

**SANYO**

No. 3781B

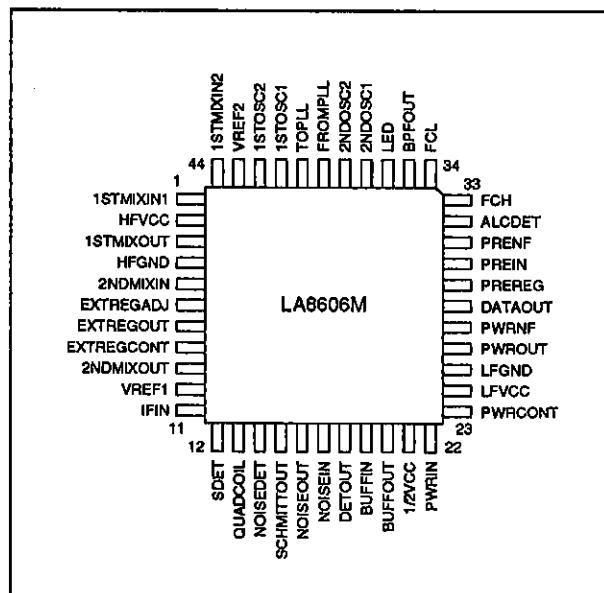
**Cordless Telephone Signal Processor****OVERVIEW**

The LA8606M is a signal processor IC for use in cordless telephone. It incorporates 1st and 2nd IF mixers, IF amplifier, quadrature detector, noise detector and noise amplifier on-chip, greatly reducing the number of external components required. Also incorporated are an S-meter driver, preamplifier, power amplifier, battery checker, LPF buffer, BPF and data shaper.

The LA8606M operates from a 1.8 to 6.0V supply and is available in 44-pin QIPs.

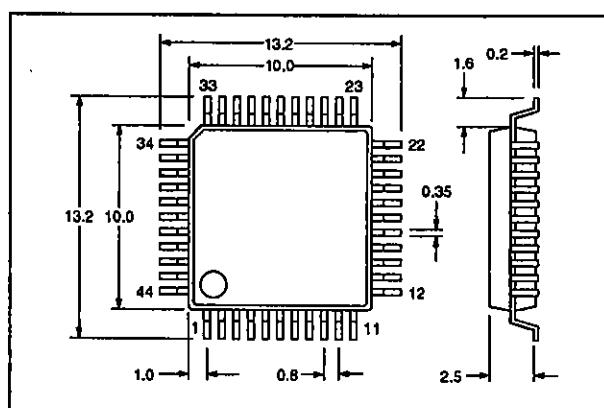
**FEATURES**

- 1st and 2nd IF mixers
- IF amplifier
- Quadrature detector
- Noise detector
- Noise amplifier
- S-meter driver
- LPF buffer, BPF and data shaper
- Controllable preamplifier for either the limiter or the ALC
- Power amplifier and power-down circuit
- Battery check function
- 1.8 to 6.0 V supply
- 44-pin QIP

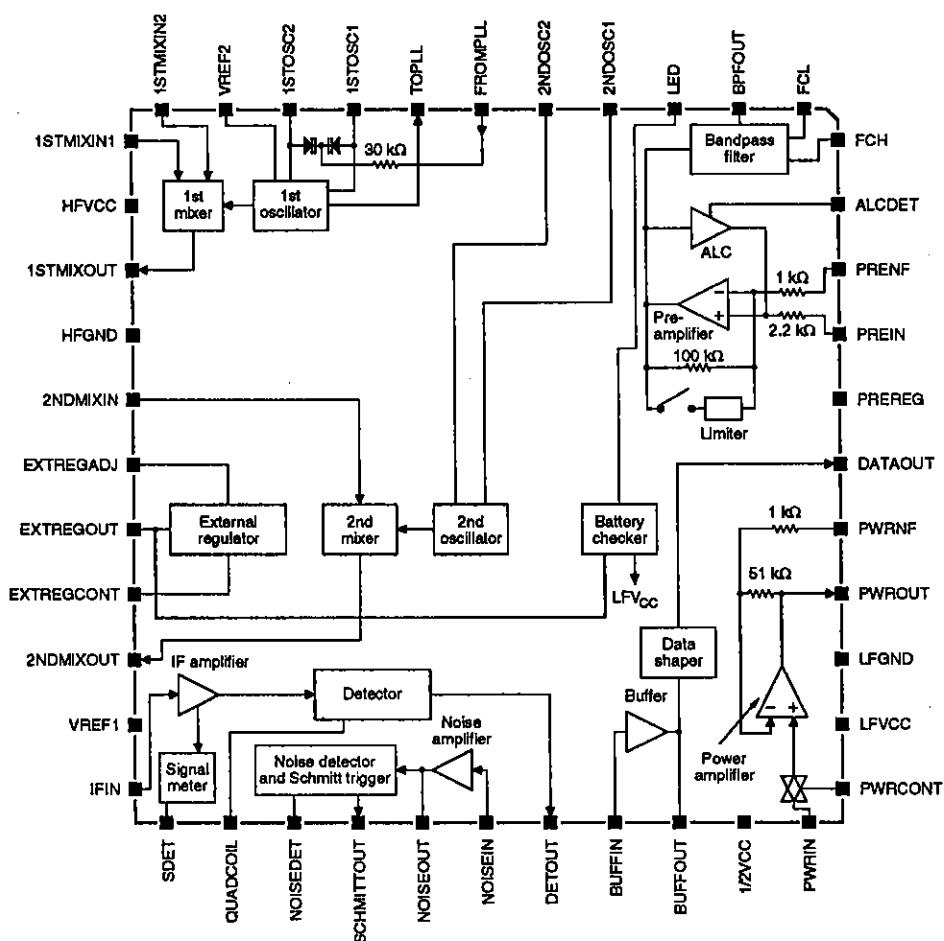
**PINOUT****PACKAGE DIMENSIONS**

Unit: mm

3148-QIP44MA



## BLOCK DIAGRAM



## PIN DESCRIPTION

Number	Name	Equivalent circuit	Description
1	1STMIXIN1		1st IF mixer inputs. Nominal voltage is 1.2 V.
44	1STMIXIN2		
2	HFVCC		Supply for high-frequency circuitry. Nominal voltage is 3.8 V.
3	1STMIXOUT		1st IF mixer output. Nominal voltage is 2.85 V.

# LA8606M

Number	Name	Equivalent circuit	Description
4	HFGND		Ground for high-frequency circuitry
5	2NDMIXIN		2nd IF mixer input. Nominal voltage is 1.2 V.
6	EXTREGADJ		
7	EXTREGOUT		Voltage regulator external PNP transistor connections. The nominal voltages of EXTREGADJ and EXTREGCONT are 1.2 V and 3.1 V, respectively.
8	EXTREGCONT		
9	2NDMIXOUT		2nd IF mixer output. Nominal voltage is 3.3 V.
10	VREF1	<p>Band gap (1.2V zener diode)</p>	Voltage reference output. Nominal voltage is 1.2 V.
11	IFIN		IF amplifier input. Nominal voltage is 1.2 V.
12	SDET		Signal meter output. Nominal voltage is 0.7 V.

# LA8606M

Number	Name	Equivalent circuit	Description
13	QUADCOIL		Quadrature detector phase coil connection. Nominal voltage is 3.8 V.
14	NOISEDET		Noise detector output. Nominal voltage is 1.4 V with no input signal.
15	SCHMITTOUT		Schmitt trigger output. Nominal voltage is 0.1 V.
16	NOISEOUT		External RC noise filter connections. Nominal voltages are 0.97 V.
17	NOISEIN		
18	DETOUT		FM detector output. Nominal output voltage is 1.2 V.
19	BUFFIN		3rd-order Butterworth filter external RC network connections. The nominal voltages of BUFFIN and BUFFOUT are 1.12 V and 1.2 V, respectively.
20	BUFFOUT		
28	DATAOUT		Data signal output
21	1/2VCC		Power amplifier voltage reference output. Nominal voltage is 1.9 V.
22	PWRIN		Power amplifier input. Nominal voltage is 1.9 V.
23	PWRCONT		Active-HIGH power amplifier control input. HIGH when 1.6 V or greater, and LOW when 0.8 V or lower
24	LFVCC		Supply for low-frequency circuitry. Nominal voltage is 3.8 V.

