



# Delphi Series L36SA, 2 x 1.6, 50W Family DC/DC Power Module: 18~75V in, 12V/4A out

The Delphi Series L36SA, 2" x 1.6", 18~75V input, single output, isolated DC/DC converter is the latest offering from a world leader in power systems technology and manufacturing - Delta Electronics, Inc. This L36SA series provides up to 50 watts of power or 15A of output current (3.3V) in an industry standard 2" x 1.6" form factor and pinout. The Delphi L36SA series operates from a wide 18~75V (4:1) input voltages. With creative design technology and optimization of component placement, these converters possess outstanding electrical and thermal performances, as well as extremely high reliability under highly stressful operating conditions. All models are fully protected from abnormal input/output voltage, current, and temperature conditions. The Delphi Series converters meet all safety requirements with basic insulation. An optional heat spreader is available for extended operation.

#### **FEATURES**

- High Efficiency: 87.5% @ 12V/4A
- Size: 49.6mm x 39.4mm x 8.9mm (1.95"x1.55"x0.35")
- Industry standard pin out
- Fixed frequency operation
- Input UVLO, OTP, Output OCP, OVP, (auto recovery)
- Pre-bias start up
- 2250V isolation and basic insulation
- No minimum load required
- 4:1 Input voltage range
- ISO 9001, TL 9000, ISO 14001, QS9000, OHSAS18001 certified manufacturing facility
- UL/cUL 60950 (US & Canada) Recognized, and TUV (EN60950) certified.
- CE mark meets 73/23/EEC and 93/68/EEC directives

#### **OPTIONS**

- Positive On/Off logic
- Sense
- Heat Spreader
- Heatsink

#### **APPLICATIONS**

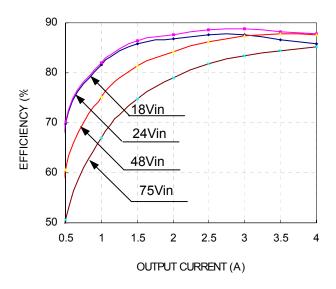
- Telecom/Datacom
- Wireless Networks
- Optical Network Equipment
- Server and Data Storage
- Industrial/Test Equipment



# **TECHNICAL SPECIFICATIONS**

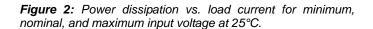
( $T_A$ =25°C, airflow rate=300 LFM,  $V_{in}$ =48Vdc, nominal Vout unless otherwise noted.)

