

06/27/2012

### High Power Factor Flyback LED Controller with HV Start-up

Rev: 00

#### **General Description**

The LD7830H is a HV start-up Flyback PFC controller, specially designed for LED lighting appliances. It operates in transition(TM) mode and integrates with complete protections required for safety and therefore it's an excellent solution to minimize the components counts. Those make it ideal for cost- effective applications.

With HV start-up technology, high power factor and TM control, it's capable to reduce the start-up time and resistor loss greatly. The circuit can easily achieve PF>0.90 to meet most internal standard requirements.

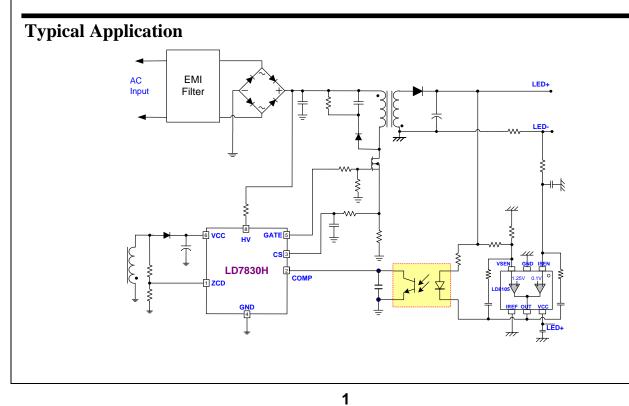
With completed protection built inside this IC, such as over voltage protection (OVP), over current protection (OCP), over load protection (OLP), over temperature protection (OTP), and short circuit protection (SCP), It enable the circuit to meet most safety requirements either in normal and abnormal test.

#### **Features**

- High voltage (600V) startup circuit
- High Power Factor Flyback PFC controller
- High-Efficiency Transition mode operation
- Wide UVLO (16V<sub>ON</sub> and 7.5 V<sub>OFF</sub>)
- VCC OVP (Over Voltage Protection)
- OLP(Over Load Protection)
- OCP (Cycle by Cycle Current Limiting) Vcs=0.5V
- Internal OTP (Over Temperature Protection)
- 500/-800mA Driving Capability
- Internal OTP function

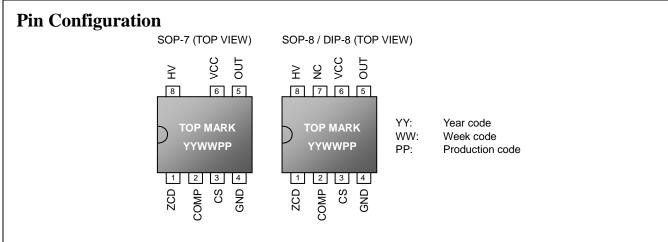
#### **Applications**

- LED Power Supply
- Flyback PFC Power Supply



Leadtrend Technology Corporation www.leadtrend.com.tw LD7830H-DS-00 June 2012





### **Ordering Information**

Part number	Package	Top Mark	Shipping
LD7830H GR	SOP-7	LD7830HGR	2500 /tape & reel
LD7830H GS	SOP-8	LD7830HGS	2500 /tape & reel
LD7830H GN	DIP-8	LD7830HGN	3600 /tube /Carton

The LD7830H is RoHs compliant/ Green Packaged.

