

# FAN4603 500mA, Fully Integrated, Buck Power Supply Module

### Features

- Solder and Play DC/DC Converter; No External Components Required
- Up to 91% Efficiency
- 500mA Output Current Capability
- 2.3V to 5.5V Input Voltage Range
- Fixed Output Voltages from 1.2V to 1.8V
- 35µA PFM Quiescent Current
- Best-in-Class Load and Line Transient Response
- ±4% PWM DC Voltage Accuracy
- No External Components Required
- High-Efficiency, Low-Ripple, Light-Load PFM
- Thermal Shutdown (TSD), Under-Voltage Lockout, (UVLO), and Short-Circuit (SCP) Protection
- 4.0 x 2.5mm MLP Package
- Maximum Height: 1.1mm

### Applications

- POL and Distributed DC-DC Module Applications
- Small Form Factor, Battery-Powered Applications
- POL Core Power for FPGA, DSP, CPU, and GPU with Fast-Transient, Wide Dynamic Load Requirements
- Wireless Cards, Meters, Hearing Aids, Bluetooth Headsets, POS Equipment, VOIP, PDAs, MIDs, Netbooks, and Servers

### Description

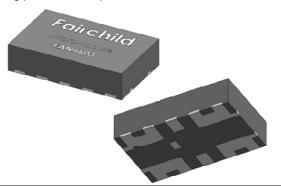
The FAN4603 is a fully integrated synchronous DC/DC buck converter that provides up to 500mA of output current over an input voltage ranging from 2.3V to 5.5V. It provides a fixed output voltage level ranging from 1.2V to 1.8V. Other voltage options are available on request.

The FAN4603 converter is offered as an ultra-miniature "Solder and Play" solution that requires no external components and is able to achieve a DC accuracy of  $\pm 4\%$  PWM and an output ripple less than 12mV.

Total footprint is 4.0 x 2.5mm with a maximum height of 1mm. It can be used in small battery-powered devices and applications with distributed DC POL requirements.

At moderate and light loads, pulse frequency modulation is used to operate the device in power-save mode with a typical quiescent current of 35µA. Even with such a low quiescent current, the part exhibits excellent transient response during large load swings. At higher loads, the system automatically switches to fixed-frequency control.

In shutdown mode, the supply current drops below  $2\mu A$ , reducing power consumption.

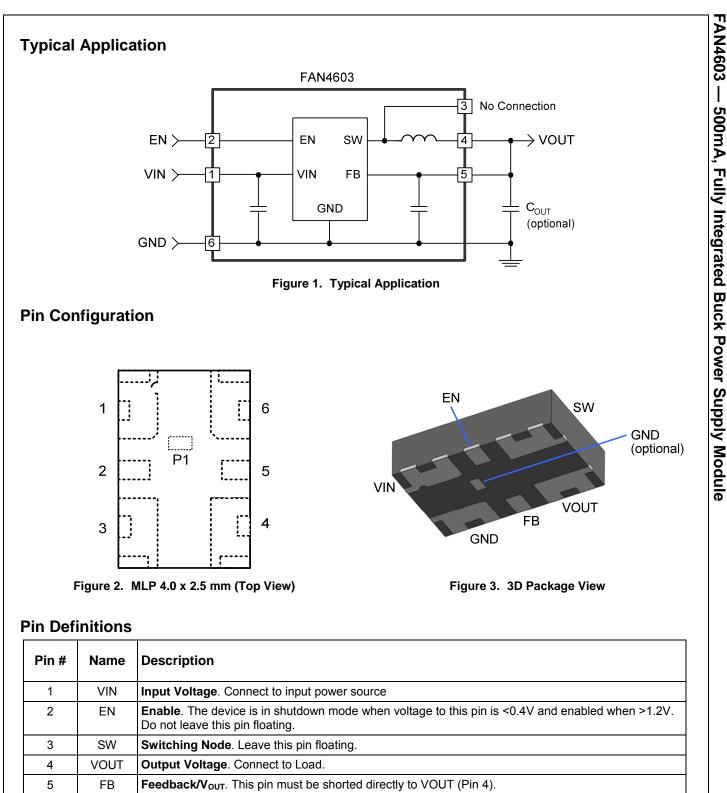


### **Ordering Information**

•				
Part Number	Output Voltage <sup>(1)</sup>	Package	Temperature Range	Packing
FAN4603MM18X	1.82V	6 Lead Molded Leadless Package (MLP), 4 x 2.5 x 1mm	–40 to 85°C	Tape and Reel
FAN4603MM15X	1.5V	6 Lead Molded Leadless Package (MLP), 4 x 2.5 x 1mm	–40 to 85°C	Tape and Reel
FAN4603MM12X	1.23V	6 Lead Molded Leadless Package (MLP), 4 x 2.5 x 1mm	–40 to 85°C	Tape and Reel