NOTES and CONDITIONS	L36SA12004 (Standard)				
	Min. Typ. Max.				
			80	Vdc	
				Vdc	
Refer to Figure21 for measuring point	-40		130	°C	
· · · · · · · · · · · · · · · · · · ·	-55		125	°C	
			2250	Vdc	
	40		7.5	) / al a	
	18		75	Vdc	
	16	17	18	Vdc	
				Vdc	
	0.75	1	1.5	Vdc	
100% Load, 18Vin			4	Α	
		60		mA	
		4		mA	
D.D. # 40 H. F. T. F. F. T. F. T. F. T. F. F. T. F. F. T. F. F. T. F.		6.5	1	A <sup>2</sup> s	
				mA	
120 HZ		60		dB	
Vin=48V Io=Io may To=25°C	11 880	12,000	12 120	Vdc	
VIII-40V, 10-10.111aX, 10-20 C	11.000	12.000	12.120	Vuc	
lo=lo.min to lo.max		±5	±10	mV	
2 24 24 2				mV	
Ta=-40°C to 85°C		±30		mV	
Over sample load, line and temperature	11.82		12.18	V	
5Hz to 20MHz bandwidth					
				mV	
Full Load, 1µF ceramic, 10µF tantalum		15		mV	
				A %	
	110		150	70	
48V 10uF Tan & 1uF Ceramic load can, 0.14/us					
		100		mV	
		100		mV	
		200		us	
				ms	
		8		ms	
Full load; 5% overshoot of Vout at startup			470	μF	
		06		%	
				<del>%</del>	
		88		/0	
			2250	Vdc	
	100			МΩ	
		1500		pF	
		300		kHz	
V-1-1-1-1-1-1-1-1-1-1-1-1-1-1-1-1-1-1-1			0		
Von/off at Ion/off=1.0mA	2		0.7	V	
νοπ/οπ αι ιοπ/οπ=υ.υ μΑ	2		18	V	
Von/off at lon/off=1 0m∆			0.7	V	
	2			V	
				mA	
Logic High, Von/off=15V				uA	
Vin = 18V ~ 60V	-10		10	%	
Vin = 61V ~ 75V	-5		10	%	
Pout ≦ max rated power			10	%	
Over full temp range; % of nominal Vout		14.4		V	
, , , , , , , , , , , , , , , , , , , ,					
lo=80% of lo, max; Ta=25°C		2.98 24.2		M hour	
	Refer to Figure 21 for measuring point  100% Load, 18Vin  P-P thru 12µH inductor, 5Hz to 20MHz 120 Hz  Vin=48V, lo=lo.max, Tc=25°C  lo=lo,min to lo,max Vin=18V to 75V Ta=-40°C to 85°C  Over sample load, line and twingth Full Load, 1µF ceramic, 10µF tantalum  48V, 10µF Tan & 1µF Ceramic load cap, 0.1A/µs 50% lo.max to 75% lo.max 75% lo.max to 50% lo.max  Full load; 5% overshoot of Vout at startup  Von/off at lon/off=0.0 µA Von/off at lon/off=0.0 µA Von/off at lon/off=0.0 V Logic High, Von/off=15V Vin = 61V ~ 75V Pout ≤ max rated power	Refer to Figure21 for measuring point	Min.   Typ.	Min.       Typ.       Max.         80       30         100       130       130         -55       125       2250         18       75       125         16       17       18         15       16       17         100% Load, 18Vin       4       60         4       4       1         P-P thru 12µH inductor, 5Hz to 20MHz       20       60         Vin=48V, Io=lo.max, Tc=25°C       11.880       12.000       12.120         Vin=48V, Io=lo.max, Tc=25°C       11.880       12.000       12.120         Vin=48V, Io=lo.max, Tc=25°C       11.880       12.000       12.120         Type Industrian       15       ±5       ±10         Vin=48V, Io=lo.max, Tc=25°C       11.880       12.000       12.120         Type Industrian       15       ±5       ±10         Type Industrian       15       ±5       ±10         Type Industrian       16       17       18       10       10       10       10       10       10       10       10       10       10       10       10	



9 75Vin 8 7 48Vin 6 LOSS (W) 5 4 24Vin 3 18Vin 2 2 2.5 3 3.5 0.5 1.5 OUTPUT CURRENT (A)

Figure 1: Efficiency vs. load current for minimum, nominal, and maximum input voltage at 25°C



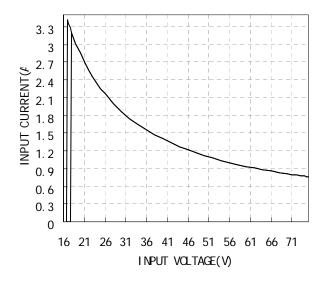
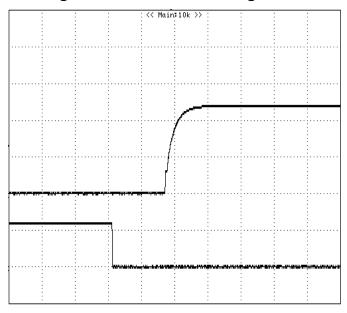


Figure 3: Typical full load input characteristics at room temperature

## For Negative Remote On/Off Logic



**Figure 4:** Turn-on transient at full rated load current (resistive load) (5 ms/div). Vin=48V.Top Trace: Vout, 5V/div; Bottom Trace: ON/OFF input, 5V/div

