

FCAS50SN60

Smart Power Module for SRM

Features

- Very low thermal resistance due to using DBC
- 600V-50A single-phase asymmetric bridge IGBT converter for SRM drive including control ICs for gate driving and protection
- Divided negative dc-link terminals for inverter current sensing applications
- Single-grounded power supply due to built-in HVIC
- Switching frequency of 2.2~8kHz
- Isolation rating of 2500Vrms/min.

Applications

- AC 200V ~ 242V single-phase SRM drives for home application vacuum cleaner.

General Description

FCAS50SN60 is an advanced smart power module for SRM drive that Fairchild has newly developed and designed to provide very compact and high performance SRM motor drives mainly targeting low-power inverter-driven SRM application especially for a vacuum air cleaner. It combines optimized circuit protection and drive matched to low-loss IGBTs. System reliability is further enhanced by the integrated under-voltage lock-out and short-circuit protection. The high speed built-in HVIC provides opto-coupler-less IGBT gate driving capability that further reduce the overall size of the inverter system design. In addition the incorporated HVIC facilitates the use of single-supply drive topology enabling the FCAS50SN60 to be driven by only one drive supply voltage without negative bias. Each phase current of inverter can be monitored separately due to the divided negative dc terminals.



Figure 1.

Integrated Power Functions

- 600V-50A IGBT asymmetric converter for single-phase SRM drives (Please refer to Figure 3)

Integrated Drive, Protection and System Control Functions

- For high-side IGBTs: Gate drive circuit, High voltage isolated high-speed level shifting
Control circuit under-voltage (UV) protection
Note) Available bootstrap circuit example is given in Figures 10.
- For low-side IGBTs: Gate drive circuit, Short circuit protection (SC)
Control supply circuit under-voltage (UV) protection
- Fault signaling: Corresponding to a UV fault (Low-side supply)
- Input interface: 5V CMOS/LSTTL compatible, Schmitt trigger input

Pin Configuration

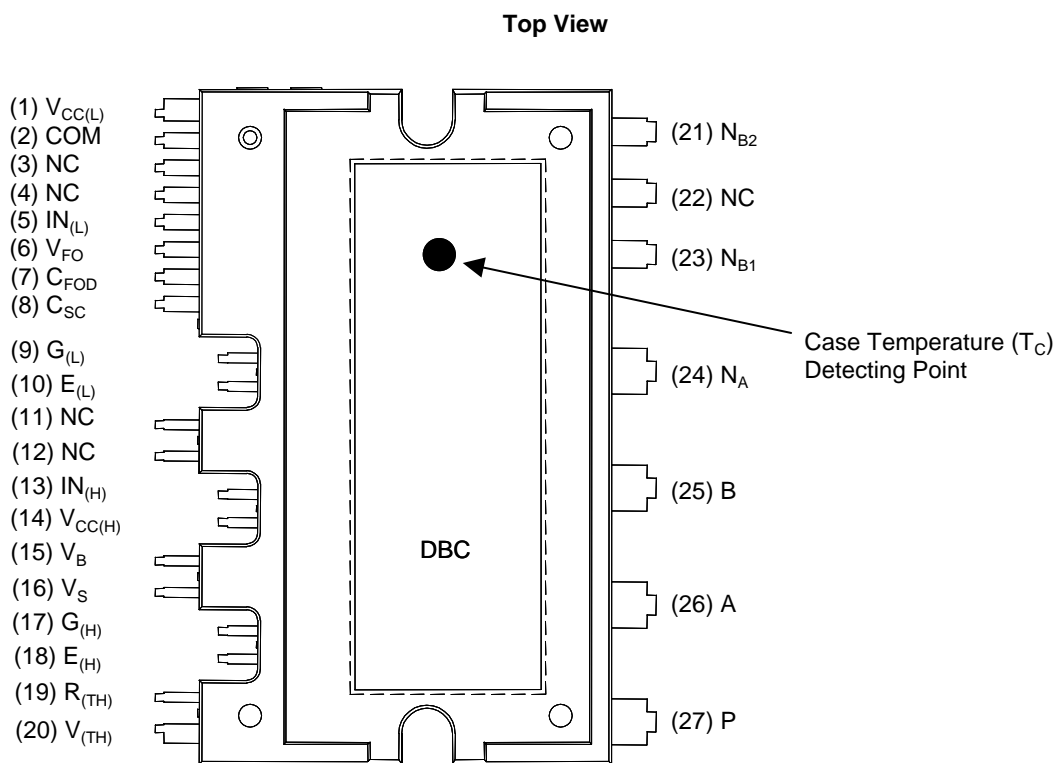
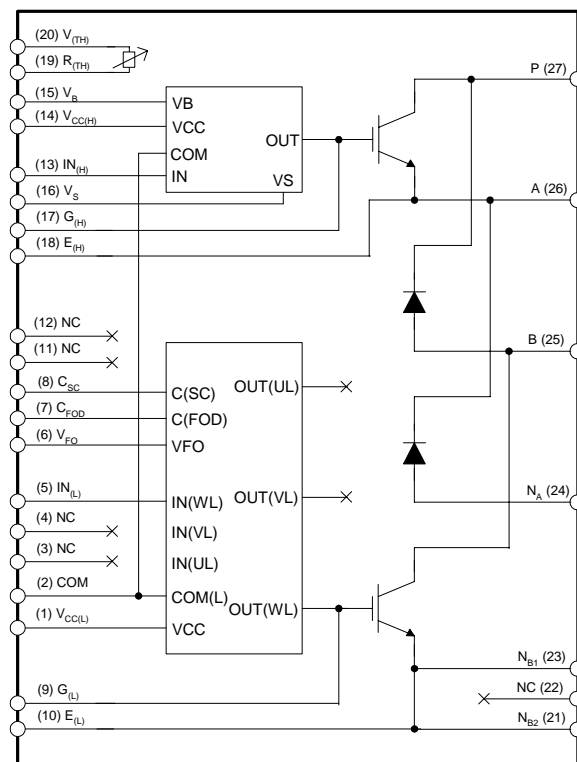