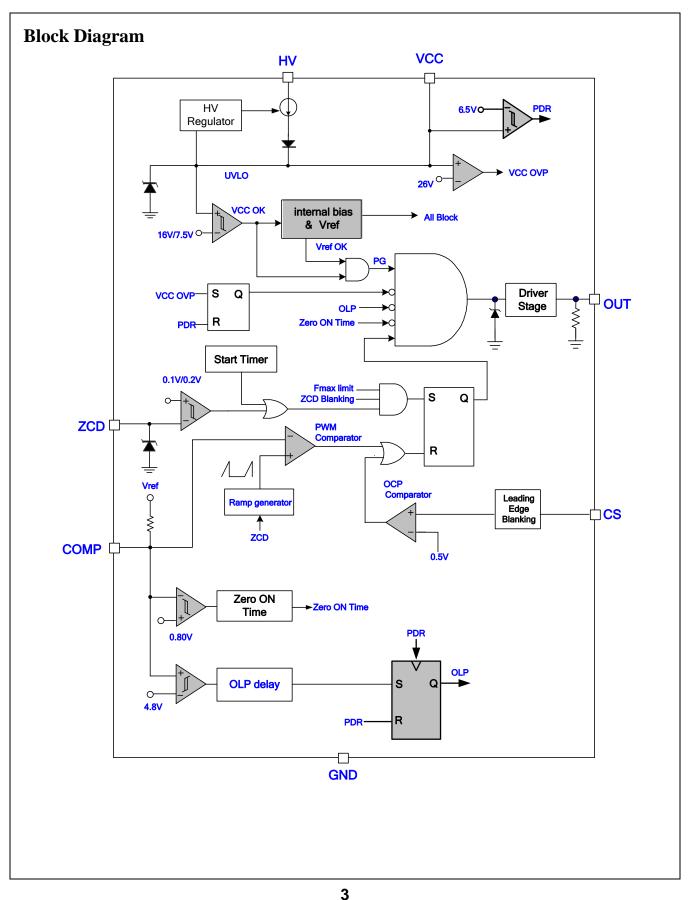
#### **Protection Mode**

Part number	VCC OVP	OLP
LD7830H	Latch	Latch

### **Pin Descriptions**

Pin	NAME	FUNCTION
1	ZCD	Quasi resonance detector and programmable maximum ON-time.
	COMP	Feedback pin. Connect with a photo-coupler to close the control loop and achieve regulation.
3	CS	Current sense pin, to sense the MOSFET current for OCP
4	GND	Ground
5	OUT	Gate drive output to drive the external MOSFET
6	VCC	Power source VCC pin
7	NC	Not connected.
8	HV	Connect this pin to positive terminal of main bulk cap to provide startup current for controller. Once Vcc is UVLO on , the HV loop will open and turn off internal current source to minimize the power loss.







### **Absolute Maximum Ratings**

Supply Voltage VCC	-0.3 ~30V
High voltage pin, HV	-0.3~600V
OUT	-0.3 ~VCC +0.3V
COMP, CS, ZCD	-0.3 ~6V
Maximum Junction Temperature	150°C
Operating Junction Temperature Range	-40°C to 125°C
Operating Ambient Temperature Range	-40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Package Thermal Resistance (SOP-8, SOP-7 $\theta_{JA}$ )	160°C/W
Package Thermal Resistance (DIP-8, θ <sub>JA</sub> )	100°C/W
Power Dissipation (SOP-8, SOP-7, at Ambient Temperature = 85°C)	250mW
Power Dissipation (DIP-8, at Ambient Temperature = 85°C)	400mW
Lead temperature (Soldering, 10sec)	260°C
ESD Voltage Protection, Human Body Model (except HV pin)	2.5KV
ESD Voltage Protection, Machine Model	250 V

#### Caution:

Stress exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stress above Recommended Operating Conditions may affect device reliability.

### **Recommended Operating Conditions**

ltem	Min.	Max.	Unit
Vcc pin capacitor	4.7	47	μF
Comp pin capacitor	1	10	μF



### **Electrical Characteristics**

(V<sub>CC</sub>=15.0V,  $T_A = 25^{\circ}C$  unless otherwise specified.)

