MIL-PRF-38534 CERTIFIED



NEGATIVE, 3 AMP, LOW DROPOUT VOLTAGE REGULATOR

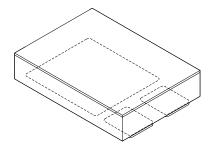
5330 SERIES

4707 Dey Road Liverpool, N.Y. 13088

(315) 701-6751

FEATURES:

- -5V, -5.2V, -10V, -12V and -15V Standard Versions
- · Low Dropout Voltage
- Output Current to 3 Amps
- Output Voltage Internally Set to $\pm 1\%$ MAX.
- . Internal Short Circuit Current Limit
- . Internal Thermal Overload Protection
- . Hermetic Surface Mount Package
- Alternate Output Voltages Available

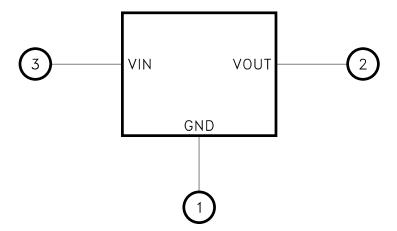


DESCRIPTION:

The MSK 5330 is a fixed, negative voltage regulator which offers low dropout and output voltage accuracy to $\pm 1\%$ maximum. The low θ jc combined with low dropout allows increased output current and excellent device efficiency. The MSK 5330 offers both internal current limit and thermal overload protection. The device is packaged in a space efficient 3 pin power surface mount ceramic package.

EQUIVALENT SCHEMATIC

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TYPICAL APPLICATIONS

- · High Efficiency Linear Regulators
- · Constant Voltage/Current Regulators
- · System Power Supplies
- · Switching Power Supply Post Regulators

PIN-OUT INFORMATION

- 1 GND
- 2 VOUT
- 3 VIN

ABSOLUTE MAXIMUM RATINGS



-VIN	Input Voltage (WRT Vouт)30V	Tst	Storage Temperature Range65°C to +150°C
P_D	Power Dissipation Internally Limited	T_LD	Lead Temperature Range 300°C
louт	Output Current3A		(10 Seconds)
TJ	Junction Temperature + 175°C	Tc	Case Operating Temperature
			MSK 5330H/E55°C to +125°C
			MSK 533040°C to +85°C

ELECTRICAL SPECIFICATIONS

Parameter	Test Conditions ③	Group A MSK 5330H/E		MSK 5330			Units		
T didinote:	rest conditions	Subgroup	Min.	Typ.	Max.	Min.	Typ.	Max.	Units
Output Voltage Tolerance	IOUT = 10mA; VIN = VOUT + 3V	1	-	0.1	1.0	-	0.1	2.0	%
Output voitage rolerance		2,3	-	0.1	2.0	-	-	-	%
Dropout Voltage ②	$0A \le Iout \le 3A$; $\Delta Vout = 1\%$	1	-	0.8	1.2	-	0.8	1.3	V
Lood Domilation	100mA≤lo∪т≤3A	1	-	0.2	1.0	-	0.2	2.0	%
Load Regulation	VIN = VOUT + 3V	2,3	-	0.3	2.0	-	-	-	%
Line Demulation	Iout = 10mA	1	-	0.1	0.5	-	0.1	0.6	%
Line Regulation	$(Vout + 3V) \le Vin \le (Vout + 15V)$	2,3	-	0.2	0.75	-	-	-	%
Quiescent Current	VIN = VOUT + 3V; IOUT = OA	1,2,3	-	4.5	10	-	4.5	10	mA
Short Circuit Current ②	VIN = VOUT + 5V	-	3.3	3.6	-	3.0	3.6	-	Α
Ripple Rejection ②	IOUT = 3A; COUT = 25μ F; f = 120Hz	-	60	75	-	60	75	-	dB
Thermal Resistance ② JUNCTION TO CASE @ 125°C		-	-	4.2	4.6	-	4.2	5.0	°C/W
Thermal Shutdown TJ		-	-	150	-	-	150	-	°C

PART ⑦ NUMBER	OUTPUT VOLTAGE
MSK5330-5.0	-5.0V
MSK5330-5.2	-5.2V
MSK5330-10	-10.0V
MSK5330-12	-12.0V
MSK5330-15	-15.0V

NOTES:

- ① Output is decoupled to ground using 33µF minimum, low ESR capacitors unless otherwise specified. ② Guaranteed by design but not tested. Typical parameters are representative of actual device
- performance but are for reference only.
- ③ All output parameters are tested using a low duty cycle pulse to maintain T_J = T_C.
 ④ Industrial grade and "E" suffix devices shall be tested to subgroup 1 unless otherwise specified.
 ⑤ Military grade devices ("H" suffix) shall be 100% tested to subgroups 1,2 and 3.
 ⑥ Subgroup 1 T_A = T_C = +25°C
- - $T_A = T_C = +125 \, ^{\circ}C$
 - 3 $T_A = T_C = -55 \,^{\circ}C$

- ⑦ Please consult the factory if alternate output voltages are required.
 ⑧ Input voltage (VIN = VOUT + a specified voltage) is implied to be more negative than VOUT.
 ⑨ Continuous operation at or above absolute maximum ratings may adversely effect the device performance and/or life cycle

BYPASS CAPACITORS

For most applications a 33uF minimum, low ESR (0.5-2 ohm) tantalum capacitor should be attached as close to the regulator's output as possible. This will effectively lower the regulator's output impedance, increase transient response and eliminate any oscillations that are normally associated with low dropout regulators. Additional bypass capacitors can be used at the remote load locations to further improve regulation. These can be either of the tantalum or the electrolytic variety. Unless the regulator is located very close to the power supply filter capacitor(s), a 4.7uF minimum low ESR (0.5-2 ohm) tantalum capacitor should also be added to the regulator's input. An electrolytic may also be substituted if desired. When substituting electrolytic in place of tantalum capacitors, a good rule of thumb to follow is to increase the size of the electrolytic by a factor of 10 over the tantalum value.

LOAD REGULATION

For best results the ground pin should be connected directly to the load as shown below, this effectively reduces the ground loop effect and eliminates excessive voltage drop in the sense leg. It is also important to keep the output connection between the regulator and the load as short as possible since this directly affects the load regulation. For example, if 20 gauge wire were used which has a resistance of about .008 ohms per foot, this would result in a drop of 8mV/ft at 1Amp of load current. It is also important to follow the capacitor selection guidelines to achieve best performance. Refer to Figure 1 for connection diagram.

MSK 5330 TYPICAL APPLICATION:

Low Dropout Negative Power Supply

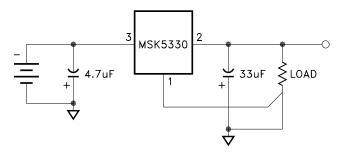


FIGURE 1

OVERLOAD SHUTDOWN

The MSK 5330 features both power and thermal overload protection. When the maximum power dissipation is not exceeded, the regulator will current limit slightly above its 3 amp rating. As the Vin-Vout voltage increases, however, shutdown occurs in relation to the maximum power dissipation curve. If the device heats enough to exceed its rated die junction temperature due to excessive ambient temperature, improper heat sinking etc., the regulator will shutdown until an appropriate junction temperature is maintained. It should also be noted that in the case of an extreme overload, such as a sustained direct short, the device may not be able to recover. In these instances, the device must be shut off and power reapplied to eliminate the shutdown condition.

HEAT SINKING

To determine if a heat sink is required for your application and if so, what type, refer to the thermal model and governing equation below.

Governing Equation: $Tj = Pd x (R_{\theta}jc + R_{\theta}cs + R_{\theta}sa) + Ta$

WHERE

Tj = Junction Temperature

Pd = Total Power Dissipation

Rejc = Junction to Case Thermal Resistance

Recs = Case to Heat Sink Thermal Resistance

Resa = Heat Sink to Ambient Thermal Resistance

Tc = Case Temperature

Ta = Ambient Temperature

Ts = Heat Sink Temperature

EXAMPLE:

This example demonstrates an analysis where the regulator is at one-half of its maximum rated power dissipation, which occurs when the output current is at 1.5 amps.