Figure 2.

Pin Descriptions

Pin Number	Pin Name	Pin Description
1	$V_{CC(L)}$	Low-side Common Bias Voltage for IC and IGBTs Driving
2	COM	Common Supply Ground
3	NC	Dummy Pin
4	NC	Dummy Pin
5	$IN_{(L)}$	Signal Input for Low-side IGBT
6	V_{FO}	Fault Output
7	C_{FOD}	Capacitor for Fault Output Duration Time Selection
8	C_{SC}	Capacitor (Low-pass Filter) for Short-Current Detection
9	$G_{(L)}$	Gate terminal of low-side IGBT
10	$E_{(L)}$	Emitter terminal of low-side IGBT
11	NC	Dummy Pin
12	NC	Dummy Pin
13	$IN_{(H)}$	Signal Input for High-side IGBT
14	$V_{CC(H)}$	High-side Bias Voltage
15	V_B	High-side Bias Voltage for Gate Driving
16	V_S	High-side Bias Voltage Ground for Gate Driving
17	$G_{(H)}$	Gate terminal of the High-side IGBT
18	$E_{(H)}$	Emitter terminal of the High-side IGBT
19	$R_{(TH)}$	Thermistor Series Resistor
20	$V_{(TH)}$	Thermistor Bias Voltage
21	N_{B2}	Negative DC–Link Input for B Leg (Should be shorted with N_{B1} externally)
22	NC	Dummy Pin
23	N_{B1}	Negative DC–Link Input for B Leg (Should be shorted with N_{B2} externally)
24	N_A	Negative DC–Link Input for A Leg
25	B	Output for B Leg
26	A	Output for A Leg
27	P	Positive DC–Link Input

Internal Equivalent Circuit and Input/Output Pins



Note:

1. The low-side is composed of one IGBT and freewheeling diode and one control IC which has gate driving and protection functions.
2. The power side is composed of four dc-link input terminals and two output terminals.
3. The high-side is composed of one IGBT and freewheeling diode and one drive IC for high-side IGBT.

Figure 3.

Absolute Maximum Ratings ($T_J = 25^\circ\text{C}$, Unless Otherwise Specified)**Inverter Part**

Symbol	Parameter	Conditions	Rating	Units
$V_{PN(\text{Surge})}$	Supply Voltage (Surge)	Applied between P- N_A , N_{B1} , N_{B2}	550	V
V_{CES}	Collector-emitter Voltage		600	V
$\pm I_C$	Each IGBT Collector Current	$T_C = 25^\circ\text{C}$	50	A
$\pm I_{CP}$	Each IGBT Collector Current (Peak)	$T_C = 25^\circ\text{C}$, Under 1ms Pulse Width	100	A
P_C	Collector Dissipation	$T_C = 25^\circ\text{C}$ per One IGBT	110	W
T_J	Operating Junction Temperature	(Note 1)	-20 ~ 125	$^\circ\text{C}$

Note:

1. The maximum junction temperature rating of the power chips integrated within the module is 150°C ($@T_C \leq 100^\circ\text{C}$). However, to insure safe operation, the average junction temperature should be limited to $T_{J(\text{ave})} \leq 125^\circ\text{C}$ ($@T_C \leq 100^\circ\text{C}$)

