# ICS8S89831I

## **DATA SHEET**

### **General Description**

DT



The ICS8S89831I is a high speed 1-to-4 Differentialto-LVPECL/ECL Fanout Buffer. The ICS8S89831I is optimized for high speed and very low output skew, making it suitable for use in demanding applications such as SONET, 1 Gigabit and 10 Gigabit Ethernet,

**Fanout Buffer** 

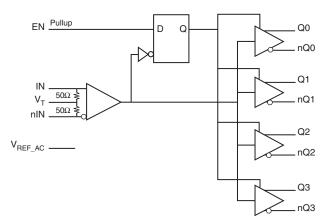
and Fibre Channel. The internally terminated differential input and VREF\_AC pin allow other differential signal families such as LVDS, LVHSTL and CML to be easily interfaced to the input with minimal use of external components. The device also has an output enable pin which may be useful for system test and debug purposes. The ICS8S898311 is packaged in a small 3mm x 3mm 16-pin VFQFN package which makes it ideal for use in space-constrained applications.

### **Features**

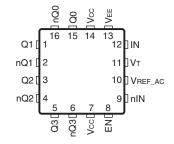
**Differential LVPECL-To-LVPECL/ECL** 

- Four LVPECL/ECL outputs
- IN, nIN input can accept the following differential input levels: LVPECL, LVDS, CML, SSTL
- 50  $\Omega$  internal input termination to V\_T
- Output frequency: >2.1GHz
- Output skew: 30ps (maximum)
- Part-to-part skew: 185ps (maximum)
- Additive phase jitter, RMS: 0.31ps (typical)
- Propagation Delay: 570ps (maximum)
- LVPECL mode operating voltage supply range:  $V_{CC} = 2.5V\pm5\%$ ,  $3.3V\pm5\%$ ,  $V_{EE} = 0V$
- ECL mode operating voltage supply range:  $V_{CC} = 0V$ ,  $V_{EE} = -3.3V \pm 5\%$ , -2.5V  $\pm 5\%$
- -40°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) package

## **Block Diagram**



## **Pin Assignment**



### ICS8S898311

16-Lead VFQFN 3mm x 3mm x 0.925mm package body K Package Top View

Number	Name	Ту	ре	Description
1, 2	Q1, nQ1	Output		Differential output pair. LVPECL/ECL interface levels.
3, 4	Q2, nQ2	Output		Differential output pair. LVPECL/ECL interface levels.
5, 6	Q3, nQ3	Output		Differential output pair. LVPECL/ECL interface levels.
7, 14	V <sub>cc</sub>	Power		Power supply pins.
8	EN	Input	Pullup	Synchronizing clock enable. When LOW, Qx outputs will go LOW and nQx outputs will go HIGH on the next LOW transition at IN input. Input threshold is $V_{CC}/2$ . Includes a 37k $\Omega$ pull-up resistor. Default state is HIGH when left floating. The internal latch is clocked on the falling edge of the input signal IN. LVTTL / LVCMOS interface levels.
9	nIN	Input		Inverting differential LVPECL clock input. RT = $50\Omega$ termination to V <sub>T</sub> .
10	V <sub>REF_AC</sub>	Output		Reference voltage for AC-coupled applications.
11	V <sub>T</sub>	Input		Termination input. I <sub>REF_AC</sub> (max.) < ±2mA.
12	IN	Input		Non-inverting LVPECL differential clock input. RT = $50\Omega$ termination to V <sub>T</sub> .
13	V <sub>EE</sub>	Power		Negative supply pin.
15, 16	Q0, nQ0	Output		Differential output pair. LVPECL/ECL interface levels.

## Table 1. Pin Descriptions

NOTE: Pullup refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

## **Table 2. Pin Characteristics**

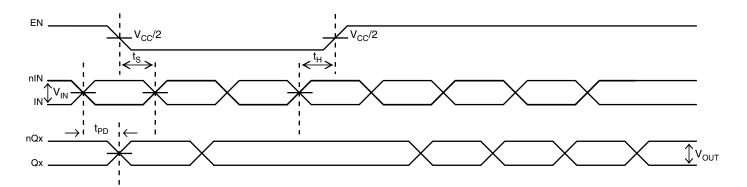
Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
R <sub>PULLUP</sub>	Input Pullup Resistor			37		kΩ

## **Function Tables**

Table 3A. Control Input Function Table

Input	Input Outputs			
EN	Q0:Q3	nQ0:nQ3		
0	Disabled; LOW	Disabled; HIGH		
1	Enabled	Enabled		

NOTE: After EN switches, the clock outputs are disabled or enabled following a falling input clock edge as shown in *Figure 1*.