Conditions for MSK 5330-5:

$$Vin = -7.0V$$
; $Iout = -1.5A$

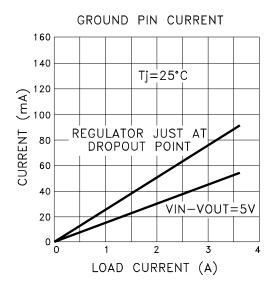
- 1.) Assume 45° heat spreading model.
- 2.) Find regulator power dissipation:

- 3.) For conservative design, set Tj = +125 °C Max.
- 4.) For this example, worst case $Ta = +90^{\circ}C$.
- 5.) $R_{\theta jc} = 4.6^{\circ} C/W$ from the Electrical Specification Table.
- 6.) $R_{\theta}cs = 0.15^{\circ}C/W$ for most thermal greases.
- 7.) Rearrange governing equation to solve for $R\theta sa$

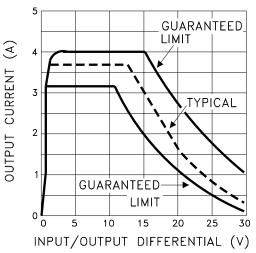
$$\begin{array}{lll} R_\theta s a = & ((Tj - Ta)/Pd) - (R_\theta j c) - (R_\theta c s) \\ & = & (125\,^\circ C - 90\,^\circ C)/3.0W - 4.6\,^\circ C/W - 0.15\,^\circ C/W \\ & = & 6.9\,^\circ C/W \end{array}$$

In this case the result is 6.9° C/W. Therefore, a heat sink with a thermal resistance of no more than 6.9° C/W must be used in this application to maintain the regulator junction temperature under 125° C.

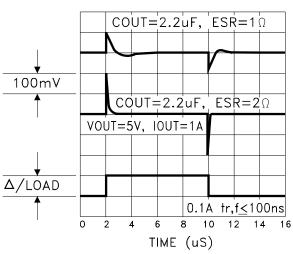
TYPICAL PERFORMANCE CURVES



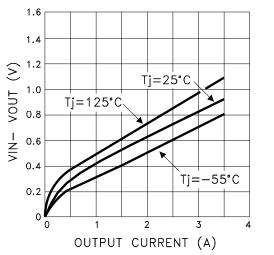
INTERNAL CURRENT LIMIT vs. VIN-VOUT



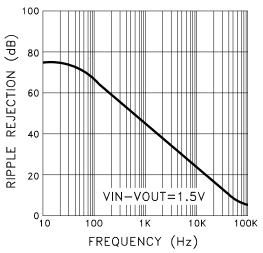
LOAD TRANSIENT RESPONSE



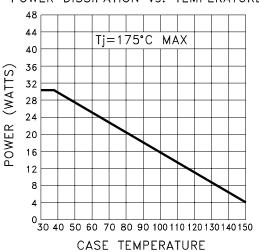
DROPOUT VOLTAGE vs. OUTPUT CURRENT

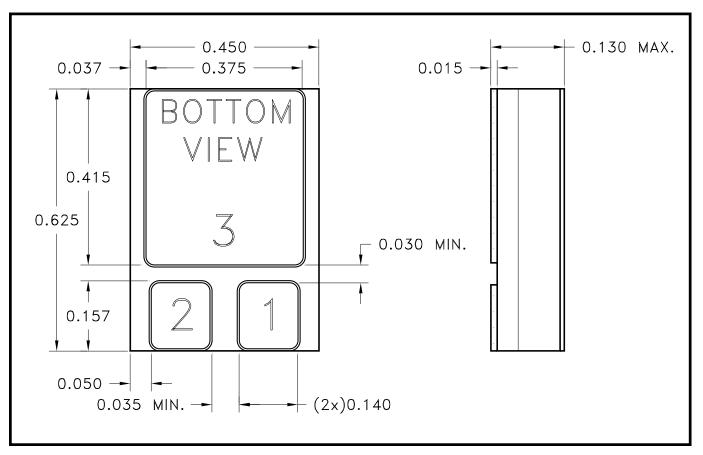


RIPPLE REJECTION vs. FREQUENCY



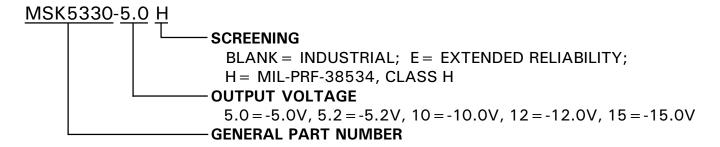
POWER DISSIPATION vs. TEMPERATURE





NOTE: ALL DIMENSIONS ARE ±0.010 INCHES UNLESS OTHERWISE LABELED.

ORDERING INFORMATION



The above example is a -5.0V, Military regulator.

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