FAIRCHILD
SEMICONDUCTOR®

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FAN5346
Series Boost LED Driver with PWM Dimming Interface

Features

- Asynchronous Boost Converter
- Drives LEDs in Series:
  - FAN5346S20X: 20V Output
  - FAN5346S30X: 30V Output
- 2.5V to 5.5V Input Voltage Range
- PWM Dimming for LED Brightness Control
- 5kHz to 100kHz PWM Dimming Frequency Range
- 1.2MHz Fixed Switching Frequency
- Soft-Start Capability
- Input Under-Voltage Lockout (UVLO)
- Output Over-Voltage Protection (OVP)
- Short-Circuit Detection
- Thermal Shutdown (TSD) Protection
- Small Form-Factor 6-Lead SSOT23 Package

Applications

- Cellular Mobile Handsets
- Mobile Internet Devices
- Portable Media Players
- PDA, DSC, MP3 Players

Description

The FAN5346 is an asynchronous constant-current LED driver that drives LEDs in series to ensure equal brightness for all the LEDs. FAN5346S20X has an output voltage of 20V and can drive up to 5 LEDs in series. FAN5346S30X has an output voltage of 30V and up to 8 LEDs in series. Optimized for small form-factor applications, the 1.2MHz fixed switching frequency allows the use of small inductors and capacitors.

The FAN5346 uses a PWM dimming control interface to set the brightness levels of the LEDs. A PWM signal of 5kHz to 100kHz is applied to the EN pin.

For safety, the device features integrated over-voltage, over-current, short-circuit detection, and thermal-shutdown protections. In addition, input under-voltage lockout protection is triggered if the battery voltage is too low.

The FAN5346 is available in a 6-lead SSOT23 package. It is "green" and RoHS compliant. *(Please see http://www.fairchildsemi.com/company/green/index.html for Fairchild's definition of green).*

Ordering Information

Part Number	Output Voltage Option	Temperature Range	Package
FAN5346S20X	20V	-40 to 85°C	6-Lead, SuperSOT™-6, JEDEC MO-193, 1.6mm Wide (MA06A)
FAN5346S30X	30V		

.

Typical Application Diagram

L = 10µH

2.5V to 5.5V

VIN

C_{IN} = 10µF

GND

EN

FAN5346

1

2

4

6

5

3

SW

D

VOUT

C_{OUT} = 1µF

FB

3 – 8 LEDs

R_{SET} = 12.4Ω @ I_{LED} = 20mA

R_{SET} = 10.0Ω @ I_{LED} = 25mA
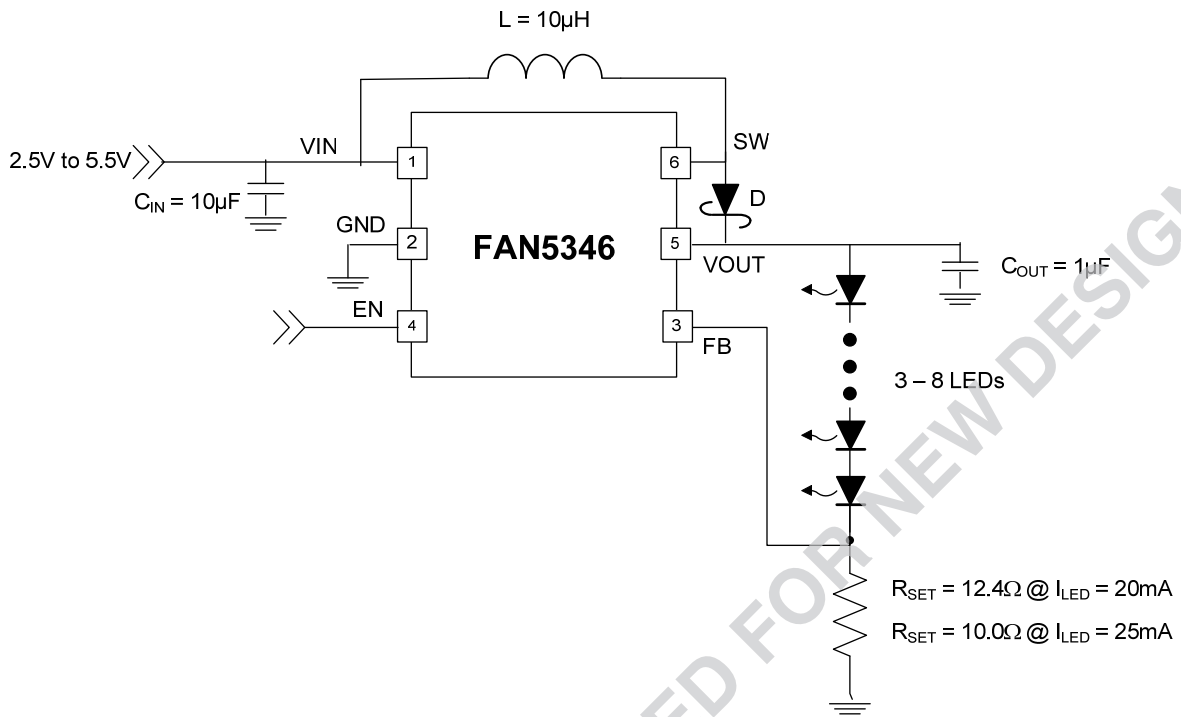
Figure 1. Typical Application

Block Diagram

VIN

1

GND

6

EN

4

GATE DRIVE

R_{EN}

BOOST CONTROL

OVP/ SCD

PWM I/F

6

SW

5

VOUT

3

FB
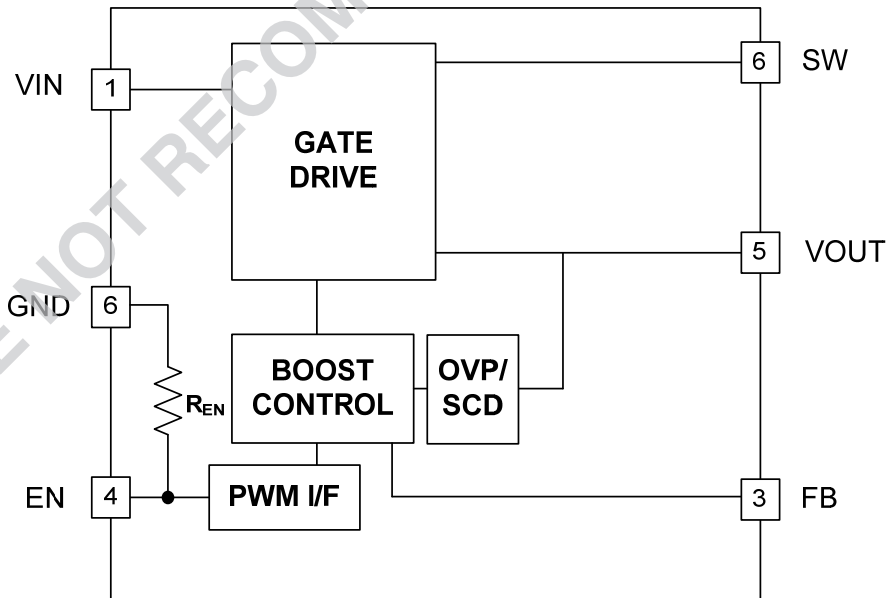
Figure 2. Functional Block Diagram

Pin Configuration
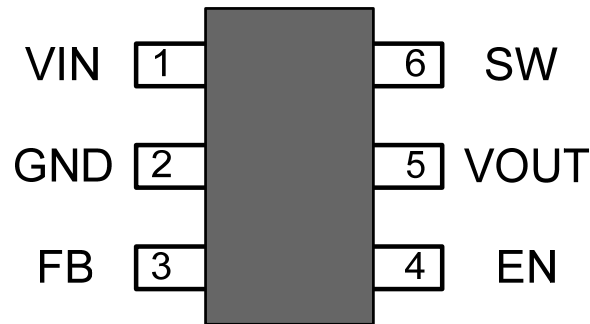
Figure 3. Pin Assignments, Top View