### Figure 1. EN Timing Diagram

	Inputs		0	utputs
IN	nIN	EN	Q0:Q3	nQ0:nQ3
0	1	1	0	1
1	0	1	1	0
Х	Х	0	0 (NOTE 1)	1(NOTE 1)

NOTE 1: On the next negative transition of the input signal (IN).

## **Absolute Maximum Ratings**

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, V <sub>CC</sub>	4.6V (LVPECL mode, V <sub>EE</sub> = 0V)
Negative Supply Voltage, V <sub>EE</sub>	-4.6V (ECL mode, V <sub>CC</sub> = 0V)
Inputs, V <sub>I</sub> (LVPECL mode)	-0.5V to V <sub>CC</sub> + 0.5V
Inputs, V <sub>I</sub> (ECL mode)	0.5V to V <sub>EE</sub> – 0.5V
Outputs, I <sub>O</sub> Continuous Current Surge Current	50mA 100mA
Input Current, IN, nIN	±50mA
V <sub>T</sub> Current, I <sub>VT</sub>	±100mA
V <sub>REF_AC</sub> Input Sink/Source Current, I <sub>REF_AC</sub>	± 2mA
Operating Temperature Range, T <sub>A</sub>	-40°C to +85°C
Package Thermal Impedance, $\theta_{JA}$ , (Junction-to-Ambient)	74.7°C/W (0 mps)
Storage Temperature, T <sub>STG</sub>	-65°C to 150°C

## **DC Electrical Characteristics**

Table 4A. Power Supply DC Characteristics,  $V_{CC} = 2.5V \pm 5\%$ ,  $3.3V \pm 5\%$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>CC</sub>	Positive Supply Voltage		2.375	3.3	3.465	V
I <sub>EE</sub>	Power Supply Current				45	mA

#### Table 4B. LVCMOS/LVTTL DC Characteristics, V\_{CC} = 2.5V $\pm$ 5%, 3.3V $\pm$ 5%, T\_A = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>IH</sub>	Input High Voltage		2.2		V <sub>CC</sub> + 0.3	V
V <sub>IL</sub>	Input Low Voltage		0		0.8	V
I <sub>IH</sub>	Input High Current	$V_{CC} = V_{IN} = 3.465 V$			10	μA
I <sub>IL</sub>	Input Low Current	$V_{CC} = 3.465 V, V_{IN} = 0 V$	-150			μA

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
R <sub>IN</sub>	Differential Input Resistance	(IN, nIN)	IN to VT, nIN to VT	40	50	60	Ω
V <sub>IH</sub>	Input High Voltage	(IN, nIN)		1.2		V <sub>CC</sub>	V
V <sub>IL</sub>	Input Low Voltage	(IN, nIN)		0		V <sub>IH</sub> – 0.15	V
V <sub>IN</sub>	Input Voltage Swing			0.15		1.2	V
$V_{\text{DIFF}_{\text{IN}}}$	Differential Input Voltage Swin	g		0.3			V
I <sub>IN</sub>	Input Current; NOTE 1	(IN, nIN)				35	mA
$V_{REF}_{AC}$	Bias Voltage			V <sub>CC</sub> – 1.45	V <sub>CC</sub> – 1.37	V <sub>CC</sub> – 1.32	V

### Table 4C. Differential DC Characteristics, $V_{CC}$ = 2.5V ± 5%, 3.3V ± 5%, T<sub>A</sub> = -40°C to 85°C

NOTE 1: Guaranteed by design.