Control Part

Symbol	Parameter	Conditions	Rating	Units
V_{CC}	Control Supply Voltage	Applied between $V_{CC(H)}$, $V_{CC(L)}$ - COM	20	V
V_{BS}	High-side Control Bias Voltage	Applied between V_B - V_S	20	V
V_{IN}	Input Signal Voltage	Applied between $IN_{(H)}$, $IN_{(L)}$ - COM	-0.3~5.5	V
V_{FO}	Fault Output Supply Voltage	Applied between V_{FO} - COM	-0.3~ $V_{CC}+0.3$	V
I_{FO}	Fault Output Current	Sink Current at V_{FO} Pin	5	mA
V_{SC}	Current Sensing Input Voltage	Applied between C_{SC} - COM	-0.3~ $V_{CC}+0.3$	V

Total System

Symbol	Parameter	Conditions	Rating	Units
$V_{PN(\text{PROT})}$	Self Protection Supply Voltage Limit (Short Circuit Protection Capability)	$V_{CC} = V_{BS} = 13.5 \sim 16.5\text{V}$ $T_J = 125^\circ\text{C}$, Non-repetitive, less than $6\mu\text{s}$	400	V
T_C	Module Case Operation Temperature		-20 ~ 95	$^\circ\text{C}$
T_{STG}	Storage Temperature		-40 ~ 125	$^\circ\text{C}$
V_{ISO}	Isolation Voltage	60Hz, Sinusoidal, AC 1 minute, Connection Pins to DBC	2500	V_{rms}

Thermal Resistance

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$R_{th(j-c)Q}$	Junction to Case Thermal Resistance	Each IGBT under Operating Condition	-	-	0.90	$^\circ\text{C/W}$
$R_{th(j-c)F}$		Each FWDi under Operating Condition	-	-	2.2	$^\circ\text{C/W}$

Note:

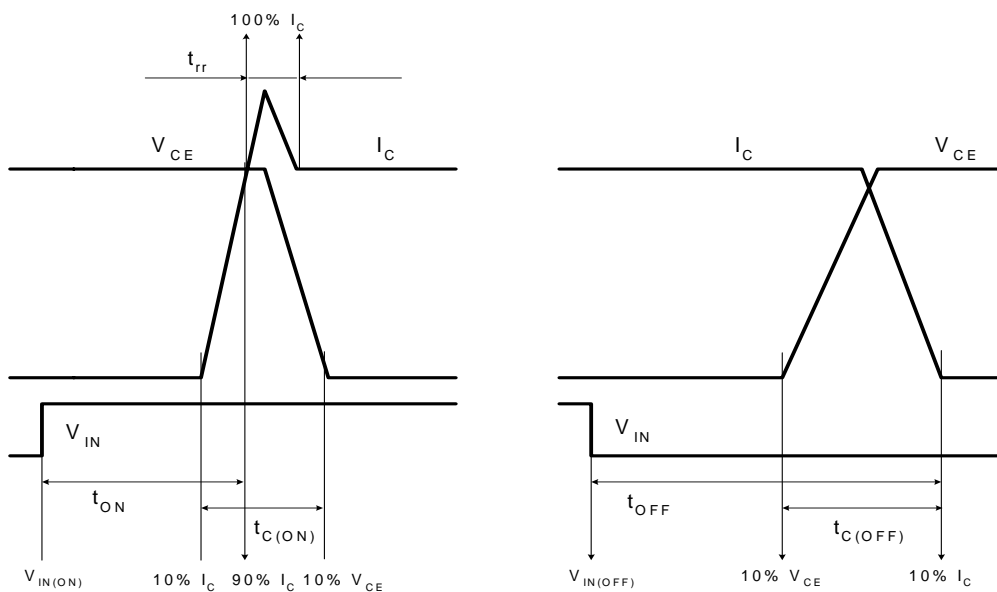
2. For the measurement point of case temperature (T_C), please refer to Figure 2.