# LA8606M

Number	Name	Equivalent circuit	Description
25	LFGND		Ground for low-frequency circuitry
26	PWROUT		Power amplifier output. Nominal voltage is 1.6 V.
27	PWRNF		Power amplifier negative feedback input. Nominal voltage is 1.9 V.
29	PREREC		Preamplifier reference voltage output. Nominal voltage is 2.6 V.
30	PREIN		Preamplifier input. Nominal voltage is 2.6 V.
31	PRENF		Preamplifier negative feedback input. Nominal voltage is 2.6 V.
32	ALCDET		ALC detector external RC network connection and ALC or limiter mode select input. Nominal voltage is 1.0 V with no input signal in ALC mode.
33	FCH		Highpass filter capacitor connection. Nominal voltage is 2.6 V.
34	FCL		Lowpass filter capacitor connection. Nominal voltage is 2.6 V.
35	BPFOUT		Bandpass filter output. Nominal voltage is 2.6 V.

# LA8606M

Number	Name	Equivalent circuit	Description
36	LED		LED driver output
37	2NDOSC1		2nd IF oscillator external crystal oscillator connection. The nominal voltages of 2NDOSC1 and 2NDOSC2 are 3.1 V and 3.8 V, respectively.
38	2NDOSC2		
39	FROMPLL		VCO control input. The input voltage range is 0.7 to 5.0 V.
40	TOPLL		1st IF oscillator buffer output. Nominal voltage is 2.2 V.
41	1STOSC1		1st IF oscillator external oscillator network connections. Nominal voltages are 0.7 V.
42	1STOSC2		
43	VREF2		1st IF oscillator and mixer reference voltage output. Nominal voltage is 1.2 V.

## SPECIFICATIONS

### Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
Supply voltage	$V_{DD}$	8	V
Output current	$I_O$	30	mA
Power dissipation	$P_D$	600	mW
Operating temperature range	$T_{OPR}$	-20 to 75	°C
Storage temperature range	$T_{STG}$	-40 to 150	°C

### Note

The device should not be completely immersed during the soldering process.

**Recommended Operating Conditions** $T_a = 25^\circ\text{C}$ 

Parameter	Symbol	Rating	Unit
Supply voltage	$V_{DD}$	3.8	V
Supply voltage range	$V_{DD}$	1.8 to 6.0	V

**Electrical Characteristics** $V_{CC} = 3.8 \text{ V}, T_a = 25^\circ\text{C}$ 

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Supply current	$I_{CC}$	No input signal. PWRCONT is HIGH.	6	9	12	mA
		No input signal. PWRCONT is LOW.	5	8	11	
LOW-level input voltage	$V_{IL}$		-	-	0.8	V
HIGH-level input voltage	$V_{IH}$		1.6	-	-	V
LOW-level output voltage	$V_{OL}$		-	-	0.5	V
HIGH-level output voltage	$V_{OH}$		3.0	-	-	V

**External voltage regulator** $V_{CC} = 3.8 \text{ V}, T_a = 25^\circ\text{C}$  unless otherwise noted

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Output voltage range	$V_O$	$V_{CC} = 1.8 \text{ to } 6.0 \text{ V}$	1.6	-	5.8	V
Line regulation	$V_{LINE}$	$V_{EXTREGOUT} = 3 \text{ V}, V_{LINE} = V_{CC} - V_{EXTREGOUT}$	$V_{CC} = 3.2 \text{ to } 6.0 \text{ V}$	-	-	mV
			$V_{CC} = 1.8 \text{ to } 3.2 \text{ V}$	-	0.2	-
Load regulation	$V_{LOAD}$	$I_O = 1 \text{ to } 30 \text{ mA}$	-	-	60	mV

**High-frequency block** $V_{CC} = 3.8 \text{ V}, T_a = 25^\circ\text{C}, f_C = 49.830 \text{ MHz}, f_{mod} = 1 \text{ kHz}, \Delta f = \pm 3 \text{ kHz}$ . For AM signals,  $f_{mod} = 400 \text{ Hz}$ , 30% modulation unless otherwise noted

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Maximum input frequency	$f_I \text{ max}$		-	50	-	MHz
1st IF mixer conversion gain	$VG_{MIX1}$		-	18	-	dB
2nd IF mixer conversion gain	$VG_{MIX2}$		-	22	-	dB
Demodulator output voltage	$V_O$	1st mixer input voltage = 80 dB $\mu$	70	110	140	mV <sub>ms</sub>
Total harmonic distortion	$THD_{HF}$		-	0.5	2.0	%
AM rejection	AMR	$V_I = 80 \text{ dB}\mu$ , amplitude modulated	30	40	-	dB

# LA8606M

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Signal-to-noise ratio	S/N	$V_I = 80 \text{ dB}\mu$ , no modulation	54	58	-	dB
		$V_I = 8 \text{ dB}\mu$ , no modulation	14	20	-	
Noise detector output voltage	$V_{ND}$	$V_I = 10 \text{ dB}\mu$	-	1.4	-	V
		$V_I = 40 \text{ dB}\mu$	-	0	-	
Signal meter output voltage	$V_{SM}$	$V_I = 10 \text{ dB}\mu$	0.2	0.6	0.9	V
		$V_I = 50 \text{ dB}\mu$	0.8	1.1	1.4	
Schmitt-trigger output voltage	$V_{SH}$	$V_I = 21 \text{ dB}\mu$	-	-	0.5	V
		$V_I = 37 \text{ dB}\mu$	3.5	-	-	
Schmitt-trigger level	SH		21	29	37	$\text{dB}\mu$
-3 dB limiting sensitivity	-3dBS		-	-	10	$\text{dB}\mu$
First mixer third-order intercept	3rdIM			-23		dBm

## Data shaper

$V_{CC} = 3.8 \text{ V}$ ,  $T_a = 25^\circ\text{C}$ ,  $f_I = 1 \text{ kHz}$

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Duty cycle	Duty		-	50	-	%

## Filters

$V_{CC} = 3.8 \text{ V}$ ,  $T_a = 25^\circ\text{C}$ ,  $f_I = 1 \text{ kHz}$ ,  $V_I = 200 \text{ mV}_{\text{rms}}$  unless otherwise noted

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
LPF attenuation	$\alpha_{LPF}$	$f_I = 3.7 \text{ kHz}$	1.5	3.0	4.5	dB
		$f_I > 3.7 \text{ kHz}$	-	18	-	dB/octave
BPF gain	$G_{BPF}$		-	0	-	dB
BPF attenuation	$\alpha_{BPF}$	$f_I = 250 \text{ Hz}$	0.5	2.5	5.5	dB
		$f_I = 3.7 \text{ kHz}$	1.0	3.1	6.0	
			-	6	-	dB/octave

## Preamplifier

$V_{CC} = 3.8 \text{ V}$ ,  $T_a = 25^\circ\text{C}$ ,  $f_I = 1 \text{ kHz}$ ,  $V_I = 2 \text{ mV}_{\text{rms}}$  unless otherwise noted

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Voltage gain	$VG_{PRE}$		37	39	41	dB
Total harmonic distortion	$THD_{PRE}$		-	0.1	0.5	%
Input conversion noise voltage	$V_{NI}$	$R_g = 0 \Omega$	-	1	3	$\mu\text{V}$
ALC range	ALC	3 dB ALC error	30	40	-	dB

# LA8606M

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
ALC level	V <sub>ALC</sub>		1.6	1.9	2.2	V
Limiter level	V <sub>LIM</sub>		2.0	2.3	2.6	V

## Power amplifier

V<sub>CC</sub> = 3.8 V, T<sub>a</sub> = 25 °C, f<sub>I</sub> = 1 kHz, V<sub>I</sub> = 10 mV<sub>rms</sub> unless otherwise noted

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
Voltage gain	V <sub>GPOWER</sub>	R <sub>L</sub> = 100 Ω, V <sub>I</sub> = -40 dBV	32	34	36	dB
Total harmonic distortion	THD <sub>PWR</sub>	R <sub>L</sub> = 100 Ω, V <sub>I</sub> = -40 dBV	-	1.5	3.0	%
Maximum output power	P <sub>O</sub> max	R <sub>L</sub> = 100 Ω, THD = 10%	10	14	-	mW
Output noise voltage	V <sub>NO</sub>	R <sub>g</sub> = 0 Ω, DIN audio filter	-	0.2	0.4	mV <sub>rms</sub>
Crosstalk rejection	CT	PWRCONT connected to ground	45	63	-	dB

