March 2007

LME49720
Dual High Performance, High Fidelity Audio Operational Amplifier

General Description

The LME49720 is part of the ultra-low distortion, low noise, high slew rate operational amplifier series optimized and fully specified for high performance, high fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LME49720 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LME49720 combines extremely low voltage noise density ($2.7nV/\sqrt{Hz}$) with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LME49720 has a high slew rate of $\pm20V/\mu s$ and an output current capability of $\pm26mA$. Further, dynamic range is maximized by an output stage that drives $2k\Omega$ loads to within 1V of either power supply voltage and to within 1.4V when driving 600Ω loads.

The LME49720's outstanding CMRR (120dB), PSRR (120dB), and V_{OS} (0.1mV) give the amplifier excellent operational amplifier DC performance.

The LME49720 has a wide supply range of $\pm2.5V$ to $\pm17V$. Over this supply range the LME49720's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low input bias current. The LME49720 is unity gain stable. This Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with values as high as 100pF.

The LME49720 is available in 8–lead narrow body SOIC, 8–lead Plastic DIP, and 8–lead Metal Can TO-99. Demonstration boards are available for each package.

Key Specifications

- Power Supply Voltage Range $\pm2.5V$ to $\pm17V$
-
  THD+N ($A_V = 1$, $V_{OUT} = 3V_{RMS}$, $f_{IN} = 1kHz$)

|	
$R_L = 2k\Omega$	0.00003% (typ)
$R_L = 600\Omega$	0.00003% (typ)
■ Input Noise Density	$2.7nV/\sqrt{Hz}$ (typ)
■ Slew Rate	$\pm20V/\mu s$ (typ)
■ Gain Bandwidth Product	55MHz (typ)
■ Open Loop Gain ($R_L = 600\Omega$)	140dB (typ)
■ Input Bias Current	10nA (typ)
■ Input Offset Voltage	0.1mV (typ)
■ DC Gain Linearity Error	0.000009%

Features

- Easily drives 600Ω loads
- Optimized for superior audio signal fidelity
- Output short circuit protection
- PSRR and CMRR exceed 120dB (typ)
- SOIC, DIP, TO-99 metal can packages

Applications

- Ultra high quality audio amplification
- High fidelity preamplifiers
- High fidelity multimedia
- State of the art phono pre amps
- High performance professional audio
- High fidelity equalization and crossover networks
- High performance line drivers
- High performance line receivers
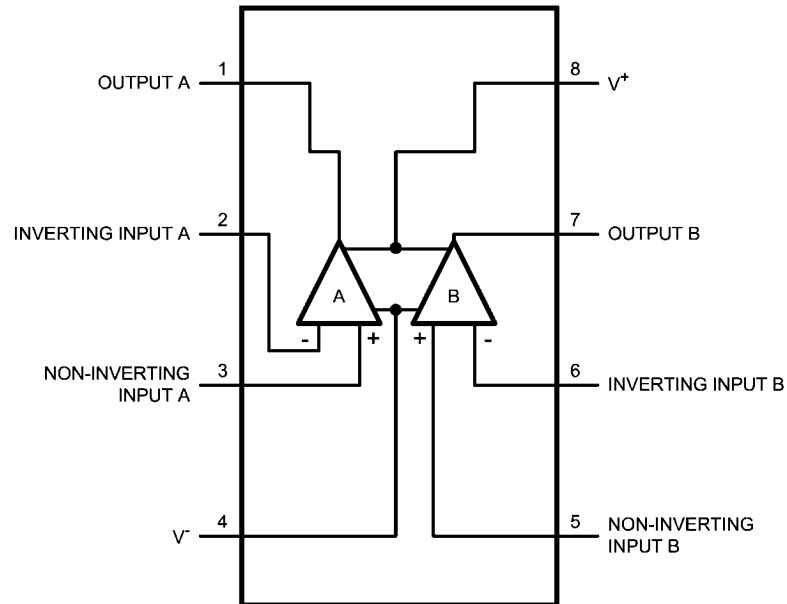- High fidelity active filters

Typical Application
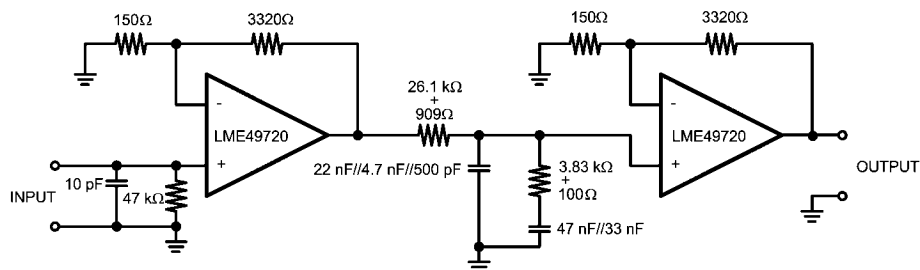
Note: 1% metal film resistors, 5% polypropylene capacitors

Passively Equalized RIAA Phono Preamplifier
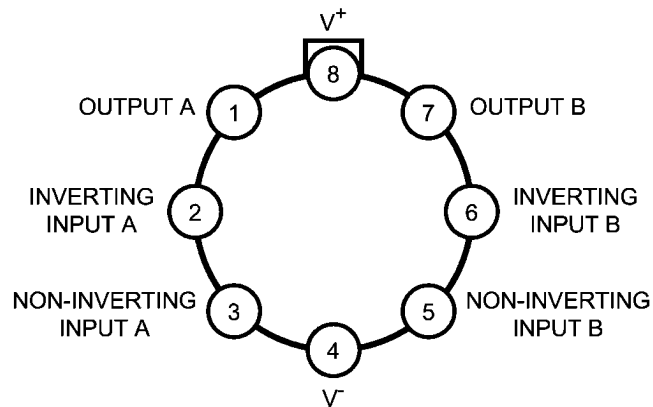
300038k5

Connection Diagrams

Dual-In-Line Package

OUTPUT A — 1

8 — V$^+$

INVERTING INPUT A — 2

7 — OUTPUT B

A

B

NON-INVERTING INPUT A — 3

6 — INVERTING INPUT B

V$^-$ — 4

5 — NON-INVERTING INPUT B

30003855

Order Number LME49720MA
See NS Package Number — M08A
Order Number LME49720NA
See NS Package Number — N08E

Metal Can

V$^+$

8

OUTPUT A — 1

7 — OUTPUT B

INVERTING INPUT A — 2

6 — INVERTING INPUT B

NON-INVERTING INPUT A — 3

5 — NON-INVERTING INPUT B

4

V$^-$

300038f3

Order Number LME49720HA
See NS Package Number — H08C

Absolute Maximum Ratings (Notes 1, 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Power Supply Voltage	
($V_S = V^+ - V^-$)	36V
Storage Temperature	−65°C to 150°C
Input Voltage	(V-) - 0.7V to (V+) + 0.7V
Output Short Circuit (Note 3)	Continuous
Power Dissipation	Internally Limited
ESD Susceptibility (Note 4)	2000V
ESD Susceptibility (Note 5)	

Pins 1, 4, 7 and 8	200V
Pins 2, 3, 5 and 6	100V
Junction Temperature	150°C
Thermal Resistance	
θ_{JA} (SO)	145°C/W
θ_{JA} (NA)	102°C/W
θ_{JA} (HA)	150°C/W
θ_{JC} (HA)	35°C/W
Temperature Range	
$T_{MIN} \leq T_A \leq T_{MAX}$	−40°C ≤ T_A ≤ 85°C
Supply Voltage Range	±2.5V ≤ V_S ≤ ± 17V

Electrical Characteristics for the LME49720

Symbol	Parameter	Conditions	LME49720		Units (Limits)
			Typical	Limit	
			(Note 6)	(Note 7)	
THD+N	Total Harmonic Distortion + Noise	$A_V = 1$, $V_{OUT} = 3V_{rms}$ $R_L = 2k\Omega$ $R_L = 600\Omega$	0.00003 0.00003	 0.00009	% (max)
IMD	Intermodulation Distortion	$A_V = 1$, $V_{OUT} = 3V_{RMS}$ Two-tone, 60Hz & 7kHz 4:1	0.00005		%
GBWP	Gain Bandwidth Product		55	45	MHz (min)
SR	Slew Rate		±20	±15	V/µs (min)
FPBW	Full Power Bandwidth	$V_{OUT} = 1V_{P-P}$, −3dB referenced to output magnitude at f = 1kHz	10		MHz
t_s	Settling time	$A_V = -1$, 10V step, $C_L = 100pF$ 0.1% error range	1.2		µs
e_n	Equivalent Input Noise Voltage	f_{BW} = 20Hz to 20kHz	0.34	0.65	µV$_{RMS}$ (max)
	Equivalent Input Noise Density	f = 1kHz f = 10Hz	2.7 6.4	4.7	nV/√Hz (max)
i_n	Current Noise Density	f = 1kHz f = 10Hz	1.6 3.1		pA/√Hz
V_{OS}	Offset Voltage		±0.1	±0.7	mV (max)
$\Delta V_{OS}/\Delta Temp$	Average Input Offset Voltage Drift vs Temperature	−40°C ≤ T_A ≤ 85°C	0.2		µV/°C
PSRR	Average Input Offset Voltage Shift vs Power Supply Voltage	ΔV_S = 20V (Note 8)	120	110	dB (min)
ISO_{CH-CH}	Channel-to-Channel Isolation	f_{IN} = 1kHz f_{IN} = 20kHz	118 112		dB
I_B	Input Bias Current	V_{CM} = 0V	10	72	nA (max)
$\Delta I_{OS}/\Delta Temp$	Input Bias Current Drift vs Temperature	−40°C ≤ T_A ≤ 85°C	0.1		nA/°C
I_{OS}	Input Offset Current	V_{CM} = 0V	11	65	nA (max)
V_{IN-CM}	Common-Mode Input Voltage Range		+14.1 −13.9	(V+) − 2.0 (V-) + 2.0	V (min)
CMRR	Common-Mode Rejection	−10V<Vcm<10V	120	110	dB (min)
Z_{IN}	Differential Input Impedance		30		kΩ
	Common Mode Input Impedance	−10V<Vcm<10V	1000		MΩ

Symbol	Parameter	Conditions	LME49720		Units (Limits)
			Typical	Limit	
			(Note 6)	(Note 7)	
A_{VOL}	Open Loop Voltage Gain	$-10V<Vout<10V$, $R_L = 600\Omega$	140	125	dB (min)
		$-10V<Vout<10V$, $R_L = 2k\Omega$	140		
		$-10V<Vout<10V$, $R_L = 10k\Omega$	140		
V_{OUTMAX}	Maximum Output Voltage Swing	$R_L = 600\Omega$	±13.6	±12.5	V (min)
		$R_L = 2k\Omega$	±14.0		
		$R_L = 10k\Omega$	±14.1		
I_{OUT}	Output Current	$R_L = 600\Omega$, $V_S = ±17V$	±26	±23	mA (min)
$I_{OUT\text{-}CC}$	Instantaneous Short Circuit Current		+53 −42		mA
R_{OUT}	Output Impedance	$f_{IN} = 10kHz$ Closed-Loop Open-Loop	0.01 13		Ω
C_{LOAD}	Capacitive Load Drive Overshoot	100pF	16		%
I_S	Total Quiescent Current	$I_{OUT} = 0mA$	10	12	mA (max)

Note 1: *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur.

Note 2: Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

Note 3: Amplifier output connected to GND, any number of amplifiers within a package.

Note 4: Human body model, 100pF discharged through a 1.5kΩ resistor.

Note 5: Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under 50Ω).

Note 6: Typical specifications are specified at +25ºC and represent the most likely parametric norm.

Note 7: Tested limits are guaranteed to National's AOQL (Average Outgoing Quality Level).

Note 8: PSRR is measured as follows: V_{OS} is measured at two supply voltages, ±5V and ±15V. PSRR = | 20log($\Delta V_{OS}/\Delta V_S$) |.