Electrical Characteristics ($T_J = 25^\circ\text{C}$, Unless Otherwise Specified)**Inverter Part**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$V_{CE(SAT)}$	Collector-Emitter Saturation Voltage	$V_{CC} = V_{BS} = 15\text{V}$ $V_{IN} = 5\text{V}$ $I_C = 50\text{A}$, $T_J = 25^\circ\text{C}$	-	1.6	2.3	V
V_{FM}	FWDi Forward Voltage	$V_{IN} = 0\text{V}$ $I_C = 50\text{A}$, $T_J = 25^\circ\text{C}$	-	2.1	3.0	V
HS	t_{ON}	$V_{PN} = 300\text{V}$, $V_{CC} = V_{BS} = 15\text{V}$ $I_C = 50\text{A}$ $V_{IN} = 0\text{V} \leftrightarrow 5\text{V}$, Inductive Load $R_{E(H)} = 10\Omega$ (Note 3)	-	0.8	-	μs
	$t_{C(ON)}$		-	0.6	-	μs
	t_{OFF}		-	1.5	-	μs
	$t_{C(OFF)}$		-	0.8	-	μs
	t_{rr}		-	0.08	-	μs
LS	t_{ON}	$V_{PN} = 300\text{V}$, $V_{CC} = V_{BS} = 15\text{V}$ $I_C = 50\text{A}$ $V_{IN} = 0\text{V} \leftrightarrow 5\text{V}$, Inductive Load (Note 3)	-	1.1	-	μs
	$t_{C(ON)}$		-	0.9	-	μs
	t_{OFF}		-	1.5	-	μs
	$t_{C(OFF)}$		-	0.8	-	μs
	t_{rr}		-	0.05	-	μs
I_{CES}	Collector - Emitter Leakage Current	$V_{CE} = V_{CES}$	-	-	250	μA

Note:

3. t_{ON} and t_{OFF} include the propagation delay time of the internal drive IC. $t_{C(ON)}$ and $t_{C(OFF)}$ are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.

**Figure 4. Switching Time Definition**

Electrical Characteristics ($T_J = 25^\circ\text{C}$, Unless Otherwise Specified)**Control Part**

Symbol	Parameter	Conditions		Min.	Typ.	Max.	Units
I _{QCCL}	Quiescent V _{CC} Supply Current	V _{CC} = 15V IN _(L) = 0V	V _{CC(L)} - COM	-	-	40	mA
I _{QCCH}		V _{CC} = 15V IN _(H) = 0V	V _{CC(H)} - COM	-	-	80	μA
I _{QBS}	Quiescent V _{BS} Supply Current	V _{BS} = 15V IN _(H) = 0V	V _B - V _S	-	-	100	μA
V _{FOH}	Fault Output Voltage	V _{SC} = 0V, V _{FO} Circuit: 4.7kΩ to 5V Pull-up		4.5	-	-	V
V _{FOL}		V _{SC} = 1V, V _{FO} Circuit: 4.7kΩ to 5V Pull-up		-	-	0.8	V
V _{SC(ref)}	Short Circuit Trip Level	V _{CC} = 15V (Note 4)		0.45	0.5	0.55	V
UV _{CCD}	Supply Circuit Under-Voltage Protection	Detection Level	Applied between V _{CC(L)} - COM	10.5	-	12.5	V
UV _{CCR}		Reset Level		11.0	-	13	V
UV _{BSD}		Detection Level	Applied between V _B - V _S	10.0	-	12.5	V
UV _{BSR}		Reset Level		10.5	-	13.0	V
t _{FOD}	Fault-out Pulse Width	C _{FOD} = 33nF (Note 5)		1.4	1.8	2.0	ms
V _{IH}	ON Threshold Voltage	Logic'1' input voltage	Applied between IN _(H) , IN _(L) - COM	3.0	-	-	V
V _{IL}	OFF Threshold Voltage	Logic'0' input voltage		-	-	0.8	V
I _{INH(ON)}	Input Bias Current	IN _(H) = 5V	Applied between IN _(H) , IN _(L) - COM	0.9	-	2.2	mA
I _{INL(ON)}		IN _(L) = 5V		0.9	-	2.4	mA
R _{TH}	Resistance of Thermistor	@ T _C = 25°C (Note Fig. 10)		-	50	-	kΩ
		@ T _C = 80°C (Note Fig. 10)		-	5.76	-	kΩ

Note:

4. Short-circuit current protection is functioning only at the low-sides.

5. The fault-out pulse width t_{FOD} depends on the capacitance value of C_{FOD} according to the following approximate equation : $C_{FOD} = 18.3 \times 10^{-6} \times t_{FOD}[\text{F}]$ **Recommended Operating Conditions**

Symbol	Parameter	Conditions	Value			Units
			Min.	Typ.	Max.	
V_{PN}	Supply Voltage	Applied between P - N_A , N_{B1} , N_{B2}	-	300	450	V
V_{CC}	Control Supply Voltage	Applied between $V_{CC(H)}$, $V_{CC(L)} - \text{COM}$	13.5	15	16.5	V
V_{BS}	High-side Bias Voltage	Applied between $V_B - V_S$	13.5	15	18.5	V
f_{PWM}	PWM Input Signal	$T_C \leq 100^\circ\text{C}$, $T_J \leq 125^\circ\text{C}$	-	3	-	kHz
$V_{IN(ON)}$	Input ON Voltage	Applied between $I_{N(H)}$, $I_{N(L)} - \text{COM}$	4 ~ 5.5			V
$V_{IN(OFF)}$	Input OFF Voltage	Applied between $I_{N(H)}$, $I_{N(L)} - \text{COM}$	0 ~ 0.65			V