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS
High voltage Supply (HV Pin)					
High voltage ourrent Source	V <sub>CC</sub> =4V , HV=80V	1.0	1.3	1.6	mA
High-voltage current Source	V <sub>CC</sub> =9V,HV=80V	2.6	3.1	3.6	mA
Off-state Leakage current	V <sub>CC</sub> >UVLO <sub>(ON)</sub> , HV=500V	0		35	μA
Supply Voltage (VCC Pin)			_		
Startup Current	V <sub>CC</sub> <uvlo on<="" td=""><td>55</td><td>75</td><td>95</td><td>μA</td></uvlo>	55	75	95	μA
	V <sub>COMP</sub> =0V, ZCD=0		1		mA
Operating Current	V <sub>COMP</sub> =3V, ZCD=0		1.45		mA
(with 1nF load on OUT pin)	V <sub>CC</sub> OVP		0.3		mA
	OLP		0.3		mA
UVLO (OFF)		6.9	7.5	8.1	V
UVLO (ON)		15.0	16	17.0	V
HV Self Bias (Linear Regulator)		8.1	8.8	9.5	V
De-Latch VCC Voltage	PDR (Power Down Reset)	6.0	6.5	7.0	V
VCC OVP Level		25	26	27	V
Voltage Feedback(Comp Pin)					
Short circuit current	V <sub>COMP</sub> =0	0.4	0.5	0.6	mA
Open loop voltage		5.0	5.3	5.8	V
OLP Trip Level		4.65	4.8	4.95	V
Zero ON-time Threshold			0.8		V
Zero ON-time Hysteresis			25		mV
OLP Delay Time		225	270	315	ms
Current Sensing (CS Pin)					
	I <sub>ZCD</sub> <110μA	0.47	0.5	0.53	V
Current Sense Input Threshold Voltage	I <sub>ZCD</sub> >490μA	0.322	0.35	0.378	V
Soft Start Time	*		8		ms
Input bias current	V <sub>CS</sub> =0V~0.5V	0		1.5	μA
LEB time		240	300	360	ns



PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS
Zero Current Detector (ZCD Pin)					
Upper Clamp Voltage	I <sub>DET</sub> =100μA	5.0	5.6	6.0	V
Lower Clamp Voltage	I <sub>DET=</sub> -2mA	0		-0.35	V
		0.05	0.1	0.15	V
Input Voltage Threshold	Hysteresis	0.05	0.15	0.2	V
Input bias current	V <sub>ZCD</sub> =1V~4V, OUT=OFF	0.0		1.0	μA
Programming Maximum ON-Time,	Fon-max (ZCD Pin)		1		
Programming Maximum ON-time	ZCD R <sub>ZCD</sub> =6kΩ,	4.8	5.82	6.8	μS
Maximum ON-time	ZCD R <sub>ZCD</sub> ≥34kΩ,	13.6	16	18.4	μS
Minimum (ON+OFF)-Time					•
Minimum (ON+OFF)-Time	F <sub>MAX</sub> (250kHz),	3.2	4	4.8	μS
Minimum OFF-Time		1.2	1.6	2.0	μS
Gate Drive Output (OUT Pin)					
Output Low Level	V <sub>CC</sub> =15V, I <sub>SINK</sub> =20mA	0		0.5	V
Output High Level	V <sub>CC</sub> =15V, I <sub>SOURCE</sub> =20mA	10		VCC	V
Output High Clamp Level	V <sub>CC</sub> =16V		13		V
Rising Time	V <sub>CC</sub> =15V, CL=1000pF		75	150	ns
Falling Time	V <sub>CC</sub> =15V, CL=1000pF		25	50	ns
Starter			•		•
Start Timer Period		100	150	200	μS
Internal OTP (Over Temp. Protectio	n)	•	•	•	•
OTP Trip level	*		140		°C
OTP Hysteresis	*		30		°C

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\*: These parameters are guaranteed by design.



#### **Application Information**

#### **Operation Overview**

The LD7830H is an excellent single-stage Flyback PFC controller for LED lighting applications. With several functions integrated, it's an ideal solution to minimize external components counts and the size.

In addition, the LD7830H is a voltage-mode TM PFC controller. The turn-on time of switch is fixed while the turn-off time is varied in the steady condition. Therefore, the switching frequency changes in accordance with the input voltage variation. The LD7830H provides functions of over-load protection, over-voltage protection, over-current protection, under-voltage lockout and integrated LEB of the current sensing. Also, the LD7830H requires no main-voltage sensing, superior to what the other traditional current mode PFC controllers behave for power saving.