#### Note:

1. Other voltage options are available on request. Contact a Fairchild representative.



GND

P1

## **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter			Max.	Units
V	Input Voltage with Respect to GND		-0.3	6.0	V
V <sub>IN</sub>	Voltage on Any Other Pin with Respect to GND		-0.3	V <sub>IN</sub>	V
TJ	T <sub>J</sub> Junction Temperature		-40	150	°C
T <sub>STG</sub>	Storage Temperature		-65	+150	°C
TL	Lead Temperature (Soldering, 10 Seconds)			+260	°C
ESD	Electrostatic Discharge Capability	Human Body Model JESD22-A114	6		kV
ESD		Charged Device Model, JESD22-C101	2		

## **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter		Max.	Units
V <sub>CC</sub>	Supply Voltage Range		5.5	V
Ι <sub>ουτ</sub>	Output Current		500	mA
T <sub>A</sub>	Operating Ambient Temperature		+85	°C
TJ	Operating Junction Temperature	-40	+125	°C

## **Thermal Properties**

Symbol	Parameter	Typical	Units
Θ <sub>JA</sub>	Junction-to-Ambient Thermal Resistance <sup>(2)</sup>	120	°C/W

Note:

 Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with a twolayer 2s0p board in accordance to the JESD51- JEDEC standard. Special attention must be paid not to exceed junction temperature T<sub>J(max)</sub> at a given ambient temperature T<sub>A</sub>.

## **Electrical Specifications**

Unless otherwise noted,  $V_{IN}$  = 2.5 to 5.5V, EN =  $V_{IN}$ .  $T_A$  = -40°C to +85°C, using circuit of Figure 1. Typical values are at 3.6 $V_{IN}$ ,  $T_A$  = 25°C.

Symbol	Parameter		Conditions	Min.	Тур.	Max.	Units
Power Sup	oplies					1	
Ι <sub>Q</sub>	Quiescent Current				35	55	μA
I <sub>SD</sub>	Shutdown Supply Current		V <sub>IN</sub> = 3.6V, EN = GND		0.1	2.0	μA
V <sub>UVLO</sub>	Under-Voltage Lockout Thre	shold	Rising V <sub>IN</sub>		2.15	2.25	V
VUVHYST	Under-Voltage Lockout Hys	teresis			150		mV
$V_{\text{ENH}}$	Enable HIGH-Level Input Vo	oltage		1.2			V
V <sub>ENL</sub>	Enable LOW-Level Input Vo	Itage				0.4	V
$V_{\text{EN}_{\text{HYS}}}$	Enable Logic Input Hysteresis Enable Input Leakage Current				100		mV
I <sub>EN</sub>			EN = V <sub>IN</sub> or GND		0.01	1.00	μA
Oscillator							
f <sub>OSC</sub>	Oscillator Frequency		PWM Mode	5.4	6.0	6.6	MHz
Regulation	n						
	O Output Voltage Accuracy	4.001/	$I_{LOAD} = 0$ to 500mA	1.750	1.820	1.890	V
		1.82V	PWM Mode	1.750	1.820	1.856	V
N		1.50V	$I_{LOAD} = 0$ to 500mA	1.440	1.500	1.560	V
Vo			PWM Mode	1.444	1.500	1.530	V
1		1.001/	$I_{LOAD} = 0$ to 500mA	1.185	1.233	1.283	V
		1.23V	PWM Mode	1.185	1.233	1.259	V
t <sub>ss</sub>	Soft-Start Time	•	Rising EN to VOUT Regulation		180	300	μs
Protection	I						
I <sub>LIM</sub>	Peak Input Current Limit			850	1050	1250	mA
TSD	Thermal Shutdown		CCM Only		+150		°C
HYSTSD	Thermal Shutdown Hysteresis				15		°C

## **Typical Performance**

Unless otherwise specified,  $V_{IN}$  = 3.7V,  $T_A$  = 25°C.

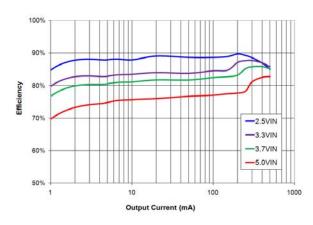


Figure 4. 1.82 VOUT Efficiency

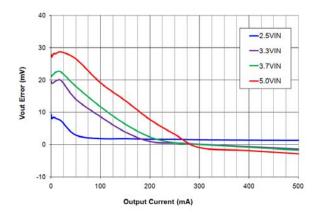


Figure 6. 1.82V<sub>OUT</sub> Regulation (Normalized)

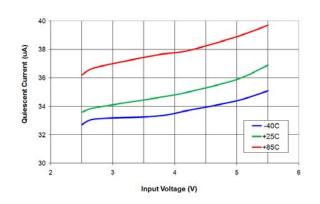


Figure 8. Quiescent Current Over Temperature (EN=V<sub>IN</sub>)