## Battery checker

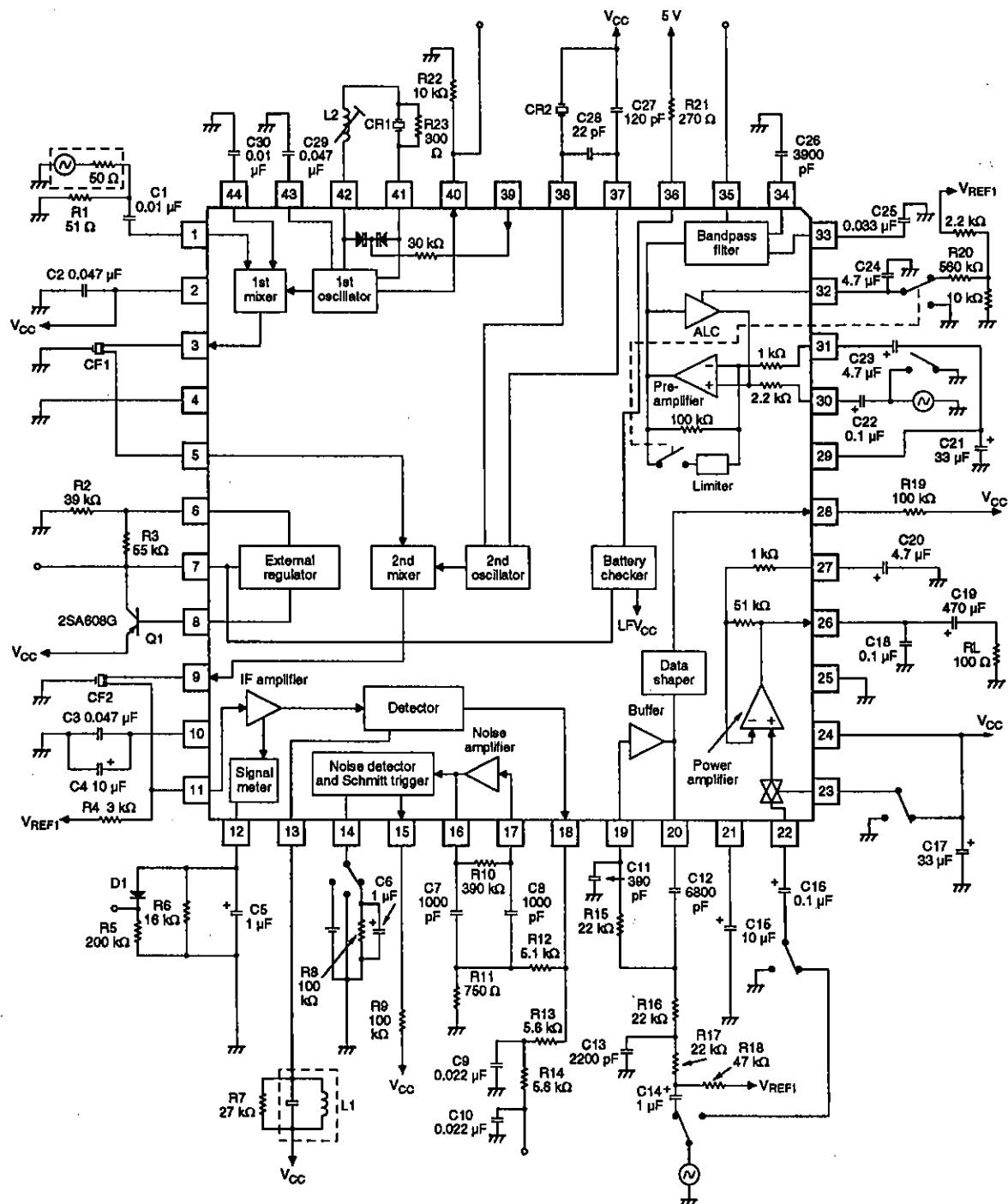
V<sub>CC</sub> = 3.8 V, T<sub>a</sub> = 25 °C

Parameter	Symbol	Condition	Rating			Unit
			min	typ	max	
LED ON voltage	V <sub>ON</sub>	See note.	0.2	0.3	0.4	V
Hysteresis	LED <sub>HYS</sub>		-	100	-	mV

### Note

$$V_{ON} = V_{CC} - V_{EXTREGOUT}$$

## Measurement Circuit



CF1 Murata SFE10.7MS2  
 CF2 Murata SFU455A  
 L1 Mitsumi WK40337  
 L2 Mitsumi YD0103  
 CR1 39.130 MHz  
 CR2 10.245 MHz

## DESIGN NOTES

### Voltage Regulator and Battery Checker

The external regulator output voltage  $V_{EXTREGOUT}$  on EXTREGOUT is determined by the resistor divider network between EXTREGADJ, EXTREGOUT and GND.

The battery checker turns ON when the voltage on EXTREGOUT reaches the threshold voltage, shown as  $V_A$  in figure 1, and when the supply voltage is  $V_A + 0.3$  V. Hysteresis is fixed at 100 mV.

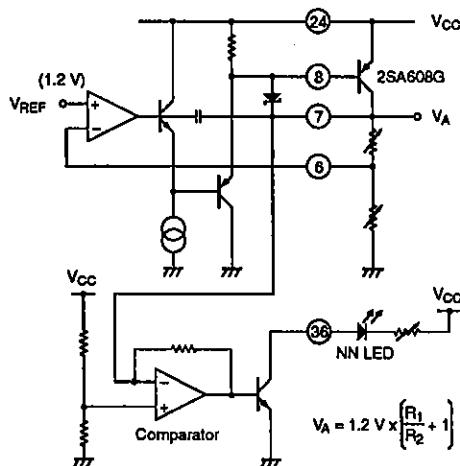


Figure 1. Battery checker threshold

### ALC Detector

When ALCDET is tied LOW as shown in figure 2, the ALC circuitry becomes a limiter. When ALCDET is not tied LOW, the ALC detector functions normally and an external RC network should be connected to ALCDET as shown in figure 3.



Figure 2. Limiter-mode connection

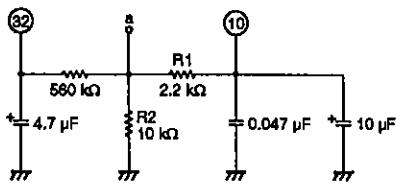


Figure 3. ALC-mode connection

#### Note

The ALC time constant resistors, R1 and R2, should be chosen such that the potential at point *a* is approxi-

mately 1.00 V for an input level below the ALC threshold, and the output current from VREF1 is 300 μA or lower.

### Highpass Filter

The highpass cutoff point of the highpass filter is determined by the RC network formed by the internal resistor  $R_H$  and an external capacitor.

### Lowpass Filter

The lowpass cutoff point of the lowpass filter is determined by the RC network formed by the internal resistor  $R_L$  and an external capacitor.

### First IF Oscillator

An external crystal, inductor and capacitor should be connected to 1STOSC1 and 1STOSC2.

### Second IF Oscillator

An external crystal and capacitor should be connected to 2NDOSC1 and 2NDOSC2 to form a Colpitts oscillator.

### VCO

The oscillator frequency is determined by the value of an internal varactor. The varactor value is controlled by the level on FROMPLL.

### Power Amplifier

PWRCONT is used to reduce power consumption by decoupling the bias currents that normally flow into the power amplifier inputs. PWRCONT is HIGH during normal operation and LOW during the power mute and battery save modes.

### IF Amplifier

To ensure the correct biasing of the IF amplifier inputs, a bias resistor should be connected between IFIN and VREF1.

### Signal Meter Output

The voltage on the signal meter output is 0.7 V, with an external 16 kΩ resistor and 1 μF capacitor connected. This output voltage, however, is temperature dependent. The output voltage can be compensated by a level shift using either an external diode, as shown in figure 4, or transistor. This circuit uses the temperature dependence of resistor  $R_A$  to compensate for the signal meter output voltage.

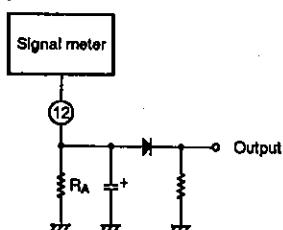


Figure 4. Compensation network

The affect of the level shift compensation is shown in figure 5, and the corresponding temperature coefficients, in figure 6.

