

Electronic Volume Control System for Car Audio

Overview

The LC7538JM is a fully equipped electronic volume IC which permits significant reductions in externally connected components while providing ample volume, balance, loudness, fader, bass and treble control functions.

- On-chip op amplifier for caching applications reduces external components.
- Reduced switching noise with silicon gate CMOS processor.
- All controls performed using serial data input (CCB).

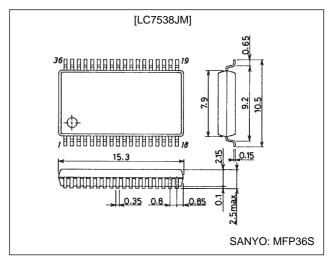
Features

- Volume : 81 positions ranging from 0 dB to -79
 - dB (in 1 dB increments) plus -∞. Separate left and right control provides
 - excellent balance function.
- Loudness : Loudness operation provided by externally attached CR to activate tap at the -20 dB position of the volume
 - ladder resistor.
- Fader : Fader function traversing 16 positions
 - with rear or front attenuated output only (these 16 positions consist of 2 dB step intervals ranging from 0 dB to -20 dB, 5 dB step intervals ranging from -20 dB to -45 dB, plus the end
 - settings of -60 dB and -∞).
- Bass and Treble: Using externally attached C (capacitor), the LC7538JM provides
 - bass-treble mutual 15-position control and formats a NF-form tone control circuit (LUX form).

Package Dimensions

unit: mm

3204-MFP36S



Specifications

Absolute Maximum Ratings at $Ta = 25^{\circ}C$, $V_{SS} = 0$ V

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V _{DD} max	V _{DD}	11	V
	V _{IN} max1	CL, DI, CE	$V_{SS} - 0.3$ to $V_{DD} + 0.3$	V
Maximum input voltage	V _{IN} max2	LTIN, RTIN, L5dBIN, R5dBIN, L1dBIN, R1dBIN, LFIN, RFIN	V _{SS} – 0.3 to V _{DD} + 0.3	V
Allowable power dissipation	Pd max	Ta ≤ 85°C	210	mW
Operating temperature	Topr		-40 to +85	°C
Storage temperature	Tstg		-50 to +125	°C

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Allowable Operation Conditions at $Ta=25^{\circ}C,\,V_{SS}=0~V$

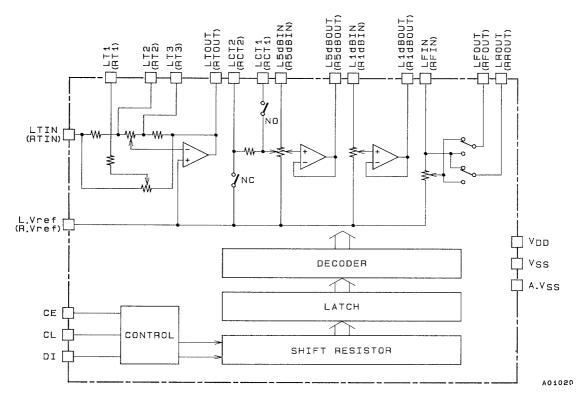
Description	Symbol	Conditions	Ratings			
Parameter			min	typ	max	Unit
Supply voltage	V _{DD}	*1	7.0		10.0	V
Input high level voltage	V _{IH}	CL, DI, CE	4.0		V _{DD}	V
Input low level voltage	V _{IL}	CL, DI, CE	V _{SS}		1.0	V
Input amplitude voltage	V _{IN}	LTIN, RTIN, L5dBIN, R5dBIN, L1dBIN, R1dBIN, LFIN, RFIN	V _{SS}		V _{DD}	Vp-p
Input pulse width	t _{øW}	CL	1			μs
Setup time	tSETUP	CL, DI, CE	1			μs
Hold time	tHOLD	CL, DI, CE	1			μs
Operating Frequency	fopg	CL			500	kHz

Note: 1. A capacitor rated at 2000 pF or less should be installed between all power supply pins and V_{SS}.