### Internal High-Voltage Startup Circuit and Under Voltage Lockout (UVLO)

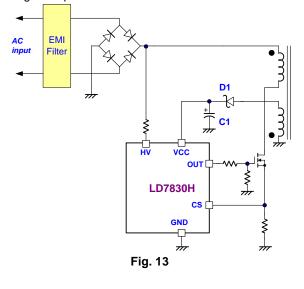
The traditional circuit provides startup current through a startup resistor to power on the PWM controller. However, it consumes significant power for it. In most cases, startup resistors carry large resistance. And, larger resistance spends more startup time.

To achieve the optimized topology, as shown in figure 13, LD7830H is implemented with a high-voltage startup circuit for such requirement. During startup, a high-voltage current source sinks from the full-bridge rectifier to provide startup current to charge Vcc capacitor C1. On condition of VCC below PDR, it consumes only 1.3mA for charge current. This protects the IC from damage in case the VCC pin is shorted to ground. In short start-up time, the charge current will increases to 3.1mA once VCC rises above PDR voltage threshold during start up. Meanwhile, it consumes only 75 $\mu$ A for Vcc supply current, so most of the HV current is reserved to charge

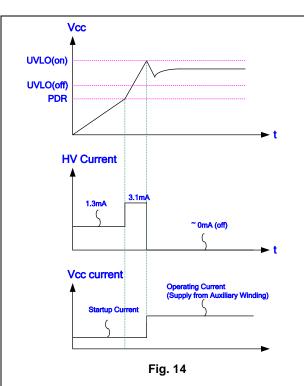
the Vcc capacitor. In using such configuration, the turn-on delay time will be almost no difference either in low-line or high-line conditions.

Once the Vcc voltage rises over UVLO(on), it will power on the LD7830H and further to deliver the gate drive signal. Then the high-voltage current source will be disabled and the supply current is provided from the auxiliary winding of the transformer. Therefore, the power loss is eliminated and power saving is enhanced.

An UVLO comparator is embedded to detect the voltage on Vcc pin to ensure the supply voltage enough to power on the LD7830H PWM controller and in addition to drive the power MOSFET. As shown in Fig. 14, a hysteresis is provided to prevent the shutdown from voltage dip during startup.







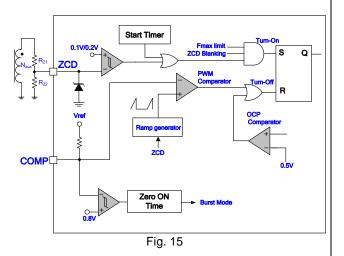
# Ramp Generator Block and Zero Current Detection (ZCD)

Fig. 15 shows typical ramp generator block and ZCD block. The COMP pin voltage and the output of the ramp generator block are compared to determine the MOSFET ON-time, as shown in Fig. 16.

A greater COMP voltage produces more ON-time. Using an external resistor to connect with ZCD pin to set the desired slope of the internal ramp, the user may program the maximum ON-time. Alternatively, the ON-time will also achieve its maximum when COMP pin voltage trip OLP trigger point.

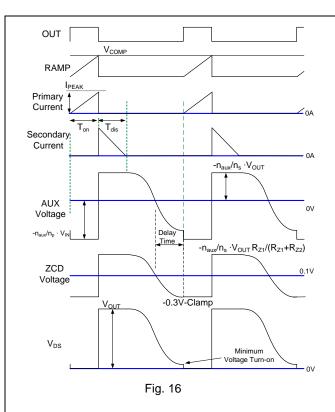
The maximum ON-time should be set according to the condition of the transformer, lowest AC line voltage, and maximum output power. A choice of optimum resistor value would result in best performance.

It shuts down the drive output if COMP pin voltage falls below zero ON-time threshold. This optimizes the efficiency for power saving in most conditions. The Zero Current Detection block will detect auxiliary winding signal to drive MOSFET as ZCD pin voltage drops to 0.1V. As ZCD pin voltage drops to 0.1V, the current through the transformer is below zero. This feature enables transition-mode operation. The ZCD comparator would not operate if ZCD pin voltage remains at above 0.2V. Once it drops below 0.1V, the zero current detector will act to turn on the MOSFET.