Mechanical Characteristics and Ratings

Parameter	Conditions		Limits			Units
			Min.	Typ.	Max.	
Mounting Torque	Mounting Screw - M3		5.17	6.29	7.30	Kg•cm
			0.51	0.62	0.72	N•m
Surface Flatness		Note Figure 5.	0	-	120	um
Weight			-	15.0	-	g

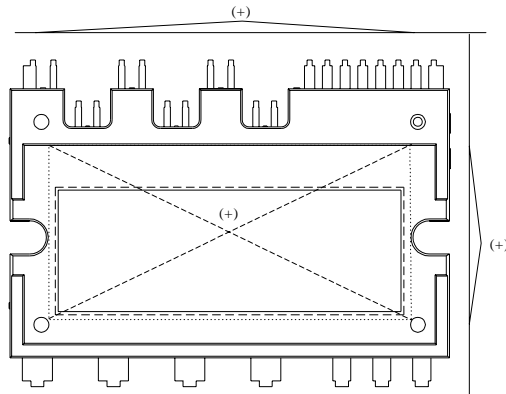
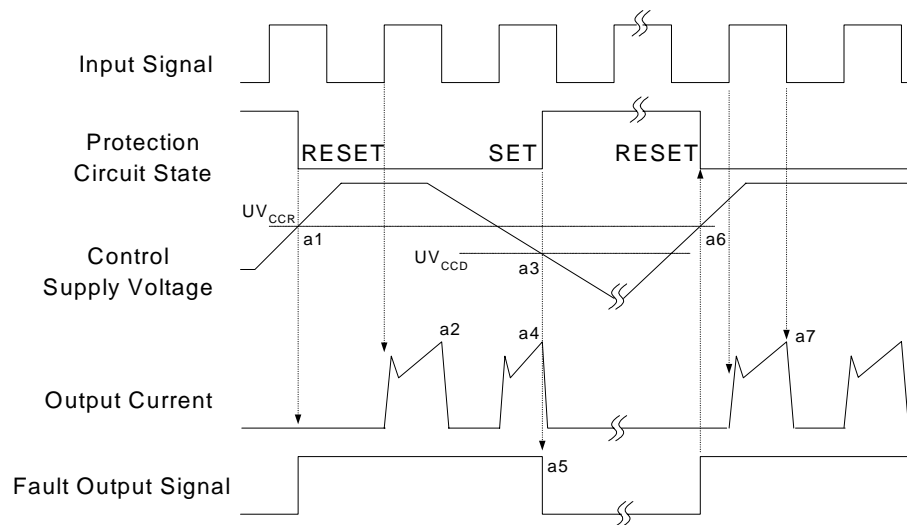