Typical Performance Characteristics

THD+N vs Output Voltage
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 2k\Omega$
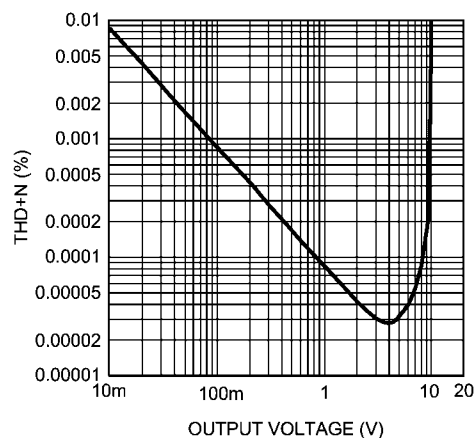
300038k6

THD+N vs Output Voltage
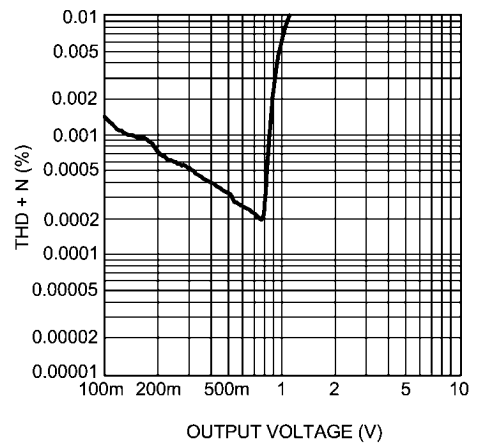$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 2k\Omega$
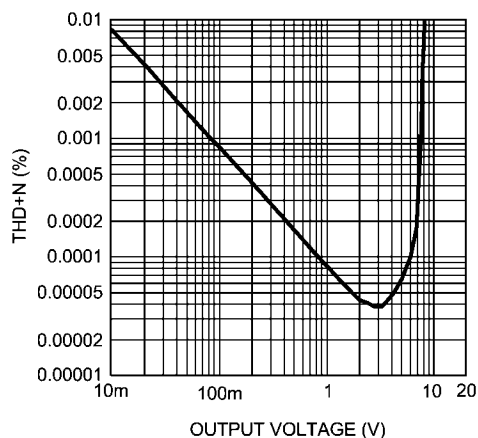
300038k7

THD+N vs Output Voltage
$V_{CC} = 17V$, $V_{EE} = -17V$
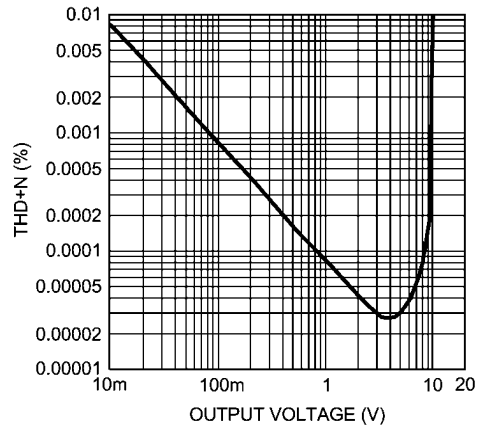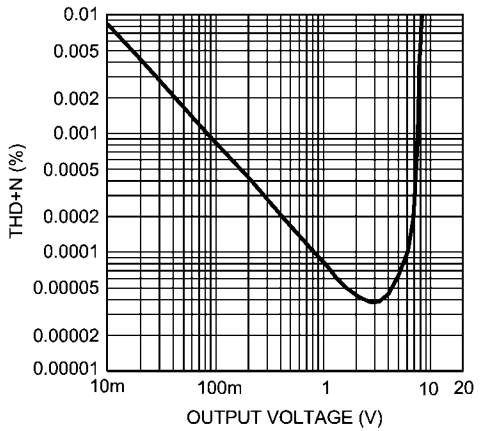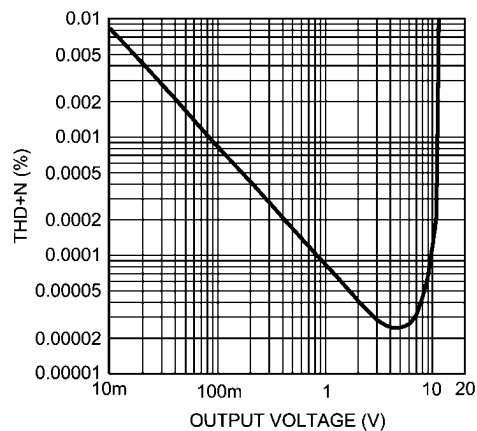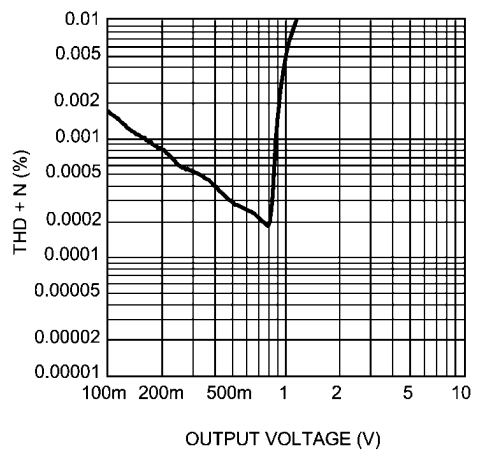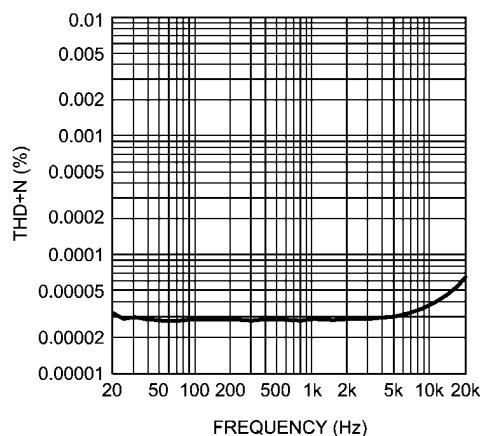$R_L = 2k\Omega$
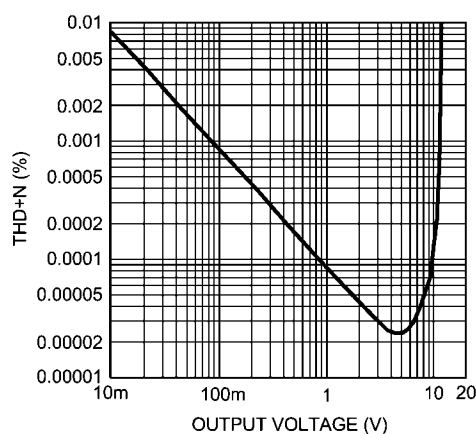
300038k8

THD+N vs Output Voltage
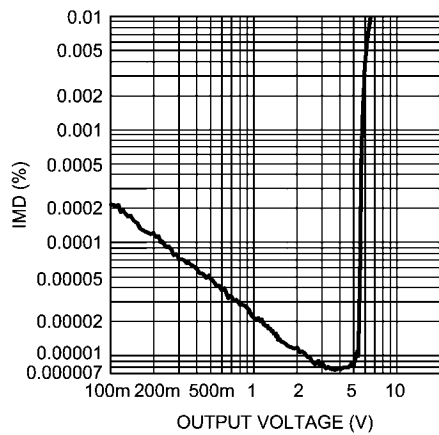$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 2k\Omega$
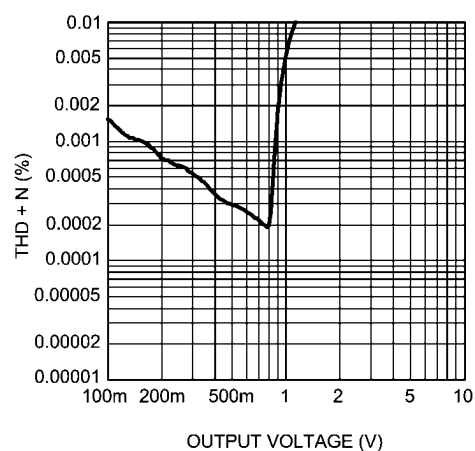
300038i4

THD+N vs Output Voltage
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 600\Omega$
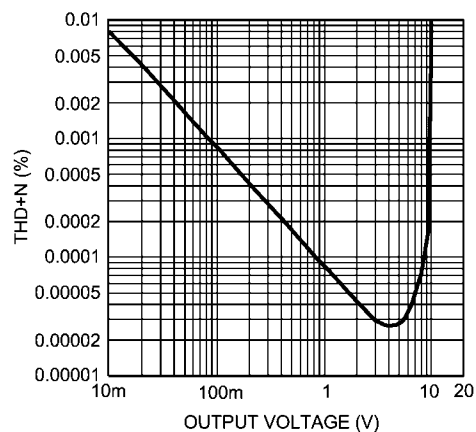
300038k9

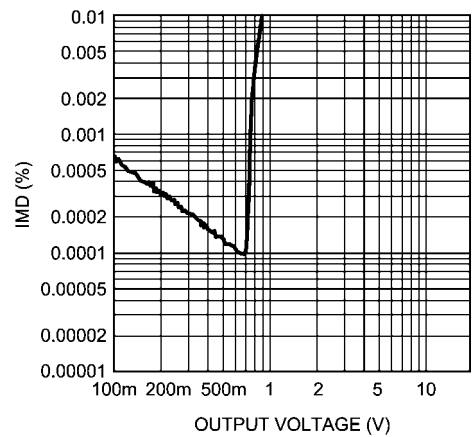**THD+N vs Output Voltage**
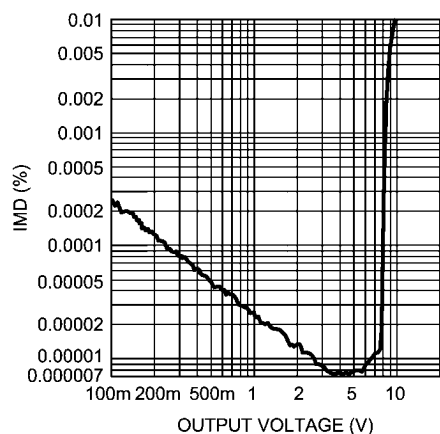$V_{CC} = 12V$, $V_{EE} = -12V$
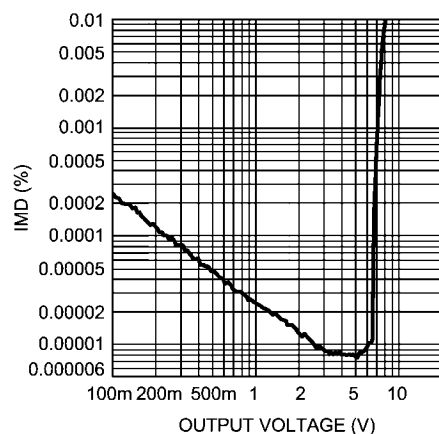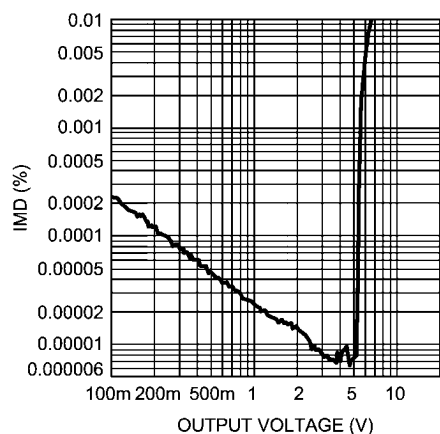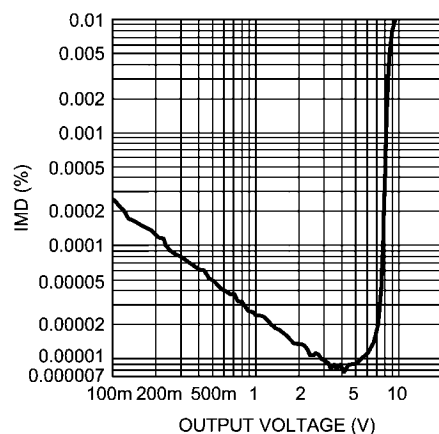$R_L = 600\Omega$

300038l0

THD+N vs Output Voltage
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 600\Omega$

THD+N (%)

OUTPUT VOLTAGE (V)

300038I1

THD+N vs Output Voltage
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 600\Omega$

THD + N (%)

OUTPUT VOLTAGE (V)

300038i6

THD+N vs Output Voltage
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 10k\Omega$

THD+N (%)

OUTPUT VOLTAGE (V)

300038I2

THD+N vs Output Voltage
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 10k\Omega$

THD+N (%)

OUTPUT VOLTAGE (V)

300038I3

THD+N vs Output Voltage
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 10k\Omega$

THD+N (%)

OUTPUT VOLTAGE (V)

300038I4

THD+N vs Output Voltage
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 10k\Omega$

THD + N (%)

OUTPUT VOLTAGE (V)

300038i5

THD+N vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$, $V_{OUT} = 3V_{RMS}$
$R_L = 2k\Omega$

30003863

THD+N vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$, $V_{OUT} = 3V_{RMS}$
$R_L = 2k\Omega$
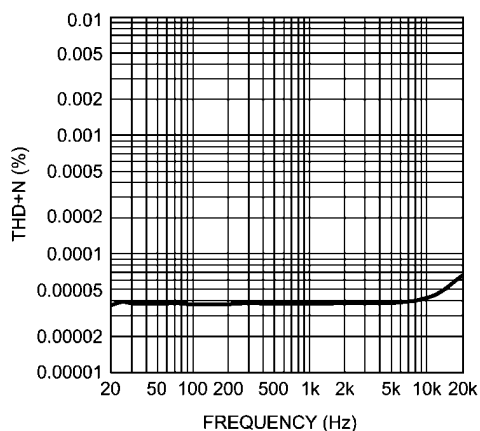
30003862

THD+N vs Frequency
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$R_L = 2k\Omega$
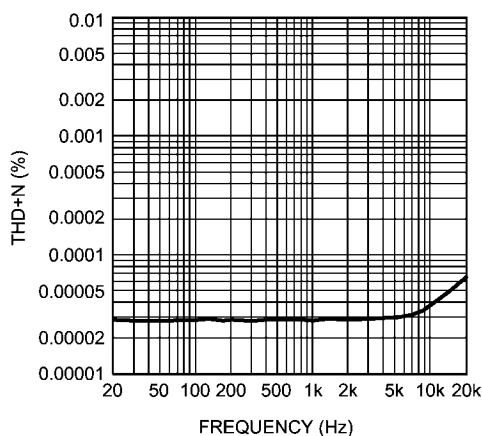
30003864

THD+N vs Frequency
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$R_L = 600\Omega$
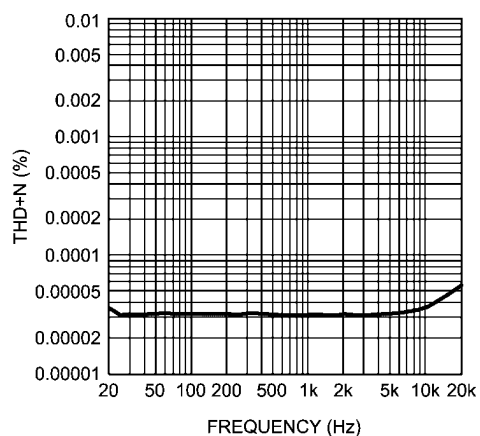
30003859

