

#### High Power Factor AC/DC LED Driver with Thermal Fold-back

### **Features**

- Constant current LED driver
- Voltage step down converter
- Built-in active power factor correction
- Quasi-resonant switching
- Thermal Fold-back Function
- Full protection: Thermal shutdown
  - V<sub>DD</sub> over voltage protection Under voltage lock-out (UVLO) LED short- circuit
- Available in MSOP8 package
- Package MSL Level : 3

## **Product Description**

Mini Small Outline Package

MBI6911 is a 85~135V input AC-DC converter designed to deliver constant current with thermal fold-back Feature. The built-in active power factor correction circuit maintains high power factor over a wide input voltage range. MBI6911 is optimized for interior power applications, and the efficiency is enhanced with quasi-resonant switching technique. MBI6911 is also featured with under voltage lock-out (UVLO), over temperature protection (OTP), over-voltage protection and LED short-circuit protection to protect the converter from being damaged accidentally. A thermal resistor can be used with an external voltage divider circuit to adjust the output current level automatically. If this feature is used to monitor an output current related heat source, a temperature balance condition should be expected.

### **Applications**

- T8 light tube LED alternative solutions
- E26/E27 light bulb LED alternative solutions
- GU10 light LED alternative solutions
- Other Interior LED power module applications

# **Typical Application Circuit**



Fig.1 Typical application circuit of MBI6911



Fig.2 Functional diagram of MBI6911

# **Pin Configuration**



Fig. 3 Pin configuration of MBI6911

# **Pin Description**

No.	Name	Description	
1	VSS	Floating ground (IC reference)	
2	COMP	Terminal to connect a compensator (EA's output)	
3	CS	Sense inductor current and output current.	
4	GAIN	Set current decaying slope of thermal fold-back	
5	VTH	Set operating point of thermal fold-back	
6	BIAS	Voltage reference of thermal fold-back	
7	GATE	Drive the gate of the external MOSFET	
8	VDD	Supply voltage (With OVP=40V)	

## **Absolute Maximum Ratings**

NOTE: Operations above the maximum ratings may cause device failure or reduce device reliability.

Characteristics	Symbol	Min	Max	Unit
Supply voltage	V <sub>DD</sub>	-0.3	44	V
Output Voltage for COMP pin	V <sub>COMP</sub>	-0.3	7	V
Output voltage for GATE pin	V <sub>GATE</sub>	-0.3	16	V
Output voltage for BIAS reference pin	V <sub>BIAS</sub>	-0.3	7	V
Input voltage for CS pin	V <sub>CS</sub>	-0.3	7	V
Input voltage for GAIN pin.	V <sub>GAIN</sub>	-0.3	7	V
Input voltage for VTH pin.	V <sub>TH</sub>	-0.3	7	V
Storage temperature	T <sub>STG</sub>	-55	150	С°
Operating ambient temperature	T <sub>OPA</sub>	-40	85	С°
Junction temperature	T <sub>JC</sub>	-40	125	С°
Power Dissipation	P <sub>DISSIP</sub>	-	3.3	W
Thermal resistance	T <sub>HR</sub>	-	40	°C /W

# **Electrical Characteristics**

Test condition:  $V_{DD}$ =24V,  $C_{IN}$ =10µF, and  $T_A$ =25°C unless otherwise specified.