Electrical Characteristics at $Ta=25^{\circ}C,\,V_{DD}=9~V,\,V_{SS}=0~V$

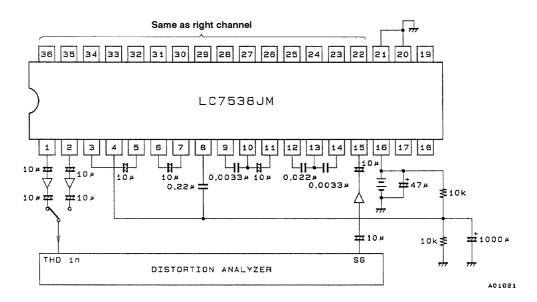
_				Ratings	Ratings	
Parameter	Symbol	Conditions min	min	typ	max	Unit
Tatal bassassia distantias	THD (1)	V _{IN} = 1 Vrms, f = 1 kHz, total overall flat		0.04		%
Total harmonic distortion	THD (2)	V _{IN} = 1 Vrms, f = 20 kHz, total overall flat		0.06		%
Crosstalk	СТ	V_{IN} = 1 Vrms, f = 1 kHz, total overall flat, Rg = 1 k Ω	60	87		dB
Maximum Output Reduction	Vo min	$V_{IN} = 1$ Vrms, f = 1 kHz, main volume $-\infty$, fader volume $-\infty$, C = 1000 μ F between Vref and V _{SS} for L/R		82		dB
	R _{VOL} (1)	5 dB step	15	25	35	kΩ
	R _{VOL} (2)	1 dB step	12	20	28	kΩ
All Resistance Value	R _{FADER}		12	20	28	kΩ
	R _{BASS}		48	80	112	kΩ
	R _{TREBLE}		30	50	70	kΩ
Input high level current	I _{IH}	VI = 8 V (CL, CE, DI pins)			10	μA
Input low level current	I _{IL}	VI = 0 V (CL, CE, DI pins)	-10			μA
Output noise voltage	V _N	All overall flat (IHF-A), Rg = 1 k Ω		7.5	15	μV
Current dissipation	I _{DD}	V _{DD} -V _{SS} = 10 V		15	21	mA
Analog switch on resistance		CT1	1.8	3.0	4.2	kΩ
		Between CT2 and Vref	0.6	1.0	1.4	kΩ
	R _{ON}	Fader S1 to S4	1.8	3.0	4.2	kΩ
		-∞	0.6	1.0	1.4	kΩ
		All other cases	6.0	10.0	14.0	kΩ

Equivalent Circuit Block Diagram



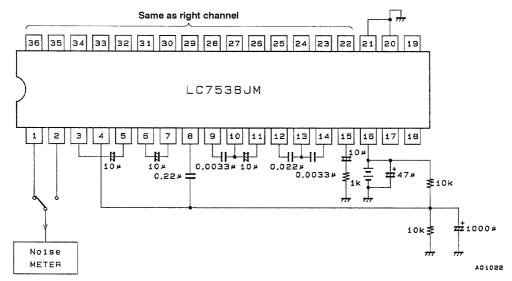
Test Circuit

a) Total Harmonic Distortion



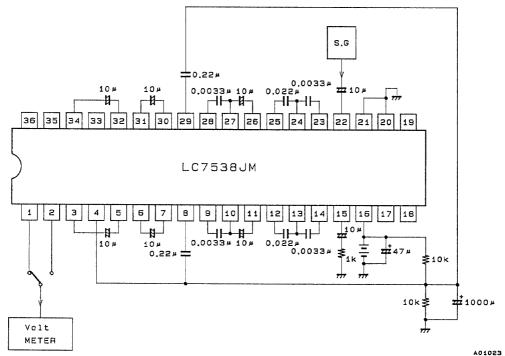
Unit (resistance: Ω , capacitance: F)

b) Output Noise Voltage



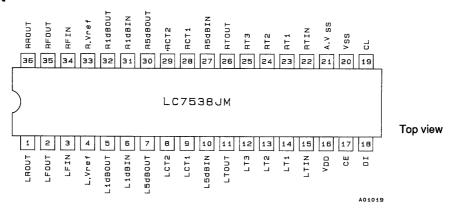
Unit (resistance: Ω , capacitance: F)

c) Crosstalk



Unit (resistance: Ω, capacitance: F)

