

High-Voltage, Current-Mode, PWM Controller

Features

- 10 to 450V input voltage range
- <1.3 mA supply current
- >1 MHz clock
- 49% maximum duty version

Applications

- Off-line high frequency power supplies
- Universal input power supplies
- High density power supplies
- Very high efficiency power supplies
- Extra wide load range power supplies

Description

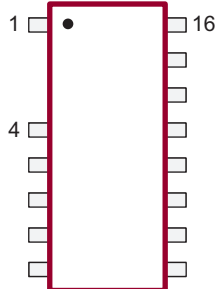
HV9120 and HV9123 are Switch-Mode Power Supply (SMPS) controllers suitable for the control of a variety of converter topologies, including flyback and forward converter.

Using an internal, high-voltage regulator, HV9120 and HV9123 can derive a bias supply for starting-up and powering a converter from a variety of power sources, such as a 12V battery or the rectified AC (230 VAC) line.

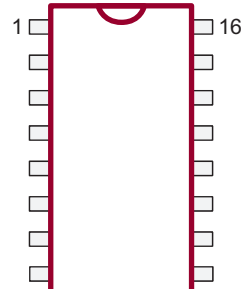
HV9120/HV9123 controllers include all essentials for a power-converter design, such as a bandgap reference, an error amplifier, a ramp generator, a high-speed PWM comparator, and a gate driver. A shutdown latch provides on/off control. Device power consumption is less than 6 mW when shutdown.

HV9120 offers 50% maximum duty and HV9123 offers nearly 100% duty.