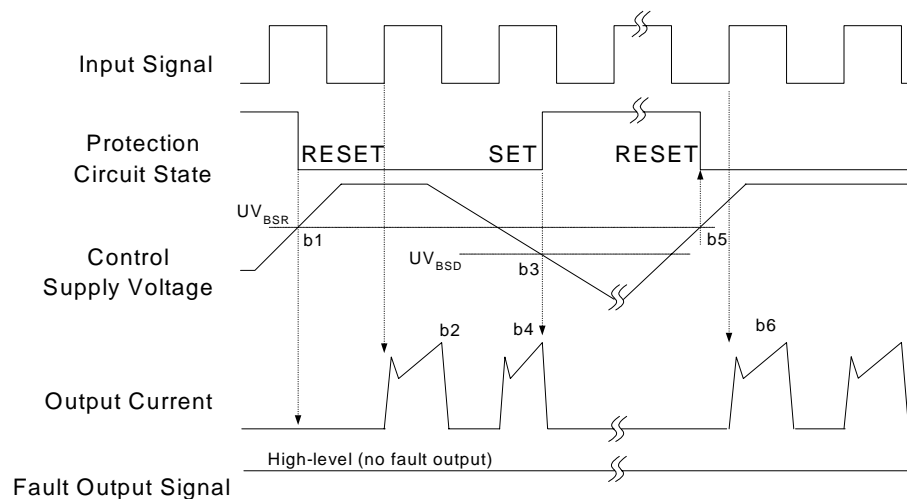
Figure 5. Flatness Measurement Position

Time Charts of Protective Function



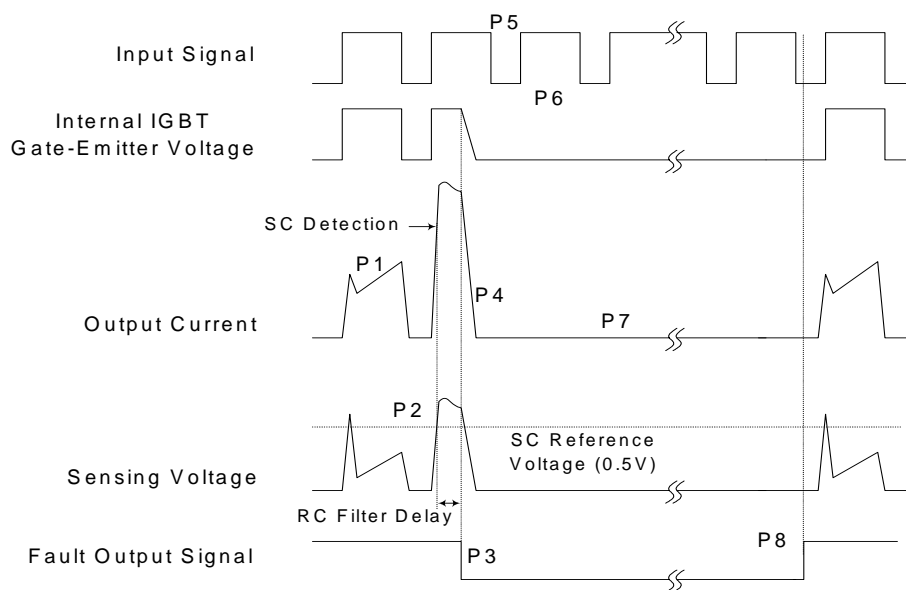
- a1 : Control supply voltage rises: After the voltage rises UV_{CCR} , the circuits start to operate when next input is applied.
- a2 : Normal operation: IGBT ON and carrying current.
- a3 : Under voltage detection (UV_{CCD}).
- a4 : IGBT OFF in spite of control input condition.
- a5 : Fault output operation starts.
- a6 : Under voltage reset (UV_{CCR}).
- a7 : Normal operation: IGBT ON and carrying current.

Fig. 6. Under-Voltage Protection (Low-side)



- b1 : Control supply voltage rises: After the voltage reaches UV_{BSR} , the circuits start to operate when next input is applied.
- b2 : Normal operation: IGBT ON and carrying current.
- b3 : Under voltage detection (UV_{BSD}).
- b4 : IGBT OFF in spite of control input condition, but there is no fault output signal.
- b5 : Under voltage reset (UV_{BSR})
- b6 : Normal operation: IGBT ON and carrying current

Fig. 7. Under-Voltage Protection (High-side)



(with the external shunt resistance and CR connection)

c1 : Normal operation: IGBT ON and carrying current.

c2 : Short circuit current detection (SC trigger).

c3 : Hard IGBT gate interrupt.

c4 : IGBT turns OFF.

c5 : Fault output timer operation starts: The pulse width of the fault output signal is set by the external capacitor C_{FO} .

c6 : Input "L" : IGBT OFF state.

c7 : Input "H": IGBT ON state, but during the active period of fault output the IGBT doesn't turn ON.

c8 : IGBT OFF state

Fig. 8. Short-Circuit Current Protection (Low-side Operation only)