THD+N vs Frequency
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$R_L = 600\Omega$
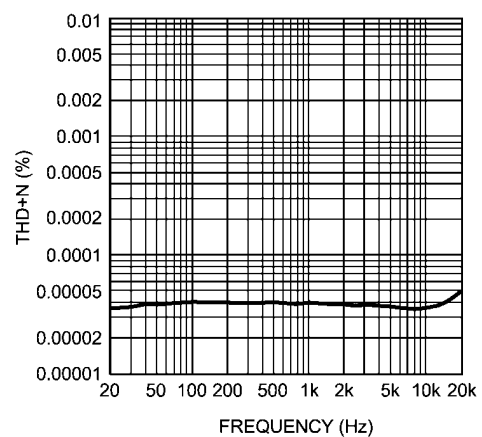
300038k3

THD+N vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 3V_{RMS}$
$R_L = 600\Omega$
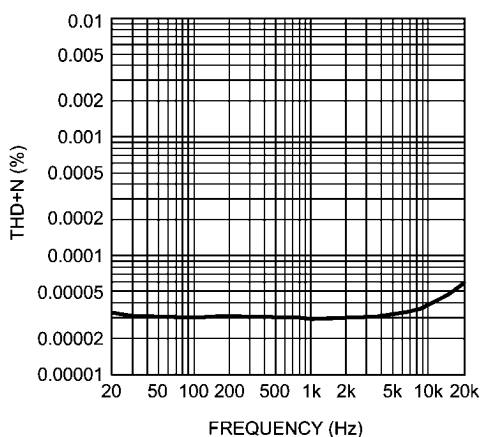
30003860

THD+N vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$, $V_{OUT} = 3V_{RMS}$
$R_L = 10k\Omega$

THD+N (%) vs FREQUENCY (Hz)
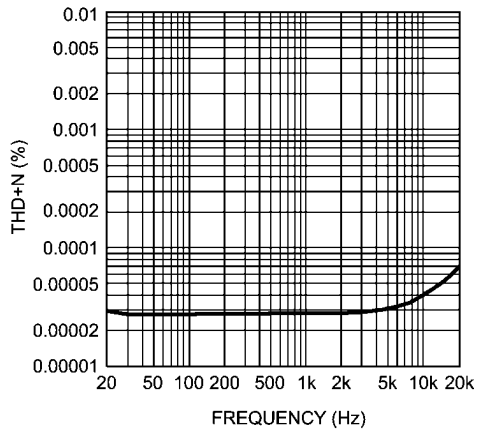
30003867

THD+N vs Frequency
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$R_L = 10k\Omega$

THD+N (%) vs FREQUENCY (Hz)
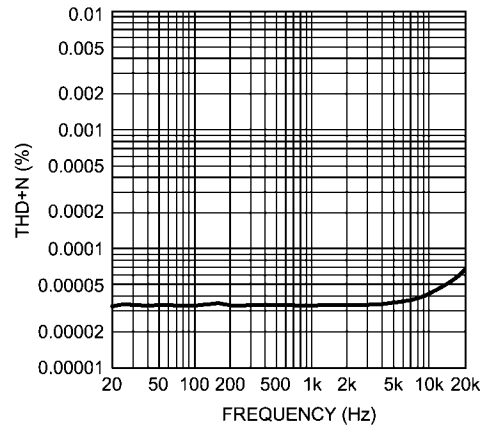
30003866

THD+N vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 3V_{RMS}$
$R_L = 10k\Omega$

THD+N (%) vs FREQUENCY (Hz)
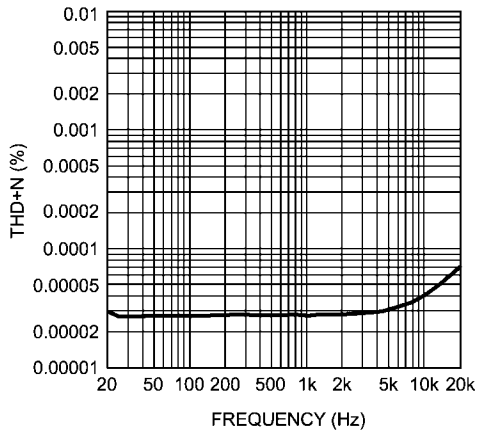
30003868

IMD vs Output Voltage
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 2k\Omega$

IMD (%) vs OUTPUT VOLTAGE (V)
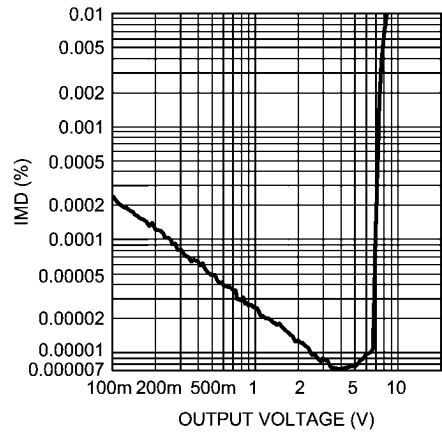
300038e6

IMD vs Output Voltage
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 2k\Omega$

IMD (%) vs OUTPUT VOLTAGE (V)

300038e5

IMD vs Output Voltage
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 2k\Omega$

IMD (%) vs OUTPUT VOLTAGE (V)

300038e4

IMD vs Output Voltage
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 2k\Omega$

300038e7

IMD vs Output Voltage
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 600\Omega$

300038e2

IMD vs Output Voltage
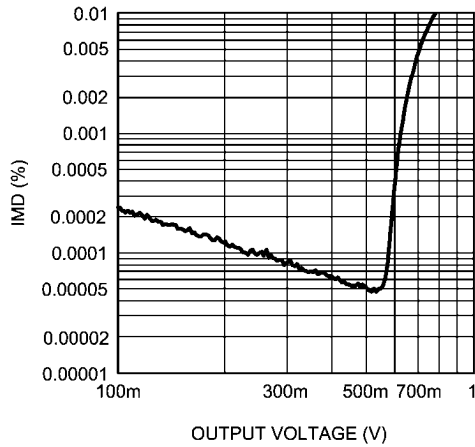$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 600\Omega$

300038e0

IMD vs Output Voltage
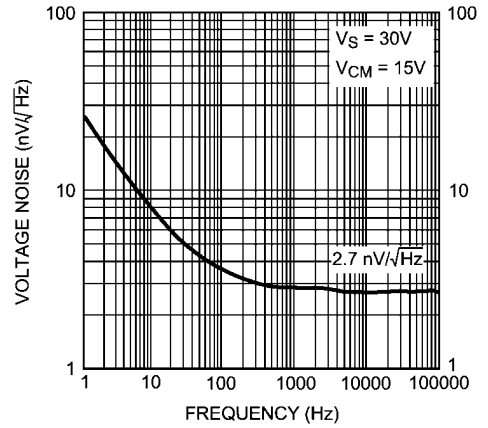$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 600\Omega$

300038e3

IMD vs Output Voltage
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 600\Omega$
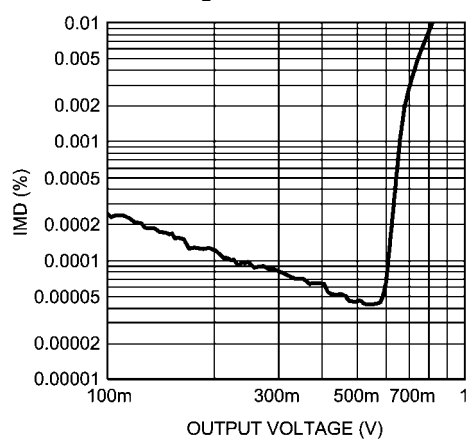
300038e1

IMD vs Output Voltage
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$R_L = 10k\Omega$
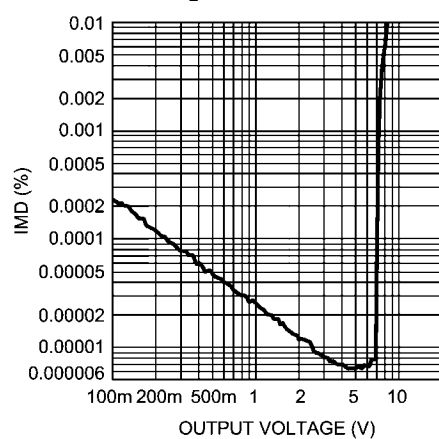
300038f1

IMD vs Output Voltage
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 10k\Omega$
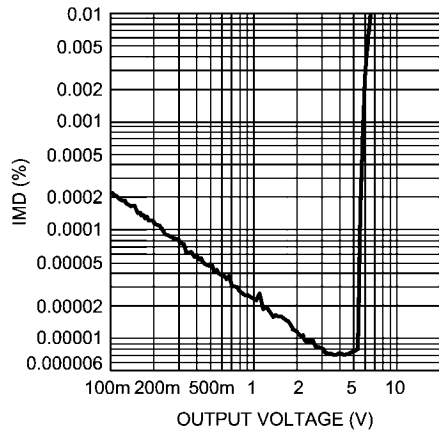
300038f0

IMD vs Output Voltage
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 10k\Omega$
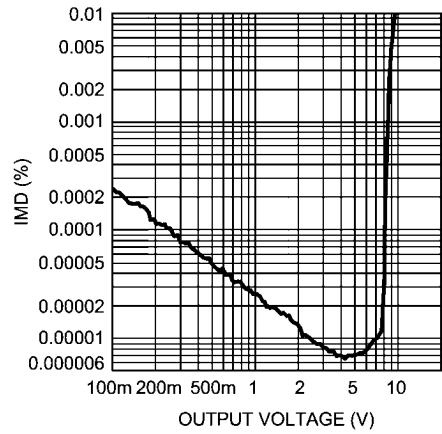
300038f2

IMD vs Output Voltage
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 10k\Omega$

300038l6

Voltage Noise Density vs Frequency

300038h6

Current Noise Density vs Frequency
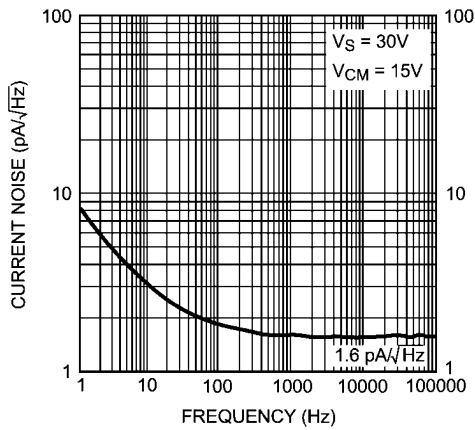
300038h7

Crosstalk vs Frequency
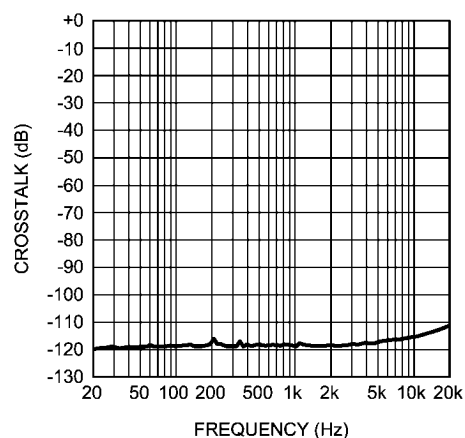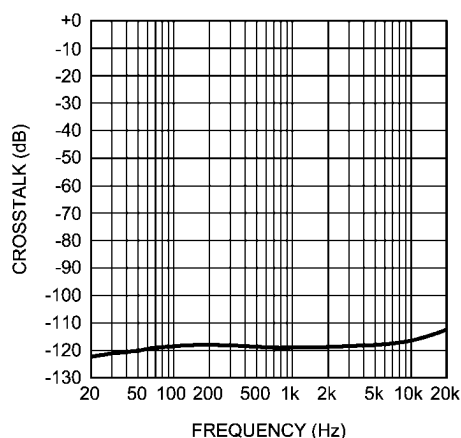$V_{CC} = 15V$, $V_{EE} = -15V$, $V_{OUT} = 3V_{RMS}$
$A_V = 0dB$, $R_L = 2k\Omega$
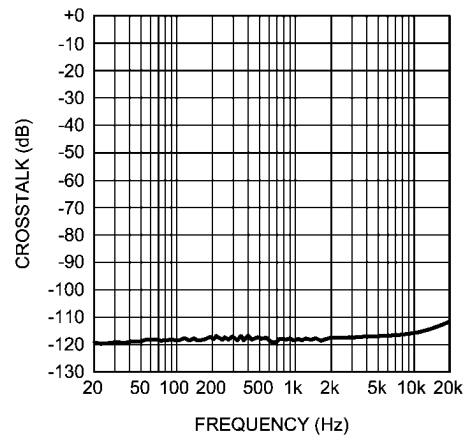
300038c8

Crosstalk vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$, $V_{OUT} = 10V_{RMS}$
$A_V = 0dB$, $R_L = 2k\Omega$

300038c9

Crosstalk vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$, $V_{OUT} = 3V_{RMS}$
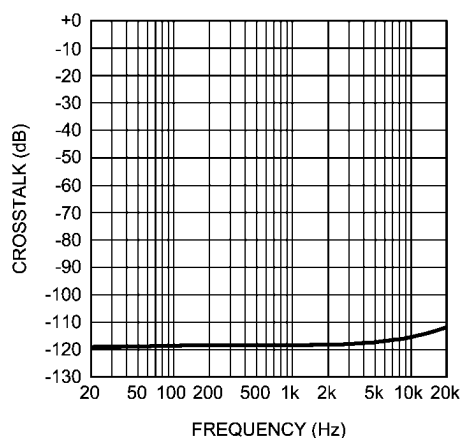$A_V = 0dB$, $R_L = 2k\Omega$