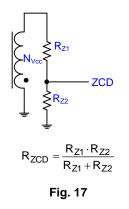
The 150 $\mu$ s timer generates a MOSFET turn-on signal if the driver output drops to low level for more than 150 $\mu$ s from the falling edge of the driver output. Fig. 16 shows typical ZCD-related waveforms. Since ZCD pin carries some capacitance, it produces some delay to the turn-on time caused from Rz1. During delay time, the junction capacitor of the MOSFET resonates with the primary inductor of the transformer and the drain-source voltage (V<sub>DS</sub>) decreases accordingly. So, the MOSFET consumes less voltage to turn on and it therefore minimizes the power dissipation.





#### **Programming Maximum ON-time**

LD7830H features adjustable maximum ON-time to limit power output in abnormal operation. The selection of maximum ON-time is subject to ZCD resistance as shown in Fig. 17. ZCD resistance can be obtained from below.



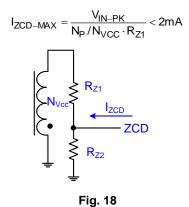
The following table is a suggestion for maximum ON-time setting.

LD7830H

R <sub>zcp</sub>	Max. Ton	Suggestion	
NZCD	(Тур.)	Caggeotion	
32k <r<sub>ZCD</r<sub>	16µs	36k	
28k <r<sub>ZCD&lt;32k</r<sub>	12.8µs	30k	
24k <r<sub>ZCD&lt;28k</r<sub>	10.7µs	26k	
20k <r<sub>ZCD&lt;24k</r<sub>	9.1µs	22k	
16k <r<sub>ZCD&lt;20k</r<sub>	8.0µs	18k	
12k <r<sub>ZCD&lt;16k</r<sub>	7.1µs	14k	
8k <r<sub>ZCD&lt;12k</r<sub>	6.4µs	10k	
R <sub>ZCD</sub> <8k	5.8µs	6k	

#### High/ Low Line OCP Compensation

LD7830H supports high/low line OCP compensation through aux-winding to reflect the magnitude of input voltage. When gate turns on, ZCD pin will source the current  $I_{ZCD}$  to clamp the voltage to zero as shown in Fig.18. According to this current information, OCP trigger level V<sub>CS-OFF</sub> will be determined by R<sub>Z1</sub> and input voltage. In order to avoid ZCD pin over rating,  $I_{ZCD}$  must be set less than 2mA.  $I_{ZCD-max}$  can be calculated as below:





The relation (typical) of  $V_{CS-OFF}$  and  $I_{ZCD}$  is shown in Table 1.

I <sub>ZCD</sub> (μA)	V <sub>CS-OFF</sub> (V)
<160	0.5
160~200	0.475
200~240	0.4625
240~290	0.45
290~350	0.425
350~390	0.4
390~450	0.375
>450	0.35

#### Table 1

#### **Output Drive Stage**

With typical 500mA/-800mA driving capability, an output stage of a CMOS buffer is incorporated to drive a power MOSFET directly. The output voltage is clamped at 13V to protect the MOSFET gate even when the VCC voltage is larger than 13V.

# Current Sensing and Leading-edge Blanking

The LD7830H detects the primary MOSFET current over CS pin for pulse-by-pulse current limit. The maximum voltage threshold of the current sensing pin is set at 0.5V. From above, the MOSFET peak current can be obtained from below.

$$I_{\text{PEAK}(\text{MAX})} = \frac{0.5\text{V}}{\text{R}_{\text{S}}}$$

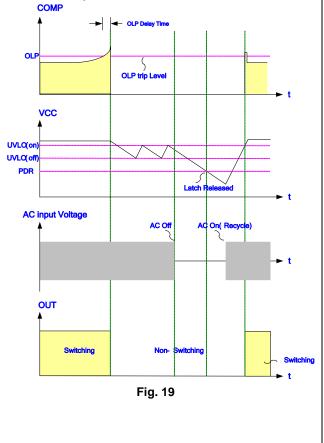
A leading-edge blanking (LEB) time is imbedded in the input of CS pin to prevent false-triggering from the current spike. In those low power applications, if the total pulse width of the turn-on spikes is less than LEB and the negative spike on the CS pin isn't over -0.3V, it is free to eliminate the R-C filter.