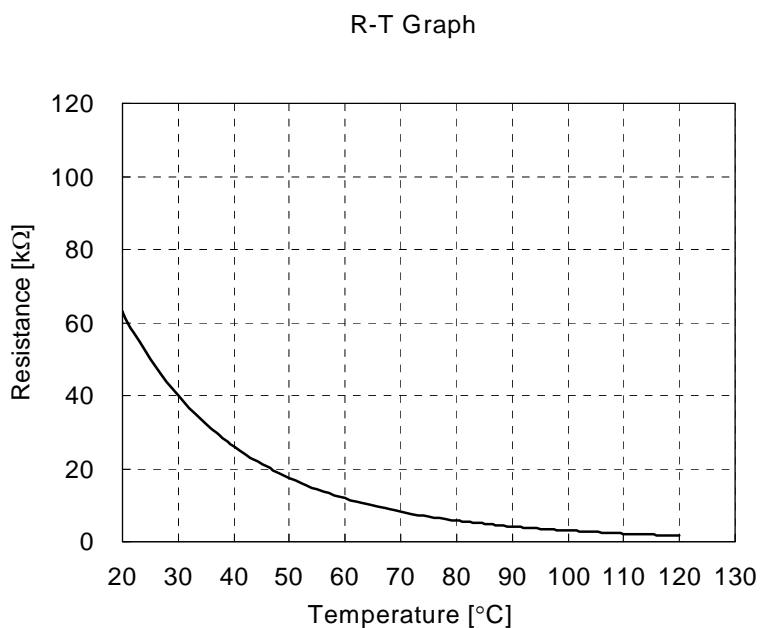
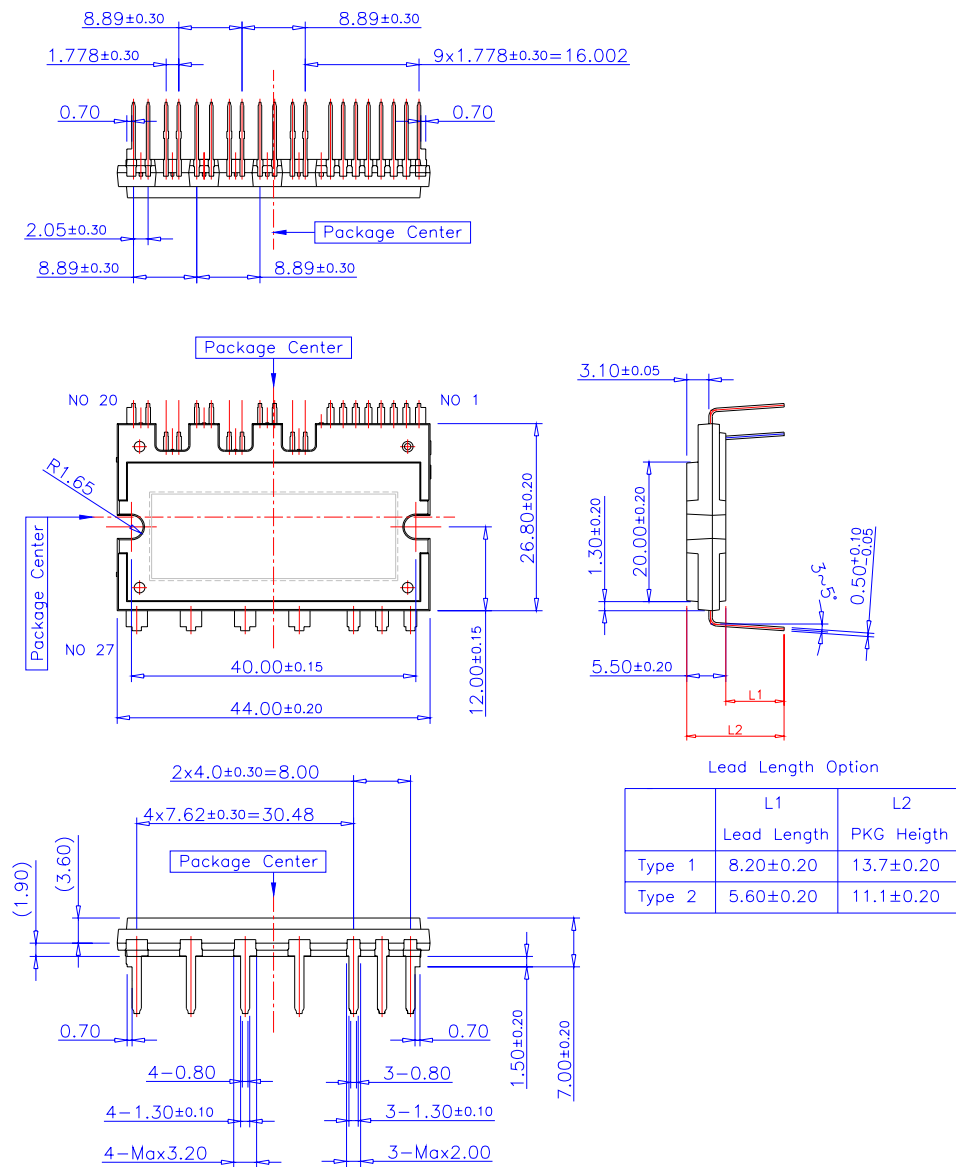