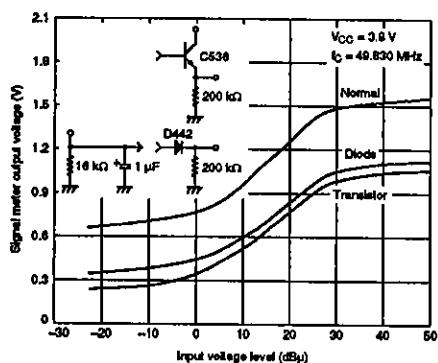


Figure 5. Signal meter output voltage vs. input voltage

### Noise Detector

The noise component of the FM modulator output, after it has been amplified and detected, is input to a Schmitt-trigger circuit which has hysteresis. A noise

component of approximately 0.5 to 1 V on NOISEOUT will activate the noise detector, depending on the Schmitt-trigger levels.

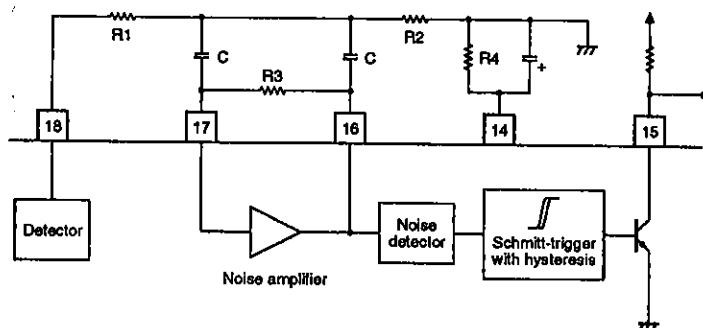


Figure 7. Noise detector circuit

The random noise level varies when a front end or filter is connected. Accordingly, the Schmitt-trigger levels will need to be checked after making such a connection. The Schmitt-trigger level on the weak input side is adjusted by reducing the resistance of R4 or by reducing the noise amplifier gain.

The bandpass noise amplifier gain,  $G_0$ , the center frequency,  $f_c$ , and the quality factor,  $Q$ , are determined from the following equations, respectively.

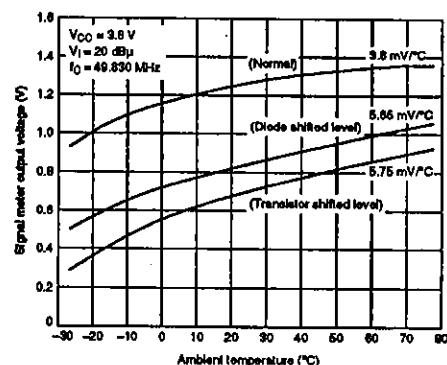


Figure 6. Signal meter output voltage vs. ambient temperature

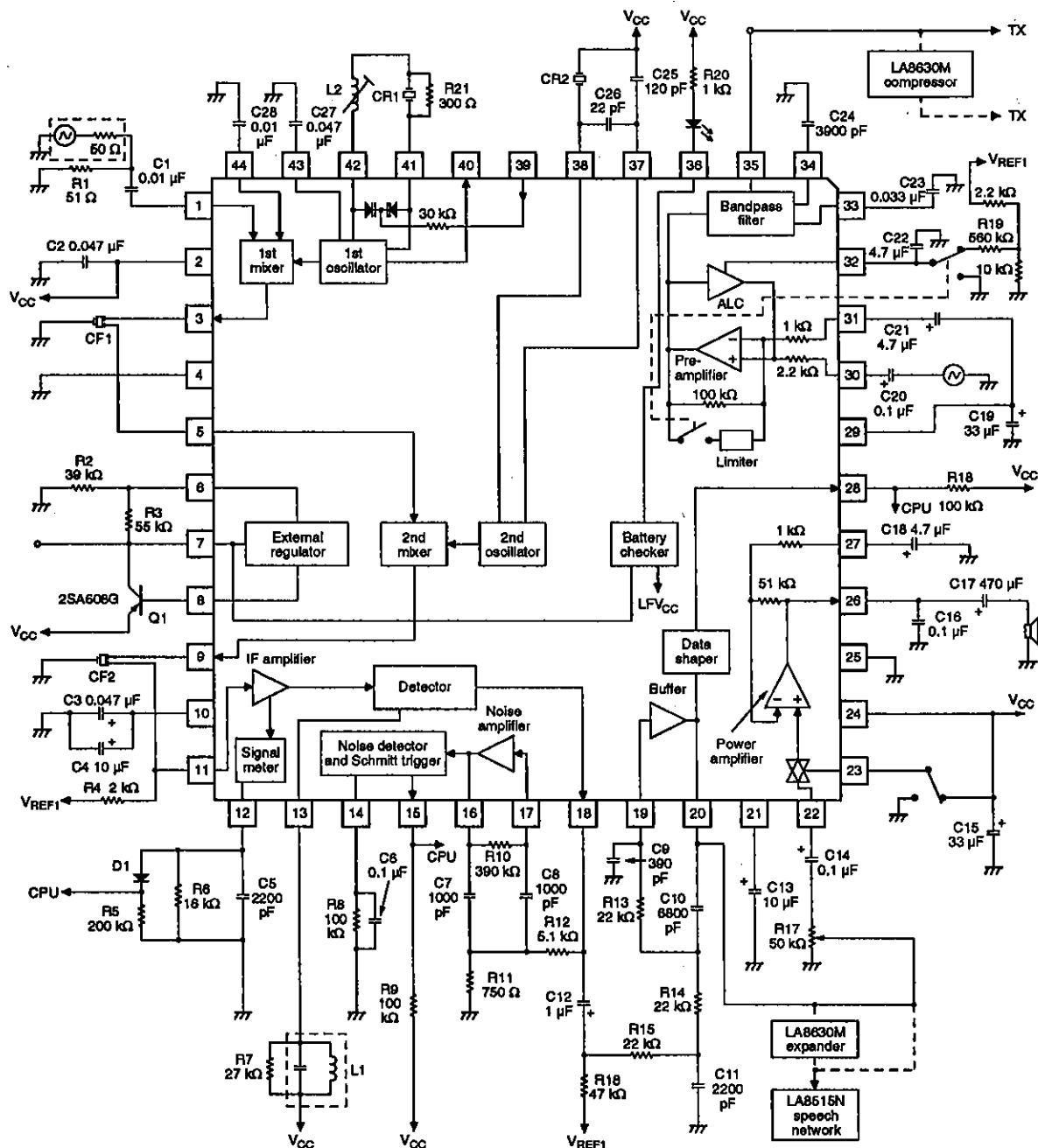
$$G_0 = \frac{1}{2} \times \frac{R_3}{R_1}$$

$$f_c = \frac{1}{2\pi C} \times \sqrt{\frac{R_1 + R_2}{R_1 \times R_2 \times R_3}}$$