Package Types



16-lead SOIC

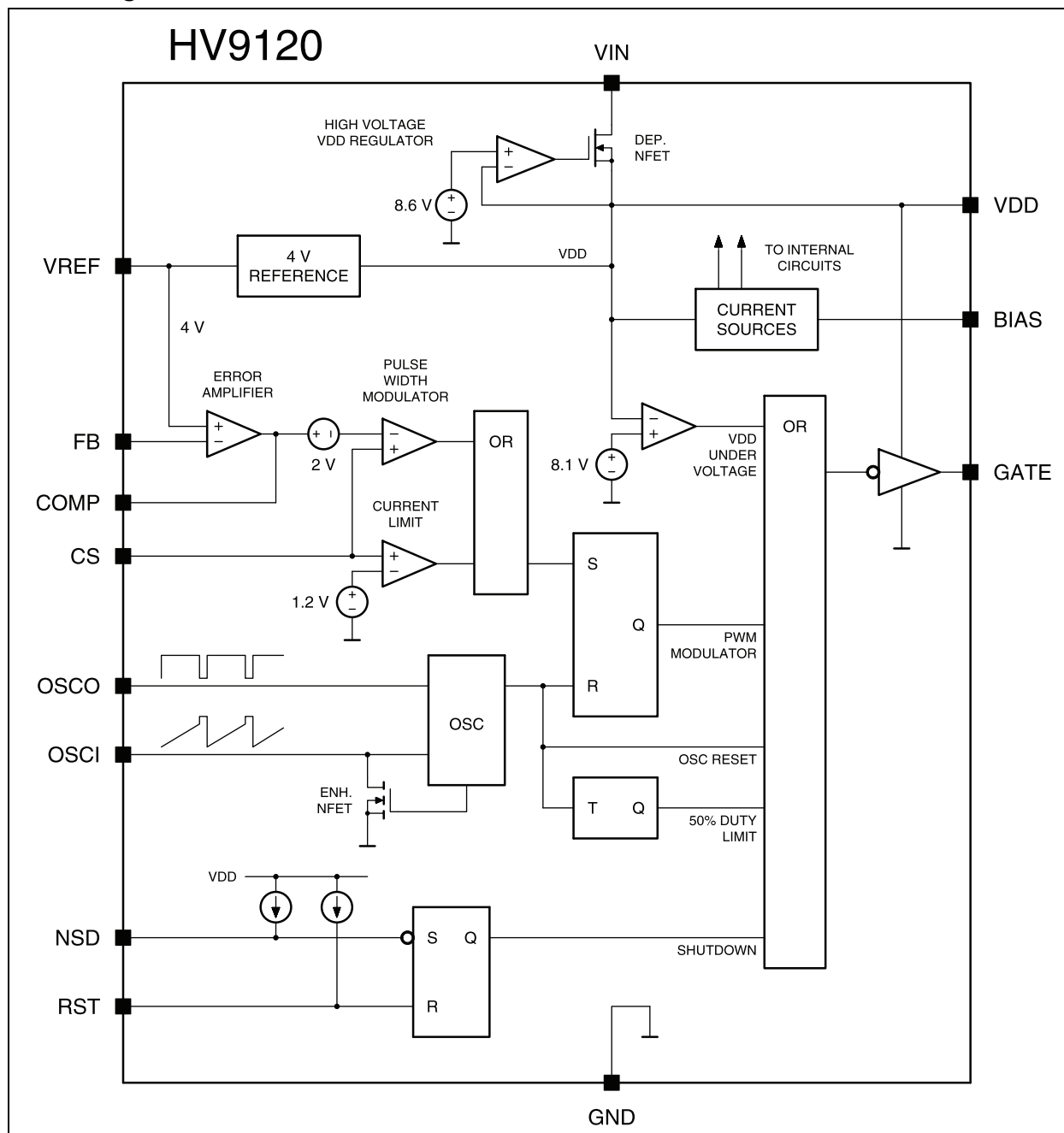


16-lead PDIP

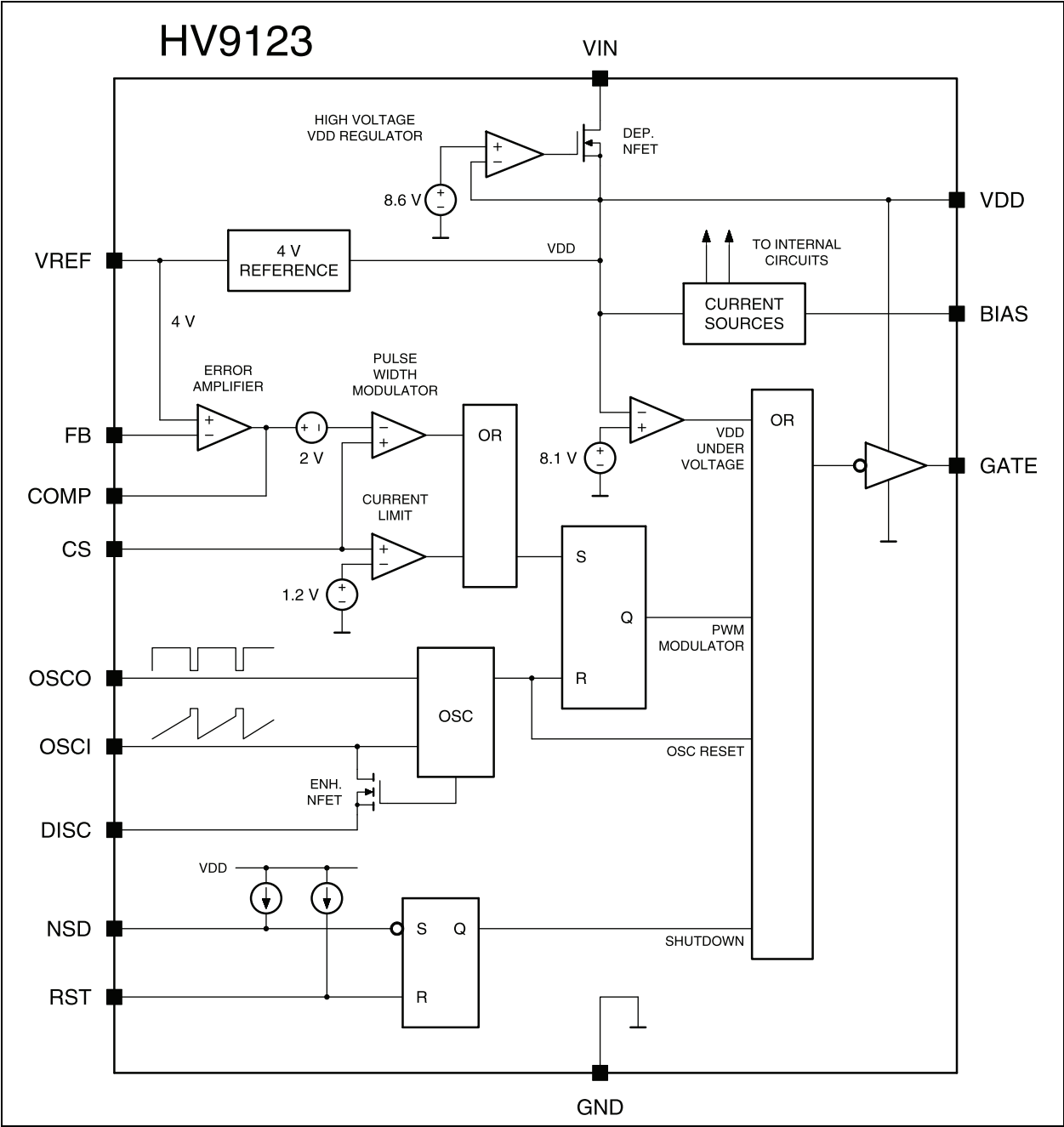
See [Table 3-1](#) for pin information

HV9120/HV9123

Block Diagram HV9120



Block Diagram HV9123



HV9120/HV9123

1.0 ELECTRICAL CHARACTERISTICS

ABSOLUTE MAXIMUM RATINGS†

Input voltage, V_{IN}	450V
Device supply voltage, V_{DD}	15.5V
Logic input voltage	-0.3V to $V_{DD} + 0.3V$
Linear input voltage	-0.3V to $V_{DD} + 0.3V$
High-voltage regulator input current (continuous), I_{IN}	2.5 mA
Operating temperature range	-40°C to +125°C
Storage temperature range	-65°C to +150°C
Power dissipation: 16-Lead SOIC	900 mW
16-Lead PDIP	1000 mW

† **Notice:** Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

Electrical Specifications: $V_{DD} = 10V$, $V_{IN} = 48V$, $V_{DISC} = 0V$, $R_{BIAS} = 390\text{ k}\Omega$, $R_{OSC} = 330\text{ k}\Omega$, $T_A = 25^\circ\text{C}$, unless otherwise noted.