### Table 4D. LVPECL DC Characteristics, $V_{CC}$ = 2.5V $\pm$ 5%, 3.3V $\pm$ 5%, $T_{A}$ = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OH</sub>	Output High Voltage; NOTE 1		V <sub>CC</sub> – 1.175		V <sub>CC</sub> – 0.85	V
V <sub>OL</sub>	Output Low Voltage; NOTE 1		V <sub>CC</sub> – 2.0		V <sub>CC</sub> – 1.575	V
V <sub>OUT</sub>	Output Voltage Swing		0.6		1.0	V
V <sub>DIFF_OUT</sub>	Differential Output Voltage Swing		1.2		2.0	V

NOTE 1: Outputs terminated with 50  $\Omega$  to V\_{CC} – 2V.

## **AC Electrical Characteristics**

**Table 5. AC Characteristics,**  $V_{CC} = 0V$ ;  $V_{EE} = -3.3V \pm 5\%$ ,  $-2.5V \pm 5\%$  or  $V_{CC} = 2.5V \pm 5\%$ ,  $3.3V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f <sub>MAX</sub>	Output Frequence	су	Output Swing $\geq$ 450mV	2.1			GHz
+	Propagation Dela	ay; (Differential);	Input Swing: 150mV	300		570	ps
t <sub>PD</sub>	NOTE 1		Input Swing: 800mV	255		510	ps
<i>t</i> sk(o)	Output Skew; NOTE 2, 4					30	ps
<i>t</i> sk(pp)	Part-to-Part Skew; NOTE 3, 4					185	ps
<i>t</i> jit	Buffer Additive Jitter; RMS; refer to Additive Phase Jitter Section		155.52MHz, Integration Range: 12kHz – 20MHz		0.31		ps
t <sub>S</sub>	Clock Enable Setup Time	EN to IN/nIN		300			ps
t <sub>H</sub>	Clock Enable Hold Time	EN to IN/nIN		300			ps
t <sub>R</sub> / t <sub>F</sub>	Output Rise/Fall	Time	20% to 80%	100		250	ps

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE: All parameters characterized at  $\leq$  1GHz unless otherwise noted.

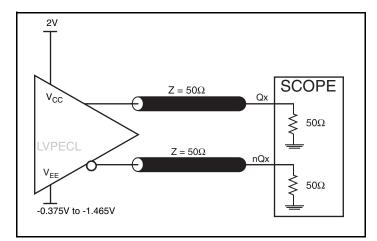
NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the output differential cross points.

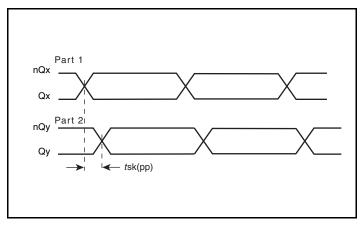
NOTE 3: Defined as skew between outputs on different devices operating at the same supply voltage and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.

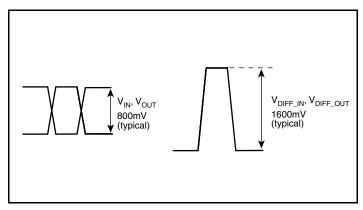
## **Parameter Measurement Information**



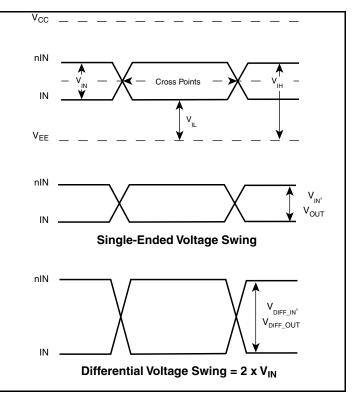
### **Output Load AC Test Circuit**



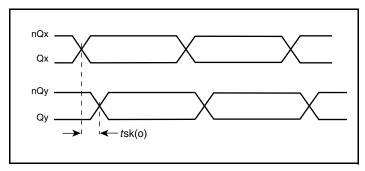
Part-to-Part Skew



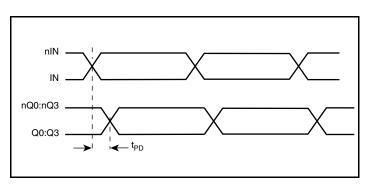
Single-ended & Differential Input Voltage Swing