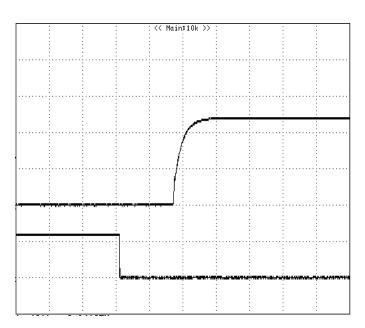
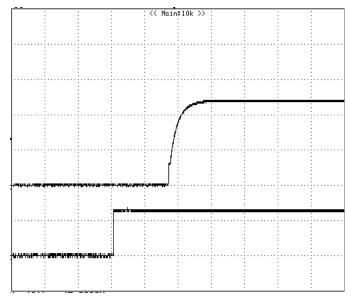


Figure 5: Turn-on transient at zero load current (5 ms/div). Vin=48V.Top Trace: Vout, 5V/div; Bottom Trace: ON/OFF input, 5V/div

# For Positive Remote On/Off Logic



**Figure 6:** Turn-on transient at full rated load current (resistive load) (5 ms/div). Vin=48V.Top Trace: Vout, 5V/div; Bottom Trace: ON/OFF input, 5V/div

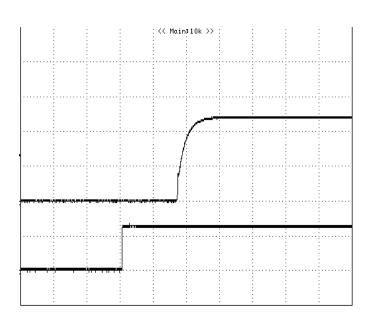
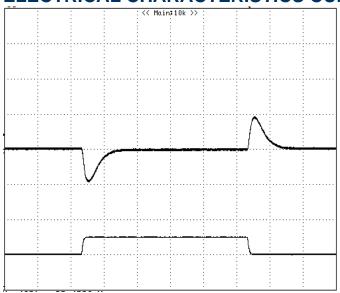
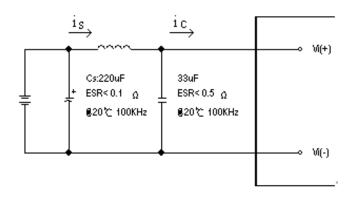


Figure 7: Turn-on transient at zero load current (5 ms/div). Vin=48V.Top Trace: Vout, 5V/div, Bottom Trace: ON/OFF input, 5V/div



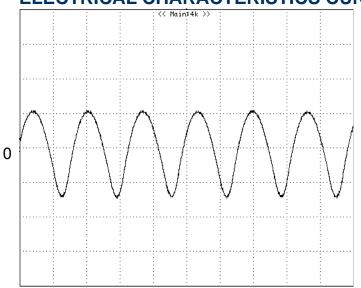
**Figure 8:** Output voltage response to step-change in load current (75%-50%-75% of Io, max; di/dt = 0.1A/ $\mu$ s). Load cap:  $10\mu$ F, tantalum capacitor and  $1\mu$ F ceramic capacitor. Top Trace: Vout (100mV/div,500us/div), Bottom Trace: I out (2A/div). Scope measurement should be made using a BNC cable (length shorter than 20 inches). Position the load between 51 mm to 76 mm (2 inches to 3 inches) from the module.

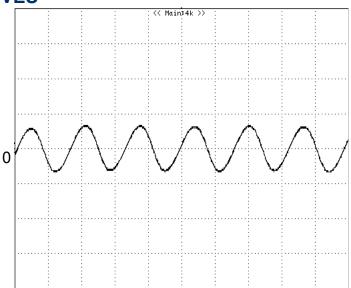
**Figure 9:** Output voltage response to step-change in load current (75%-50%-75% of Io, max; di/dt =  $2.0A/\mu s$ ). Load cap:  $330\mu F$ ,  $35m\Omega$  ESR solid electrolytic capacitor and  $1\mu F$  ceramic capacitor. Top Trace: Vout (100mV/div, 500us/div), Bottom Trace: I out (2A/div). Scope measurement should be made using a BNC cable (length shorter than 20 inches). Position the load between 51 mm to 76 mm (2 inches to 3 inches) from the module.