Parameter	Symbol	Min	Typ	Max	Units	Conditions	
Reference							
Output voltage	V _{REF}	3.92	4.00	4.08	V	R _L = 10 MΩ	
		3.84	4.00	4.16		R _L = 10 MΩ, T _A = -40°C to +125°C	
Output impedance	Z _{OUT}	15	30	45	kΩ	(Note 1)	
Short circuit current	I _{SHORT}	-	125	250	μA	V _{REF} = GND	
Change in V _{REF} with temperature	ΔV _{REF}	-	0.25	-	mV/°C	T _A = -40°C to +125°C (Note 1)	
Oscillator							
Oscillator frequency	f _{MAX}	1.0	3.0	-	MHz	R _{OSC} = 0Ω	
Initial accuracy	f _{OSC}	80	100	120	kHz	R _{OSC} = 330 kΩ (Note 2)	
		160	200	240		R _{OSC} = 150 kΩ (Note 2)	
VDD regulation	-	-	-	15	%	9.5V< V _{DD} <13.5V	
Temperature coefficient	-	-	170	-	ppm/°C	T _A = -40°C to +125°C (Note 1)	
PWM							
Maximum duty cycle	HV9120	D _{MAX}	49.0	49.4	49.6	%	(Note 1)
	HV9123		95	97	99		
Dead time	HV9123	D _{MIN}	-	225	-	ns	HV9123 only (Note 1)
Minimum duty cycle			-	-	0	%	—
Pulse width where pulse drops out			-	80	125	ns	(Note 1)
Current Limit							
Maximum input signal	V _{LIM}	1.0	1.2	1.4	V	V _{FB} = 0V	
Delay to output	t _D	-	80	120	ns	V _{CS} = 1.5V, V _{COMP} ≤ 2.0V (Note 1)	

ELECTRICAL CHARACTERISTICS (CONTINUED)

Electrical Specifications: $V_{DD} = 10V$, $V_{IN} = 48V$, $V_{DISC} = 0V$, $R_{BIAS} = 390\text{ k}\Omega$, $R_{OSC} = 330\text{ k}\Omega$, $T_A = 25^\circ\text{C}$, unless otherwise noted.

Parameter		Symbol	Min	Typ	Max	Units	Conditions
Error Amplifier							
Feedback voltage		V _{FB}	3.92	4.00	4.08	V	FB shorted to COMP
Input bias current		I _{IN}	-	25	500	nA	V _{FB} = 4.0V
Input offset voltage		V _{OS}	nulled during trim			-	–
Open loop voltage gain		A _{VOL}	60	80	-	dB	(Note 1)
Unity gain bandwidth		GB	1.0	1.3	-	MHz	(Note 1)
Output source current		I _{SOURCE}	-1.4	-2.0	-	mA	V _{FB} = 3.4V
Output sink current		I _{SINK}	0.12	0.15	-	mA	V _{FB} = 4.5V
High-voltage Regulator and Start-up							
Input voltage		V _{IN}	10	-	450	V	I _{IN} < 10 μA; V _{CC} > 9.4V
Input leakage current		I _{IN}	-	-	10	μA	V _{DD} > 9.4V
Regulator turn-off threshold voltage		V _{TH}	8.0	8.7	9.4	V	I _{IN} = 10 μA
Undervoltage lockout		V _{LOCK}	7.0	8.1	8.9	V	–
Supply							
Supply current		I _{DD}	-	0.75	1.3	mA	C _L < 75 pF
Quiescent supply current		I _Q	-	0.55	-	mA	V _{NSD} = 0V
Nominal bias current		I _{BIAS}	-	20	-	μA	–
Operating range		V _{DD}	9.0	-	13.5	V	–
Shutdown Logic							
Shutdown delay		t _{SD}	-	50	100	ns	C _L = 500 pF, V _{CS} = 0V (Note 1)
NSD pulse width		t _{SW}	50	-	-	ns	(Note 1)
RST pulse width		t _{RW}	50	-	-	ns	(Note 1)
Latching pulse width		t _{LW}	25	-	-	ns	V _{NSD} , V _{RST} =0V(Note 1)
Input low voltage		V _{IL}	-	-	2.0	V	–
Input high voltage		V _{IH}	7.0	-	-	V	–
Input current, input high voltage		I _{IH}	-	1.0	5.0	μA	V _{IN} = V _{DD}
Input current, input low voltage		I _{IL}	-	-25	-35	μA	V _{IN} = 0V
Output							
Output high voltage		V _{OH}	V _{DD} -0.25	-	-	V	I _{OUT} = 10 mA
			V _{DD} -0.3	-	-		I _{OUT} = 10 mA, T _A = -40°C to 125°C
Output low voltage		V _{OL}	-	-	0.2	V	I _{OUT} = -10 mA
			-	-	0.3		I _{OUT} = -10 mA, T _A = -40°C to 125°C
Output resistance	Pull up	R _{OUT}	-	15	25	Ω	I _{OUT} = ±10 mA
	Pull down		-	8.0	20		
	Pull up		-	20	30	Ω	I _{OUT} = ±10 mA, T _A = -40°C to 125°C
	Pull down		-	10	30		
Rise time		t _R	-	30	75	ns	C _L = 500 pF (Note 1)
Fall time		t _F	-	20	75	ns	C _L = 500 pF(Note 1)

Note 1: Design guidance only; Not 100% tested in production.

2: Stray capacitance on OSC in pin must be $\leq 5\text{ pF}$.