Pin Definitions

Pin #	Name	Description
5	VOUT	**Boost Output Voltage**. Output of the boost regulator. Connect the LEDs to this pin. Connect C_{OUT} (output capacitor) to GND.
1	VIN	**Input Voltage**. Connect to the power source and decouple with C_{IN} to GND.
4	EN	**Enable Brightness Control**. Program dimming levels by driving pin with the PWM signal.
3	FB	**Voltage Feedback**. The boost regulator regulates this pin to 0.250V to control the LED string current. Tie this pin to a current setting resistor (R_{SET}) between GND and the cathode of the LED string.
6	SW	**Switching node**. Tie inductor L1 from VIN to SW pin.
2	GND	**Ground**. Tie directly to a GND plane.

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter		Min.	Max.	Unit
V_{IN}	VIN Pin		-0.3	6.0	V
V_{FB}, V_{EN}	FB, EN Pins		-0.3	V_{IN} + 0.3	V
V_{SW}	SW Pin	FAN5346S20X	-0.3	22.0	V
		FAN5346S30X	-0.3	33.0	V
V_{OUT}	VOUT Pin	FAN5346S20X	-0.3	22.0	V
		FAN5346S30X	-0.3	33.0	V
ESD	Electrostatic Discharge Protection	Human Body Model per JESD22-A114		1.5	kV
		Charged Device Model per JESD22-C101		1.5	
T_J	Junction Temperature		-40	+150	°C
T_{STG}	Storage Temperature		-65	+150	°C
T_L	Lead Soldering Temperature, 10 Seconds			+260	°C

Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to absolute maximum ratings.

Symbol	Parameter		Min.	Max.	Unit
V_{IN}	V_{IN} Supply Voltage		2.5	5.5	V
V_{OUT}	V_{OUT} Voltage[1]	FAN5346S20X	6.2	18.5	V
		FAN5346S30X	6.2	28.5	
I_{OUT}	V_{OUT} Load Current		5	25	mA
T_A	Ambient Temperature		-40	+85	°C
T_J	Junction Temperature		-40	+125	°C

Note:
1. Application should guarantee that minimum and maximum duty cycle fall between 20-85% to meet the specified range.

Thermal Properties

Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with four-layer 2s2p boards in accordance to JEDEC standard JESD51. Special attention must be paid not to exceed junction temperature $T_{J(max)}$ at a given ambient temperature T_A.

Symbol	Parameter	Typical	Unit
θ_{JA}	Junction-to-Ambient Thermal Resistance, SSOT23-6 Package	151	°C/W

Electrical Specifications

V_{IN} = 2.5V to 5.5V and T_A = -40°C to +85°C unless otherwise noted. Typical values are at T_A = +25°C and V_{IN} = 3.6V.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
Power Supplies						
I_{SD}	Shutdown Supply Current	EN = GND		0.30	0.90	µA
$I_{Q(ACTIVE)}$	Quiescent Current at I_{LOAD} = 0mA	Device Not Switching, No Load		300		µA
V_{UVLO}	Under-Voltage Lockout Threshold	V_{IN} Rising	2.10	2.35	2.60	V
		V_{IN} Falling	1.80	2.05	2.30	
V_{UVHYST}	Under-Voltage Lockout Hysteresis			250		mV
EN: Enable Pin						
V_{IH}	HIGH-Level Input Voltage		1.2			V
V_{IL}	LOW-Level Input Voltage				0.4	V
R_{EN}	EN Pull-Down Resistance		200	300	400	kΩ
f_{PWM}	PWM Dimming Frequency[3]		5		100	kHz
t_{SD}	EN LOW, Shutdown Pulse Width	V_{IN} = 3.6V; from Falling Edge of EN			1	ms
Feedback and Reference						
V_{FB}	Feedback Voltage	I_{LED} = 20mA from -40°C to +85°C, 2.5V ≤ V_{IN} ≤ 5.5V	230	250	270	mV
I_{FB}	Feedback Input Current	V_{FB} = 250mV		0.1	1.0	µA
Power Outputs						
$R_{DS(ON)\_Q1}$	Boost Switch On Resistance	V_{IN} = 3.6V, I_{SW} = 100mA		600		mΩ
		V_{IN} = 2.5V, I_{SW} = 100mA		650		
$I_{SW(OFF)}$	SW Node Leakage[2]	EN = 0, V_{IN} = V_{SW} = V_{OUT} = 5.5V, V_{LED} = 0V		0.1	2.0	µA
I_{LIM-PK}	Boost Switch Peak Current Limit	FAN5346S20X: V_{IN} = 3.2V to 4.3V, T_A = -20°C to +60°C, V_F = 3.4V, 4 LEDs	200	300	400	mA
		FAN5346S30X	500	750	1000	
Oscillator						
f_{SW}	Boost Regulator Switching Frequency		0.95	1.15	1.35	MHz
Output and Protection						
V_{OVP}	Boost Output Over-Voltage Protection	FAN5346S20X	18.0	20.0	21.5	V
		FAN5346S30X	27.5	30.0	32.5	
	OVP Hysteresis	FAN5346S20X		0.8		
		FAN5346S30X		1.0		
V_{TLSC}	V_{OUT} Short-Circuit Detection Threshold	V_{OUT} Falling		V_{IN} – 1.4		V
V_{THSC}	V_{OUT} Short-Circuit Detection Threshold	V_{OUT} Rising		V_{IN} – 1.2		V
D_{MAX}	Maximum Boost Duty Cycle[3,4]		85			%
D_{MIN}	Minimum Boost Duty Cycle[3,4]				20	
T_{TSD}	Thermal Shutdown			150		°C
T_{HYS}	Thermal Shutdown Hysteresis			35		°C

Notes:
2. SW leakage current includes the leakage current of two internal switches; SW to GND and SW to V_{OUT}.
3. Not tested in production; guaranteed by design.
4. Application should guarantee that minimum and maximum duty cycle fall between 20-85% to meet the specified range.