300038c6

Crosstalk vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$, $V_{OUT} = 10V_{RMS}$
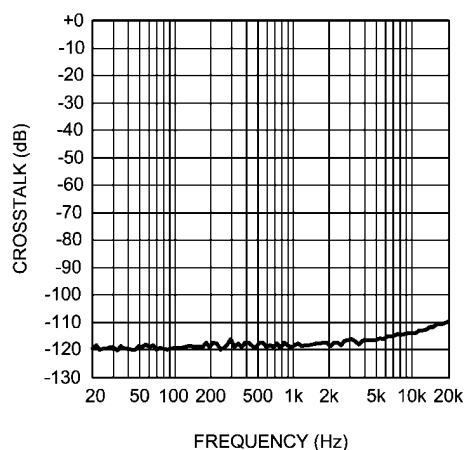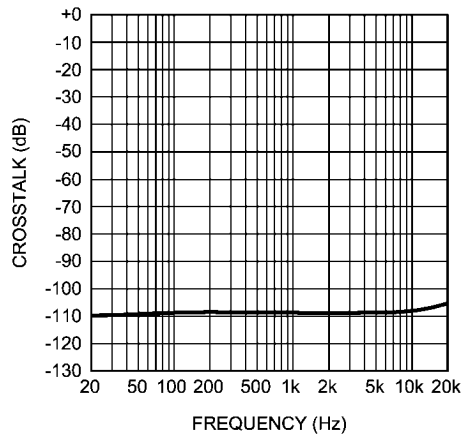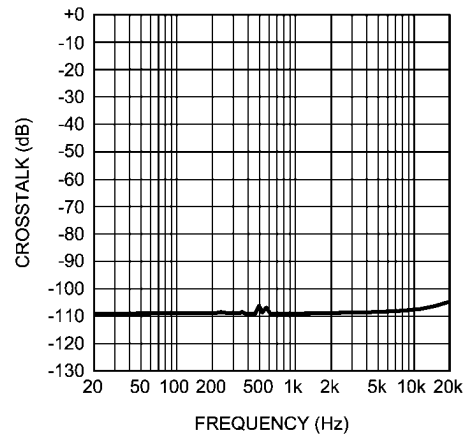$A_V = 0dB$, $R_L = 2k\Omega$

300038c7

Crosstalk vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 3V_{RMS}$
$A_V = 0dB$, $R_L = 2k\Omega$
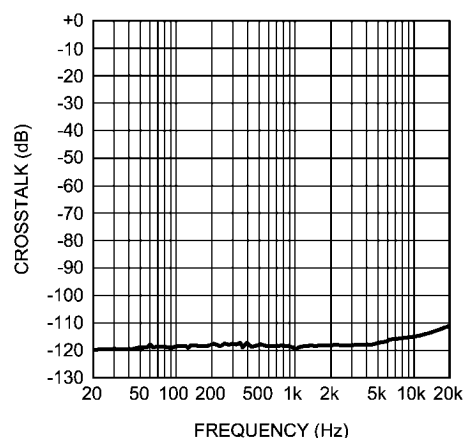
300038d0

Crosstalk vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 10V_{RMS}$
$A_V = 0dB$, $R_L = 2k\Omega$

300038d1

Crosstalk vs Frequency
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$, $V_{OUT} = 1V_{RMS}$
$A_V = 0dB$, $R_L = 2k\Omega$

300038n8

Crosstalk vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$, $V_{OUT} = 3V_{RMS}$
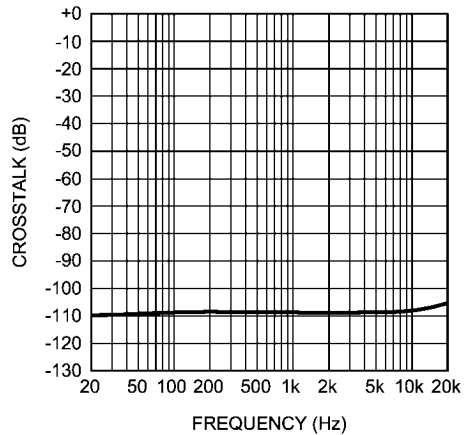$A_V = 0dB$, $R_L = 600\Omega$

300038d6

Crosstalk vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$, $V_{OUT} = 10V_{RMS}$
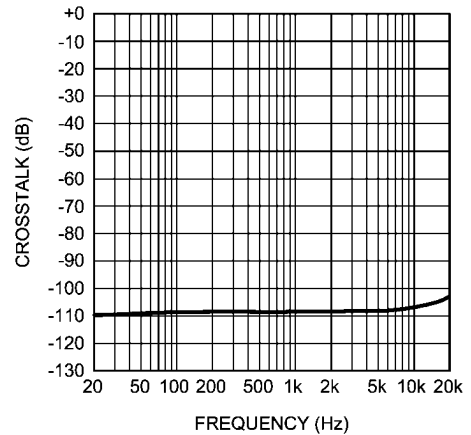$A_V = 0dB$, $R_L = 600\Omega$

300038d7

Crosstalk vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$, $V_{OUT} = 3V_{RMS}$
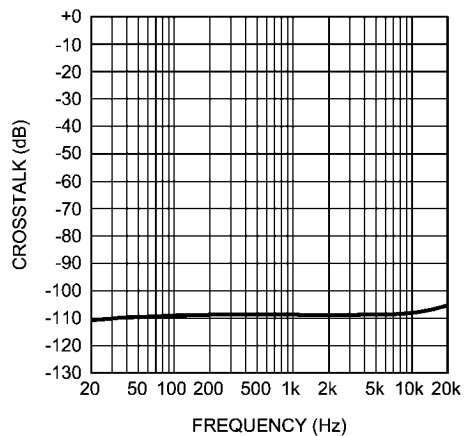$A_V = 0dB$, $R_L = 600\Omega$

300038d4

Crosstalk vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$, $V_{OUT} = 10V_{RMS}$
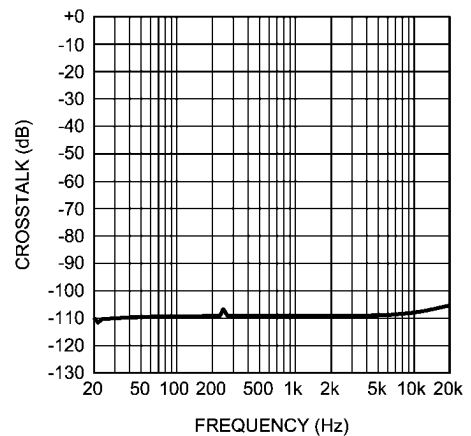$A_V = 0dB$, $R_L = 600\Omega$

300038d5

Crosstalk vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 3V_{RMS}$
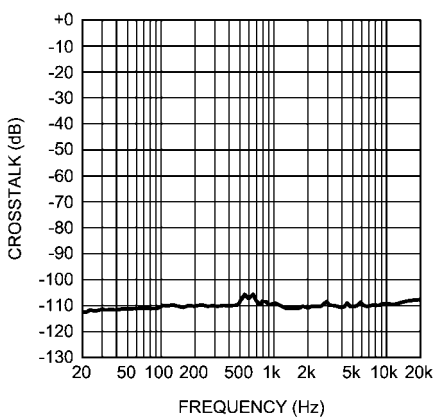$A_V = 0dB$, $R_L = 600\Omega$

300038d8

Crosstalk vs Frequency
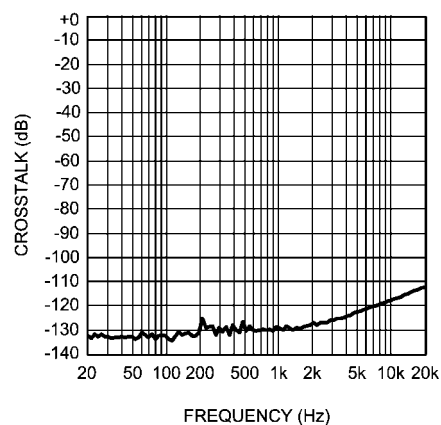$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 10V_{RMS}$
$A_V = 0dB$, $R_L = 600\Omega$

300038d9

Crosstalk vs Frequency
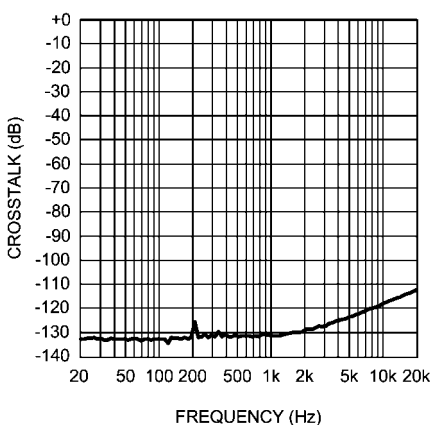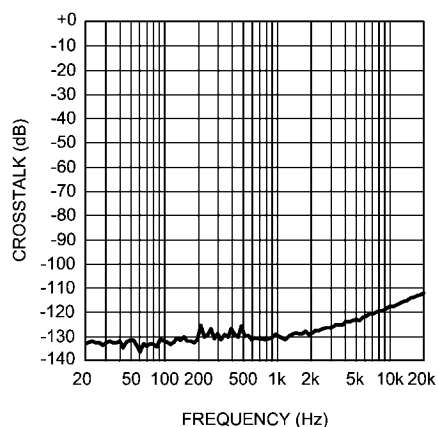$V_{CC} = 2.5V, V_{EE} = -2.5V, V_{OUT} = 1V_{RMS}$
$A_V = 0dB, R_L = 600\Omega$

Crosstalk vs Frequency
$V_{CC} = 15V, V_{EE} = -15V, V_{OUT} = 3V_{RMS}$
$A_V = 0dB, R_L = 10k\Omega$

Crosstalk vs Frequency
$V_{CC} = 15V, V_{EE} = -15V, V_{OUT} = 10V_{RMS}$
$A_V = 0dB, R_L = 10k\Omega$

Crosstalk vs Frequency
$V_{CC} = 12V, V_{EE} = -12V, V_{OUT} = 3V_{RMS}$
$A_V = 0dB, R_L = 10k\Omega$

Crosstalk vs Frequency
$V_{CC} = 12V, V_{EE} = -12V, V_{OUT} = 10V_{RMS}$
$A_V = 0dB, R_L = 10k\Omega$

Crosstalk vs Frequency
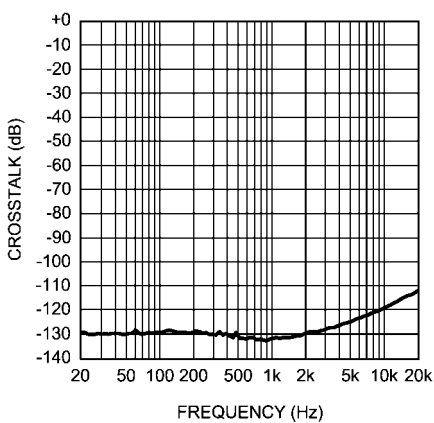$V_{CC} = 17V, V_{EE} = -17V, V_{OUT} = 3V_{RMS}$
$A_V = 0dB, R_L = 10k\Omega$
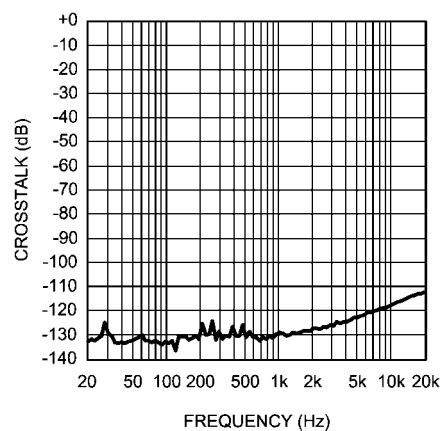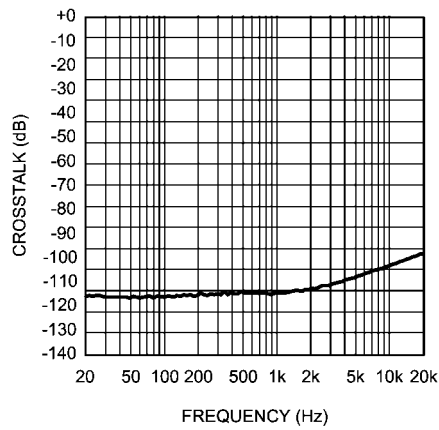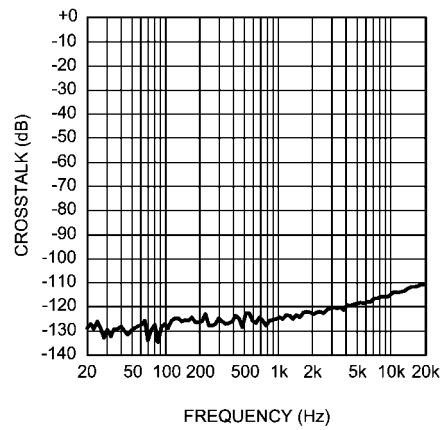
300038d2
300038o0
300038n7
300038n9
300038n6
300038n5

Crosstalk vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$, $V_{OUT} = 10V_{RMS}$
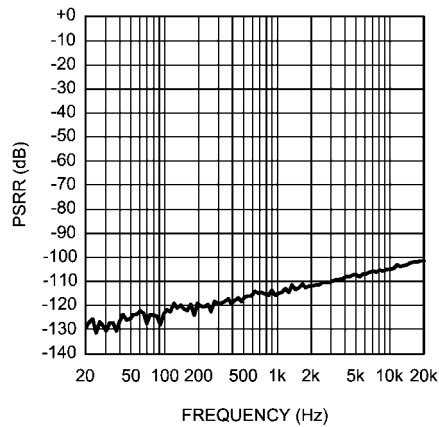$A_V = 0dB$, $R_L = 10k\Omega$