Characteristics	Symbol	Conditions	Min	Тур	Max	Unit
Supply voltage						
Input Voltage	V <sub>DD</sub>		18	. 4	37	V 0/
Line regulation		RCS Variation $\pm 1\%$	-	±1	- ⊥1	% 0/_
				-	±1 ±1	-70 0/2
Supply voltage		Avo-11370 vo			± 1	70
		Recommended operating voltage				
voltage-AUX VDD	V <sub>DD-AUX</sub>	not a testable spec.	-	30	-	V
Turn-on threshold Voltage	V <sub>DD-ON</sub>	@ Start-up	14.4	16	17.6	V
Turn-off threshold Voltage	V <sub>DD-OFF</sub>	@ UVLO	6.3	7	7.7	V
V <sub>DD</sub> over voltage protection	V <sub>DD-OVP</sub>		37	30	43	V
UVLO Hysteresis	V <sub>DD-HYS</sub>		8.1	9	9.9	V
Operating current	I <sub>DD-OP</sub>	$V_{DD}$ =24V, C <sub>L</sub> =1nF, No switching (Protection with dummy load)	1.5		2.3	mA
Starting current		$v_{DD}$ =24V, C <sub>L</sub> =IIIF, ISW=30KHZ	0.0	- 20	1.0	
Error amplifier	IDD-ST		-	20	40	uA
	V		20	3.2	35	V
Current sense reference voltage			0.19	0.2	0.21	V
Gate driver	V REF	<u> </u>	0.10	0.2	0.21	v
	Vou	V <sub>22</sub> =30V	10.0	10	40.0	N/
	V OH		10.8	12	13.2	V
	V <sub>OL</sub>		-	-	0.5	V
Rising time	I <sub>R</sub>	V <sub>DD</sub> =30V, C <sub>L</sub> =1nF 20%~80%	-	-	100	ns
Falling time	T <sub>F</sub>	V <sub>DD</sub> =30V, C <sub>L</sub> =1nF 80%~20%	-	-	100	ns
Maximum frequency	f <sub>MAX</sub>		170	200	230	kHz
Maximum on time	t <sub>ON-MAX</sub>		21.6	24	26.4	us
Minimum on time	t <sub>on-MIN</sub>		180	200	220	ns
Maximum off time	t <sub>OFF-MAX</sub>		29.97	33.3	36.63	us
Thermal foldback	•					
BIAS voltage	BIAS		0.95	1	1.05	V
BIAS maximum output current	I <sub>BIAS_MAX</sub>				2	mA
Thermal foldback operation point	VTH		0.475	0.5	0.525	V
Thermal foldback trans-conductance	$GM_{GAIN}$		900	1100	1300	uA/V
Over-current protection(OCP)						
Current Limit threshold CS	V <sub>oc</sub>		1.18	1.2	1.32	V
Short protection threshold CS	V <sub>SCP</sub>		1.35	1.5	1.65	V
Startup						
Pre-charge COMP Level	V <sub>PRE</sub>		0.9	1	1.1	V
Over-temperature protection(O	ГР)					
Thermal shutdown temperature	T <sub>SD</sub>		140	155	170	°C

### **Application Information**

MBI6911 is an 85~135V ac input constant current LED driver designed for high PF LED applications. In the application circuit, there are distinct grounds, namely VSS and GNDE. VSS is the reference ground for internal circuit while GNDE is the earth ground. Users should be aware that VSS and GND **CAN NOT** be directly connected together to avoid IC damage and system malfunction.

### **Operation Principle**

#### **Buck Converter**

The typical application circuit of MBI6911 is shown as Fig.4. In the initial state, the  $V_{DD}$  is supplied from  $V_{IN}$  through  $R_{ST1}$  and  $R_{ST2}$ , and then turns on  $Q_1$ . In this duration, the energy of  $V_{IN}$  stores in inductor and also supplies the output loading. When  $Q_1$  turns off,  $Q_1$  will be turned off, and the output loading is supplied by the energy stored in inductor.

#### **Quasi-Resonant Switching**

When power module works in DCM (Discontinuous Conduction Mode), the self-resonance occurs in the duration of MOSFET turns off. The self-resonance will induce the ringing waveform at the Drain terminal of MOSFET, which is the major ingredient of switching loss and EMI. To improve these problems, MBI6911 detects the valley voltage of main switch's  $V_{DS}$ . Moreover,  $Q_1$  will be turned on when the MBI6911 detects the valley voltage of  $V_{DS}$  after 5us, the minimum switching period. The resonant frequency should be limited less than 500KHz. The illustration of QR is shown as Fig.4.



Fig.4 The illustration of Quasi-Resonant switching

#### Start-Up

In the initial state, the  $V_{DD}$  is supplied from the AC source through  $R_{ST}$ , once the  $V_{DD}$  reaches the start-up voltage,  $V_{DD_ON}$ , MBI6911 starts working and consumes the energy stored in  $C_{DD}$ . In the meantime,  $V_{COMP}$  is charged to 1V and enable the fast start-up function. The fast start-up function, whose duty cycle is 3 times of normal operation, and will be terminated until the  $V_{DD}$  reaches 17.3V.

In general the power module with active power factor correction needs large output capacitance to suppress the output ripple current. However, the larger output capacitance results the slower start-up time. To speed up the start-up time, the increasing  $C_{DD}$  and decreasing  $R_{ST}$  is another method.



Fig.5 The illustration of fast start-up

## Protection

#### $V_{\text{DD}}$ Over Voltage Protection (OVP)

MBI6911 realizes input over voltage protection by sensing the input voltage at pin VDD. Once any abnormal spike occurs and exceeds the  $V_{DD}$  OVP threshold  $V_{DD_OVP}$ , numerically 40V, GATE signal ceases switching and  $V_{DD}$  drops accordingly. The converter is fully turned off when  $V_{DD}$  drop below  $V_{UVLO}$ .