Typical Characteristics

V_{IN} = 3.6V, T_A = 25°C, I_{LED} = 25mA, L = 10µH, C_{OUT} = 1.0µF, and C_{IN} = 10.0µF.
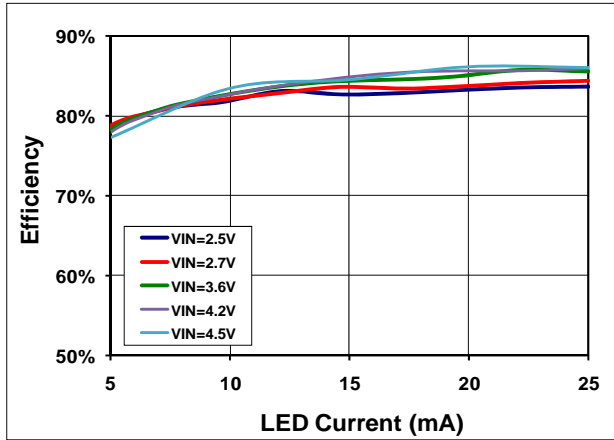

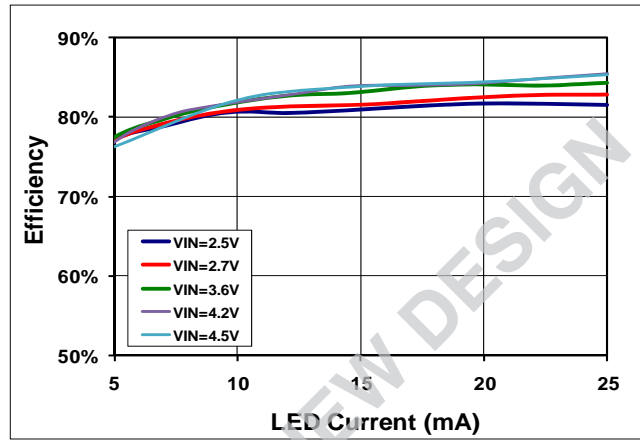Figure 4. 3 LEDs: Efficiency vs. LED Current vs. Input Voltage

Figure 5. 4 LEDs: Efficiency vs. LED Current vs. Input Voltage
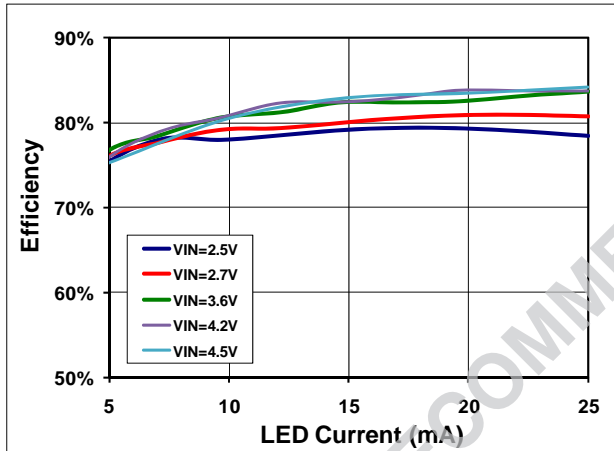

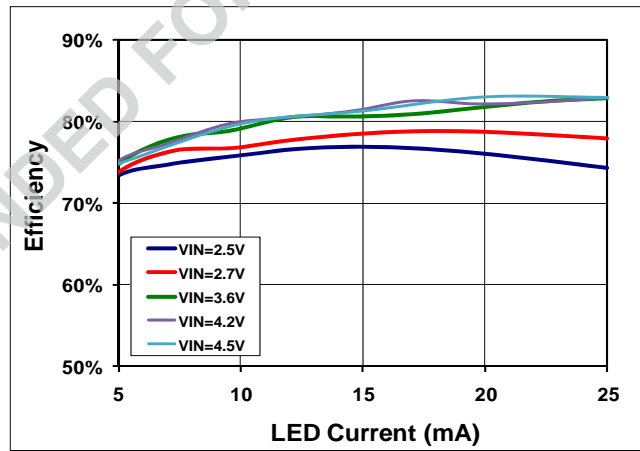Figure 6. 5 LEDs: Efficiency vs. LED Current vs. Input Voltage

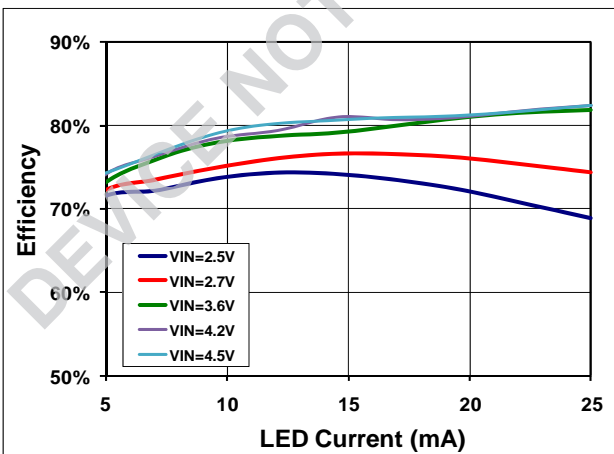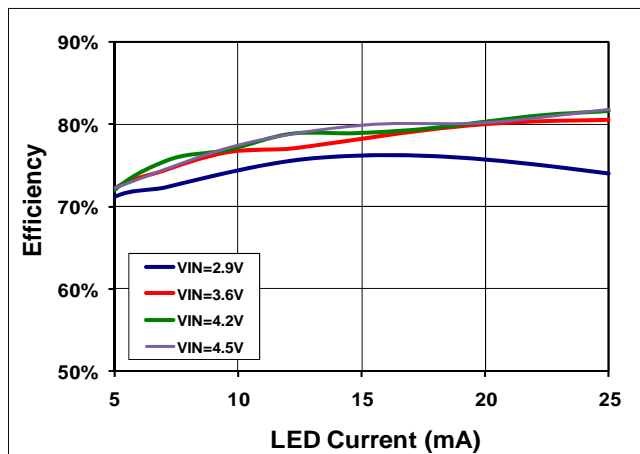Figure 7. 6 LEDs: Efficiency vs. LED Current vs. Input Voltage