### **Differential Input Level**

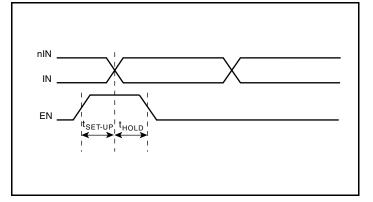


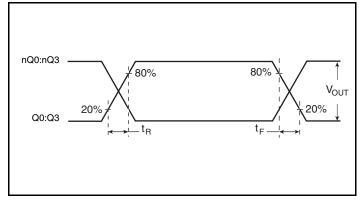
#### **Output Skew**



#### **Propagation Delay**

### Parameter Measurement Information, continued





Setup & Hold Time

**Output Rise/Fall Time** 

## **Application Information**

### **Recommendations for Unused Output Pins**

### **Outputs:**

### **LVPECL Outputs**

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

## 3.3V Differential Input with Built-In 50 $\!\Omega$ Termination Interface

The IN /nIN with built-in 50 $\Omega$  terminations accept LVDS, LVPECL, LVHSTL, CML, SSTL and other differential signals. Both signals must meet the V<sub>IN</sub> and V<sub>IH</sub> input requirements. *Figures 2A to 2D* show interface examples for the IN/nIN input with built-in 50 $\Omega$  terminations driven by the most common driver types. The input interfaces

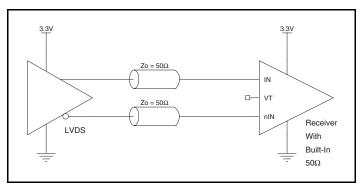


Figure 2A. IN/nIN Input with Built-In 50  $\Omega$  Driven by an LVDS Driver

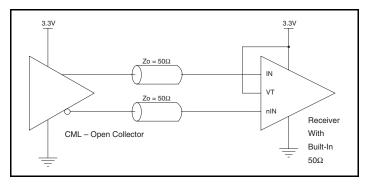


Figure 2C. IN/nIN Input with Built-In 50  $\!\Omega$  Driven by a CML Driver with Open Collector

suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.

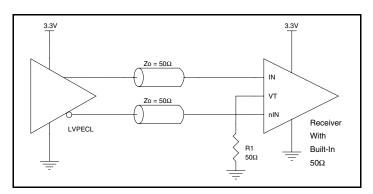


Figure 2B. IN/nIN Input with Built-In 50 $\Omega$  Driven by an LVPECL Driver

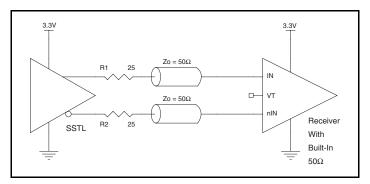
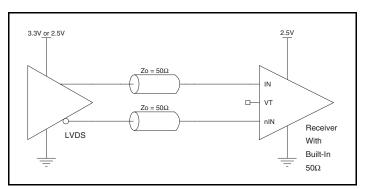
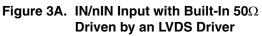


Figure 2D. IN/nIN Input with Built-In 50 $\Omega$  Driven by an SSTL Driver

## 2.5V LVPECL Input with Built-In 50 $\Omega$ Termination Interface

The IN /nIN with built-in 50 $\Omega$  terminations accept LVDS, LVPECL, CML, SSTL and other differential signals. Both signals must meet the V<sub>IN</sub> and V<sub>IH</sub> input requirements. *Figures 3A to 3D* show interface examples for the IN/nIN with built-in 50 $\Omega$  termination input driven by





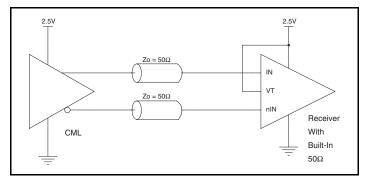


Figure 3C. IN/nIN Input with Built-In 50 $\Omega$ Driven by a CML Driver with Open Collector

the most common driver types. The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.