HV9120/HV9123

TEMPERATURE SPECIFICATIONS

Parameter	Symbol	Min	Typ	Max	Units	Conditions
Temperature Ranges						
Operating Temperature		-40		125	°C	
Storage Temperature		-65	–	150	°C	
Package Thermal Resistances						
Thermal Resistance, SOIC	θ_{ja}	–	83	–	°C/W	
Thermal Resistance, PDIP	θ_{ja}	–	51	–	°C/W	

1.1 Truth Table

TRUTH TABLE

SHUTDOWN	RESET	OUTPUT
H	H	Normal operation
H	H → L	Normal operation, no change
L	H	Off, not latched
L	L	Off, latched
L → H	L	Off, latched, no change

2.0 TYPICAL PERFORMANCE CURVES

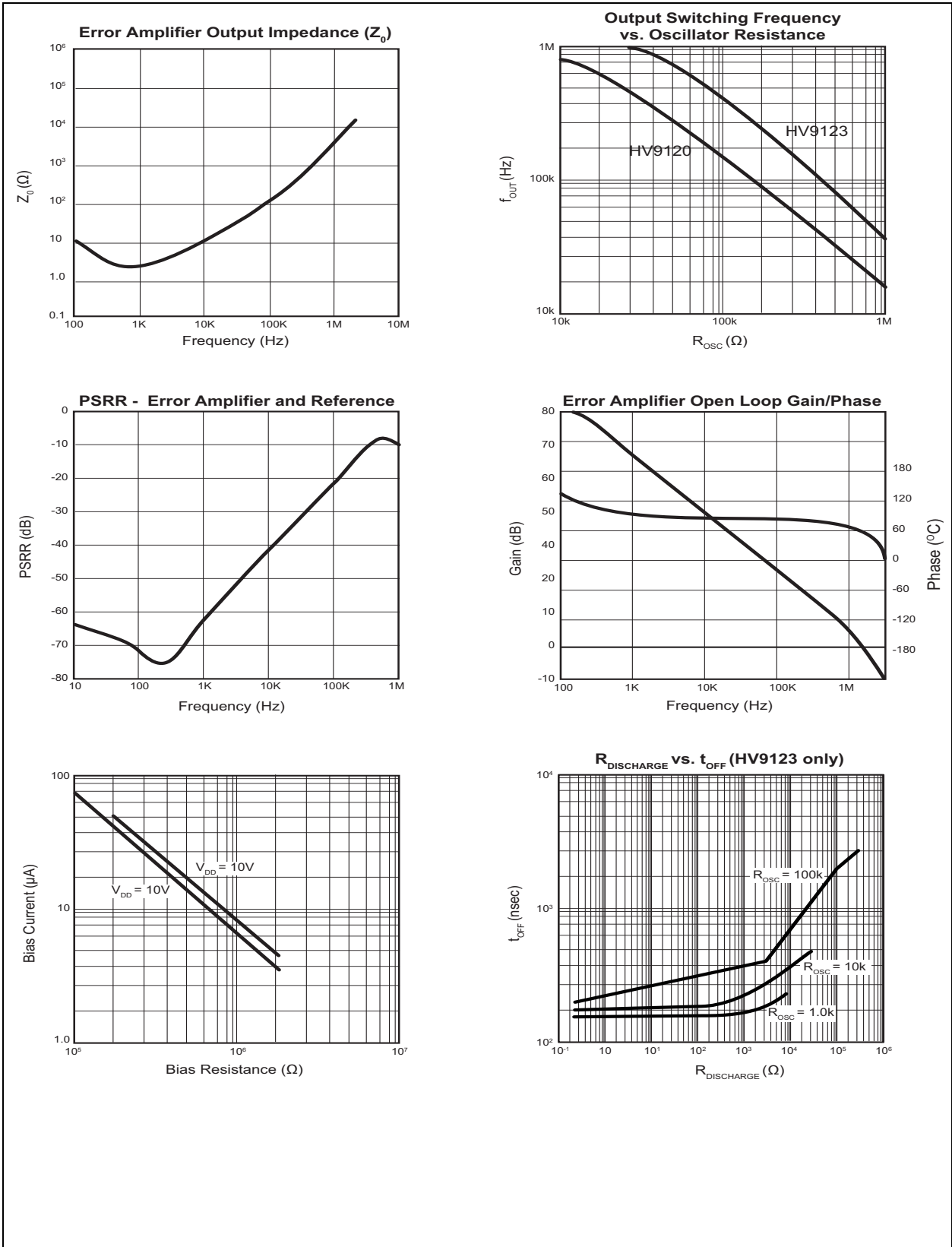


FIGURE 2-1: Typical Performance Curves

HV9120/HV9123

3.0 PIN DESCRIPTION

The locations of the pins are listed in [Features](#).

TABLE 3-1: PIN DESCRIPTION

Pin #	Symbol HV9120	Symbol HV9123	Description
1	V _{IN}	V _{IN}	High-voltage, V _{DD} regulator input
2	NC	NC	No connect
3	NC	NC	No connect
4	CS	CS	Current-sense input
5	GATE	GATE	Gate-drive output
6	GND	GND	Ground
7	VDD	VDD	High-voltage, V _{DD} regulator output
8	OSCO	OSCO	Oscillator output
9	OSCI	OSCI	Oscillator Input
10	NC	DISC	Oscillator discharge, current set
11	VREF	VREF	4V Reference output Reference voltage level can be overridden by an externally-applied voltage source.
12	NSD	NSD	Active low input to set shutdown latch
13	RST	RST	Active high input to reset shutdown latch
14	COMP	COMP	Error-amplified output
15	FB	FB	Feedback-voltage input
16	BIAS	BIAS	Internal bias, current set

4.0 TEST CIRCUITS

The test circuits for characterizing error-amplifier output impedance, Z_{OUT} , and error-amplifier, power-supply rejection ratio, PSRR, are shown in [Figure 4-1](#).