Figure 8. 7 LEDs: Efficiency vs. LED Current vs. Input Voltage

Figure 9. 8 LEDs: Efficiency vs. LED Current vs. Input Voltage

Typical Characteristics

V_{IN} = 3.6V, T_A = 25°C, I_{LED} = 25mA, L = 10µH, C_{OUT} = 1.0µF, and C_{IN} = 10.0µF.
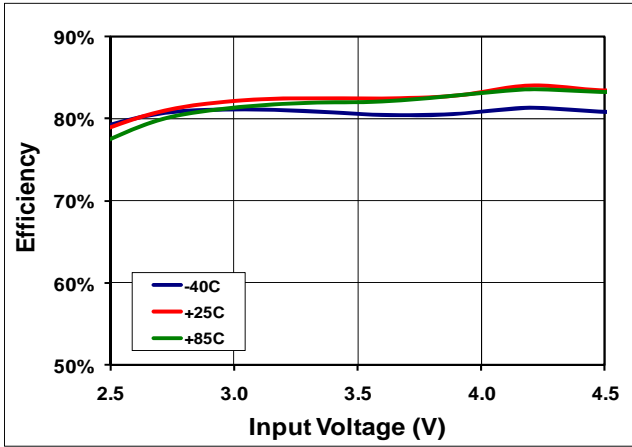
Figure 10. Efficiency vs. Input Voltage vs. Temperature for 5 LEDs in Series
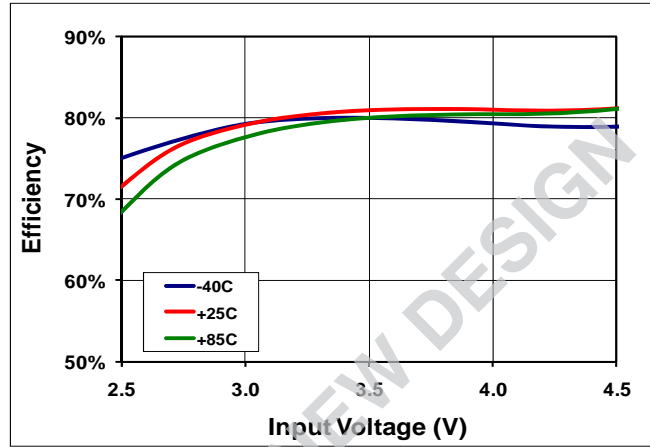
Figure 11. Efficiency vs. Input Voltage vs. Temperature for 7 LEDs in Series
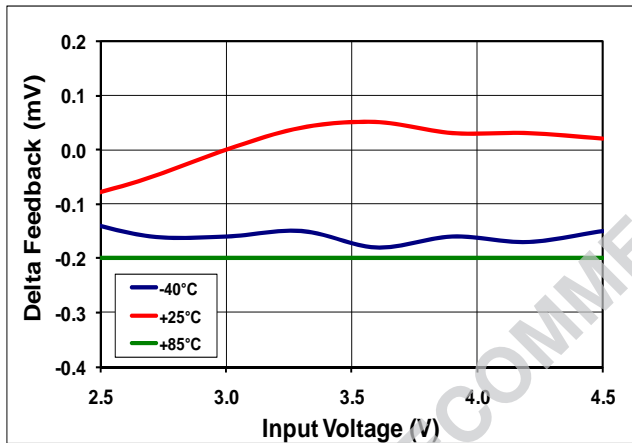
Figure 12. Delta of V_{FB} Over Input Voltage and Temperature for 7 LEDs with L=10µH and C_{OUT}=1.0µF
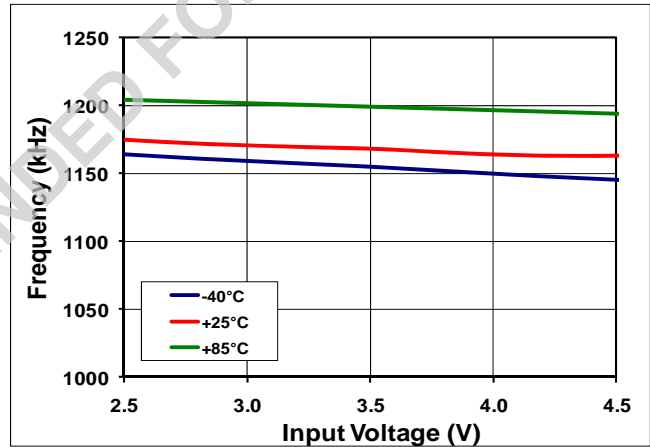
Figure 13. Frequency vs. Input Voltage vs. Temperature
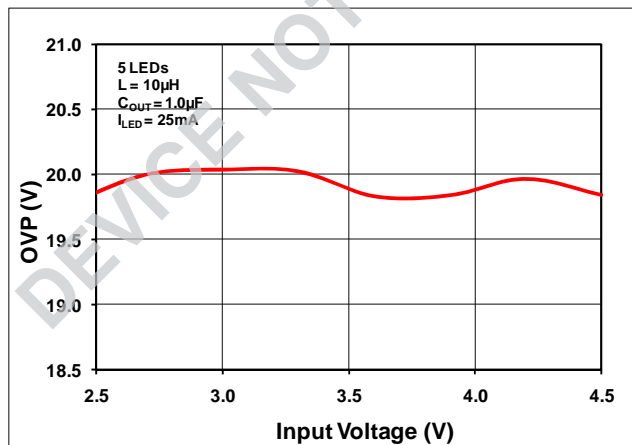
Figure 14. OVP vs. Input Voltage: FAN5346S20X
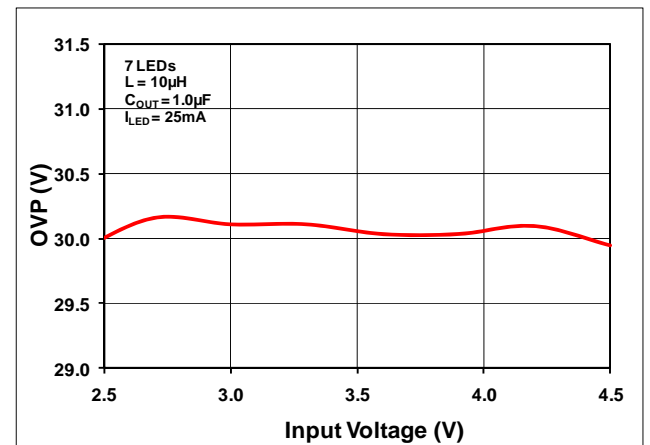
Figure 15. OVP vs. Input Voltage: FAN5346S30X

