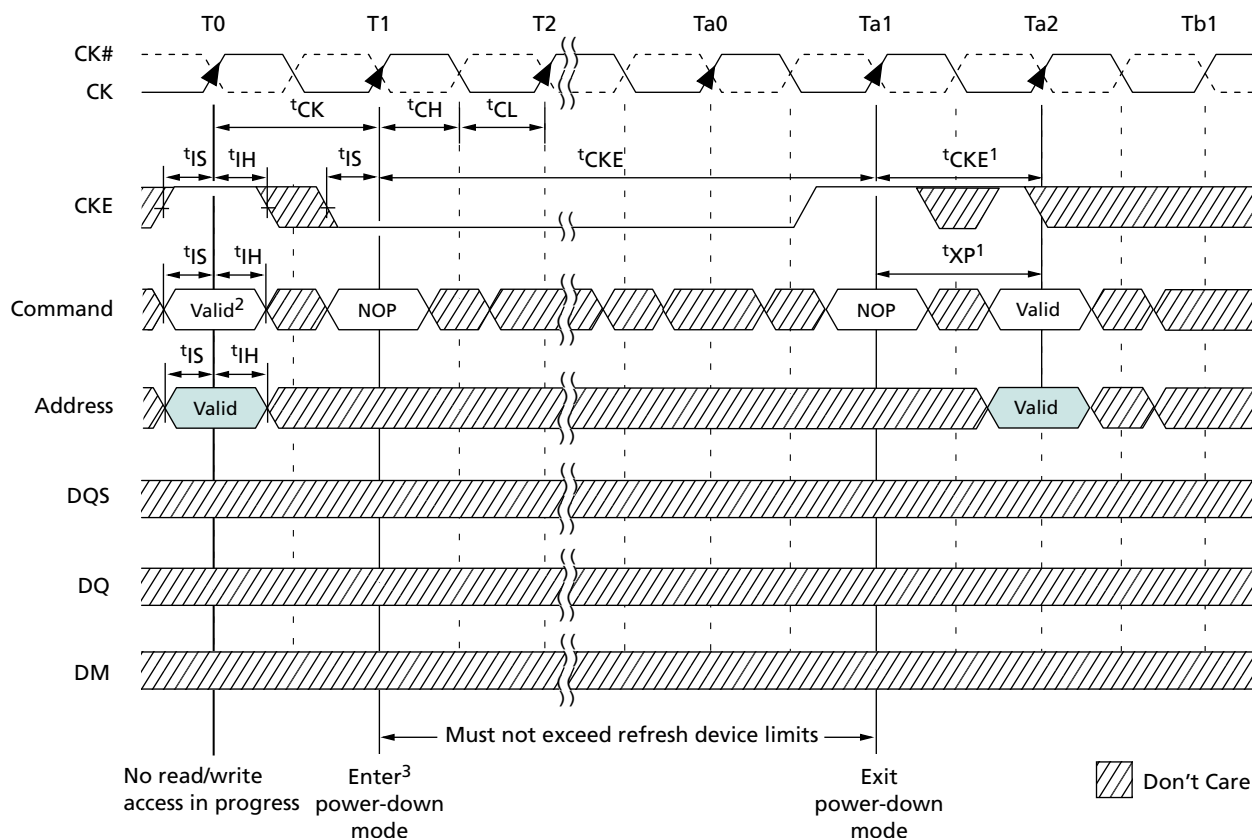
## Power-Down

Power-down is entered when CKE is registered LOW. If power-down occurs when all banks are idle, this mode is referred to as precharge power-down; if power-down occurs when there is a row active in any bank, this mode is referred to as active power-down. Entering power-down deactivates all input and output buffers, including CK and CK# and excluding CKE. Exiting power-down requires the device to be at the same voltage as when it entered power-down and received a stable clock. Note that the power-down duration is limited by the refresh requirements of the device.

When in power-down, CKE LOW must be maintained at the inputs of the device, while all other input signals are "Don't Care." The power-down state is synchronously exited when CKE is registered HIGH (in conjunction with a NOP or DESELECT command). NOP or DESELECT commands must be maintained on the command bus until t<sub>XP</sub> is satisfied. See Figure 51 (page 91) for a detailed illustration of power-down mode.