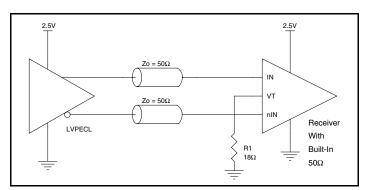


Figure 3B. IN/nIN Input with Built-In 50  $\Omega$  Driven by an LVPECL Driver

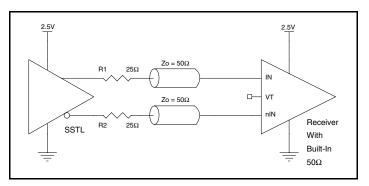


Figure 3D. IN/nIN Input with Built-In 50 $\Omega$  Driven by an SSTL Driver

### **Termination for 3.3V LVPECL Outputs**

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive  $50\Omega$ 

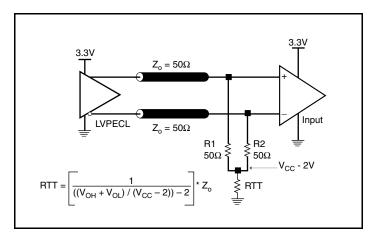


Figure 4A. 3.3V LVPECL Output Termination

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

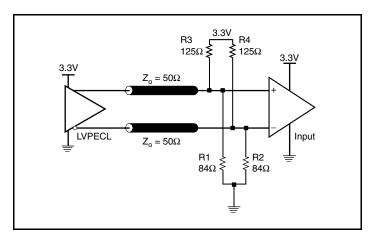


Figure 4B. 3.3V LVPECL Output Termination

## **Termination for 2.5V LVPECL Outputs**

*Figure 5A* and *Figure 5B* show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating  $50\Omega$  to  $V_{CC}$  – 2V. For  $V_{CC}$  = 2.5V, the  $V_{CC}$  – 2V is very close to ground

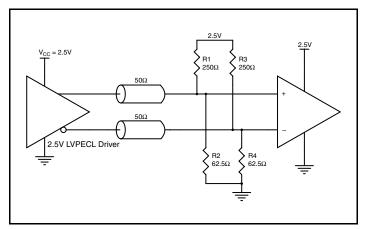


Figure 5A. 2.5V LVPECL Driver Termination Example

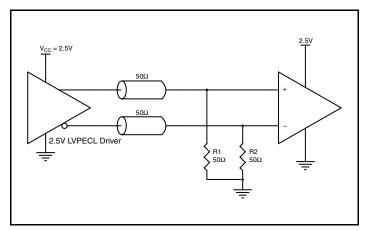


Figure 5C. 2.5V LVPECL Driver Termination Example

level. The R3 in Figure 5B can be eliminated and the termination is shown in *Figure 5C*.

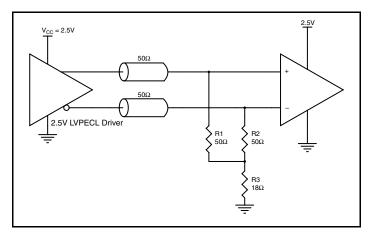


Figure 5B. 2.5V LVPECL Driver Termination Example

## VFQFN EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 6*. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.

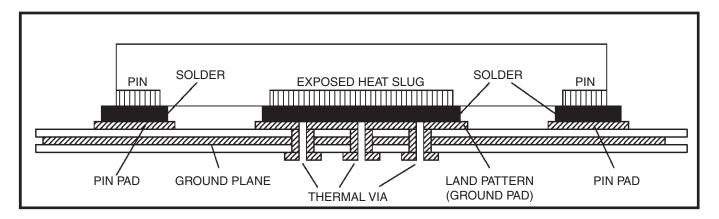


Figure 6. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)

### Schematic Example

*Figure 7* shows a schematic example of the ICS8S898311. This schematic provides examples of input and output handling. The ICS8S898311 input has built-in 50 $\Omega$  termination resistors. The input can directly accept various types of differential signal without AC couple. For AC couple termination, the ICS8S898311 also provides the VREF\_AC pin for proper offset level after the AC couple. This example shows the ICS8S898311 input driven by a 2.5V LVPECL

driver with AC couple. The ICS8S89831I outputs are LVPECL driver. In this example, we assume the traces are long transmission line and the receiver is high input impedance without built-in matched load. An example of 3.3V LVPECL termination is shown in this schematic. Additional termination approaches are shown in the LVPECL Termination Application Note.