Typical Characteristics
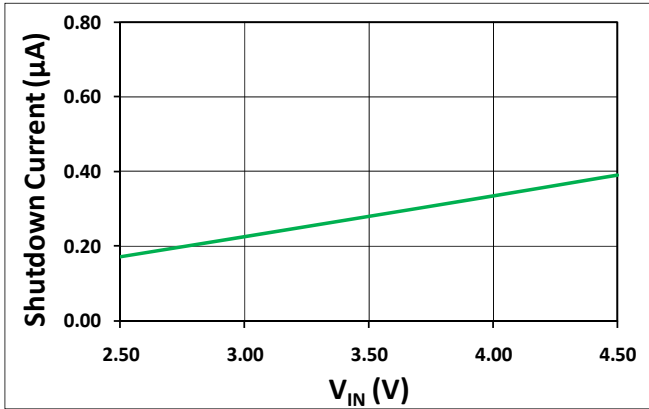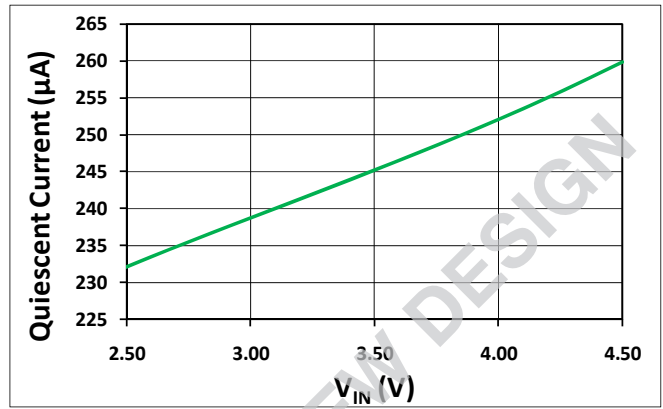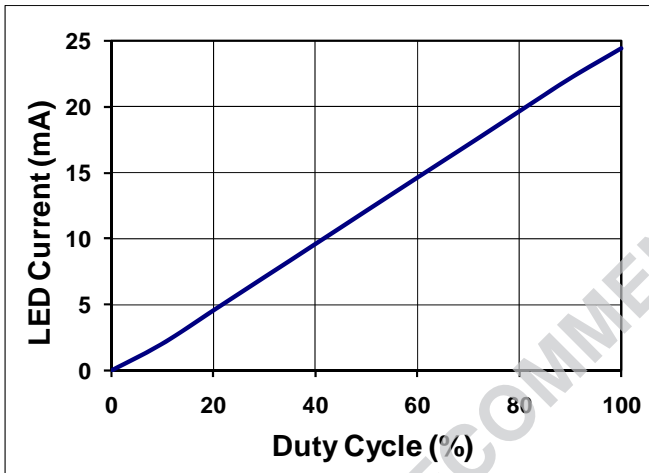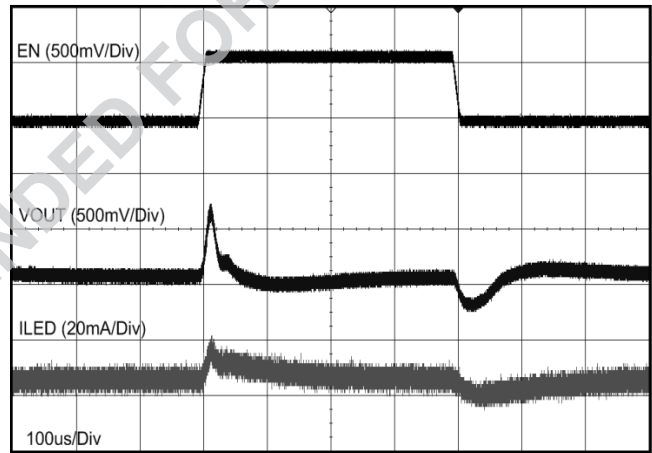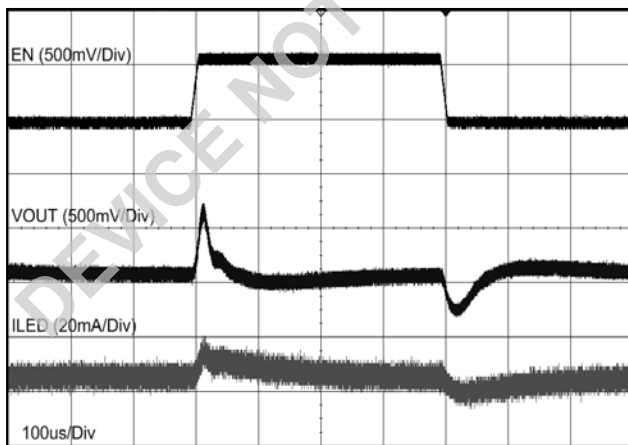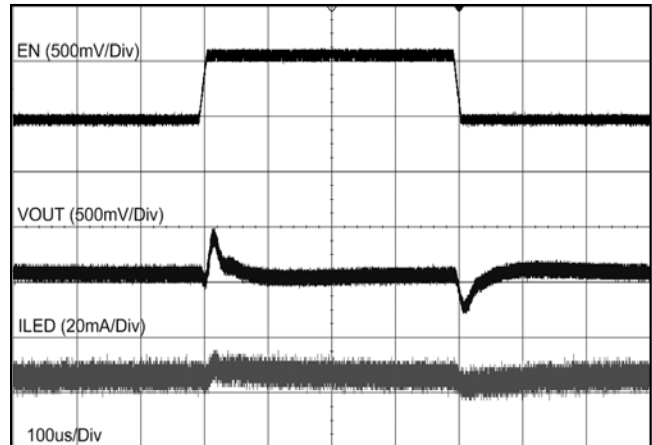
V_{IN} = 3.6V, T_A = 25°C, I_{LED} = 25mA, L = 10µH, C_{OUT} = 1.0µF, and C_{IN} = 10.0µF.
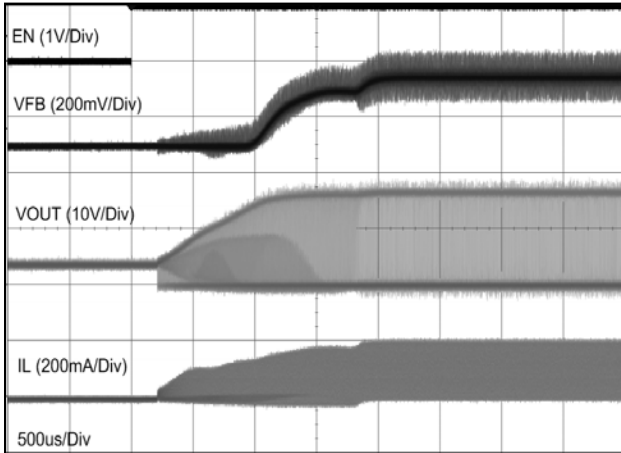