$$Q = \frac{1}{2} \times \sqrt{\frac{R_1 + R_2}{R_1 \times R_2}} \times R_3$$

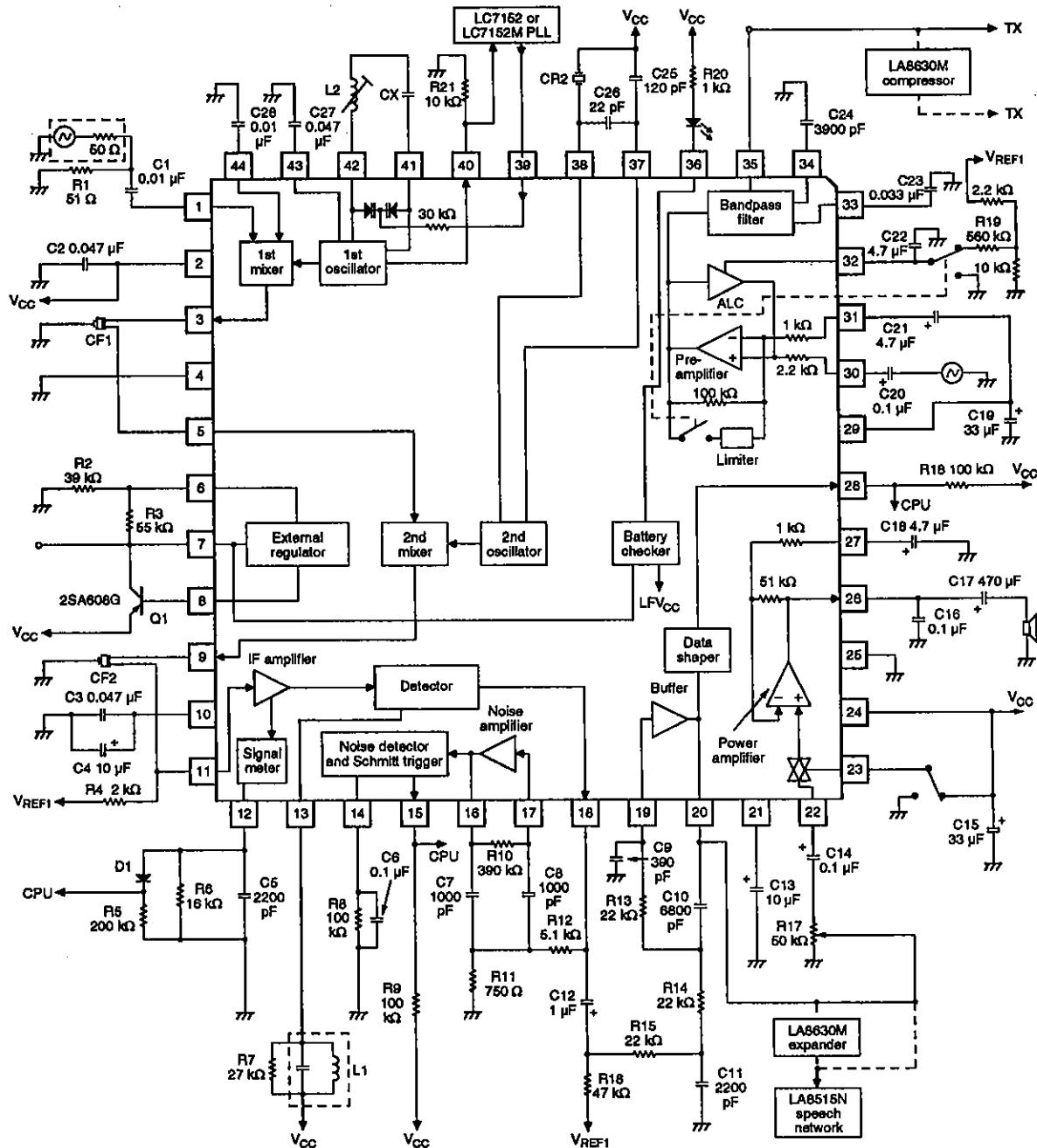
## TYPICAL APPLICATIONS

## Fixed Station Frequency Operation



CF1	Murata SFE10.7MS2
CF2	Murata CFU455E2
L1	Mitsumi WK40337
L2	Mitsumi YD0103
CR1	39.130 MHz
CR2	10.245 MHz