Pin Assignment



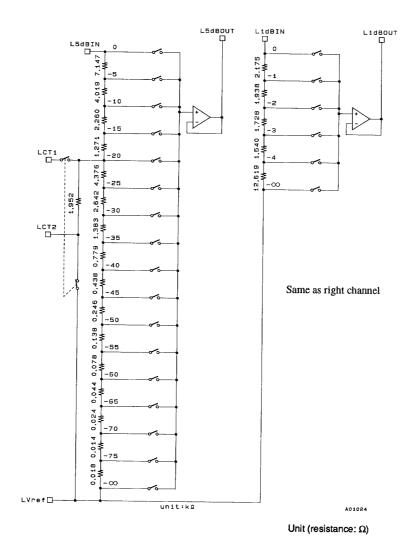
Pin Descriptions

Pin name	Pin No.	Description	Remarks		
LROUT	1	These pins function as output pins for the fader. Output reduction for rear and	۷۵۵-↔		
LFOUT	2	front is performed separately for each. Attenuation capacity is unified for	★		
RROUT	36	both left and right. Step positioning is designed using an open circuit so that	★		
RFOUT	35	reception is performed using high impedance.	VSS ,,,,		
LFIN	3	When utilizing the fader function, these pins function as input pins.	V _{DD} - Q -		
RFIN	34	Low impedance driven.	VSS 777 A02182		
LVref	4	These pins are common pins for fader volume, tone and main volume. The pattern impedance connected here should be lowered as much as possible. LVref and RVref are not connected to V _{SS} . Connections for LVref and RVref to V _{SS} should be established externally to	VDD-O		
RVref	33	match all specifications. Notably, attention should be paid to capacity since capacitors are subject to residual resistance during volume output reduction when installed between LVref (RVref) and V _{SS} as is the case with single power sources. • Normally, high voltage applied from V _{DD} .	1/2VDD ZZZ Ø		
L1dBOUT	5	These pins are output pins for the 1 dB step attenuator located in the section main volume.	OV DD		
R1dBOUT	32		AVSS A02184		
L1dBIN	6	These pins are input pins for the 1 dB step attenuator located in the section main volume.	VDD		
R1dBIN	31	Low impedance driven.	VSS ,,, A02182		
L5dBOUT	7	These pins are output pins for the 5 dB step attenuator located in the section main volume.	V DD		
R5dBOUT	30	main volume.	AVSS A02184		
LCT1	9				
LCT2	8	These pins are for loudness control. Connect a high-band compensation capacitor between CT1 to 5dB IN and a low-band compensation capacitor.			
RCT1	28	between CT2 to Vref.			
RCT2	29				
L5dBIN	10	These pins are input pins for the 5 dB step attenuator located in the section main volume.	V _{DD} -⊖-		
R5dBIN	27	Low impedance driven.	VSS ,,,, A02182		
LTOUT	11	These pins are output pins for tone control.	ovon		
RTOUT	26		Vref AVSS WVSS A02185		