CROSSTALK (dB)

FREQUENCY (Hz)

300038n3

Crosstalk vs Frequency
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$, $V_{OUT} = 1V_{RMS}$
$A_V = 0dB$, $R_L = 10k\Omega$

CROSSTALK (dB)

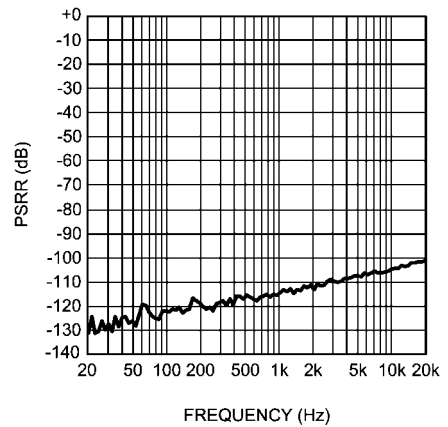FREQUENCY (Hz)

300038n4

PSRR+ vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 2k\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR (dB)

FREQUENCY (Hz)

300038o1

PSRR- vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 2k\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR (dB)

FREQUENCY (Hz)

300038n2

PSRR+ vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 2k\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR (dB)
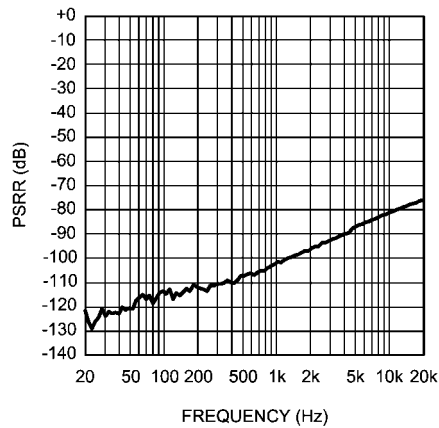
FREQUENCY (Hz)

300038n1

PSRR- vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 2k\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR (dB)

FREQUENCY (Hz)

300038n0

PSRR+ vs Frequency
V_{CC} = 17V, V_{EE} = −17V
R_L = 2kΩ, V_{RIPPLE} = 200mVpp

PSRR- vs Frequency
V_{CC} = 17V, V_{EE} = −17V
R_L = 2kΩ, V_{RIPPLE} = 200mVpp

300038m9

300038o3

PSRR+ vs Frequency
V_{CC} = 2.5V, V_{EE} = −2.5V
R_L = 2kΩ, V_{RIPPLE} = 200mVpp

PSRR- vs Frequency
V_{CC} = 2.5V, V_{EE} = −2.5V
R_L = 2kΩ, V_{RIPPLE} = 200mVpp

300038m8

300038o6

PSRR+ vs Frequency
V_{CC} = 15V, V_{EE} = −15V
R_L = 600Ω, V_{RIPPLE} = 200mVpp

PSRR- vs Frequency
V_{CC} = 15V, V_{EE} = −15V
R_L = 600Ω, V_{RIPPLE} = 200mVpp

300038o2

300038o7

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PSRR+ vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 600\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR- vs Frequency
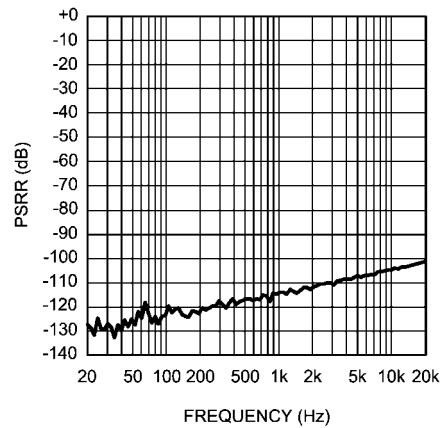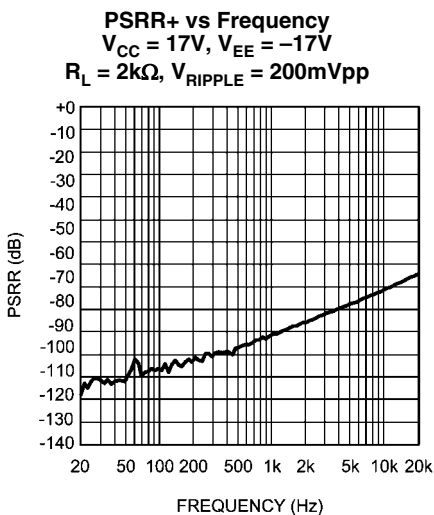$V_{CC} = 12V$, $V_{EE} = -12V$
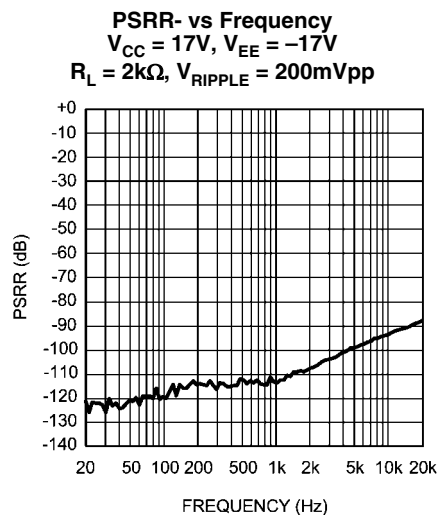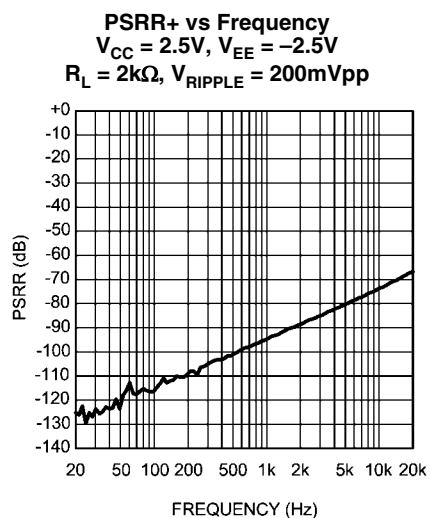$R_L = 600\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR+ vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 600\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR- vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 600\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR+ vs Frequency
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 600\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR- vs Frequency
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 600\Omega$, $V_{RIPPLE} = 200mVpp$

PSRR+ vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 10k\Omega$, $V_{RIPPLE} = 200mVpp$

300038m3

PSRR- vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 10k\Omega$, $V_{RIPPLE} = 200mVpp$

300038m2

PSRR+ vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 10k\Omega$, $V_{RIPPLE} = 200mVpp$

300038m1

PSRR- vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 10k\Omega$, $V_{RIPPLE} = 200mVpp$

300038m0

PSRR+ vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 10k\Omega$, $V_{RIPPLE} = 200mVpp$

300038l9

PSRR- vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 10k\Omega$, $V_{RIPPLE} = 200mVpp$

300038l8

PSRR+ vs Frequency
V_{CC} = 2.5V, V_{EE} = −2.5V
R_L = 10kΩ, V_{RIPPLE} = 200mVpp

300038l7

PSRR- vs Frequency
V_{CC} = 2.5V, V_{EE} = −2.5V
R_L = 10kΩ, V_{RIPPLE} = 200mVpp

300038l5

CMRR vs Frequency
V_{CC} = 15V, V_{EE} = −15V
R_L = 2kΩ

300038g0

CMRR vs Frequency
V_{CC} = 12V, V_{EE} = −12V
R_L = 2kΩ

300038f7

CMRR vs Frequency
V_{CC} = 17V, V_{EE} = −17V
R_L = 2kΩ

300038g3

CMRR vs Frequency
V_{CC} = 2.5V, V_{EE} = −2.5V
R_L = 2kΩ

300038f4

CMRR vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 600\Omega$

300038o9

CMRR vs Frequency
$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 600\Omega$
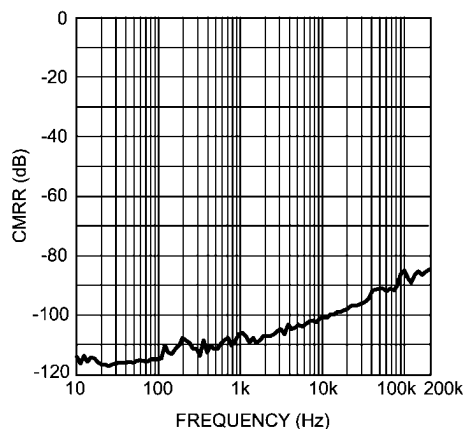
300038f9

CMRR vs Frequency
$V_{CC} = 17V$, $V_{EE} = -17V$
$R_L = 600\Omega$
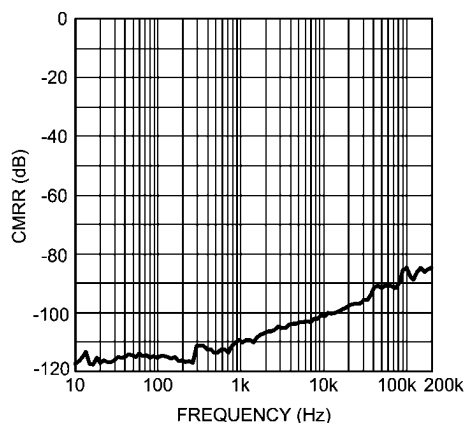
300038g5

CMRR vs Frequency
$V_{CC} = 2.5V$, $V_{EE} = -2.5V$
$R_L = 600\Omega$
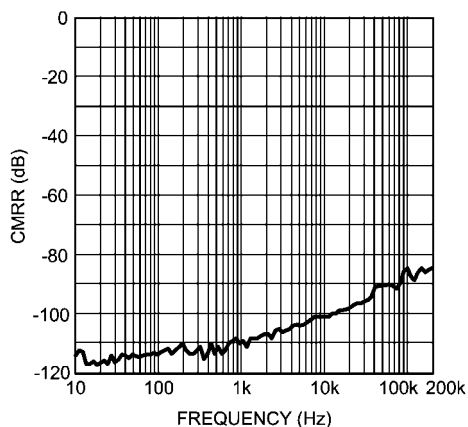
300038f6

CMRR vs Frequency
$V_{CC} = 15V$, $V_{EE} = -15V$
$R_L = 10k\Omega$
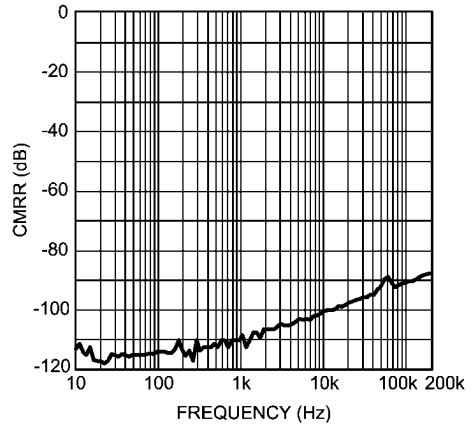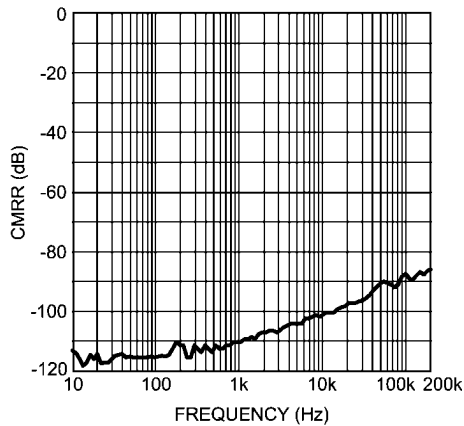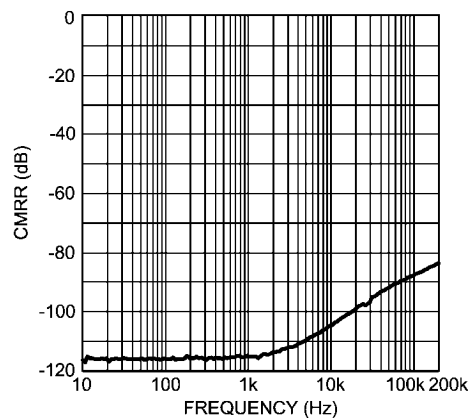

300038o8

CMRR vs Frequency
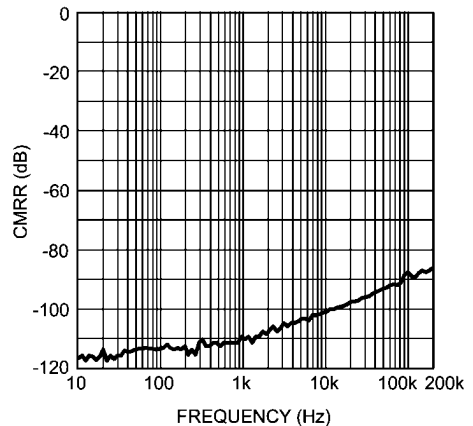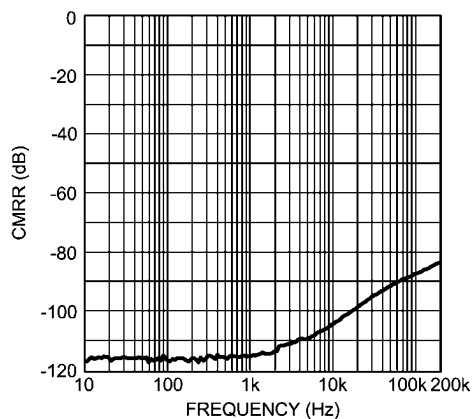$V_{CC} = 12V$, $V_{EE} = -12V$
$R_L = 10k\Omega$
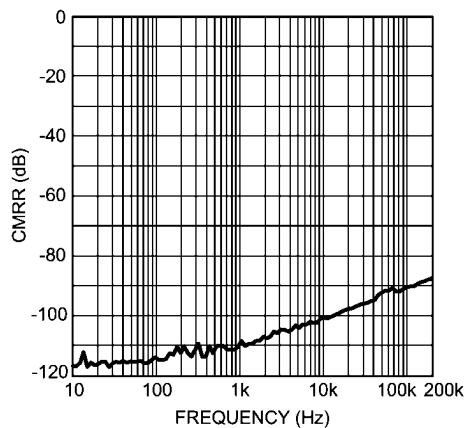
300038f8