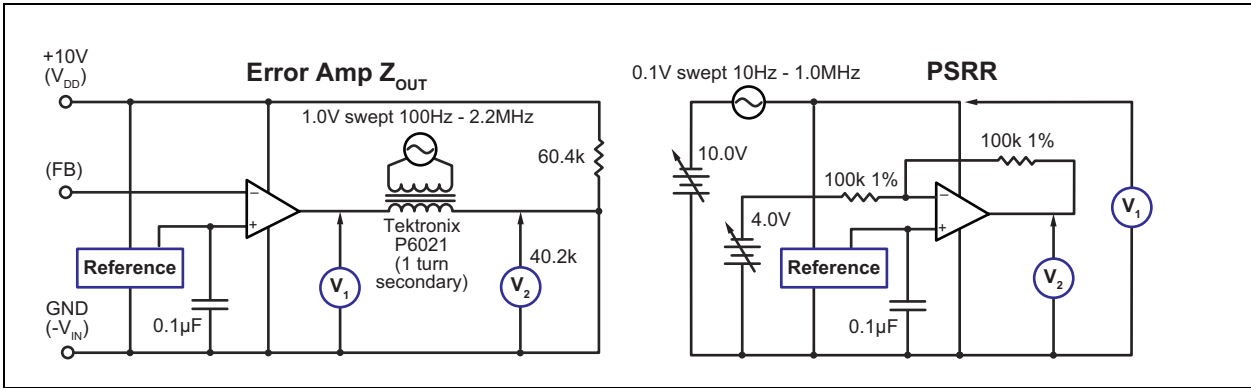


FIGURE 4-1: Test Circuits

5.0 DETAILED DESCRIPTION

5.1 High-Voltage Regulator

The high-voltage regulator included in HV9120 and HV9123 consists of a high-voltage, n-channel, depletion-mode DMOS transistor, driven by an error amplifier, providing a current path between the V_{IN} terminal and the V_{DD} terminal. The maximum current, about 20 mA, occurs when $V_{DD} = 0$, with current reducing as V_{DD} rises. This path shuts off when V_{DD} rises to somewhere between 7.8 and 9.4V. So, if V_{DD} is held at 10 or 12V by an external source, no current other than leakage is drawn through the high voltage transistor. This minimizes dissipation.

Use an external capacitor between V_{DD} and GND to store energy used by the chip in the time between shut-off of the high voltage path and the V_{DD} supply's output rising enough to take over powering the chip. This capacitor should have a value of 100X or more the effective gate capacitance of the MOSFET being driven, as well as very good high-frequency characteristics. See the equation below. Ceramic caps work well. Electrolytic capacitors are generally not suitable.

$$C_{VDD} \geq 100 \times (\text{gate charge of FET at 10V})$$

The device uses a resistor divider string to monitor V_{DD} for both the under voltage lockout circuit and the shutoff circuit of the high voltage FET. Setting the under voltage sense point about 0.6V lower on the string than the FET shutoff point guarantees that the under voltage lockout releases before the FET shuts off.

5.2 Bias Circuit

HV9120 and HV9123 require an external bias resistor, connected between the BIAS pin and GND, to set currents in a series of current mirrors used by the analog sections of the chip. The nominal external bias current requirement is 15 to 20 μA , which can be set by a 390 k Ω to 510 k Ω resistor if $V_{DD} = 10\text{V}$, or a 510 k Ω to 680 k Ω resistor if $V_{DD} = 12\text{V}$. A precision resistor is not required, $\pm 5\%$ meets the device requirements.

5.3 Clock Oscillator

The clock oscillator of the HV9120 and HV9123 consists of a ring of CMOS inverters, timing capacitors, and a capacitor-discharge FET. A single external resistor between the OSC1 and OSC0 sets the oscillator frequency (see [Figure 2-1](#), Output Switching Frequency vs Oscillator Resistance).

HV9120 includes a frequency-dividing flip-flop that allows the part to operate with a 50% duty limit. Accordingly, the effective switching frequency of the power

converter is half the oscillator frequency (see [Figure 2-1](#), Output Switching Frequency vs Oscillator Resistance).

An internal, discharge FET resets the oscillator ramp at the end of the oscillator cycle. The FET is internally connected to GND in HV9120 (50% max duty version). Whereas, the FET is externally connected to GND, by way of a resistor, in the HV9123 (100% duty version). The resistor programs the oscillator dead time at the end of the oscillator period in HV9123 applications.

The oscillator turns off during shutdown to reduce supply current by about 150 μA .

5.4 Reference

The reference of the HV9120 and HV9123 consists of a band-gap reference, followed by a buffer amplifier, which scales the voltage up to 4.0V. The scaling resistors of the buffer amplifier are trimmed during manufacture so that the output of the error amplifier, when connected in a gain of -1 configuration, is as close to 4.0V as possible. This nulls out the input offset of the error amplifier. As a consequence, even though the observed reference voltage of a specific part may not be exactly 4.0V, the feedback voltage required for proper regulation will be 4.0V.

An approximately 50 k Ω resistor is located internally between the output of the reference buffer amplifier and the circuitry it feeds—reference output pin and non-inverting input to the error amplifier. This allows overriding the internal reference with a low impedance voltage source $\leq 6.0\text{V}$. Using an external reference reinstates the input offset voltage of the error amplifier. Overriding the reference should seldom be necessary.

The reference of the HV9120 and HV9123 is a high impedance node, and usually there will be significant electrical noise nearby. Therefore, a bypass capacitor between the reference pin and GND is strongly recommended. The reference buffer amplifier is compensated to be stable with a capacitive load of 0.01 to 0.1 μF .