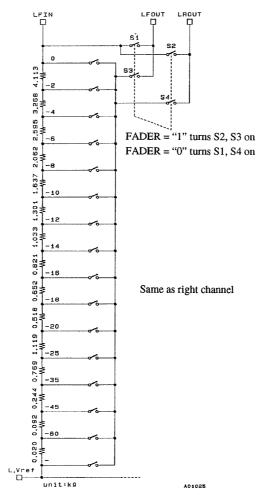
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Pin name	Pin No.	Description	Remarks		
LT3	12				
LT2	13	These pins are for connecting bass and treble compensation for the tone	VDD-O-W-		
LT1	14	circuit.			
RT3	25	Connect a high-band compensation capacitor between T1 and T2. Connect a low-band compensation capacitor between T2 and T3.	*		
RT2	24	Connect a low-particompensation capacitor between 12 and 13.	VSS ,,,		
RT1	23				
LTIN	15	These pins are tone control input pins.	V DD		
RTIN	22	Low impedance driven.	VSS ,,,,		
V _{DD}	16	Supply voltage pin.			
A. V _{SS}	21	Ground pin for on-chip op amp.	9 V D D A02187		
V _{SS}	20	Ground pin for internal logic.	0 V D D		
CE	17	This is the chip enable pin. According to the timing of the switch from high to low, data is written to an internal latch and all analog switches operate. Data transfer with high-level switches to enable.	OVDD OVSS AQ2188		
DI	18	These are input pins for the clock and serial data for control.	°V _{DD}		
CL	19		VSS Ap2188		

Equivalent Circuit for Main Volume Section



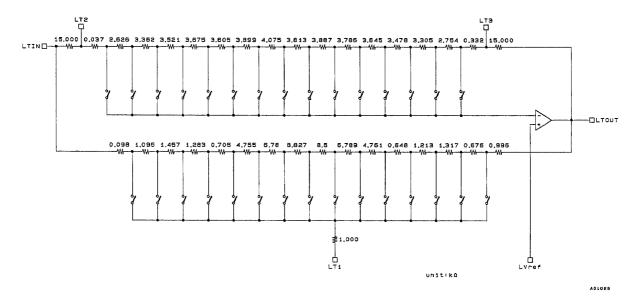
Equivalent Circuit for Fader Volume Section



When data of $-\infty$ is transferred to main volume control 1 dB step, S1 and S2 open and S3, S4 are turned on simultaneously.

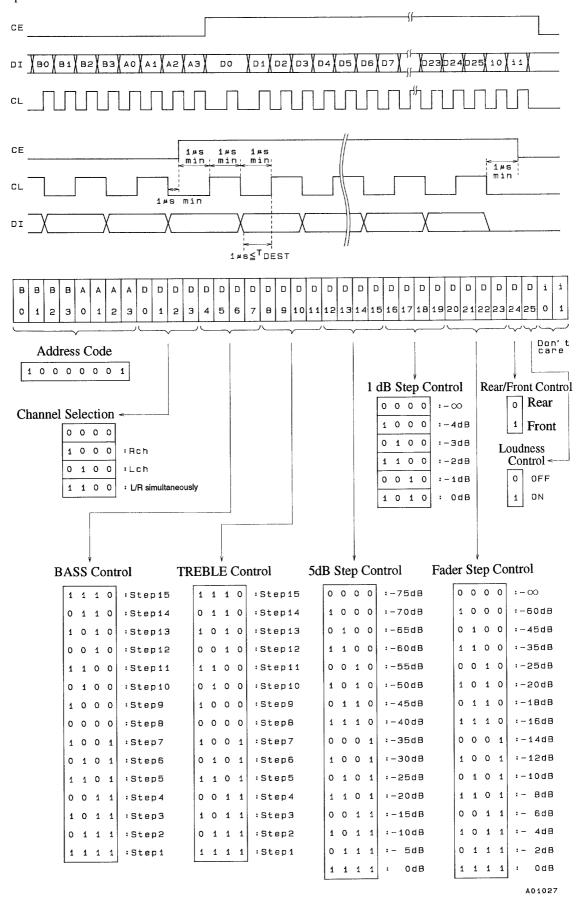
Unit (resistance: Ω)