CMRR vs Frequency
V_{CC} = 17V, V_{EE} = −17V
R_L = 10kΩ
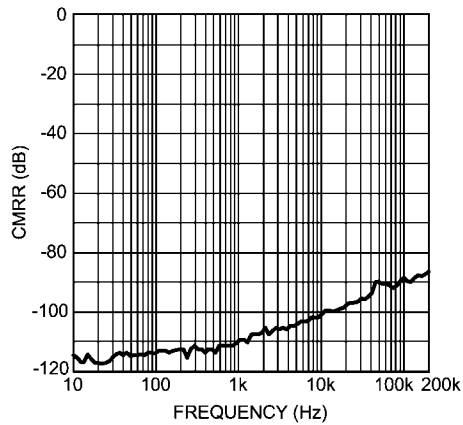
300038g4

CMRR vs Frequency
V_{CC} = 2.5V, V_{EE} = −2.5V
R_L = 10kΩ
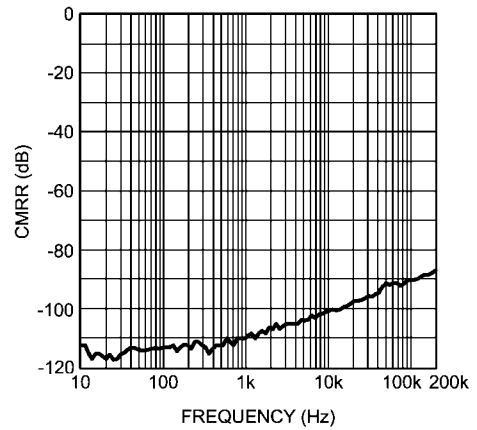
300038f5

Output Voltage vs Load Resistance
V_{DD} = 15V, V_{EE} = −15V
THD+N = 1%
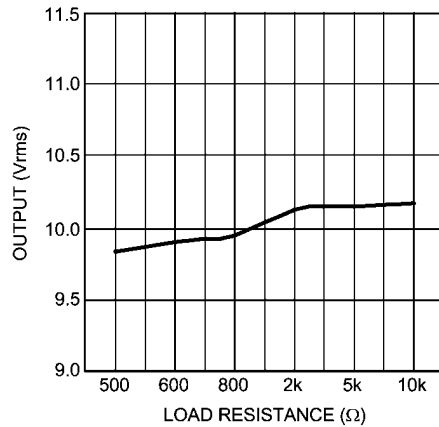
300038h1

Output Voltage vs Load Resistance
V_{DD} = 12V, V_{EE} = −12V
THD+N = 1%
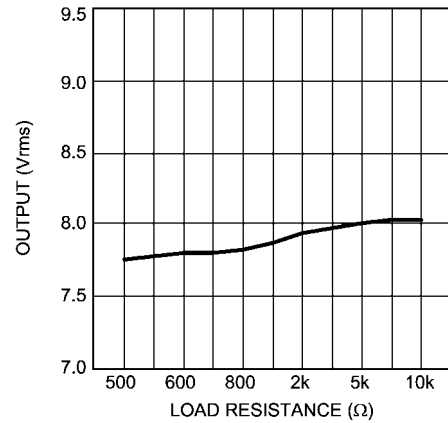
300038h0

Output Voltage vs Load Resistance
V_{DD} = 17V, V_{EE} = −17V
THD+N = 1%
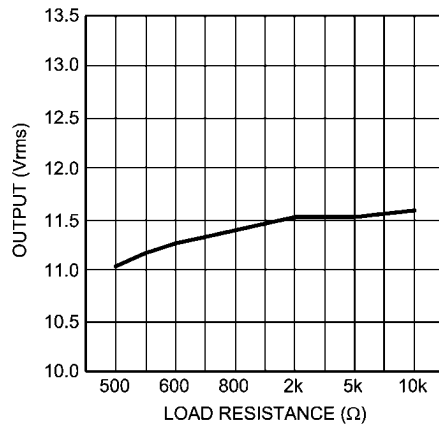
300038h2

Output Voltage vs Load Resistance
V_{DD} = 2.5V, V_{EE} = −2.5V
THD+N = 1%
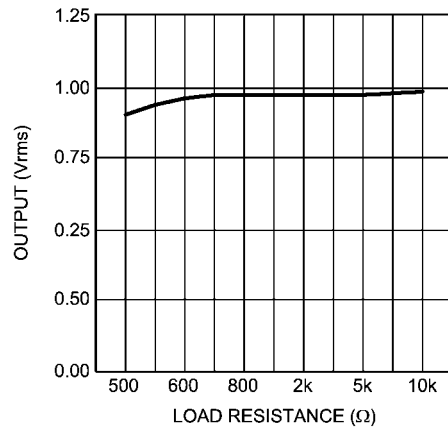
300038g9

Output Voltage vs Supply Voltage
$R_L = 2k\Omega$, THD+N = 1%
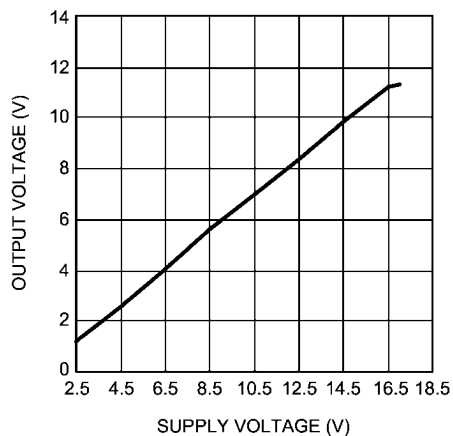
300038j9

Output Voltage vs Supply Voltage
$R_L = 600\Omega$, THD+N = 1%
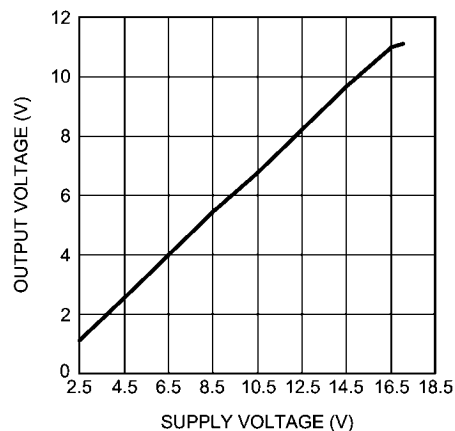
300038j8

Output Voltage vs Supply Voltage
$R_L = 10k\Omega$, THD+N = 1%
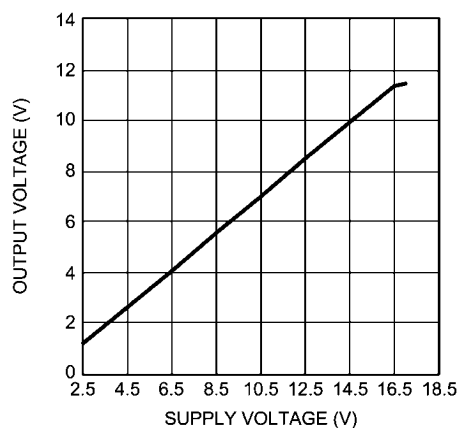
300038k0

Supply Current vs Supply Voltage
$R_L = 2k\Omega$
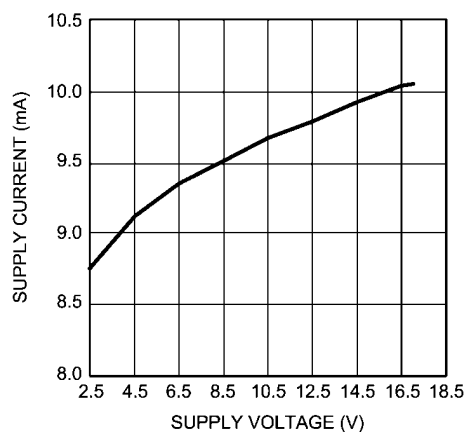
300038j6

Supply Current vs Supply Voltage
$R_L = 600\Omega$
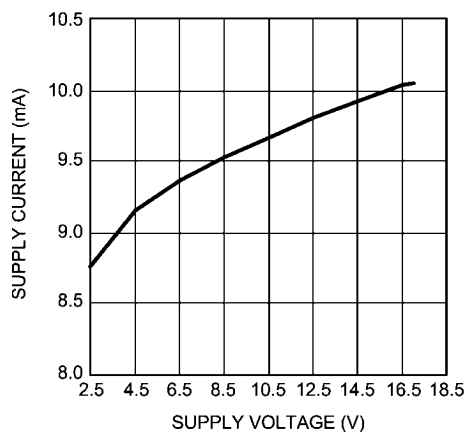
300038j5

Supply Current vs Supply Voltage
$R_L = 10k\Omega$
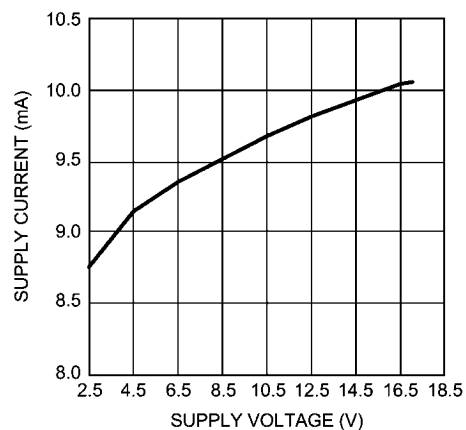
300038j7

Full Power Bandwidth vs Frequency
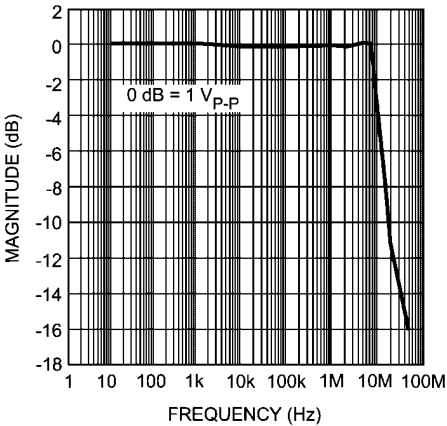
300038j0

Gain Phase vs Frequency
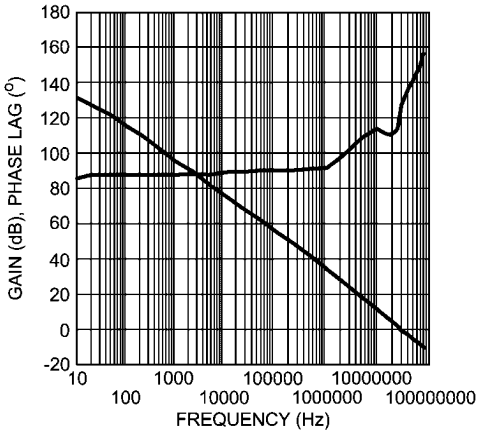
300038j1

Small-Signal Transient Response
$A_V = 1$, $C_L = 10pF$
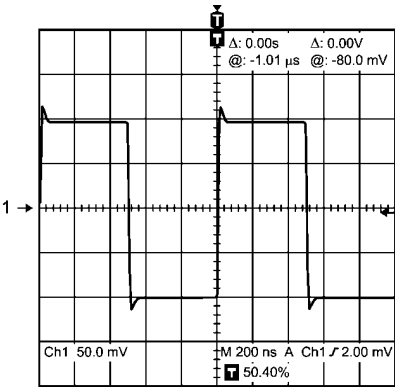
300038i7

Small-Signal Transient Response
$A_V = 1$, $C_L = 100pF$
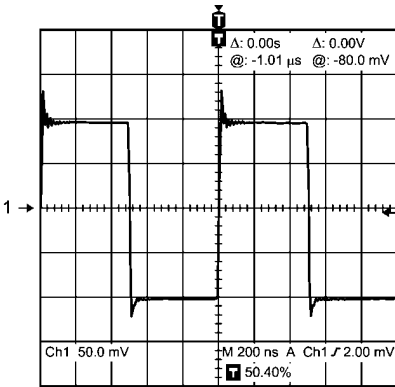
300038i8

Application Information

DISTORTION MEASUREMENTS

The vanishingly low residual distortion produced by LME49720 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LME49720's low residual distortion is an input referred internal error. As shown in Figure 1, adding the 10Ω resistor connected between the amplifier's inverting and non-inverting inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in Figure 1.