## Operation with an External PLL



CF1 Murata SFE10.7MS2

CF2 Murata CFU455E2

L1 Mitsumi WK40337

CR2 10.245 MHz

When f<sub>VCO1</sub> = 36 MHz,

CX 18 pF

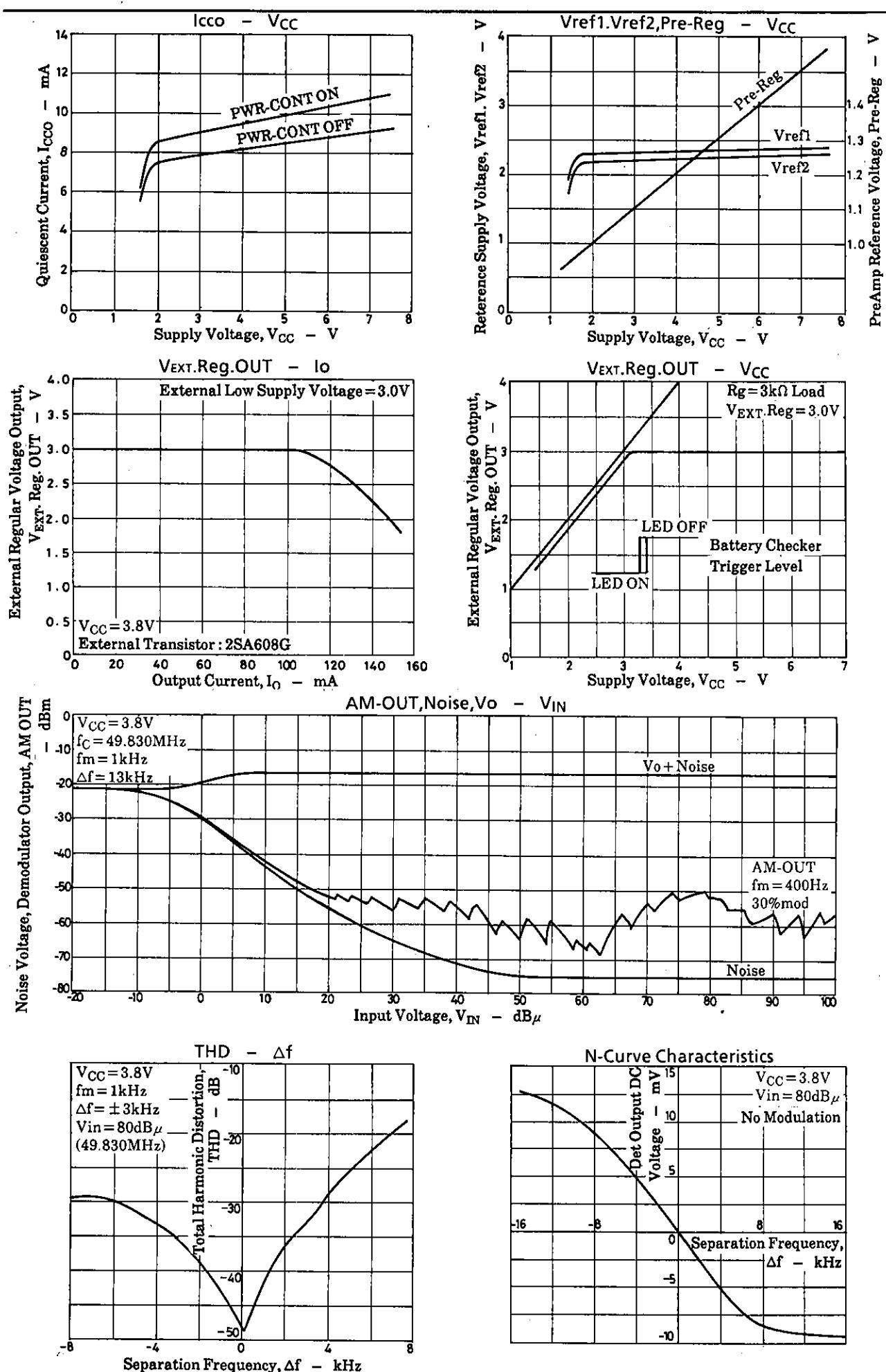
L2 Mitsumi YD1001

When f<sub>VCO2</sub> = 39 MHz,

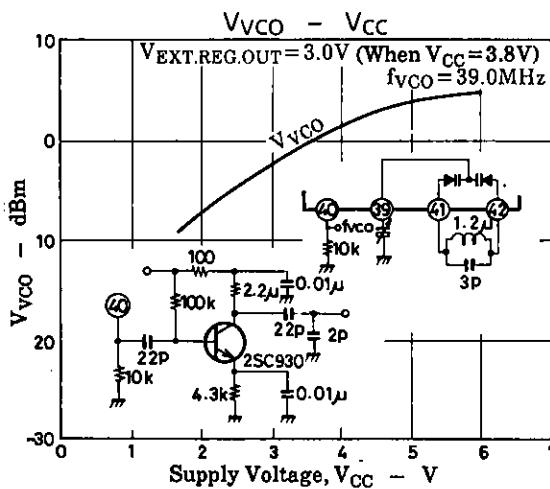
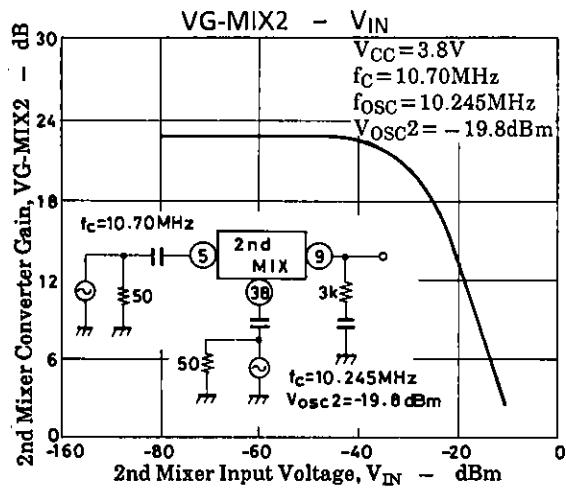
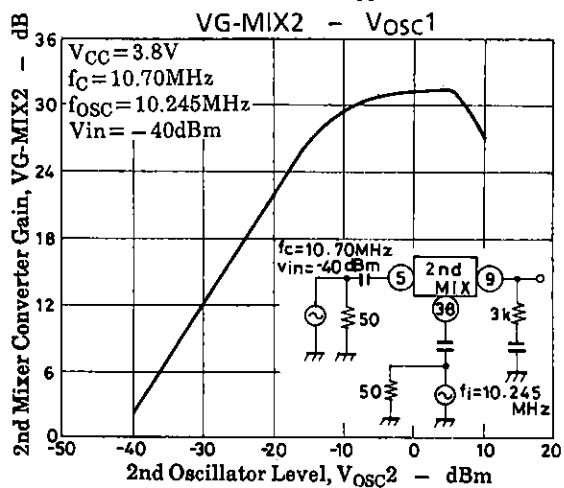
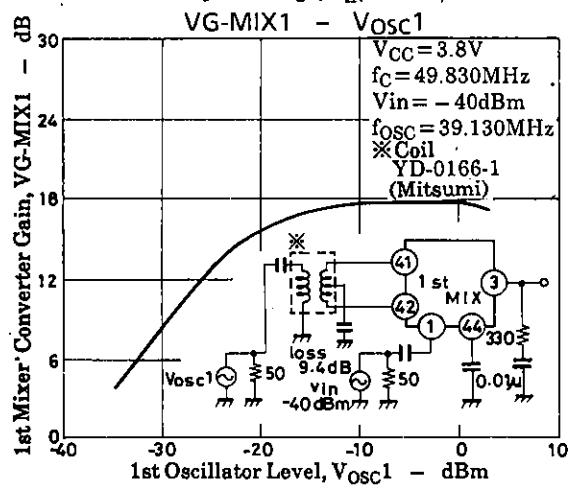
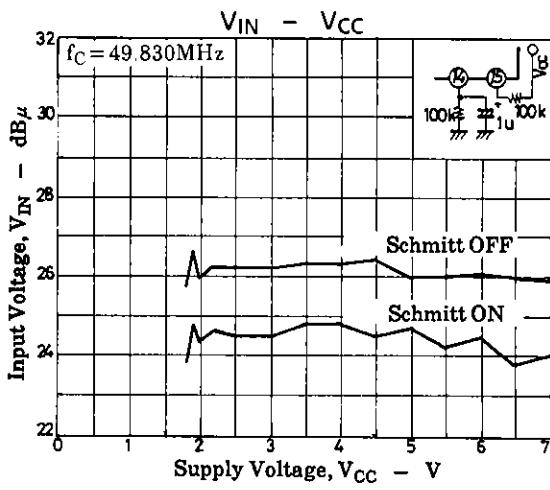
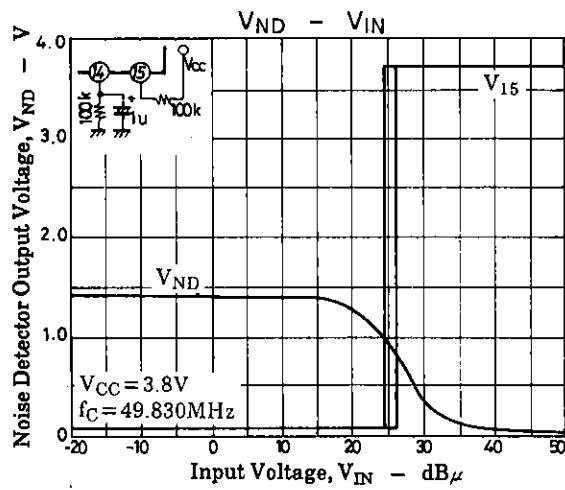
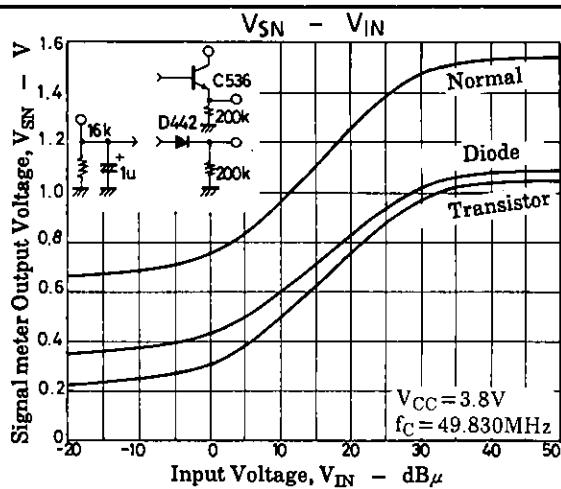
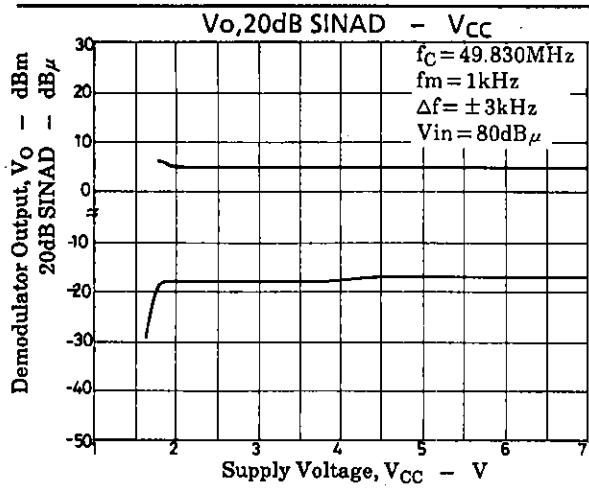
CX 18 pF