Equivalent Circuit for Tone Section

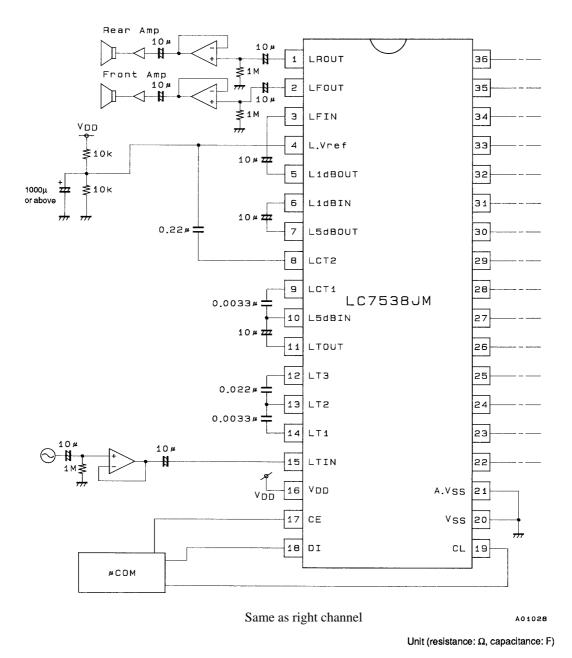


Control System Timing and Data Format

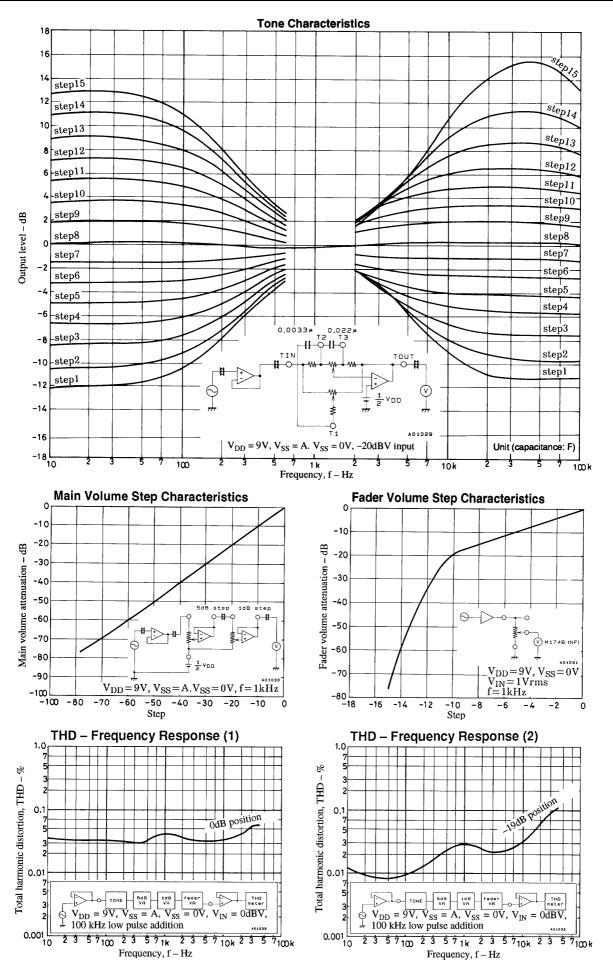
Controlling of LC7538JM involves the input of regulating serial data to CE, CL and DI pins. Data format consists of 36 bits composed of an 8-bit address and 28-bit data.

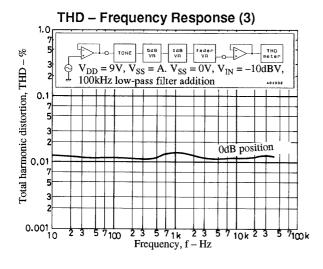


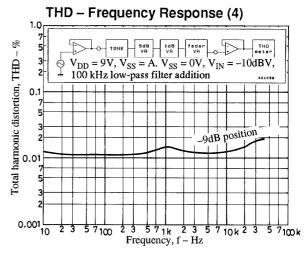
Sample Application Circuit

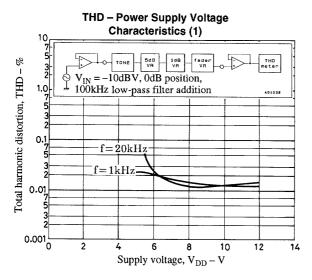


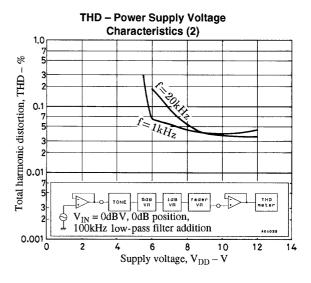
Note: Bipolar electrolytic capacitors should be used as widely as possible where others are not recommended directly.