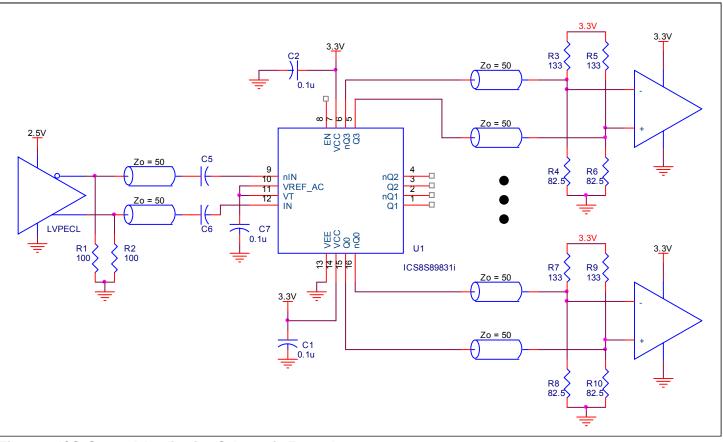


Figure 7. ICS8S89831I Application Schematic Example

### **Power Considerations**

This section provides information on power dissipation and junction temperature for the ICS8S89831I. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the ICS8S898311 is the sum of the core power plus the power dissipation in the load(s). The following is the power dissipation for  $V_{CC} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipation in the load.

- Power (core)<sub>MAX</sub> = V<sub>CC MAX</sub> \* I<sub>EE MAX</sub> = 3.465V \* 45mA = 155.925mW
- Power (outputs)<sub>MAX</sub> = 32.94mW/Loaded Output pair If all outputs are loaded, the total power is 4 \* 32.94mW = 131.76mW
- Power Dissipation for internal termination  $R_T$ Power  $(R_T)_{MAX} = (V_{IN MAX})^2 / R_T MIN = (1.2V)^2 / 80\Omega = 18mW$

Total Power\_MAX (3.3V, with all outputs switching) = 155.925mW + 131.76mW + 18mW = 305.685mW

#### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + T<sub>A</sub>

Tj = Junction Temperature

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

T<sub>A</sub> = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 74.7°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}$ C + 0.306W \* 74.7°C/W = 107.9°C. This is well below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

#### Table 6. Thermal Resistance $\theta_{JA}$ for 16 Lead VFQFN, Forced Convection

θ <sub>JA</sub> by Velocity						
Meters per Second	0	1	2.5			
Multi-Layer PCB, JEDEC Standard Test Boards	74.7°C/W	65.3°C/W	58.5°C/W			

#### 3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pairs.

The LVPECL output driver circuit and termination are shown in Figure 8.

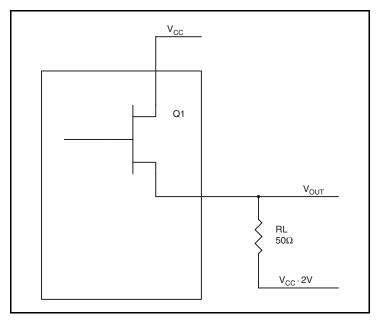


Figure 8. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50 $\Omega$  load, and a termination voltage of V<sub>CC</sub> – 2V.

- For logic high,  $V_{OUT} = V_{OH\_MAX} = V_{CC\_MAX} 0.85V$  $(V_{CC\_MAX} - V_{OH\_MAX}) = 0.85V$
- For logic low, V<sub>OUT</sub> = V<sub>OL\_MAX</sub> = V<sub>CC\_MAX</sub> 1.575V (V<sub>CC\_MAX</sub> - V<sub>OL\_MAX</sub>) = 1.575V

Pd\_H is power dissipation when the output drives high.

Pd\_L is the power dissipation when the output drives low.