5.5 Error Amplifier

The error amplifier in HV9120 and HV9123 is a low-power, differential-input, operational amplifier. A PMOS input stage is used, so the common mode range includes ground and the input impedance is high.

5.6 Current Sense Comparators

HV9120 and HV9123 use a dual-comparator system with independent comparators for modulation and current limiting. This allows the designer greater latitude in compensation design, as there are no clamps, except ESD protection, on the compensation pin.

5.7 Remote Shutdown

The NSD and RST pins control the shutdown latch. These pins have internal, current-source pull-ups so they can be driven from open drain logic. When not used they should be left open, or connected to V_{DD} .

5.8 Output Buffer

The output buffer of HV9120 and HV9123 is of standard CMOS construction—P-channel pull-up and N-channel pull-down. Thus, the body-drain diodes of the output stage can be used for spike clipping. External Schottky diode clamping of the output is not required.

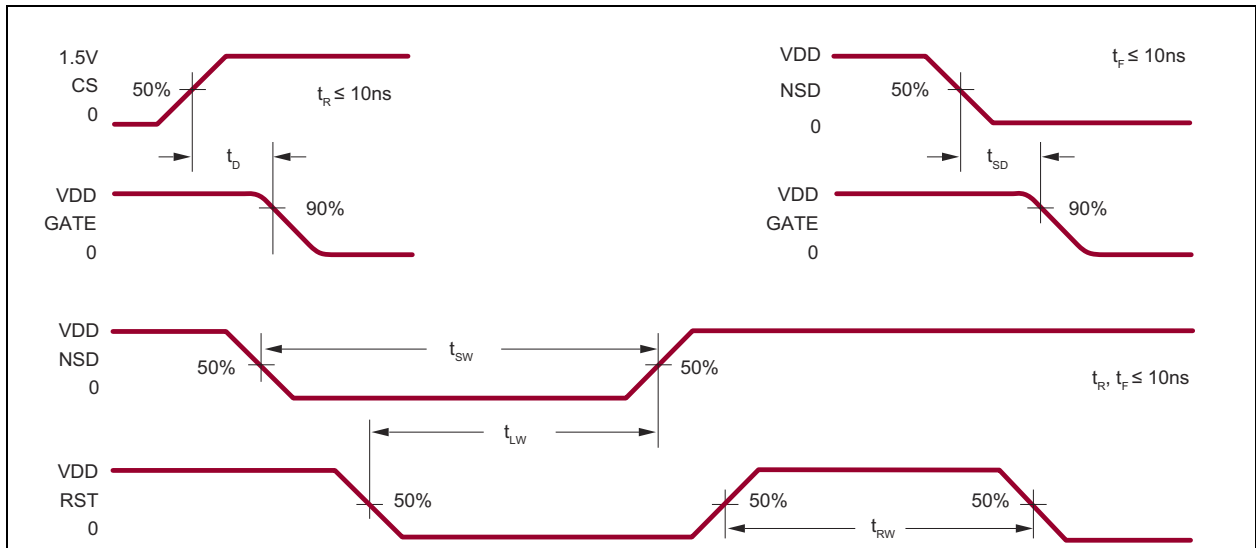
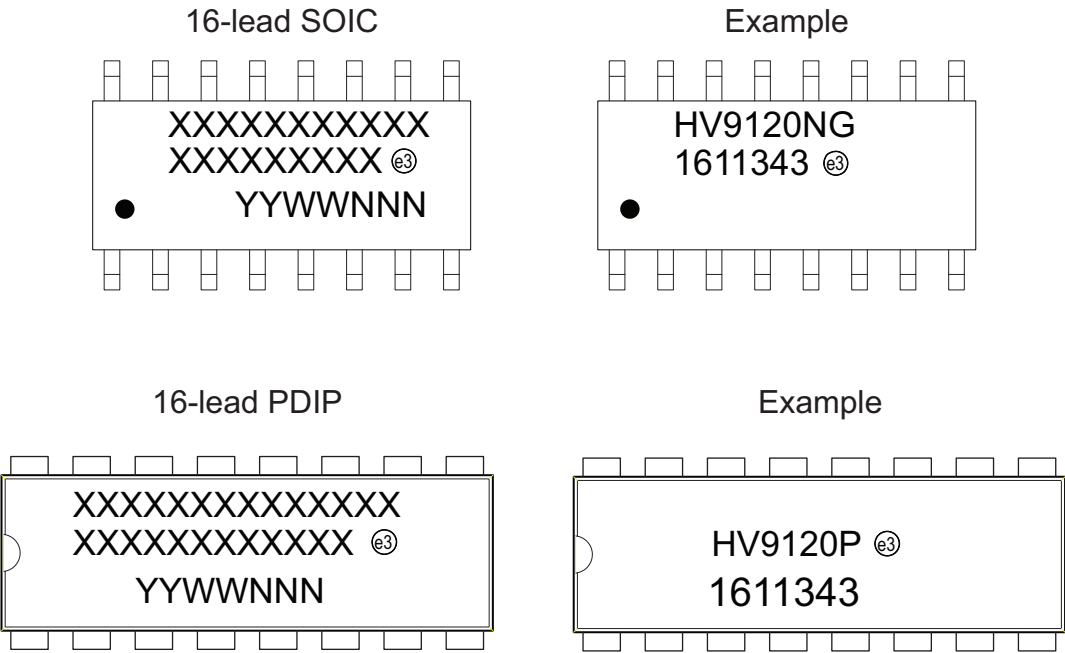