Figure 22. Startup Waveform for Switch Voltage, Inductor Current, V_{FB}, and EN for 5 LEDs
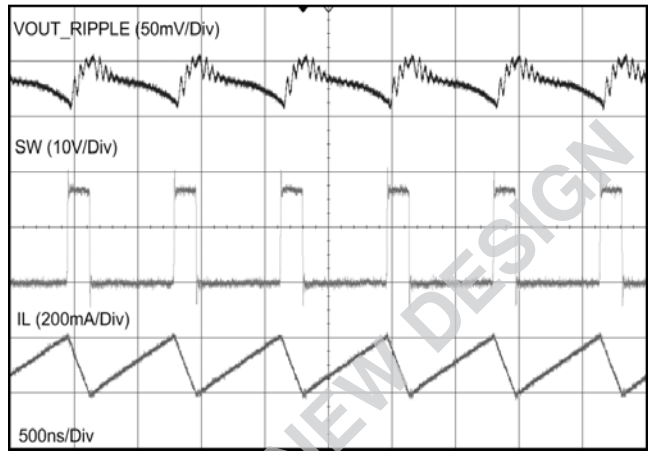
Figure 23. Steady-State Waveform for V_{OUT}, Switch Voltage, and Inductor Current for 5 LEDs
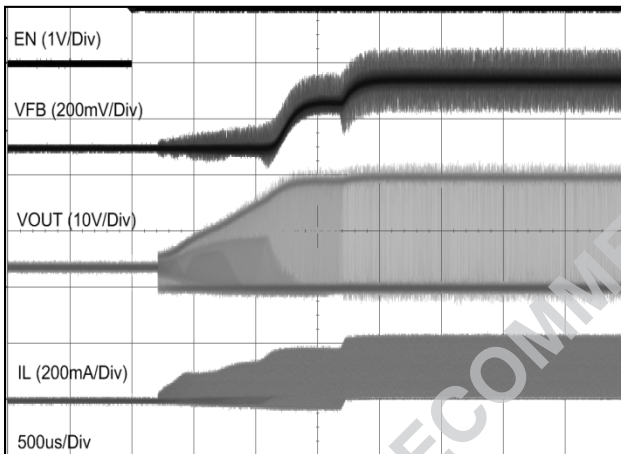
Figure 24. Startup Waveform for Switch Voltage, Inductor Current, V_{FB}, and EN for 6 LEDs
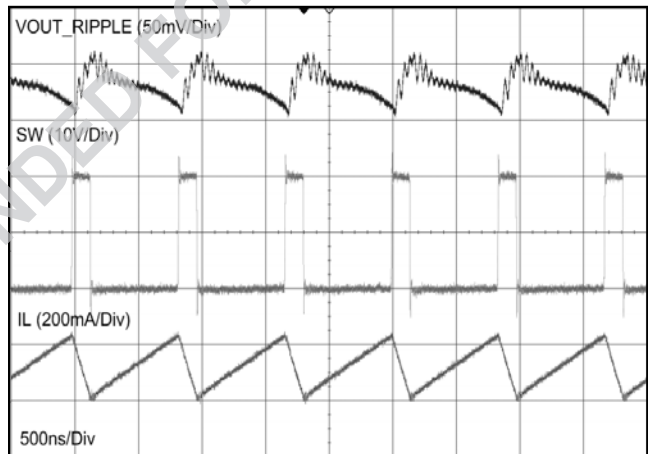
Figure 25. Steady-State Waveform for V_{OUT}, Switch Voltage, and Inductor Current for 6 LEDs
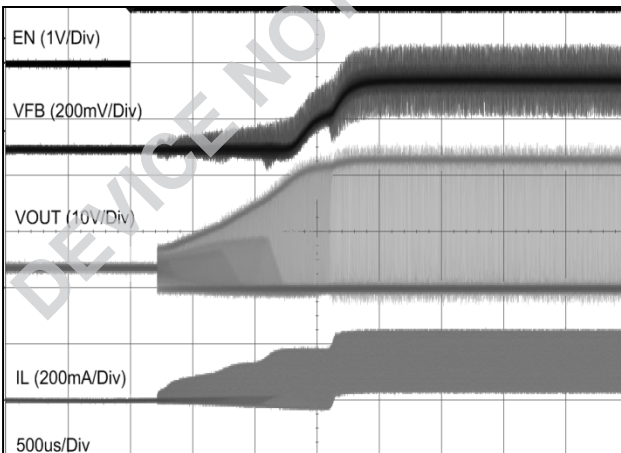
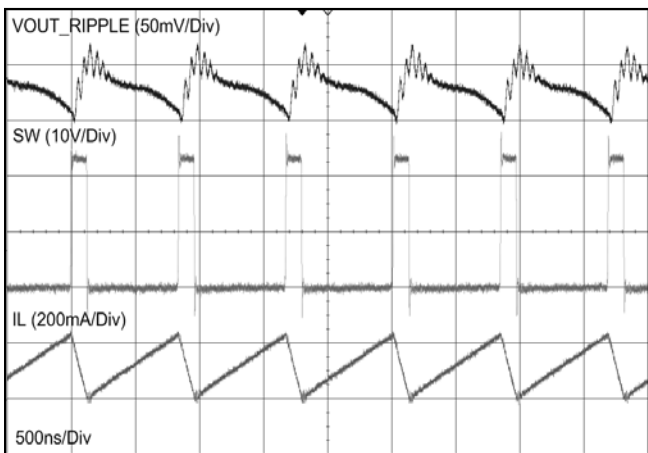**Figure 26. Startup Waveform for Switch Voltage, Inductor Current, V_{FB}, and EN for 7 LEDs**

Figure 27. Steady-State Waveform for V_{OUT}, Switch Voltage, and Inductor Current for 7 LEDs

Circuit Description

Overview

The FAN5346 is an inductive current-mode boost serial LED driver that achieves LED current regulation by maintaining 0.250V across the R_{SET} resistor. The current through the LED string (I_{LED}) is given by:

$$I_{LED} = \frac{0.250}{R_{SET}} \qquad (1)$$

The voltage V_{OUT} is determined by the sum of the forward voltages across each LED, plus the voltage across R_{SET}, which is always 250mV.

Driving Eight LEDs in Series

FAN5346S30X can drive 8 LEDs in series, but the minimum input voltage (V_{IN}) must be greater than or equal to 2.9V, while the forward voltage of the white LED should be less than or equal to 3.2V, and the maximum LED current cannot exceed 20mA to maintain stable operation.

UVLO and Soft-Start

If EN has been LOW for more than 1ms, the IC may initiate a "cold start" soft-start cycle when EN rises, provided V_{IN} is above the UVLO threshold.

PWM Dimming

The FAN5346 uses a PWM signal to directly modulate output current in the LED string to vary the perceived LED brightness. When the EN pin is held HIGH, the FB voltage is 250mV. This voltage is reduced when a PWM signal is applied to the EN pin, thereby enabling the LEDs to be dimmed. The FB voltage is given by the equation:

$$V_{FB} = DutyCycle \times 250mV \qquad (2)$$

where DutyCycle = the duty cycle of the PWM signal and 250mV is the internal reference voltage.

Figure 28 illustrates how the FAN5346 divides the internal 250mV reference voltage at the duty cycle of the PWM signal. A low-pass filter filters the PWM signal, which then is input into the error amplifier as the reference voltage for the FB pin.
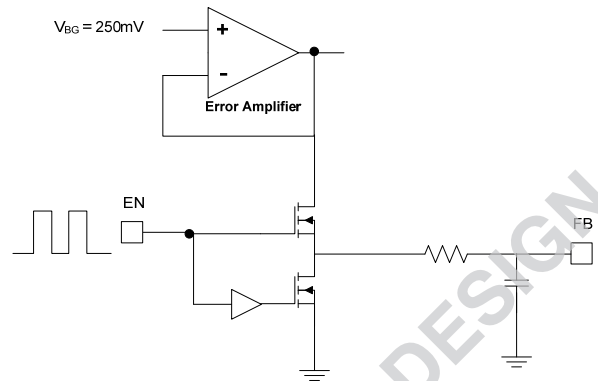

Figure 28. Block Diagram of FB and EN Circuit for PWM Dimming

Over-Current and Short-Circuit Detection

The boost regulator employs a cycle-by-cycle peak inductor current limit of 300mA (typical) and 750mA (typical) for FAN5346S20X and FAN5346S30X, respectively.