L2 Mitsumi YD1002

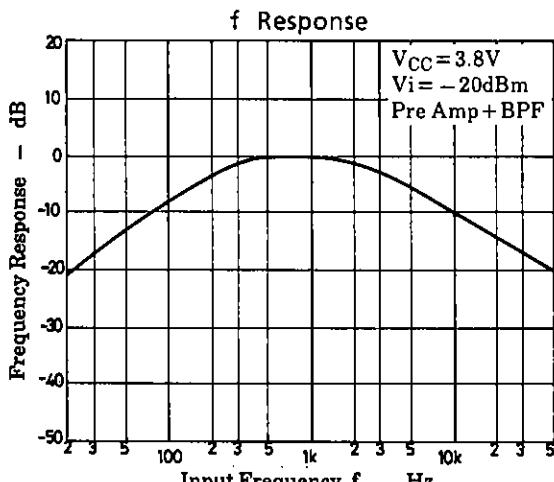
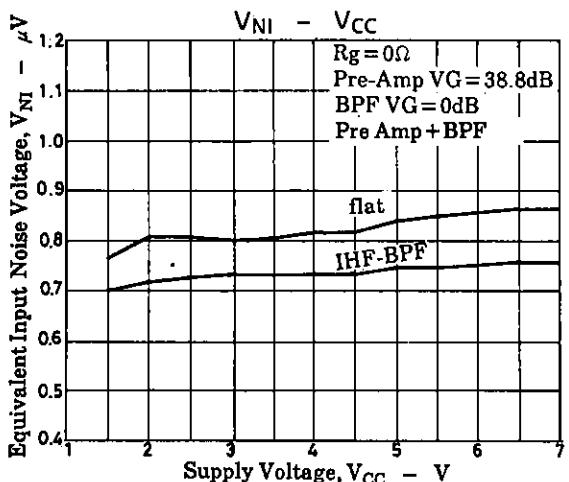
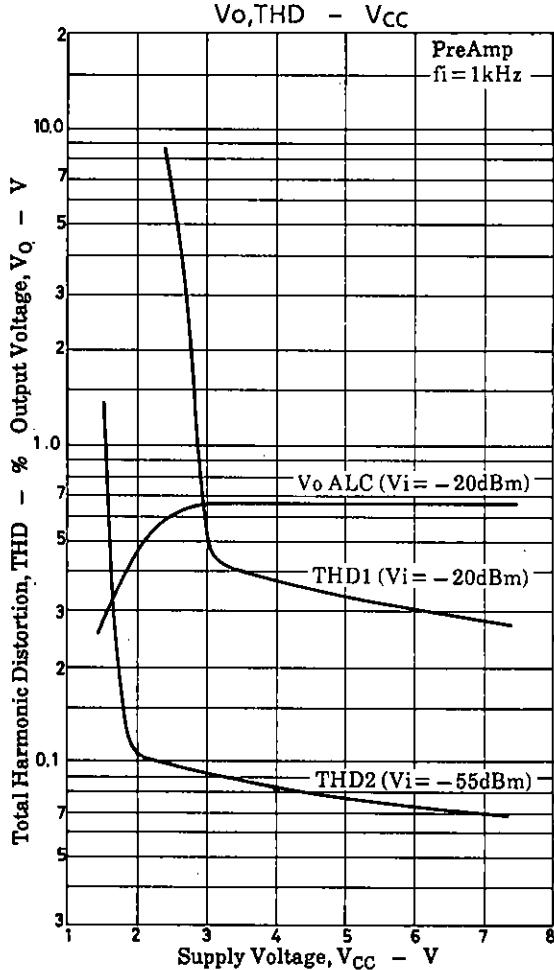
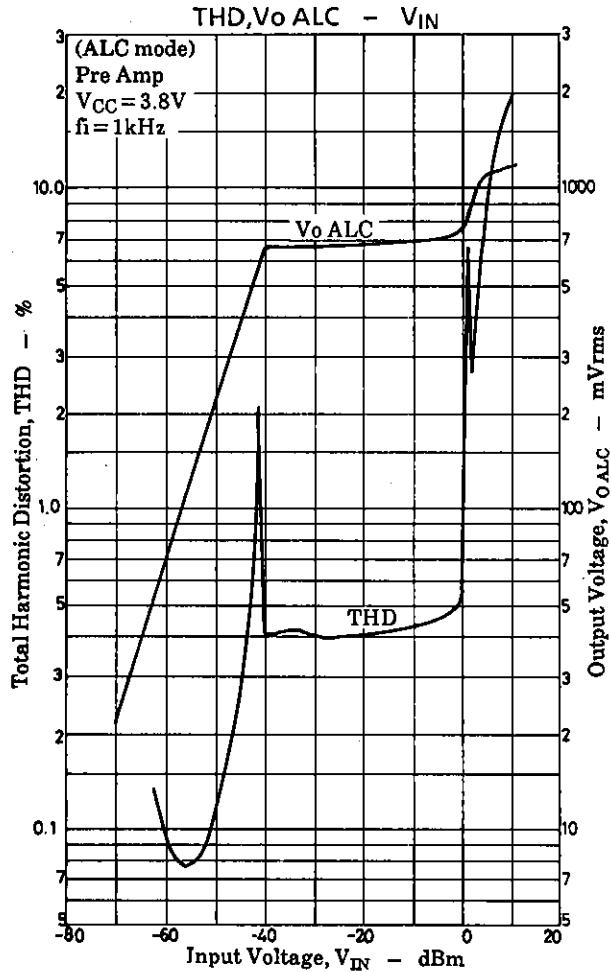
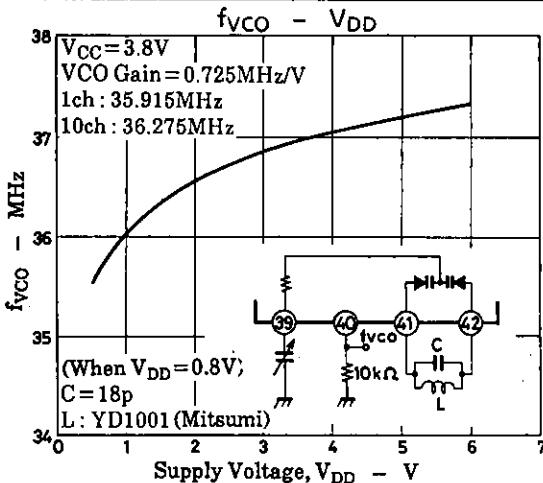
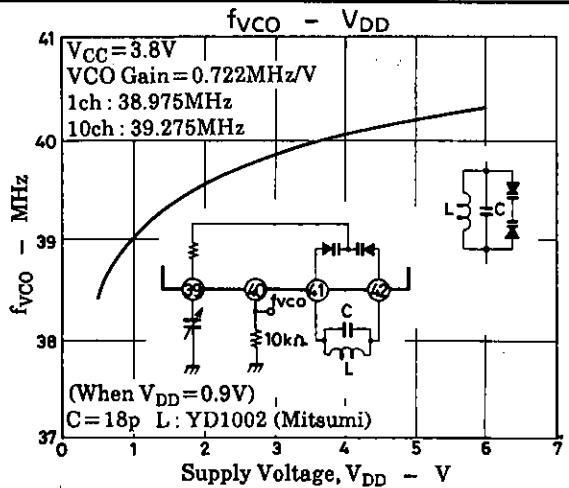
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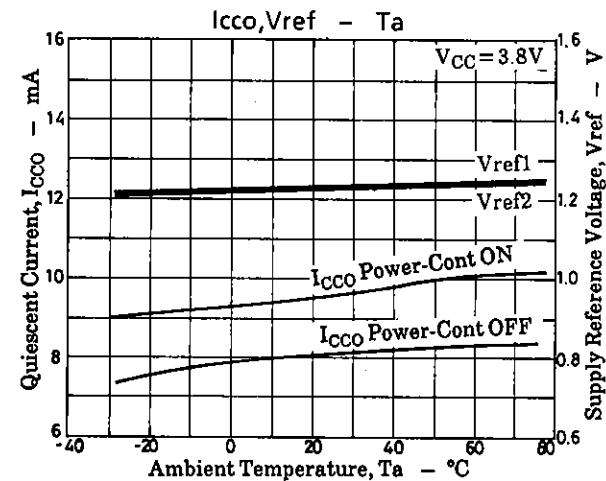
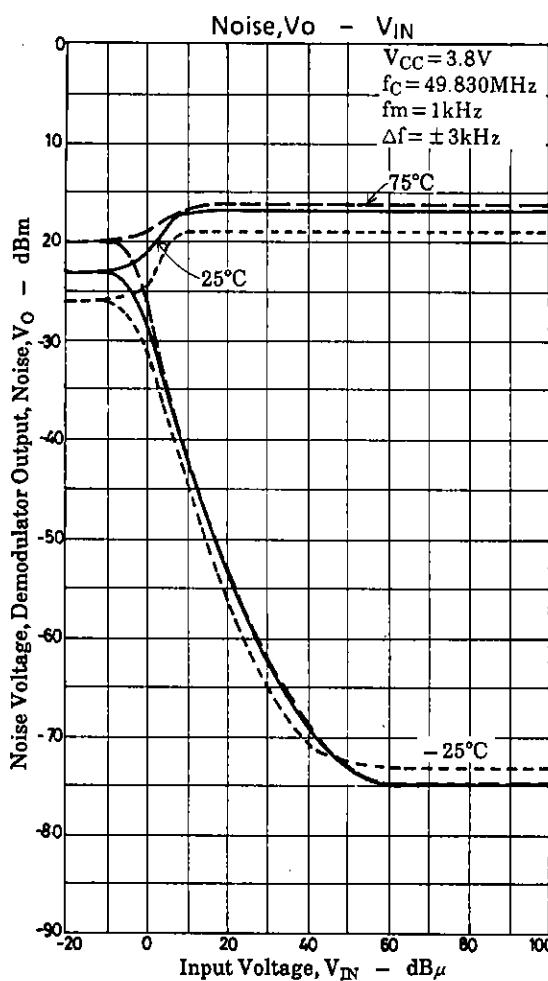
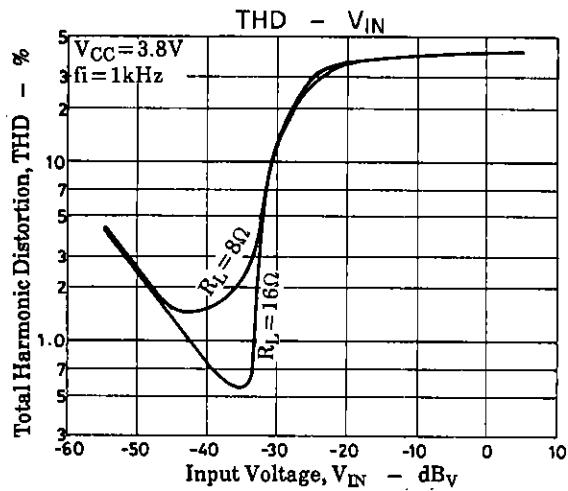
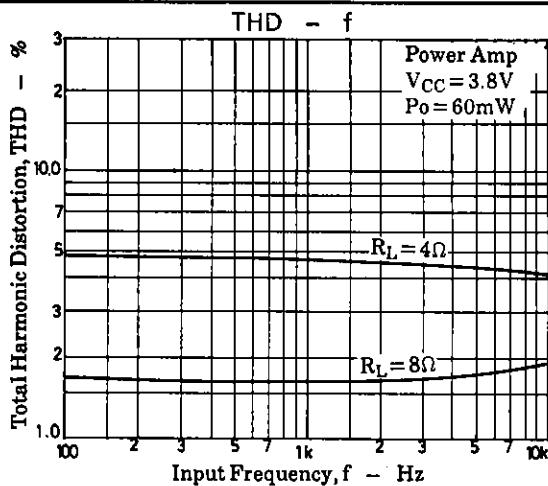
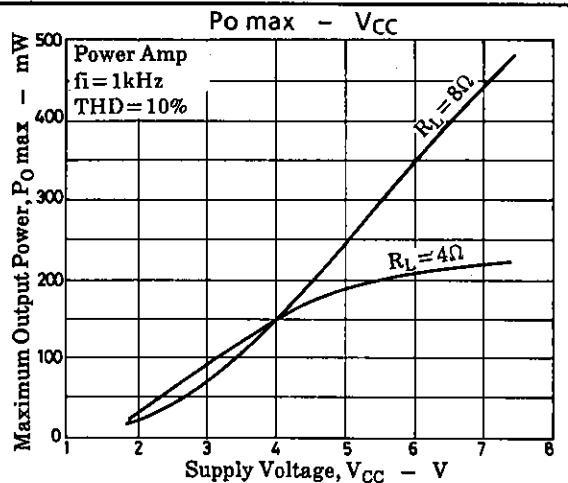