FIGURE 5-1: Shutdown Timing Waveforms

HV9120/HV9123

6.0 PACKAGING INFORMATION

6.1 Package Marking Information



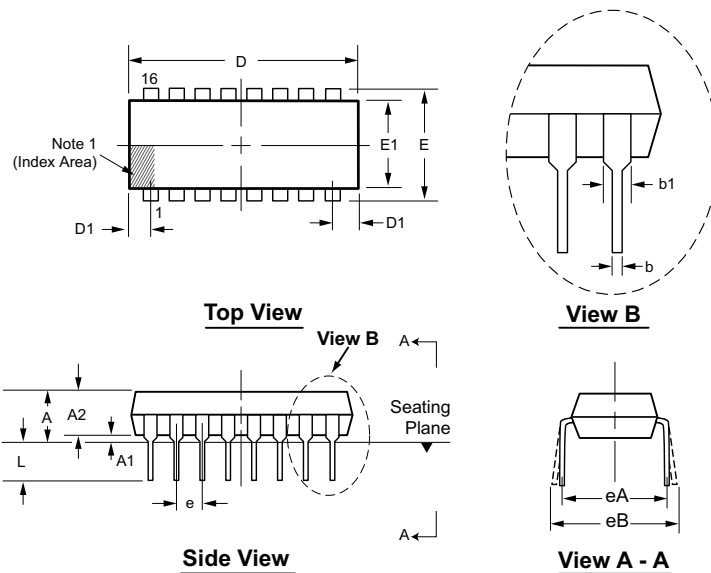
Legend:	XX...X	Product Code or Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	ⓔ3	Pb-free JEDEC® designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator (ⓔ3) can be found on the outer packaging for this package.
Note:	In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for product code or customer-specific information. Package may or not include the corporate logo.	

DS20005519A-page 13

Drawings are not to scale.

HV9120/HV9123

16-Lead PDIP (.300in Row Spacing) Package Outline (P) .790x.250in body, .210in height (max), .100in pitch



Note: For the most current package drawings, see the Microchip Packaging Specification at www.microchip.com/packaging.

Note:

1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbol		A	A1	A2	b	b1	D	D1	E	E1	e	eA	eB	L
Dimension (inches)	MIN	.130*	.015	.115	.014	.045	.745†	.005	.290†	.240	.100 BSC	.300 BSC	.300*	.115
	NOM	-	-	.130	.018	.060	.790	-	.310	.250			-	.130
	MAX	.210	.035*	.195	.023†	.070	.810†	.050*	.325	.280			.430	.150

JEDEC Registration MS-001, Variation AB, Issue D, June, 1993.

* This dimension is not specified in the JEDEC drawing.

† This dimension differs from the JEDEC drawing.

Drawings not to scale.

APPENDIX A: REVISION HISTORY

Revision A (May 2016)

- Updated file to Microchip format.
- Merged Supertex Doc #s DSFP-HV9120 and DSFP-HV9123 to Microchip DS20005519A.
- Revised [Electrical Characteristics](#) to accommodate the merged products.
- Updated Pin names to reflect new naming convention.
- Significant text changes to [Detailed Description](#)
- Minor text changes throughout.

HV9120/HV9123

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<u>PART NO.</u>	<u>XX</u>	-	<u>X</u>	-	<u>X</u>
Device	Package Options		Environmental		Media Type
Device:	HV9120	=	High Voltage Current-Mode PWM Controller, 10 to 450V input voltage range, 49% duty cycle		
	HV9123	=	High Voltage Current-Mode PWM Controller, 9 to 80V input voltage range, 99% duty cycle		
Package:	NG	=	16-lead SOIC		
	P	=	16-lead PDIP		
Environmental	G	=	Lead (Pb)-free/ROHS-compliant package		
Media Type:	(blank)	=	45/Tube for NG package 24/Tube for P package		
	M901	=	2600/Reel for NG package		
	M934	=	2600/Reel for NG package		

Examples:
a) HV9120NG-G 14-Lead SOIC package, 53/Tube
b) HV9123NG-G 14-Lead SOIC package, 53/Tube
c) HV9123NG-G-M901 14-Lead SOIC package, 2600/Reel

Note: For media types M901 and M934, the base quantity for tap and reel was standardized at 2600/reel. Both options will result in delivery of the same number of parts/reel.

Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights unless otherwise stated.

Trademarks

The Microchip name and logo, the Microchip logo, dsPIC, FlashFlex, flexPWR, JukeBlox, KEELOQ, KEELOQ logo, Klear, LANCheck, MediaLB, MOST, MOST logo, MPLAB, OptoLyzer, PIC, PICSTART, PIC³² logo, RightTouch, SpyNIC, SST, SST Logo, SuperFlash and UNI/O are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

The Embedded Control Solutions Company and mTouch are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Analog-for-the-Digital Age, BodyCom, chipKIT, chipKIT logo, CodeGuard, dsPICDEM, dsPICDEM.net, ECAN, In-Circuit Serial Programming, ICSP, Inter-Chip Connectivity, KlearNet, KlearNet logo, MiWi, motorBench, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, RightTouch logo, REAL ICE, SQI, Serial Quad I/O, Total Endurance, TSHARC, USBCheck, VariSense, ViewSpan, WiperLock, Wireless DNA, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

Silicon Storage Technology is a registered trademark of Microchip Technology Inc. in other countries.

GestIC is a registered trademark of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2016, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

ISBN: 978-1-5224-0537-5

QUALITY MANAGEMENT SYSTEM
CERTIFIED BY DNV
== ISO/TS 16949 ==

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.