Fig.6 The illustration of  $V_{\text{DD}}$  over voltage protection

#### Under Voltage Lock-Out (UVLO)

When  $V_{IN}$  drops below 7V, the GATE output will be forced low to turn off the external power MOSFET. When  $V_{IN}$  rises above 16.0V, the GATE output resumes normal operation and the external power MOSFET starts switching.

#### **Over Temperature Protection (OTP)**

When the junction temperature exceeds 155°C, the built-in over-temperature protection (OTP) is activated to force off the MOSFET. Once the junction temperature drops below 125°C, OTP is deactivated and MBI6911 resumes normal operation.

#### **LED Short-Circuit Protection**

When the output terminals of LED string are short, the output voltage will be clamped to zero, and the abnormal current of inductor is detected by CS pin. Once the CS voltage achieves the short- circuit protection threshold level (1.5V), MBI6911 will be shutdown and  $V_{DD}$  restart again through start-up resistor. The system operates in hiccup mode until the fault condition is removed; the illustration of short- circuit protection is shown as Fig.7.



Fig. 7 LED short-circuit protection

#### **Programmable Thermal Fold-back**

MBI6911 provides a novel programmable thermal fold-back function to enhance the reliability of lighting system, and the schematic is shown as Fig.8.  $V_{BIAS}$  is the reference voltage of thermal fold-back; once the  $V_{VTH}$  descends to  $V_{BIAS}$  with the rising temperature ( $T_{FB}$ ), the thermal fold-back will be enabled and descends the LED current (Fig.8). The descending slop of LED current is decided by the NTC and  $R_{GAIN}$ , the larger  $R_{GAIN}$  results the sharper slope, and vice versa.

When the LED current descends to 20% of normal's, the MBI6911 will into cut-off mode, and then the system will recover to normal operation when the temperature is lower than cut-off point ( $T_{cut-off}$ ).

If the thermal fold-back is unnecessary, please connect the BIAS and VTH pins together and shorts the GAIN pin to VSS pin.



Fig.8 The operation principle of programmable thermal fold-back

### Soldering Process of "Pb-free & Green" Package Plating\*

Macroblock has defined "Pb-Free & Green" to mean semiconductor products that are compatible with the current RoHS requirements and selected 100% pure tin (Sn) to provide forward and backward compatibility with both the current industry-standard SnPb-based soldering processes and higher-temperature Pb-free processes. Pure tin is widely accepted by customers and suppliers of electronic devices in Europe, Asia and the US as the lead-free surface finish of choice to replace tin-lead. Also, it adopts tin/lead (SnPb) solder paste, and please refer to the JEDEC J-STD-020C for the temperature of solder bath. However, in the whole Pb-free soldering processes and materials, 100% pure tin (Sn) will all require from 245 °C to 260°C for proper soldering on boards, referring to JEDEC J-STD-020C as shown below.

For managing MSL3 Package, it should refer to JEDEC J-STD-020C about floor life management & refer to JEDEC J-STD-033C about re-bake condition while IC's floor life exceeds MSL3 limitation.



Package Thickness	Volume mm <sup>3</sup> <350	Volume mm <sup>3</sup> 350-2000	Volume $mm^3 \ge 2000$
<1.6mm	260 +0 °C	260 +0 °C	260 +0 °C
1.6mm – 2.5mm	260 +0 °C	250 +0 °C	245 +0 °C
≧2.5mm	250 +0 °C	245 +0 °C	245 +0 °C

\*Note: For details, please refer to Macroblock's "Policy on Pb-free & Green Package".

## Package Power Dissipation (PD)

The maximum power dissipation,  $P_D(max)=(Tj-Ta)/R_{th(j-a)}$ , decreases as the ambient temperature increases.



## **Outline Drawing**



#### MBI6911 GMS Outline Drawing

Note: Please use the maximum dimensions for designing the thermal pad layout. To avoid the short- circuit risk, the vias or circuit traces shall not pass across the maximum area of thermal pad.

## **Product Top Mark Information**



### **Product Revision History**

Datasheet version	Device Version Code
V1.00	A

### **Product Ordering Information**

Part Number	"Pb-free" Package Type	Weight (g)
MBI6911GMS-A	GMS: MSOP-8L-118mil	0.023 g

## Disclaimer

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