**Figure 10:** Test set-up diagram showing measurement points for Input Terminal Ripple Current and Input Reflected Ripple Current.

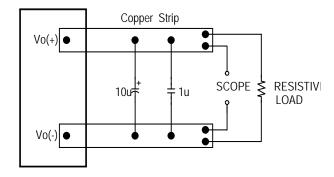
Note: Measured input reflected-ripple current with a simulated source Inductance ( $L_{TEST}$ ) of 12  $\mu$ H. Capacitor Cs offset possible battery impedance. Measure current as shown above



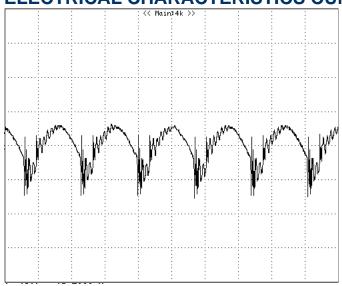


**Figure 11:** Input Terminal Ripple Current, i<sub>c</sub>, at full rated output current and nominal input voltage with 12μH source impedance and 33μF electrolytic capacitor (500 mA/div, 2us/div).

**Figure 12:** Input reflected ripple current,  $i_s$ , through a  $12\mu H$  source inductor at nominal input voltage and rated load current (20 mA/div, 2us/div).



**Figure 13:** Output voltage noise and ripple measurement test setup



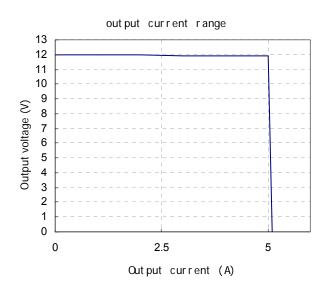


Figure 14: Output voltage ripple at nominal input voltage and rated load current (Io=4A)(20 mV/div, 2us/div)
Load capacitance: 1μF ceramic capacitor and 10μF tantalum capacitor. Bandwidth: 20 MHz. Scope measurements should be made using a BNC cable (length shorter than 20 inches). Position the load between 51 mm to 76 mm (2 inches to 3 inches) from the module.

Figure 15: Output voltage vs. load current showing typical current limit curves and converter shutdown points.

#### **DESIGN CONSIDERATIONS**

## **Input Source Impedance**

The impedance of the input source connecting to the DC/DC power modules will interact with the modules and affect the stability. A low ac-impedance input source is recommended. If the source inductance is more than a few  $\mu H,$  we advise adding a 10 to 100  $\mu F$  electrolytic capacitor (ESR < 0.7  $\Omega$  at 100 kHz) mounted close to the input of the module to improve the stability.

## **Layout and EMC Considerations**

Delta's DC/DC power modules are designed to operate in a wide variety of systems and applications. For design assistance with EMC compliance and related PWB layout issues, please contact Delta's technical support team. An external input filter module is available for easier EMC compliance design. Application notes to assist designers in addressing these issues are pending to release.

## **Safety Considerations**

The power module must be installed in compliance with the spacing and separation requirements of the end-user's safety agency standard, i.e., UL60950, CAN/CSA-C22.2 No. 60950-00 and EN60950: 2000 and IEC60950-1999, if the system in which the power module is to be used must meet safety agency requirements.

Basic insulation based on 75 Vdc input is provided between the input and output of the module for the purpose of applying insulation requirements when the input to this DC-to-DC converter is identified as TNV-2 or SELV. An additional evaluation is needed if the source is other than TNV-2 or SELV.

When the input source is SELV circuit, the power module meets SELV (safety extra-low voltage) requirements. If the input source is a hazardous voltage which is greater than 60 Vdc and less than or equal to 75 Vdc, for the module's output to meet SELV requirements, all of the following must be met:

- The input source must be insulated from the ac mains by reinforced or double insulation.
- The input terminals of the module are not operator accessible.
- If the metal baseplate is grounded, one Vi pin and one Vo pin shall also be grounded.
- A SELV reliability test is conducted on the system where the module is used, in combination with the module, to ensure that under a single fault, hazardous voltage does not appear at the module's output.