**Figure 50: Power-Down Entry (in Active or Precharge Mode)**



**Figure 51: Power-Down Mode (Active or Precharge)**


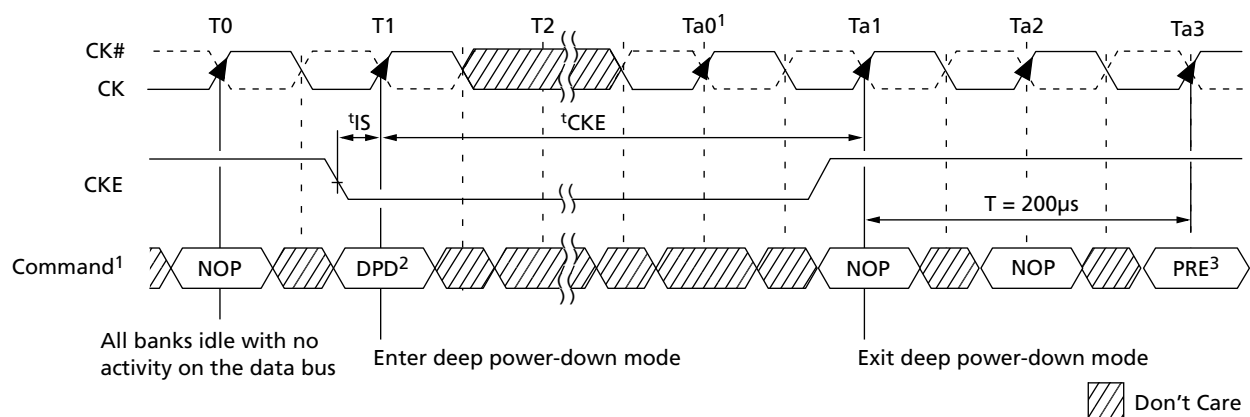
- Notes:
1.  $t_{CKE}$  applies if CKE goes LOW at Ta2 (entering power-down);  $t_{XP}$  applies if CKE remains HIGH at Ta2 (exit power-down).
  2. If this command is a PRECHARGE (or if the device is already in the idle state), then the power-down mode shown is precharge power-down. If this command is an ACTIVE (or if at least 1 row is already active), then the power-down mode shown is active power-down.
  3. No column accesses can be in progress when power-down is entered.

## Deep Power-Down

Deep power-down (DPD) is an operating mode used to achieve maximum power reduction by eliminating power to the memory array. Data will not be retained after the device enters DPD mode.

Before entering DPD mode the device must be in the all banks idle state with no activity on the data bus ( $t_{RP}$  time must be met). DPD mode is entered by holding CS# and WE# LOW with RAS# and CAS# HIGH at the rising edge of the clock while CKE is LOW. CKE must be held LOW to maintain DPD mode. The clock must be stable prior to exiting DPD mode. To exit DPD mode, assert CKE HIGH with either a NOP or DESELECT command present on the command bus. After exiting DPD mode, a full DRAM initialization sequence is required.

**Figure 52: Deep Power-Down Mode**



- Notes:
1. Clock must be stable prior to CKE going HIGH.
  2. DPD = deep power-down.
  3. Upon exit of deep power-down mode, a full DRAM initialization sequence is required.