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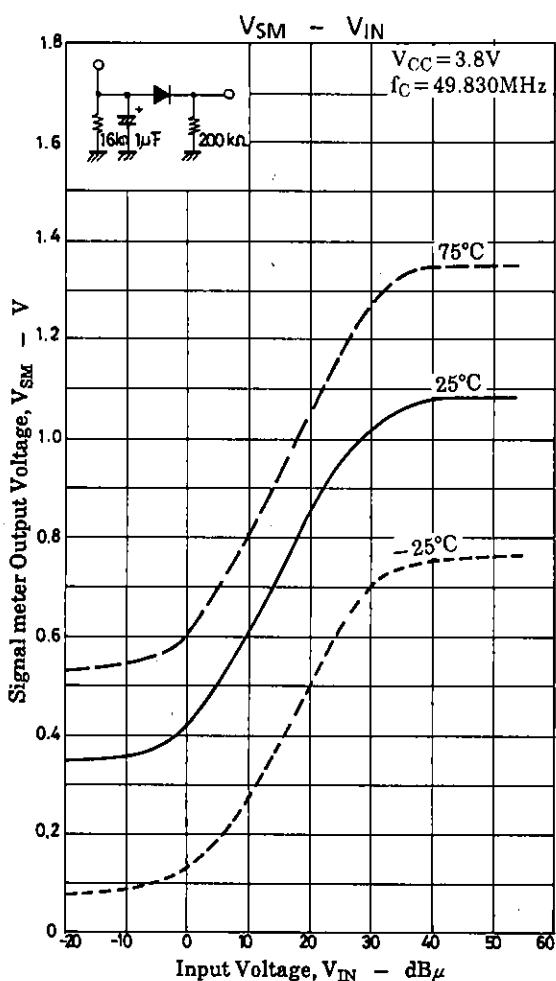
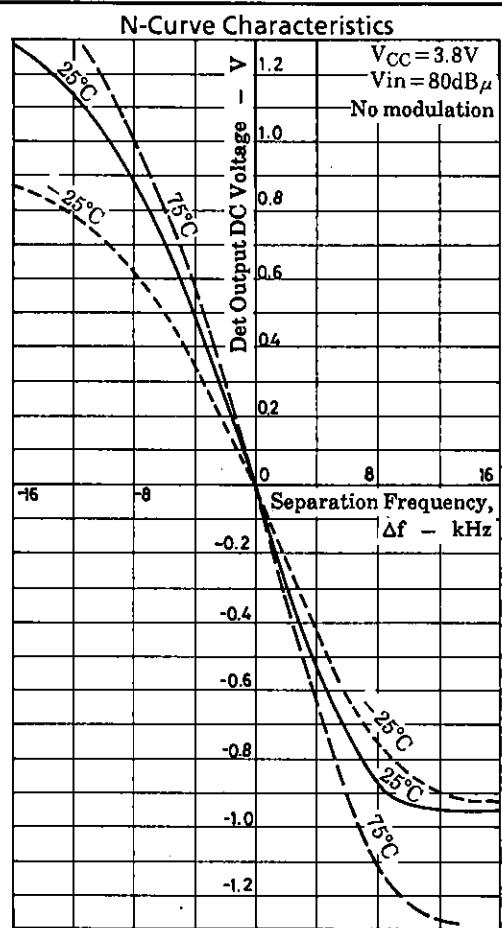
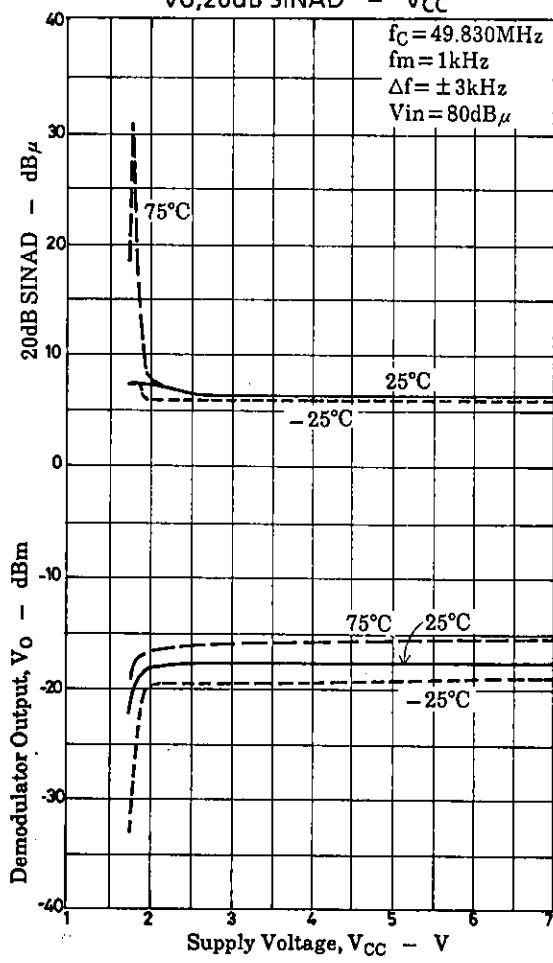
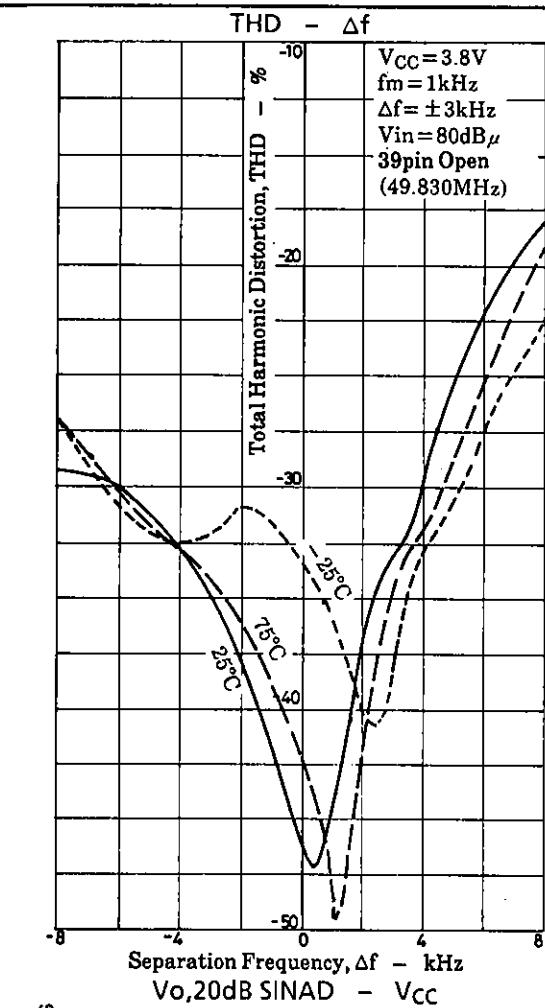


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