When installed into a Class II equipment (without grounding), spacing consideration should be given to the end-use installation, as the spacing between the module and mounting surface have not been evaluated.

The power module has extra-low voltage (ELV) outputs when all inputs are ELV.

This power module is not internally fused. To achieve optimum safety and system protection, an input line fuse is highly recommended. The safety agencies require a normal-blow fuse with 10A maximum rating to be installed in the ungrounded lead. A lower rated fuse can be used based on the maximum inrush transient energy and maximum input current.

## **Soldering and Cleaning Considerations**

Post solder cleaning is usually the final board assembly process before the board or system undergoes electrical testing. Inadequate cleaning and/or drying may lower the reliability of a power module and severely affect the finished circuit board assembly test. Adequate cleaning and/or drying are especially important for un-encapsulated and/or open frame type power modules. For assistance on appropriate soldering and cleaning procedures, please contact Delta's technical support team.

## FEATURES DESCRIPTIONS

#### **Over-Current Protection**

The modules include an internal output over-current protection circuit, which will endure current limiting for an unlimited duration during output overload. If the output current exceeds the OCP set point, the modules will automatically shut down (hiccup mode).

The modules will try to restart after shutdown. If the overload condition still exists, the module will shut down again. This restart trial will continue until the overload condition is corrected.

#### **Over-Voltage Protection**

The modules include an internal output over-voltage protection circuit, which monitors the voltage on the output terminals. If this voltage exceeds the over-voltage set point, the module will shut down (Hiccup mode). The modules will try to restart after shutdown. If the fault condition still exists, the module will shut down again. This restart trial will continue until the fault condition is corrected.

#### **Over-Temperature Protection**

The over-temperature protection consists of circuitry that provides protection from thermal damage. If the temperature exceeds the over-temperature threshold the module will shut down.

The module will try to restart after shutdown. If the over-temperature condition still exists during restart, the module will shut down again. This restart trial will continue until the temperature is within specification.

#### Remote On/Off

The remote on/off feature on the module can be either negative or positive logic. Negative logic turns the module on during a logic low and off during logic high. Positive logic turns the modules on during logic high and off during logic low.

Remote on/off can be controlled by an external switch between the on/off terminal and the Vi(-) terminal. The switch can be an open collector or open drain.

For negative logic if the remote on/off feature is not used, please short the on/off pin to Vi(-). For positive logic if the remote on/off feature is not used, please leave the on/off pin floating.

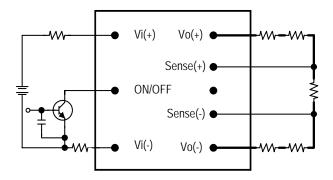


Figure 16: Remote on/off implementation

#### **Remote Sense**

Remote sense compensates for voltage drops on the output by sensing the actual output voltage at the point of load. The voltage between the remote sense pins and the output terminals must not exceed the output voltage sense range given here:

$$[Vo(+) - Vo(-)] - [SENSE(+) - SENSE(-)] \le 10\% \times Vout$$

This limit includes any increase in voltage due to remote sense compensation and output voltage set point adjustment (trim).

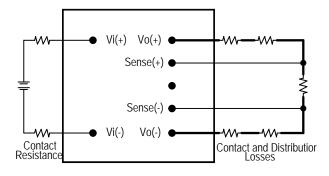


Figure 17: Effective circuit configuration for remote sense operation

If the remote sense feature is not used to regulate the output at the point of load, please connect SENSE(+) to Vo(+) and SENSE(-) to Vo(-) at the module.

The output voltage can be increased by both the remote sense and the trim; however, the maximum increase is the larger of either the remote sense or the trim, not the sum of both.

When using remote sense and trim, the output voltage of the module is usually increased, which increases the power output of the module with the same output current.