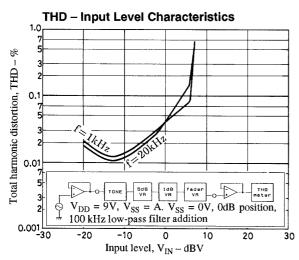


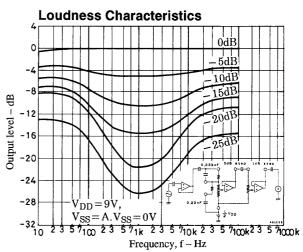












Loudness External Constant Calculation Sample

First, refer to page 7 where the 5 dB step internal equivalent circuit for the LC7538JM is shown. Using this information, an external constant for loudness can be added to establish a simplified circuit for computation as shown in Figure 1. Computations gaining a 5 dB boost with f = 100 Hz using this configuration are shown in the following. (f = 100 Hz and 5 dB boost)

Within Figure 1, when R and C are defined as:

 $R1 = R2 = 10 \text{ k}\Omega$

 $R3 = 1 k\Omega$

C1 = Z1, C2 = Z2, then the following equation can be established:

$$V_{\mbox{OUT}} = \frac{\frac{\mbox{R2} \mbox{ (R3 + Z2)}}{\mbox{R2 + R3 + Z2}}}{\frac{\mbox{R1 \cdot Z1}}{\mbox{R1 + Z1}} + \frac{\mbox{R2} \mbox{ (R3 + Z2)}}{\mbox{R2 + R3 + Z2}} = -20 \mbox{dB}$$

$$V_{OUT} = \frac{\frac{R2 (R3 + 10 \cdot Z2)}{R2 + R3 + 10 \cdot Z2}}{\frac{R1 \cdot 10 \cdot Z1}{R1 + 10 \cdot Z1} + \frac{R2 (R3 + 10 \cdot Z2)}{R2 + R3 + 10 \cdot Z2}} = -15 dB$$

thereby resulting in,

 $Z1 \neq 178.3 \text{ k}\Omega$ and $Z2 = 176 \Omega$.

Under such conditions where f=1 kHz, specifications may be satisfied if C (capacitor) having these impedances is supplied externally. The end result is that C1=893 pF and C2=0.9 μ F.

R1, R2 and R3 : LC7538JM on-chip resistors

C1 : External high-band compensation capacitorC2 : External low-band compensation capacitor

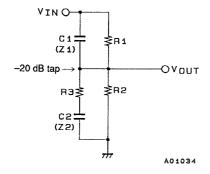


Figure 1

Notes for Above Applications

- When the power supply is turned on, the internal analog switch becomes inexact. Until data is set, counter measures such as those required for muting are performed externally.
- In order to prevent crossover into the analog system of high-frequency digital signals transferred to the CL, DI and CE pins, transfer along these signal lines should occur along shielded lines, or the signal lines should be protected by using the grounding pattern or the circuit.
- Capacitors of at least 2000 pF must be inserted between each power supply pin and the V_{SS} pin.
- For volume steps with large attenuation levels (over -20 dB), when the loudness circuit is off the high frequency region (above about 4 kHz) will be attenuated by about 3 dB relative to the low frequency region (about 400 Hz) due to the influence of the resistance of the loudness circuit analog switch. Therefore we recommend using tone control compensation together with the volume step described above.
- When sending data immediately after power on, send data as follows:
 - When sending independent left and right data, send data at least four times.

— Alternatively, when sending data that drives the left and right channels at the same time, send the data at least twice.

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