Fig. 9. R-T Curve of the Built-in Thermistor

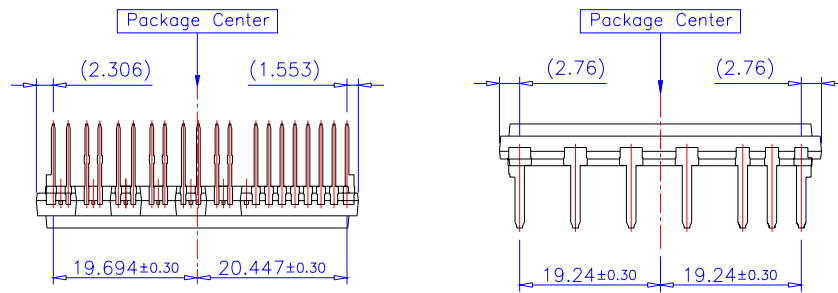
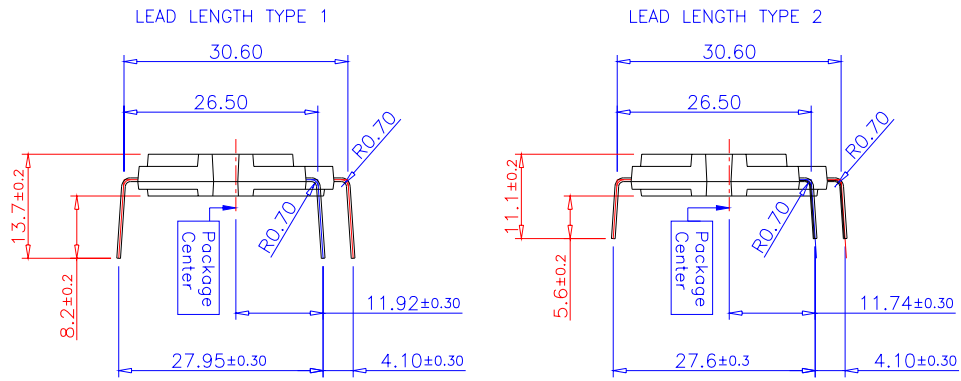
1. To avoid malfunction, the wiring of each input should be as short as possible. (less than 2-3cm)
2. By virtue of integrating an application specific type HVIC inside the Module, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.
3. V_{FO} output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7k Ω resistance. Please refer to Figure 10.
4. C_{SP15} of around 7 times larger than bootstrap capacitor C_{BS} is recommended.
5. V_{FO} output pulse width should be determined by connecting an external capacitor (C_{FOD}) between $C_{FOD}(pin7)$ and COM(pin2). (Example : if $C_{FOD} = 33$ nF, then $t_{FO} = 1.8$ ms (typ.)) Please refer to the note 6 for calculation method.
6. Input signal is High-Active type. There is a 3.3k Ω resistor inside the IC to pull down each input signal line to GND. When employing RC coupling circuits, set up such RC couple that input signal agree with turn-off/turn-on threshold voltage.
7. To prevent errors of the protection function, the wiring around R_{SC} , R_F and C_{SC} should be as short as possible.
8. In the short-circuit protection circuit, please select the $R_F C_{SC}$ time constant in the range 3~4 μ s.
9. Each capacitor should be mounted as close to the pins as possible.
10. To prevent surge destruction, the wiring between the smoothing capacitor and the P&N pins should be as short as possible. The use of a high frequency non-inductive capacitor of around 0.1~0.22 μ F between P and N pins is recommended.
11. Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays.
12. C_{SPC15} should be over 1 μ F and mounted as close to the pins of the module as possible.
13. $N_{B1}(pin23)$ and $N_{B2}(pin21)$ should be shorted externally.

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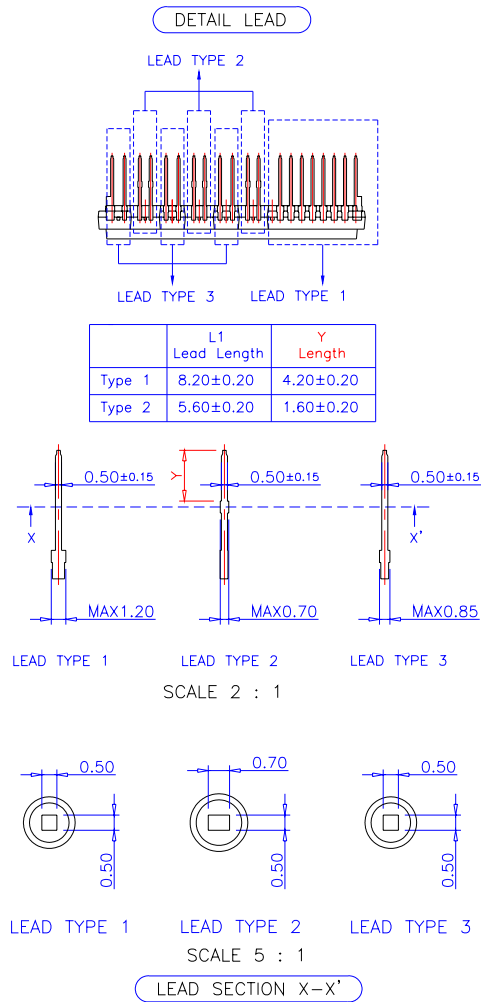
Detailed Package Outline Drawings



Detailed Package Outline Drawings (Continued)



Detailed Package Outline Drawings (Continued)



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