Worldwide Sales and Service

AMERICAS

Corporate Office
2355 West Chandler Blvd.
Chandler, AZ 85224-6199
Tel: 480-792-7200
Fax: 480-792-7277
Technical Support:
<http://www.microchip.com/support>
Web Address:
www.microchip.com

Atlanta
Duluth, GA
Tel: 678-957-9614
Fax: 678-957-1455

Austin, TX
Tel: 512-257-3370

Boston
Westborough, MA
Tel: 774-760-0087
Fax: 774-760-0088

Chicago
Itasca, IL
Tel: 630-285-0071
Fax: 630-285-0075

Cleveland
Independence, OH
Tel: 216-447-0464
Fax: 216-447-0643

Dallas
Addison, TX
Tel: 972-818-7423
Fax: 972-818-2924

Detroit
Novi, MI
Tel: 248-848-4000

Houston, TX
Tel: 281-894-5983

Indianapolis
Noblesville, IN
Tel: 317-773-8323
Fax: 317-773-5453

Los Angeles
Mission Viejo, CA
Tel: 949-462-9523
Fax: 949-462-9608

New York, NY
Tel: 631-435-6000

San Jose, CA
Tel: 408-735-9110

Canada - Toronto
Tel: 905-673-0699
Fax: 905-673-6509

ASIA/PACIFIC

Asia Pacific Office
Suites 3707-14, 37th Floor
Tower 6, The Gateway
Harbour City, Kowloon

Hong Kong
Tel: 852-2943-5100
Fax: 852-2401-3431

Australia - Sydney
Tel: 61-2-9868-6733
Fax: 61-2-9868-6755

China - Beijing
Tel: 86-10-8569-7000
Fax: 86-10-8528-2104

China - Chengdu
Tel: 86-28-8665-5511
Fax: 86-28-8665-7889

China - Chongqing
Tel: 86-23-8980-9588
Fax: 86-23-8980-9500

China - Dongguan
Tel: 86-769-8702-9880

China - Hangzhou
Tel: 86-571-8792-8115
Fax: 86-571-8792-8116

China - Hong Kong SAR
Tel: 852-2943-5100
Fax: 852-2401-3431

China - Nanjing
Tel: 86-25-8473-2460
Fax: 86-25-8473-2470

China - Qingdao
Tel: 86-532-8502-7355
Fax: 86-532-8502-7205

China - Shanghai
Tel: 86-21-5407-5533
Fax: 86-21-5407-5066

China - Shenyang
Tel: 86-24-2334-2829
Fax: 86-24-2334-2393

China - Shenzhen
Tel: 86-755-8864-2200
Fax: 86-755-8203-1760

China - Wuhan
Tel: 86-27-5980-5300
Fax: 86-27-5980-5118

China - Xian
Tel: 86-29-8833-7252
Fax: 86-29-8833-7256

ASIA/PACIFIC

China - Xiamen
Tel: 86-592-2388138
Fax: 86-592-2388130

China - Zhuhai
Tel: 86-756-3210040
Fax: 86-756-3210049

India - Bangalore
Tel: 91-80-3090-4444
Fax: 91-80-3090-4123

India - New Delhi
Tel: 91-11-4160-8631
Fax: 91-11-4160-8632

India - Pune
Tel: 91-20-3019-1500

Japan - Osaka
Tel: 81-6-6152-7160
Fax: 81-6-6152-9310

Japan - Tokyo
Tel: 81-3-6880-3770
Fax: 81-3-6880-3771

Korea - Daegu
Tel: 82-53-744-4301
Fax: 82-53-744-4302

Korea - Seoul
Tel: 82-2-554-7200
Fax: 82-2-558-5932 or
82-2-558-5934

Malaysia - Kuala Lumpur
Tel: 60-3-6201-9857
Fax: 60-3-6201-9859

Malaysia - Penang
Tel: 60-4-227-8870
Fax: 60-4-227-4068

Philippines - Manila
Tel: 63-2-634-9065
Fax: 63-2-634-9069

Singapore
Tel: 65-6334-8870
Fax: 65-6334-8850

Taiwan - Hsin Chu
Tel: 886-3-5778-366
Fax: 886-3-5770-955

Taiwan - Kaohsiung
Tel: 886-7-213-7828

Taiwan - Taipei
Tel: 886-2-2508-8600
Fax: 886-2-2508-0102

Thailand - Bangkok
Tel: 66-2-694-1351
Fax: 66-2-694-1350

EUROPE

Austria - Wels
Tel: 43-7242-2244-39
Fax: 43-7242-2244-393

Denmark - Copenhagen
Tel: 45-4450-2828
Fax: 45-4485-2829

France - Paris
Tel: 33-1-69-53-63-20
Fax: 33-1-69-30-90-79

Germany - Dusseldorf
Tel: 49-2129-3766400

Germany - Karlsruhe
Tel: 49-721-625370

Germany - Munich
Tel: 49-89-627-144-0
Fax: 49-89-627-144-44

Italy - Milan
Tel: 39-0331-742611
Fax: 39-0331-466781

Italy - Venice
Tel: 39-049-7625286

Netherlands - Drunen
Tel: 31-416-690399
Fax: 31-416-690340

Poland - Warsaw
Tel: 48-22-3325737

Spain - Madrid
Tel: 34-91-708-08-90
Fax: 34-91-708-08-91

Sweden - Stockholm
Tel: 46-8-5090-4654

UK - Wokingham
Tel: 44-118-921-5800
Fax: 44-118-921-5820