## Clock Change Frequency

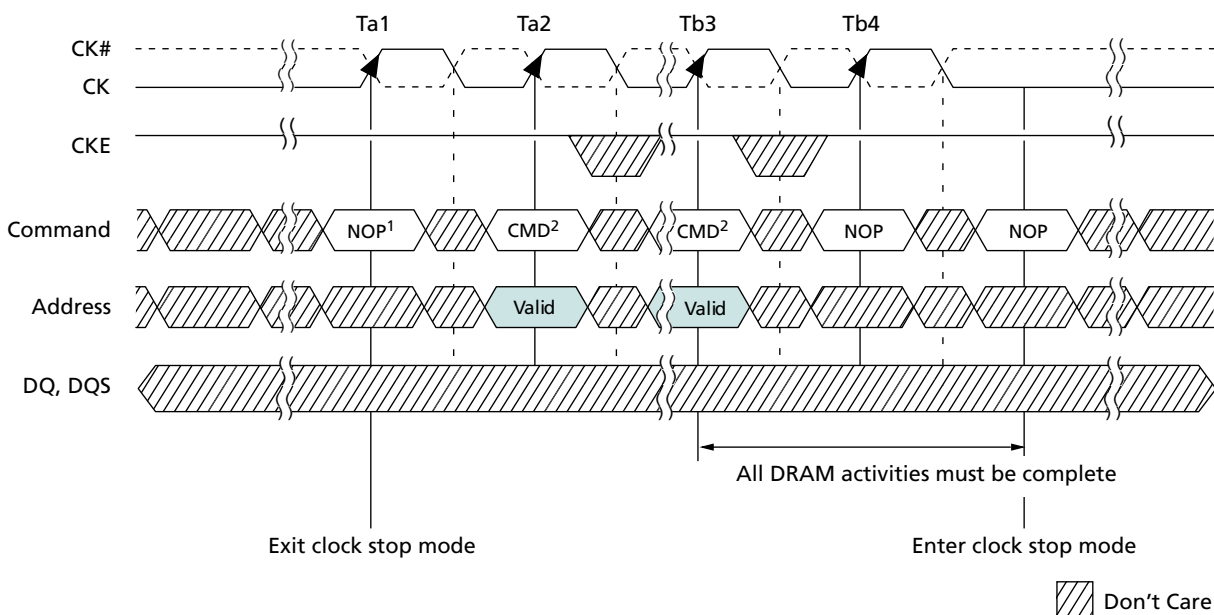
One method of controlling the power efficiency in applications is to throttle the clock that controls the device. The clock can be controlled by changing the clock frequency or stopping the clock.

The device enables the clock to change frequency during operation only if all timing parameters are met and all refresh requirements are satisfied.

The clock can be stopped altogether if there are no DRAM operations in progress that would be affected by this change. Any DRAM operation already in process must be completed before entering clock stop mode; this includes the following timings:  $t_{RCD}$ ,  $t_{RP}$ ,  $t_{RFC}$ ,  $t_{MRD}$ ,  $t_{WR}$ , and  $t_{RPST}$ . In addition, any READ or WRITE burst in progress must be complete. (See READ Operation and WRITE Operation.)

CKE must be held HIGH with CK = LOW and CK# = HIGH for the full duration of the clock stop mode. One clock cycle and at least one NOP or DESELECT is required after the clock is restarted before a valid command can be issued.

**Figure 53: Clock Stop Mode**



- Notes:
1. Prior to Ta1, the device is in clock stop mode. To exit, at least one NOP is required before issuing any valid command.
  2. Any valid command is supported; device is not in clock suspend mode.

## Revision History

### Rev. F, Production – 02/10

- Updated document format

### Rev. E, Production – 8/09

- Updated 60-ball VGBGA ball assignments figure (ball H3)

### Rev. D, Production – 04/09

- Removed ball assignment columns from ball description table per new style standard
- Electrical Specifications and Recommended AC Operating Conditions: Added new note 8

### Rev. C, Production – 10/08

- Updated 60-ball VGBGA ball assignments figure.
- Updated 90-ball VGBGA ball assignments figure.
- Reordered ball numbers in ball description table
- Updated 90-ball VGBGA package drawing
- Updated Typical Self Refresh Current vs. Temperature figure
- Target Output Drive Characteristics (Full Strength): Updated pull-down current min and pull-up current min values for 1.80V and 1.90V
- Target Output Drive Characteristics (Three-Quarter Strength): Updated pull-down current min and pull-up current min values for 1.80V and 1.90V
- Target Output Drive Characteristics (One-half Strength): Updated pull-down current min and pull-up current min values for 1.80V and 1.90V
- Changed status to production

### Rev. B, Preliminary – 05/08

- Key Timing Parameters table and Electrical Characteristics and Recommended Operating Conditions: Updated -6 <sup>t</sup>AC/<sup>t</sup>DQSCK from 5.5ns to 5.0ns (max); updated -75 <sup>t</sup>RC from 75ns to 67.5ns (<sup>t</sup>RC = <sup>t</sup>RAS + <sup>t</sup>RP). Deleted former note 21.
- I<sub>DD</sub> Table: Updated I<sub>DD5</sub> specification to tRFCmin rather than 138ns
- Changed to Preliminary status

### Rev. A, Advance – 04/08

- Initial release.

8000 S. Federal Way, P.O. Box 6, Boise, ID 83707-0006, Tel: 208-368-3900  
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This data sheet contains minimum and maximum limits specified over the power supply and temperature range set forth herein. Although considered final, these specifications are subject to change, as further product development and data characterization sometimes occur.