Care should be taken to ensure that the maximum output power does not exceed the maximum rated power.

# **FEATURES DESCRIPTIONS (CON.)**

## **Output Voltage Adjustment (TRIM)**

To increase or decrease the output voltage set point, the modules may be connected with an external resistor between the TRIM pin and either the SENSE(+) or SENSE(-). The TRIM pin should be left open if this feature is not used.

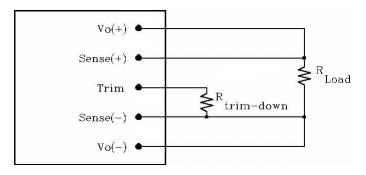


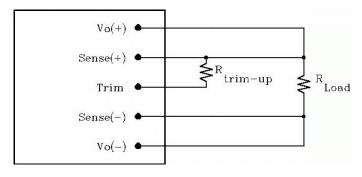
Figure 18: Circuit configuration for trim-down (decrease output voltage)

If the external resistor is connected between the TRIM and SENSE (-) pins, the output voltage set point decreases (Fig. 18). The external resistor value required to obtain a percentage of output voltage change  $\triangle$ % is defined as:

$$Rtrim - down = \frac{511}{\Lambda} - 10.2(K\Omega)$$

Ex. When Trim-down -10% (12V×0.9=10.8V)

Rtrim - down = 
$$\frac{511}{10}$$
 - 10.2 = 40.9(K\O)



**Figure 19:** Circuit configuration for trim-up (increase output voltage)

If the external resistor is connected between the TRIM and SENSE (+) the output voltage set point increases (Fig. 19). The external resistor value required to obtain a percentage output voltage change  $\triangle$ % is defined as:

Rtrim - up = 
$$\frac{5.11\text{Vo}(100 + \Delta)}{1.225 \Delta} - \frac{511}{\Delta} - 10.2(K\Omega)$$

Ex. When Trim-up +10% (12V×1.1=13.2V)

$$Rtrim - up = \frac{5.11 \times 12 \times (100 + 10)}{1.225 \times 10} - \frac{511}{10} - 10.2 = 489 (K\Omega)$$

The output voltage can be increased by both the remote sense and the trim, however the maximum increase is the larger of either the remote sense or the trim, not the sum of both.

When using remote sense and trim, the output voltage of the module is usually increased, which increases the power output of the module with the same output current.

Care should be taken to ensure that the maximum output power of the module remains at or below the maximum rated power.

## THERMAL CONSIDERATIONS

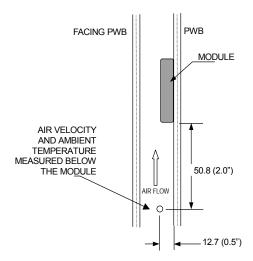
Thermal management is an important part of the system design. To ensure proper, reliable operation, sufficient cooling of the power module is needed over the entire temperature range of the module. Convection cooling is usually the dominant mode of heat transfer.

Hence, the choice of equipment to characterize the thermal performance of the power module is a wind tunnel.

#### Thermal Testing Setup

Delta's DC/DC power modules are characterized in heated vertical wind tunnels that simulate the thermal environments encountered in most electronics equipment. This type of equipment commonly uses vertically mounted circuit cards in cabinet racks in which the power modules are mounted.

The following figure shows the wind tunnel characterization setup. The power module is mounted on a test PWB and is vertically positioned within the wind tunnel. The space between the neighboring PWB and the top of the power module is constantly kept at 6.35mm (0.25").



Note: Wind Tunnel Test Setup Figure Dimensions are in millimeters and (Inches)

Figure 20: Wind tunnel test setup

#### **Thermal Derating**

Heat can be removed by increasing airflow over the module. To enhance system reliability, the power module should always be operated below the maximum operating temperature. If the temperature exceeds the maximum module temperature, reliability of the unit may be affected.