However, the total pulse width of the turn-on spike is determined by the output power, circuit design and PCB layout. It is strongly recommended to adopt a smaller

R-C filter for those large power applications to avoid CS pin being damaged from negative turn-on spike.

#### Over Load Protection (OLP) - Latch mode

The LD7830H features latch mode of smart OLP protection. Figure 19 shows the waveform under fault condition. The feedback system will force the voltage loop enter toward the saturation and thus pull the voltage high on COMP pin (VCOMP). When the VCOMP ramps up to the OLP threshold and stays there for more than OLP delay time, it will activate the protection to latch off the gate output and stop switching of the power circuit. The delay time is to prevent the false-triggering during power-on, turn-off transient or in peak load condition. Unless the over load condition is removed, the controller will remain latched until the Vcc drops below PDR. It is necessary to start another AC power-on recycling to resume the output.

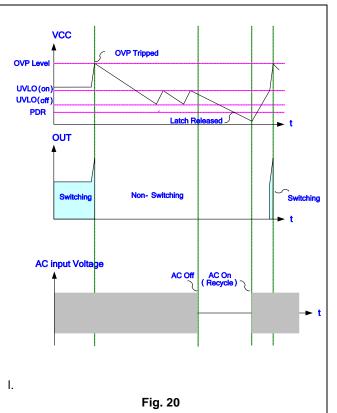




### OVP (Over Voltage Protection) on Vcc-Latched Mode Protection

The V<sub>GS</sub> ratings of the nowadays power MOSFETs is mostly for 29V maximum. To protect the V<sub>GS</sub> from fault conditions, LD7830H is implemented with OVP function on Vcc. If Vcc voltage is over the OVP threshold, the output gate drive circuit will be shut down simultaneously to stop switching of the power MOSFET.

The Vcc OVP function in LD7830H is a latch-off type protection. If the OVP condition occurs (usually caused by the feedback loop opened), Vcc will not recover until AC power turns off. Figure 20 shows its operation. The Vcc drops since there's no AC power present. The de-latch level of OVP is defined by internal PDR. Over voltage protection could be activated after Vcc is below PDR level.