Over-Voltage / Open-Circuit Protection

If the LED string is an open circuit, FB remains at 0V and the output voltage continues to increase in the absence of an over-voltage protection (OVP) circuit. The FAN5346S20X OVP circuit disables the boost regulator when V_{OUT} exceeds 20V and continues to keep the regulator off until V_{OUT} drops below 19V. For FAN5346S30X, the OVP is 30V and it turns back on when V_{OUT} is below 29V

Thermal Shutdown

When the die temperature exceeds 150°C, a reset occurs and remains in effect until the die cools to 115°C; at which time, the circuit is allowed to begin the soft-start sequence.

Application Information

The reference schematic diagram is shown in Figure 29. FAN5346 is able to drive up to eight LEDs with input voltage equal to or greater than 2.9V ($V_{IN} \geq 2.9V$). However, the number of LEDs that can be used FAN5346 depends on forward voltage. It is recommended that the forward voltage (V_F) of the white LED be no greater than 3.2V and the maximum LED current be 20mA. FAN5345 can be also used as a boost convertor by connect the V_{OUT} point to the load directly. The return trace of the load should also return to GND through a sense resistor (R1).
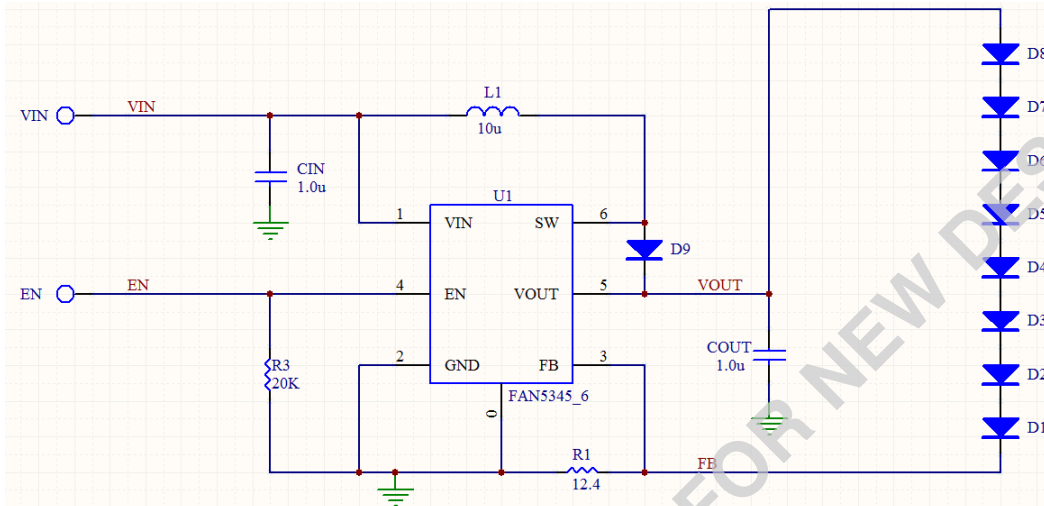

Figure 29. Reference Application Schematic Diagram

Component Placement and PCB Recommendations

FAN5346 switches at 1.2MHz to boost the output voltage. Component placement and PCB layout need to be carefully taken into consideration to ensure stable output and to prevent generation of noise. Figure 30 is a portion of the evaluation board layout. The critical layout elements are: the L1, C_{IN}, C_{IN} return trace, C_{OUT}, and the C_{OUT} return trace.

Figure 30. Reference PCB Layout

Input Capacitor and Return Trace

The input capacitor is the first priority in a switching buck or boost regulator PCB layout. A stable input source (V_{IN}) enables a switching regulator to deliver its best performance. During the regulator's operation, it is switching at a high frequency, which makes the load of C_{IN} change dynamically since it is trying to make the input source vary at the same switching frequency as the regulator. To ensure a stable input source, C_{IN} needs to hold enough energy to minimize the variation at the input pin of the regulator. For C_{IN} to have a fast response of charge / discharge, the trace from C_{IN} to the input pin of the regulator and the return trace from GND of the regulator to C_{IN} should be as short and wide as possible to minimize trace resistance, inductance, and capacitance. During operation, the current flow from C_{IN} through the regulator to the load and back to C_{IN} contains high-frequency variation due to switching. Trace resistance reduces the overall efficiency due to I^2R loss. Even a small trace inductance could effectively yield ground variation to add noise on VOUT. The input capacitor should be placed close to the VIN and GND pins of the regulator and traces should be as short as possible. Avoid routing the return trace through different layers because vias have strong inductance effect at high frequencies. If routing to other PCB layers is unavoidable, place vias next to the VIN and GND pins of the regulator to minimize the trace distance.

Output Capacitor and Return Trace

The output capacitor serves the same purpose as the input capacitor, but also maintains a stable output voltage. As explained above, the current travels to the load and back to the C_{OUT} GND terminal. C_{OUT} should be placed close to the VOUT pin. The traces of C_{OUT} to L1, VOUT, and the return
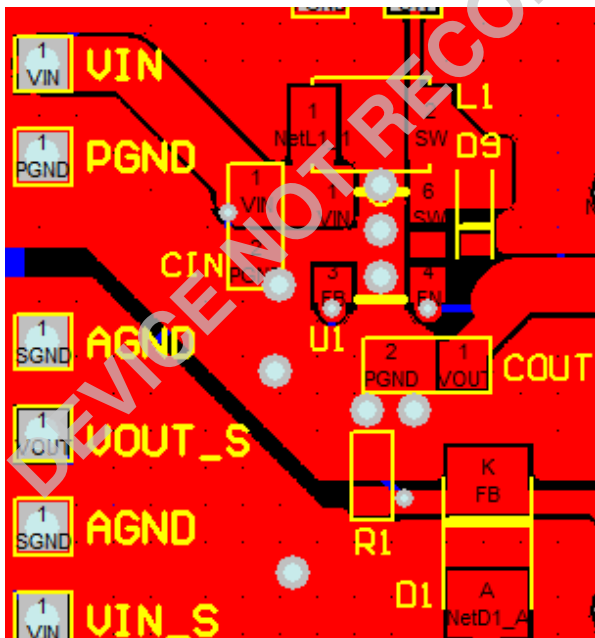
trace from load to C_{OUT} should be as short and wide as possible to minimize trace resistance and inductance. To minimize noise coupling to load, a small-value capacitor can be placed between VOUT and C_{OUT} to route high-frequency noise back to GND before it gets to the load.

Inductor

Inductor (L1) should be placed as close to the regulator as possible to minimize trace resistance and inductance for the reasons explained above.

Sense Resistor

The sense resistor provides a feedback signal for the regulator to control output voltage. A long trace from the sense resistor to the FB pin couples noise into the FB pin. If noise is coupled into the FB pin, it causes unstable operation of the switching regulator, which affects application performance. The return trace from the sense resistor to the FB pin should be short and away from any fast-switching signal traces. The ground plane under the return trace is not necessary. If the ground plane under the return trace is noisy; but not the same ground plane as the regulator; the noise could be coupled into the FB pin through PCB parasitic capacitance, yielding noisy output.