This technique is verified by duplicating the measurements with high closed loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.
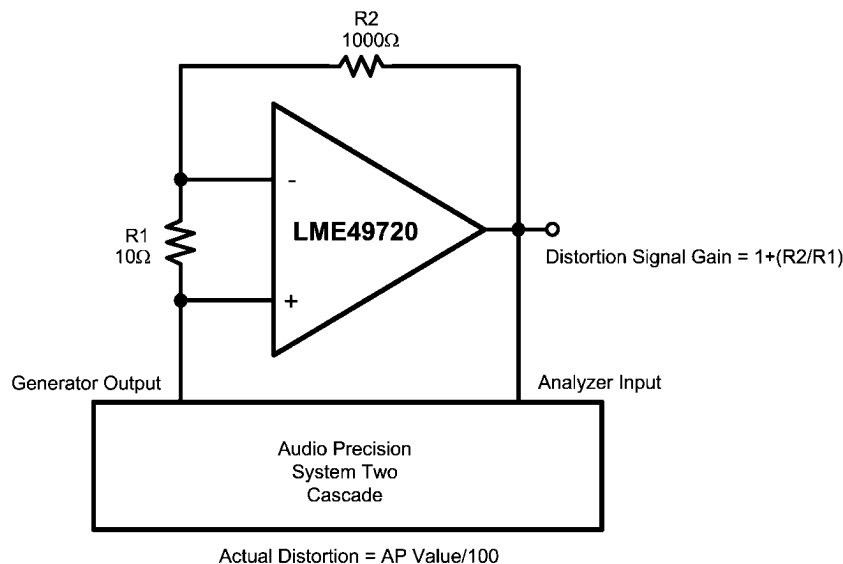
FIGURE 1. THD+N and IMD Distortion Test Circuit

The LME49720 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.
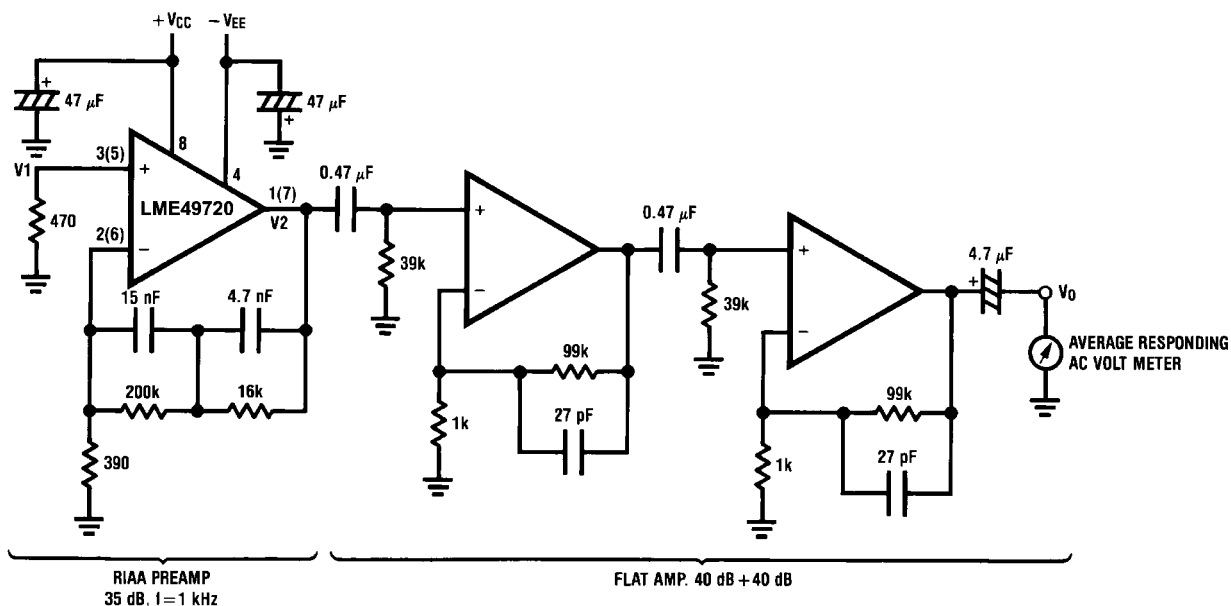
30003827

Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

Noise Measurement Circuit
Total Gain: 115 dB @f = 1 kHz
Input Referred Noise Voltage: e_n = V0/560,000 (V)

RIAA Preamp Voltage Gain, RIAA Deviation vs Frequency
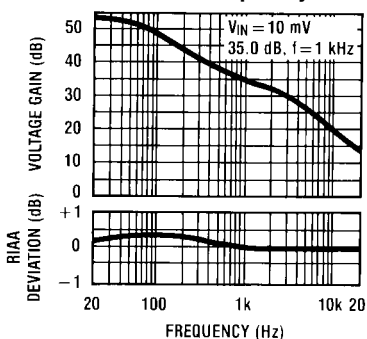
30003828

Flat Amp Voltage Gain vs Frequency
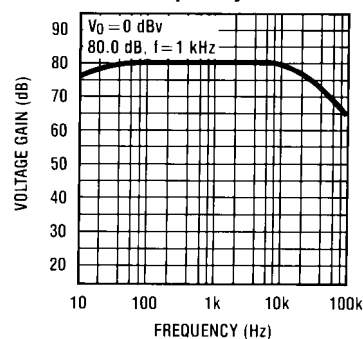
30003829

TYPICAL APPLICATIONS
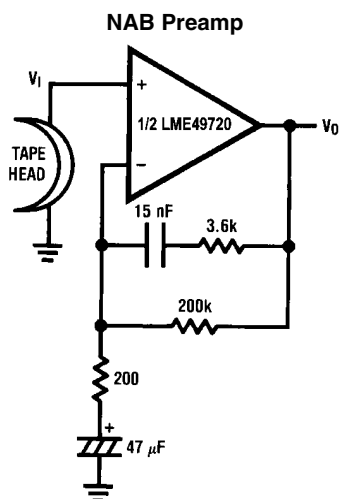
NAB Preamp

30003830

$A_V = 34.5$
$F = 1$ kHz
$E_n = 0.38$ μV
A Weighted

NAB Preamp Voltage Gain vs Frequency
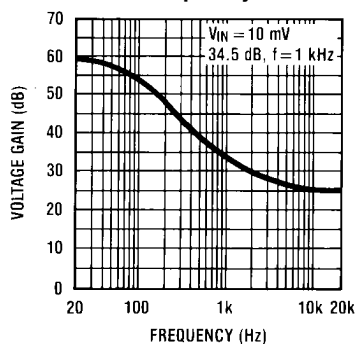
30003831

Balanced to Single Ended Converter
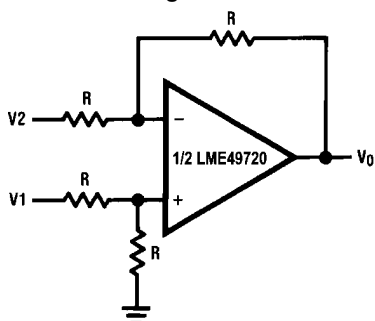
30003832

$V_O = V1 - V2$

Adder/Subtracter
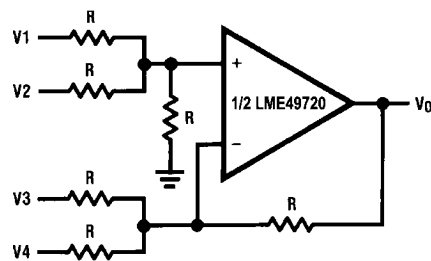
30003833

$V_O = V1 + V2 - V3 - V4$

Sine Wave Oscillator
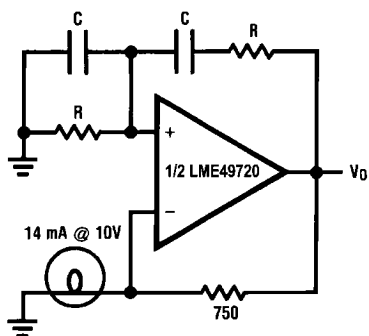
30003834

$$f_O = \frac{1}{2\pi RC}$$

Second Order High Pass Filter
(Butterworth)
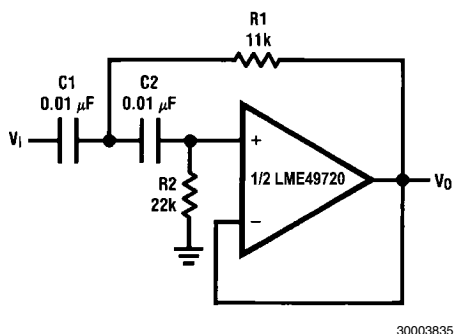
30003835

if C1 = C2 = C

$$R1 = \frac{\sqrt{2}}{2\omega_0 C}$$

$$R2 = 2 \bullet R1$$

Illustration is f_0 = 1 kHz

Second Order Low Pass Filter
(Butterworth)
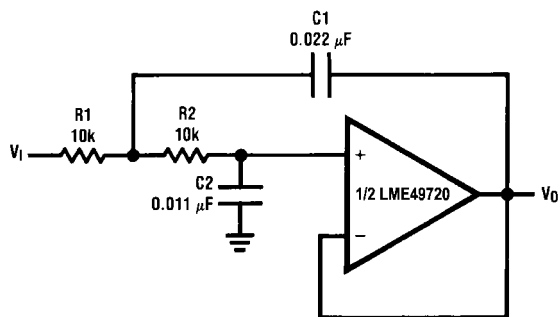
30003836

if R1 = R2 = R

$$C1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is f_0 = 1 kHz

State Variable Filter
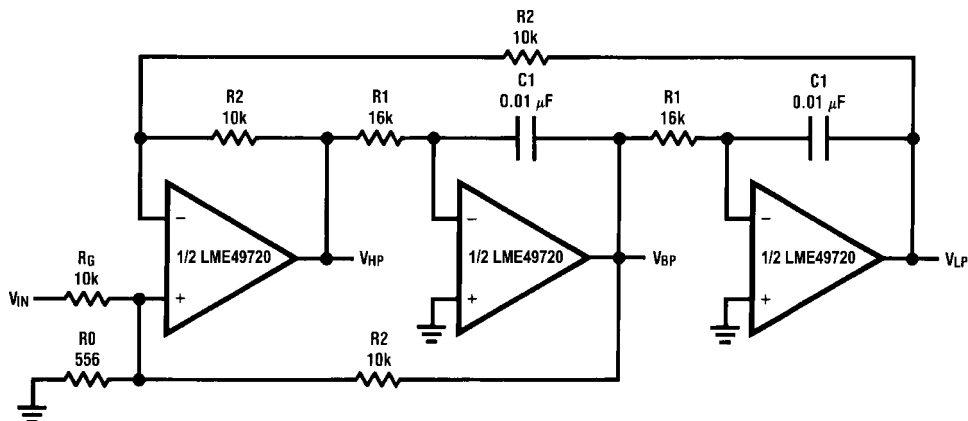
30003837

$$f_0 = \frac{1}{2\pi C1 R1}, \; Q = \frac{1}{2}\left(1 + \frac{R2}{R0} + \frac{R2}{RG}\right), A_{BP} = QA_{LP} = QA_{LH} = \frac{R2}{RG}$$

Illustration is f_0 = 1 kHz, Q = 10, A_{BP} = 1

AC/DC Converter

R5
20k

C1
10 μF

R2
20k

R3
10k

R4
20k

R1
20k

V_{IN}

1/2 LME49720

D1
1S1588

D2
1S1588

1/2 LME49720

$V_0 = |\overline{V_{IN}}|$

R6
15k

R7
6.2k
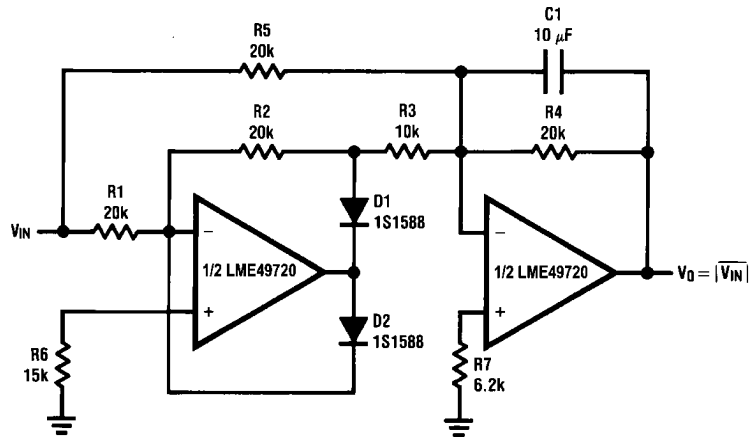
30003838

2 Channel Panning Circuit (Pan Pot)

3.41R1
51k

R1
15k

R1
15k

1/2 LME49720

V_{O1}

V_I

0.707R1
10k

R1
15k

R1
15k

1/2 LME49720

V_{O2}

3.41R1
51k

30003839

Line Driver

R2

R1

V_I

1/2 LME49720

R3
10k

V_{CC}

Q1

R9
10k

R7
33

V_0

R5
10k
BIAS

R8
33

Q2

R6
10k

$-V_{EE}$

30003840

Tone Control

BOOST←BASS→CUT

R1
11k

R2
100k

R1
11k

V_I

C1
0.05 μF

C1
0.05 μF

R3
11k

C2
0.005 μF

1/2 LME49720

V_0

R5
3.6k

R5
3.6k

R4
500k

BOOST←TREBLE→CUT
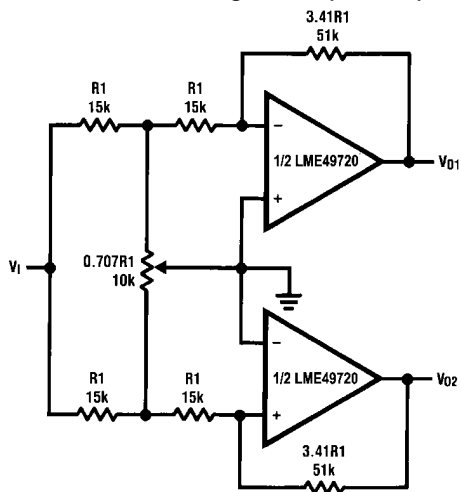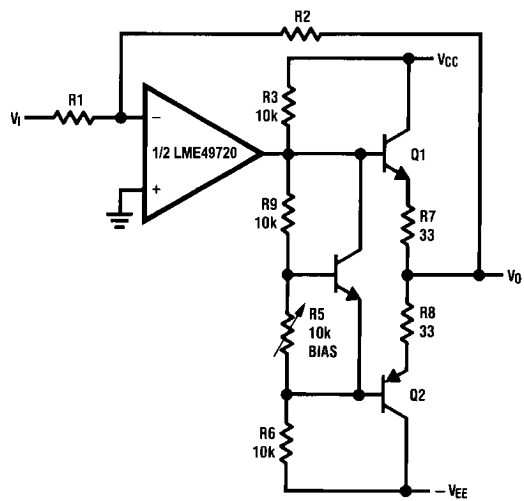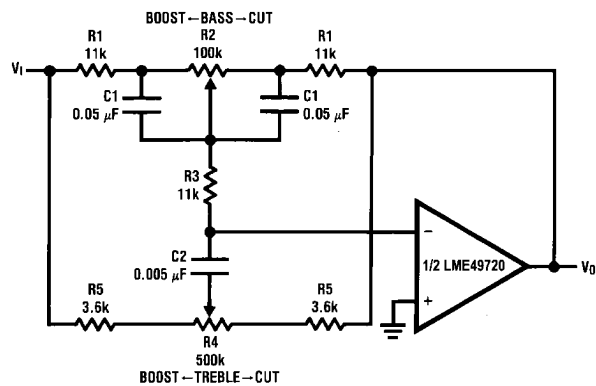
300038p0