## THERMAL CURVES

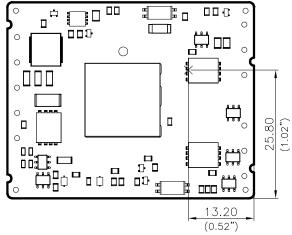
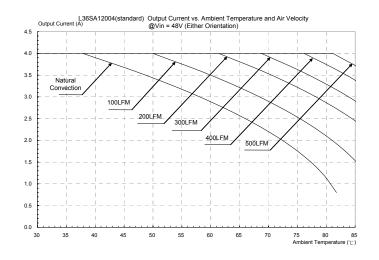


Figure 21: Hot spot temperature measured point

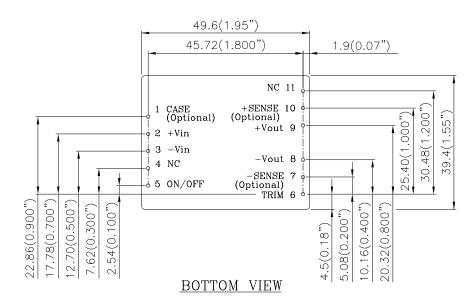
\* The allowed maximum hot spot temperature is defined at 130  ${\mathcal C}$ 

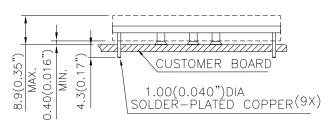


**Figure 22:** Output current vs. ambient temperature and air velocity  $@V_{in}=48V$  (Either Orientation)

# **MECHANICAL DRAWING**

#### **OPEN FRAME VERSION**





SIDE VIEW

NOTES:

DIMENSIONS ARE IN MILLIMETERS AND (INCHES)
TOLERANCES: X.Xmm±0.5mm(X.XX in.±0.02 in.)
X.XXmm±0.25mm(X.XXX in.±0.010 in.)

PIN NO.	NAME	FUNCTION
1	CASE (OPTION)	CASE
2	+VIN	POSITIVE INPUT VOLTAGE
3	-VIN	NEGATIVE INPUT VOLTAGE
4	NC	NOT CONNECTED
5	ON/OFF	REMOTE ON/OFF
6	TRIM	OUTPUT VOLTAGE TRIM
7	-SENSE (OPTION)	NEGATIVE OUTPUT VOLTAGE SENSE
8	-VO	NEGATIVE OUTPUT VOLTAGE
9	+VO	POSITVE OUTPUT VOLTAGE
10	+SENSE (OPTION)	POSITVE OUTPUT VOLTAGE SENSE
11	NC	NOT CONNECTED

#### ALL PINS ARE COPPER WITH TIN PLATING

## PART NUMBERING SYSTEM

L	36	S	Α	120	04	N	R	F	Α
Type of Product	Input Voltage	Number of Outputs	Product Series	Output Voltage	Output Current	ON/OFF Logic	Pin Length		Option Code
L- 2 X 1.6 Brick	18~75V	S- Single	Advanced	120-12.0V	04-4A	N-Negative P-Positive	R-0.170"	F- RoHS 6/6 (Lead Free)	A-Standard Functions B-With sense

#### **MODEL LIST**

MODEL NAME	INPUT		DEL NAME INPUT OUTPUT		EFF @ 100% LOAD	
L36SA3R315NRFA	18V~75V	2.1A	3.3V	15A	88%	
L36SA05010NRFA	18V~75V	1.9A	5V	10A	89%	
L36SA12004NRFA	18V~75V	1.9A	12V	4A	87.5%	

Default remote on/off logic is negative and pin length is 0.170"

For different remote on/off logic and pin length, please refer to part numbering system above or contact your local sales

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#### **WARRANTY**

Delta offers a two (2) year limited warranty. Complete warranty information is listed on our web site or is available upon request from Delta.

Information furnished by Delta is believed to be accurate and reliable. However, no responsibility is assumed by Delta for its use, nor for any infringements of patents or other rights of third parties, which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Delta. Delta reserves the right to revise these specifications at any time, without notice.