 $\begin{array}{l} \mathsf{Pd}_{-}\mathsf{H} = [(\mathsf{V}_{\mathsf{OH}\_\mathsf{MAX}} - (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - 2\mathsf{V}))/\mathsf{R}_{\mathsf{L}}] * (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - \mathsf{V}_{\mathsf{OH}\_\mathsf{MAX}}) = [(2\mathsf{V} - (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - \mathsf{V}_{\mathsf{OH}\_\mathsf{MAX}}))/\mathsf{R}_{\mathsf{L}}] * (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - \mathsf{V}_{\mathsf{OH}\_\mathsf{MAX}}) = [(2\mathsf{V} - 0.85\mathsf{V})/50\Omega] * 0.85\mathsf{V} = 19.55\mathsf{mW} \end{array}$ 

 $\begin{aligned} \mathsf{Pd}_{-L} &= [(\mathsf{V}_{\mathsf{OL}\_\mathsf{MAX}} - (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - 2\mathsf{V}))/\mathsf{R}_{L}] * (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - \mathsf{V}_{\mathsf{OL}\_\mathsf{MAX}}) = [(2\mathsf{V} - (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - \mathsf{V}_{\mathsf{OL}\_\mathsf{MAX}}))/\mathsf{R}_{L}] * (\mathsf{V}_{\mathsf{CC}\_\mathsf{MAX}} - \mathsf{V}_{\mathsf{OL}\_\mathsf{MAX}}) = [(2\mathsf{V} - 1.575\mathsf{V})/50\Omega] * 1.575\mathsf{V} = 13.39\mathsf{mW} \end{aligned}$ 

Total Power Dissipation per output pair = Pd\_H + Pd\_L = **32.94mW** 

### **Reliability Information**

### Table 7. $\theta_{\text{JA}}$ vs. Air Flow Table for a 16 Lead VFQFN

$\theta_{JA}$ by Velocity					
Meters per Second	0	1	2.5		
Multi-Layer PCB, JEDEC Standard Test Boards	74.7°C/W	65.3°C/W	58.5°C/W		

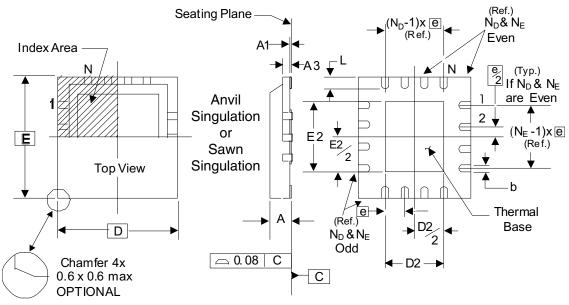
### **Transistor Count**

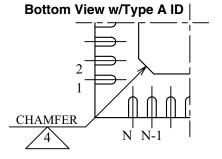
The transistor count for ICS8S898311 is: 328

This device is pin and function compatible and a suggested replacement for ICS889831.

## Package Outline and Package Dimensions







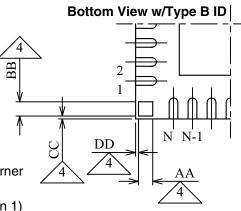
There are 3 methods of indicating pin 1 corner at the back of the VFQFN package are:

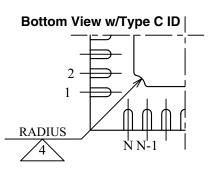
- 1. Type A: Chamfer on the paddle (near pin 1)
- 2. Type B: Dummy pad between pin 1 and N.
- 3. Type C: Mouse bite on the paddle (near pin 1)

### Table 8. Package Dimensions

JEDEC Variation: VEED-2/-4 All Dimensions in Millimeters				
Symbol	Minimum	Maximum		
Ν	16			
Α	0.80	1.00		
A1	0	0.05		
A3	0.25 Ref.			
b	0.18	0.30		
N <sub>D</sub> & N <sub>E</sub>	4			
D & E	3.00 Basic			
D2 & E2	1.00	1.80		
е	0.50 Basic			
L	0.30	0.50		

Reference Document: JEDEC Publication 95, MO-220





## **Ordering Information**

#### Table 9. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8S89831AKILF	831A	"Lead-Free" 16 Lead VFQFN	Tube	-40°C to 85°C
8S89831AKILFT	831A	"Lead-Free" 16 Lead VFQFN	2500 Tape & Reel	-40°C to 85°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

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# **Revision History Sheet**

Rev	Table	Page	Description of Change	Date
		11	Deleted Differential Input with Built-in $50\Omega$ Termination Unused Input Handling application section. This section does not apply when there is only one input.	
A		15	Power Considerations - in Power Dissipation section, corrected Power (RT) calculation. Calcuation = 18mW from 98mW. Total Power and Junction Temperature calculations have also been updated.	4/22/10



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