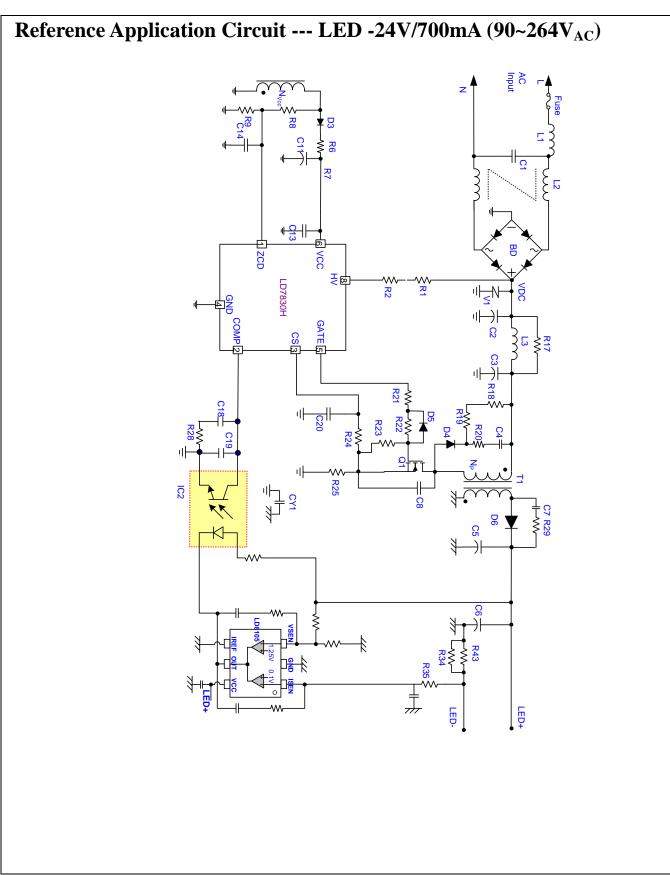
#### **Fault Protection**

There are several critical protections were integrated in the LD7830H to prevent the power supply or adapter from being damaged. Those damages usually come from open or short condition caused to LD7830H.

Once it happens in below conditions, the gate output will turn off immediately to protect the power circuit.

- 1. Comp pin floating
- 2. CS pin floating



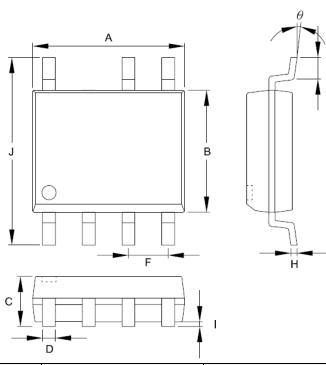




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### Package Information

SOP-7

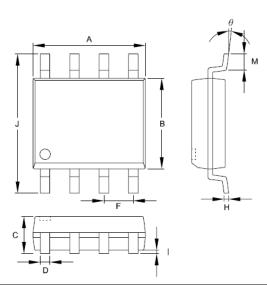


	Dimensions in Millimeters		Dimensio	ns in Inch
Symbols	MIN	МАХ	MIN	МАХ
А	4.801	5.004	0.189	0.197
В	3.810	3.988	0.150	0.157
С	1.346	1.753	0.053	0.069
D	0.330	0.508	0.013	0.020
F	1.194	1.346	0.047	0.053
н	0.178	0.254	0.007	0.010
I	0.102	0.254	0.004	0.010
J	5.791	6.198	0.228	0.244
М	0.406	1.270	0.016	0.050
θ	0°	8°	0°	8°



### Package Information

SOP-8

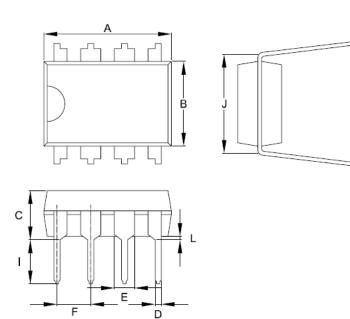


	Dimensions i	n Millimeters	Dimensio	ns in Inch
Symbols	MIN	МАХ	MIN	МАХ
А	4.801	5.004	0.189	0.197
В	3.810	3.988	0.150	0.157
С	1.346	1.753	0.053	0.069
D	0.330	0.508	0.013	0.020
F	1.194	1.346	0.047	0.053
Н	0.178	0.254	0.007	0.010
I	0.102	0.254	0.004	0.010
J	5.791	6.198	0.228	0.244
М	0.406	1.270	0.016	0.050
θ	0°	8°	0°	8°



### **Package Information**

DIP-8



Symbol	Dimension in Millimeters		Dimensions in Inches	
Symbol	Min	Мах	Min	Max
А	9.017	10.160	0.355	0.400
В	6.096	7.112	0.240	0.280
С		5.334		0.210
D	0.356	0.584	0.014	0.023
E	1.143	1.778	0.045	0.070
F	2.337	2.743	0.092	0.108
I	2.921	3.556	0.115	0.140
J	7.366	8.255	0.29	0.325
L	0.381		0.015	

#### **Important Notice**

Leadtrend Technology Corp. reserves the right to make changes or corrections to its products at any time without notice. Customers should verify the datasheets are current and complete before placing order.



### **Revision History**

Rev.	Date	Change Notice
00	6/27/2012	Original Specification