As shown in Figure 30; C_{IN}, C_{OUT}, and L1 are all placed next to the regulator. All traces are on the same layer to minimize trace resistance and inductance. Total PCB area, not including the sense resistor, is $67.2 mm^2$ (7.47mm x 8.99mm).

Table 1. Recommended External Components

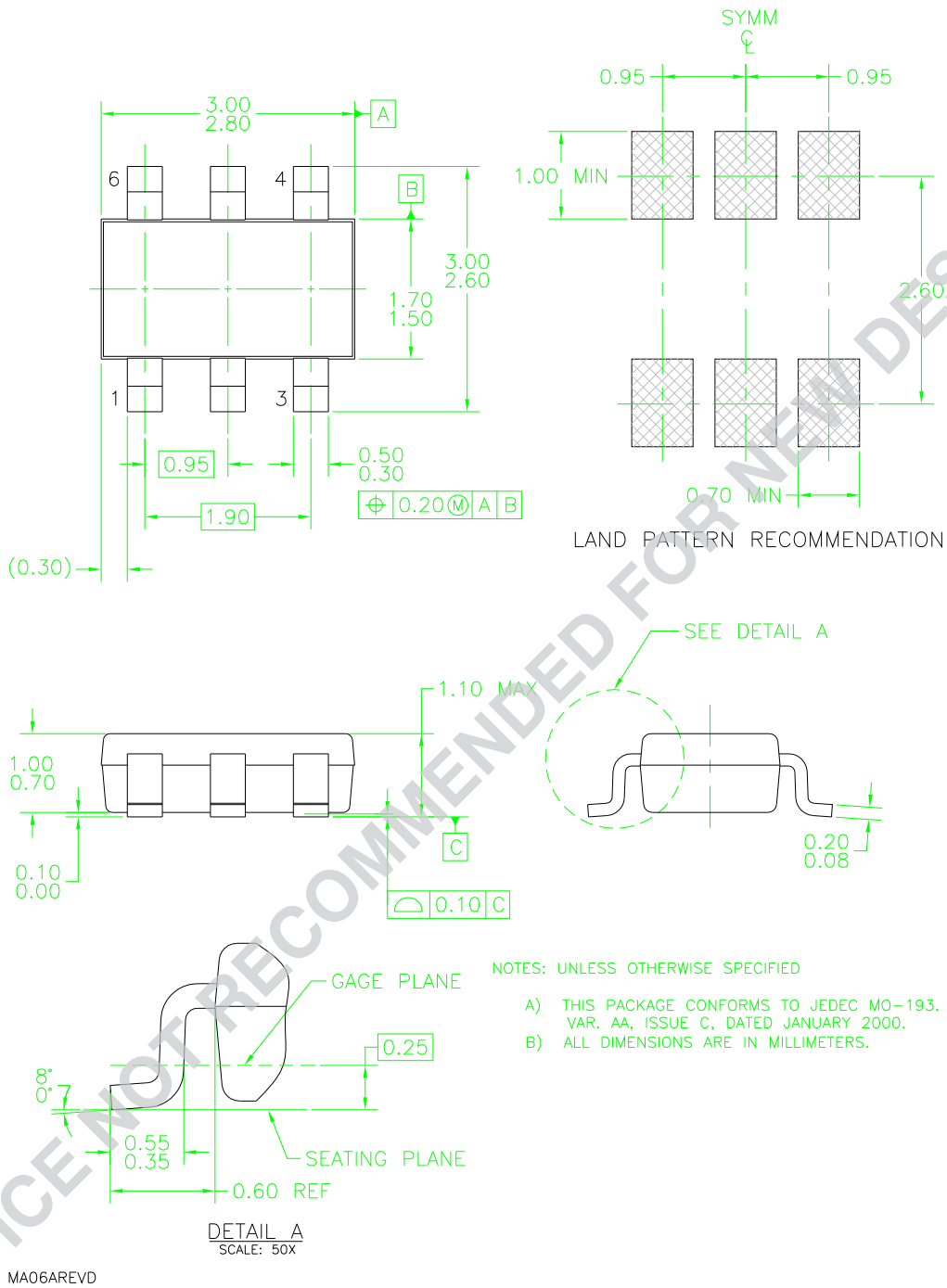| Inductor (L)	Part Number	Manufacturer
10.0µH	LQH43MN100K03	Murata
	NLCV32T-100K-PFR	TDK
	VLF3010AT-100MR49-1	TDK
	DEM2810C 1224-AS-H-100M	TOKO
Minimum C_{OUT}		
1.0µF	CV105X5R105K25AT	AVX / Kyocera
Minimum C_{IN}		
10.0µF	GRM21BR71A106KE51L	Murata
Schottky Diode		
N/A	RBS520S30	Fairchild Semiconductor
N/A	RB520S-30	Rohm

Physical Dimensions

Figure 31. 6-Lead, SuperSOT™-6, JEDEC MO-193, 1.6mm Wide

Package drawings are provided as a service to customers considering Fairchild components. Drawings may change in any manner without notice. Please note the revision and/or date on the drawing and contact a Fairchild Semiconductor representative to verify or obtain the most recent revision. Package specifications do not expand the terms of Fairchild's worldwide terms and conditions, specifically the warranty therein, which covers Fairchild products.

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http://www.fairchildsemi.com/packaging/.

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AX-CAP™*	Global Power Resource℠	Programmable Active Droop™	the power franchise*
BitSiC™	GreenBridge™	QFET®	
Build it Now™	Green FPS™	QS™	TinyBoost™
CorePLUS™	Green FPS™ e-Series™	Quiet Series™	TinyBuck™
CorePOWER™	Gmax™	RapidConfigure™	TinyCalc™
CROSSVOLT™	GTO™	○™	TinyLogic®
CTL™	IntelliMAX™		TINYOPTO™
Current Transfer Logic™	ISOPLANAR™	Saving our world, 1mW/W/kW at a time™	TinyPower™
DEUXPEED®	Making Small Speakers Sound Louder	SignalWise™	TinyPWM™
Dual Cool™	and Better™	SmartMax™	TinyWire™
EcoSPARK®	MegaBuck™	SMART START™	TranSiC™
EfficientMax™	MICROCOUPLER™	Solutions for Your Success™	TriFault Detect™
ESBC™	MicroFET™	SPM®	TRUECURRENT®*
	MicroPak™	STEALTH™	µSerDes™
F®	MicroPak2™	SuperFET™	
Fairchild®	MillerDrive™	SuperSOT™-3	µSerDes™
Fairchild Semiconductor®	MotionMax™	SuperSOT™-6	UHC®
FACT Quiet Series™	mWSaver™	SuperSOT™-8	Ultra FRFET™
FACT®	OptoHiT™	SupreMOS®	UniFET™
FAST®	OPTOLOGIC®	SyncFET™	VCX™
FastvCore™	OPTOPLANAR®	Sync-Lock™	VisualMax™
FETBench™		SYSTEM GENERAL®*	VoltagePlus™
FlashWriter®*			XS™
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Definition of Terms

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Preliminary	First Production	Datasheet contains preliminary data; supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve design.
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