$$f_L = \frac{1}{2\pi R2 C1}, \quad f_{LB} = \frac{1}{2\pi R1 C1}$$

$$f_H = \frac{1}{2\pi R5 C2}, \quad f_{HB} = \frac{1}{2\pi(R1 + R5 + 2R3)C2}$$

Illustration is:

f_L = 32 Hz, f_{LB} = 320 Hz

f_H =11 kHz, f_{HB} = 1.1 kHz
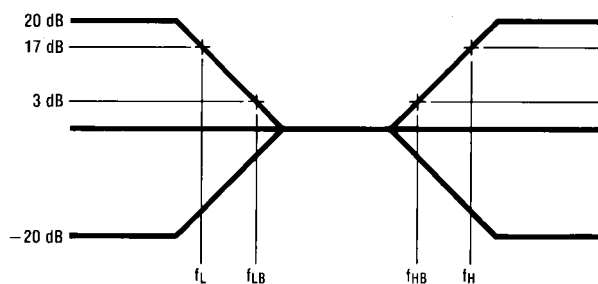
30003842

RIAA Preamp
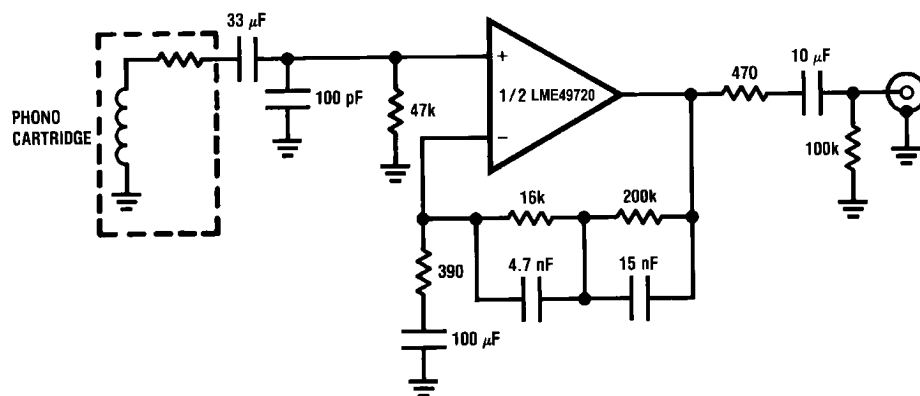
30003803

A_v = 35 dB

E_n = 0.33 µV

S/N = 90 dB

f = 1 kHz

A Weighted

A Weighted, V_{IN} = 10 mV

@f = 1 kHz

Balanced Input Mic Amp
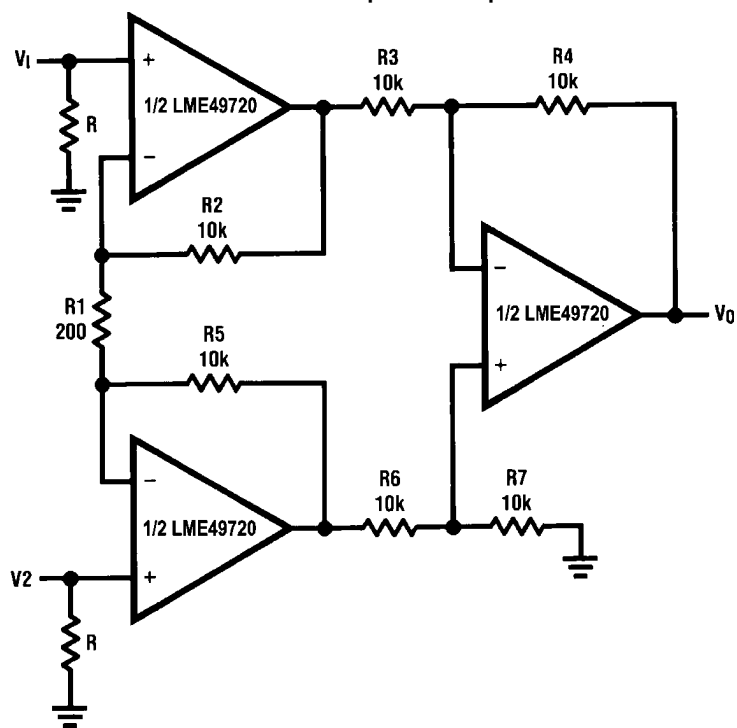
30003843

If R2 = R5, R3 = R6, R4 = R7

$$V0 = \left(1 + \frac{2R2}{R1}\right)\frac{R4}{R3}(V2 - V1)$$

Illustration is:
   V0 = 101(V2 − V1)

10 Band Graphic Equalizer

CUT 20k BOOST
C2
R2
C1
1/2 LME49720
R1
V_I 3k
fo1
fo2
fo3
fo4
fo5
fo6
fo7
fo8
fo9
fo10
1/2 LME49720
V_0
3k

30003844
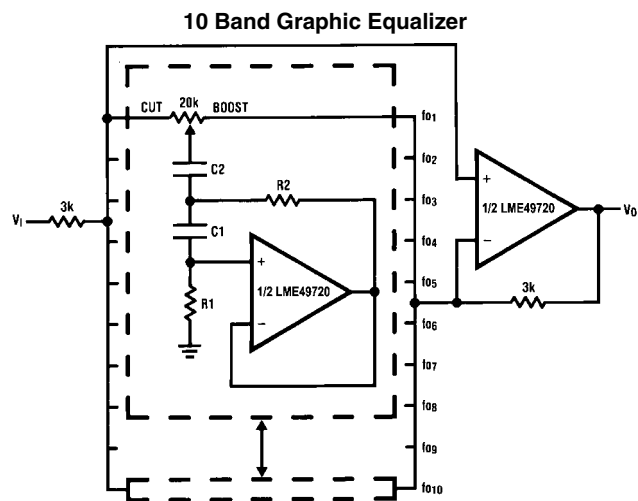
fo (Hz)	C_1	C_2	R_1	R_2
32	0.12μF	4.7μF	75kΩ	500Ω
64	0.056μF	3.3μF	68kΩ	510Ω
125	0.033μF	1.5μF	62kΩ	510Ω
250	0.015μF	0.82μF	68kΩ	470Ω
500	8200pF	0.39μF	62kΩ	470Ω
1k	3900pF	0.22μF	68kΩ	470Ω
2k	2000pF	0.1μF	68kΩ	470Ω
4k	1100pF	0.056μF	62kΩ	470Ω
8k	510pF	0.022μF	68kΩ	510Ω
16k	330pF	0.012μF	51kΩ	510Ω

Note 9: At volume of change = ±12 dB

Q = 1.7

Reference: "AUDIO/RADIO HANDBOOK", National Semiconductor, 1980, Page 2–61

Revision History

Rev	Date	Description
1.0	03/30/07	Initial release.

Physical Dimensions inches (millimeters) unless otherwise noted

4.9±0.1
[.193±.004]
NOT INCLUDING MOLD FLASH
0.15[.006] MAX PER END

6±0.2
[.236±.008]

3.9±0.1
[.154±.004]
NOT INCLUDING
MOLD FLASH

PIN 1 ID

8X (0.6)
[.02]

8X (2.2)
[.09]

(5.2)
[.20]

6X (1.27)
[.05]

RECOMMENDED LAND PATTERN

1.35-1.75
[.053-.069]
TYP

6X 1.27
[.050]

0.1[.004] C

8X 0.35-0.51
[.0138-.0200]
TYP

0.25[.010] M C A S B S

(1.45)
[.057]

0.25
[.010]
GAGE PLANE

0.10-0.25
[.004-.010]
TYP

R0.18±0.02
[.007±.0008]

R0.23±0.02
[.009±.0008]

0°-8°

0.41-1.27
[.016-.050]

(1.04)
[.041]

45° X 0.25-0.50
[.010-.020]

SEATING
PLANE

0.190-0.248
[.0075-.0098]
TYP

CONTROLLING DIMENSION IS MILLIMETER
VALUES IN [ ] ARE INCHES
DIMENSIONS IN ( ) FOR REFERENCE ONLY

M08A (Rev L)

**Narrow SOIC Package
Order Number LME49720MA
NS Package Number M08A**

0.373 - 0.400
(9.474 - 10.16)

0.090
(2.286)

0.092
(2.337) DIA

PIN NO. 1 IDENT

OPTION 1

0.250 ± 0.005
(6.35 ± 0.127)

0.032 ± 0.005
(0.813 ± 0.127)
RAD

PIN NO. 1 IDENT

OPTION 2

0.280
(7.112) MIN

0.300 - 0.320
(7.62 - 8.128)

0.030
(0.762) MAX

0.040
(1.016) TYP

20° ± 1°

0.039
(0.991)

0.145 - 0.200
(3.683 - 5.080)

0.130 ± 0.005
(3.302 ± 0.127)

0.125 - 0.140
(3.175 - 3.556)

0.020
(0.508)
MIN

95° ± 5°

0.009 - 0.015
(0.229 - 0.381)

0.125
(3.175)
DIA
NOM

0.325 +0.040
     −0.015

(8.255 +1.016
      −0.381)

0.065
(1.651)

90° ± 4°
TYP

0.018 ± 0.003
(0.457 ± 0.076)

0.100 ± 0.010
(2.540 ± 0.254)

0.045 ± 0.015
(1.143 ± 0.381)

0.050
(1.270)

0.060
(1.524)

N08E (REV F)

**Dual-In-Line Package
Order Number LME49720NA
NS Package Number N08E**

Ø.130±.015
[3.3±0.38]

Ø.360±.010
[9.14±0.25]

Ø.200±.005
[5.08±0.12]

.037±.008
[0.94±0.2]

45° TYP

.031±.003
[0.79±0.07]

REFERENCE PLANE

.500 MIN
[12.7]

.175±.010
[4.45±0.25]

SEATING PLANE

Ø.325±.010
[8.26±0.25]

.0175±.0015 TYP
[0.445±0.038]

.015-.040
[0.38-1.02]

.035 MAX
[0.89]

.025 MAX
[0.64]
UNCONTROLLED
LEAD DIA

H08C (Rev F)

CONTROLLING DIMENSION IS INCH
VALUES IN [ ] ARE IN MILLIMETERS
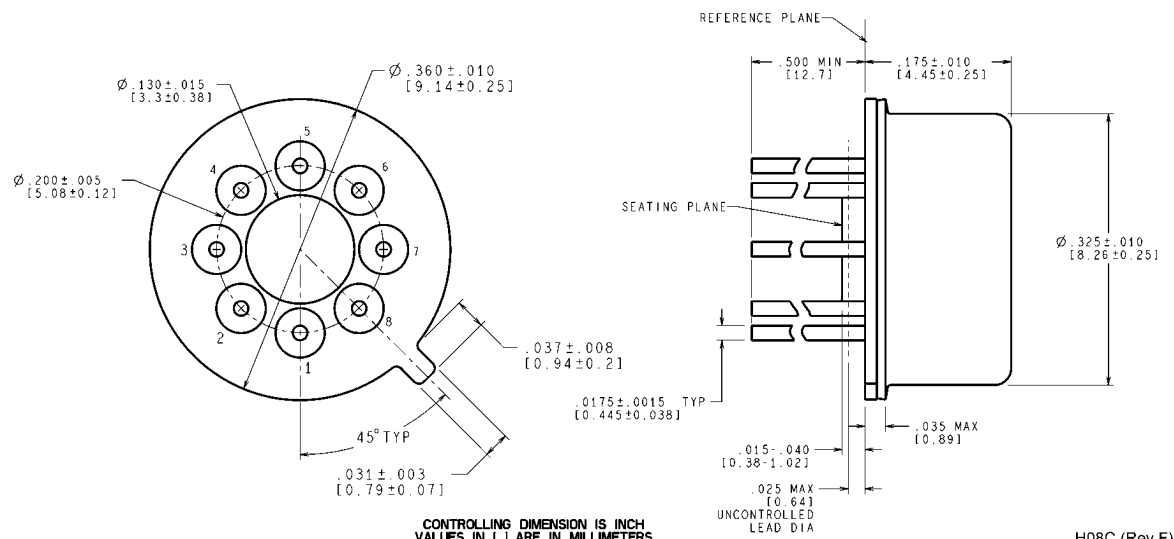
**TO-99 Metal Can Package
Order Number LME49720HA
NS Package Number H08C**

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