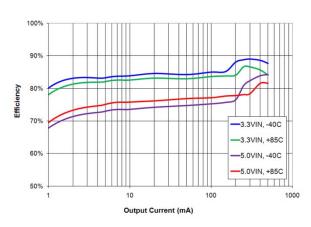


Figure 5. 1.82 VOUT Efficiency Over Temperature

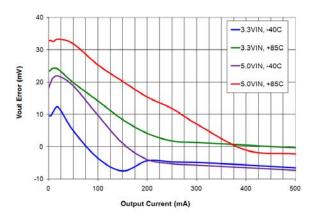


Figure 7. 1.82V<sub>OUT</sub> Regulation (Normalized) Over Temperature

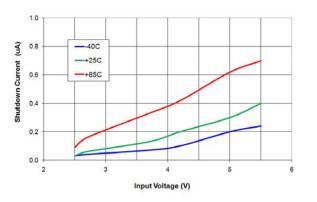
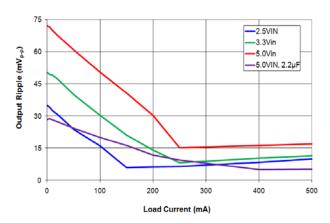


Figure 9. Shutdown Current Over Temperature (EN=0V)

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### **Typical Performance**

Unless otherwise specified,  $V_{IN}$  = 3.7V,  $T_A$  = 25°C.





1.82V<sub>OUT</sub> Output Ripple with Optional 2.2μF C<sub>OUT</sub> (1.5μF Actual)

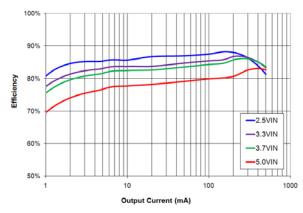


Figure 12. 1.23 VOUT Efficiency

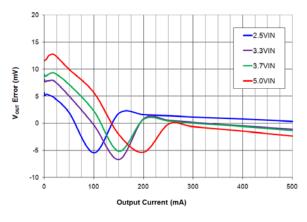


Figure 14. 1.23V<sub>OUT</sub> Regulation (Normalized)

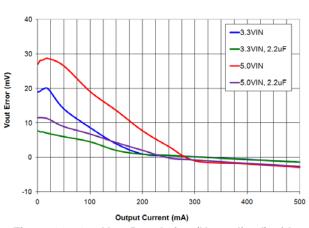


Figure 11. 1.82V<sub>OUT</sub> Regulation (Normalized) with Optional 2.2 $\mu$ F C<sub>OUT</sub> (1.5 $\mu$ F Actual)

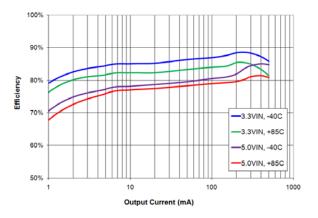


Figure 13. 1.23 VOUT Efficiency Over Temperature

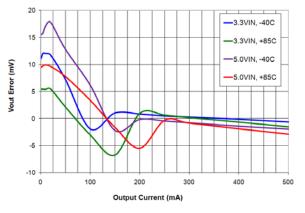


Figure 15. 1.23V<sub>OUT</sub> Regulation (Normalized) Over Temperature

FAN4603 — 500mA, Fully Integrated Buck Power Supply Module

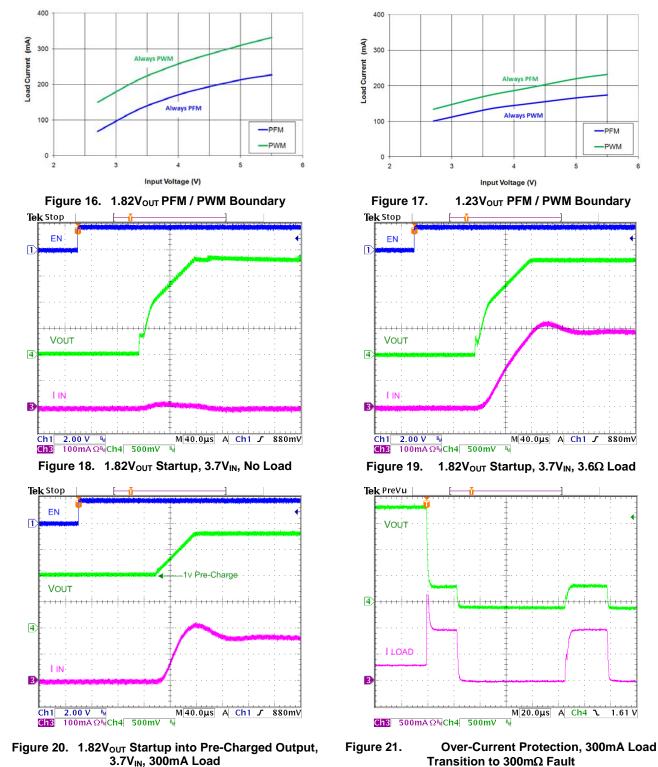
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-PFM

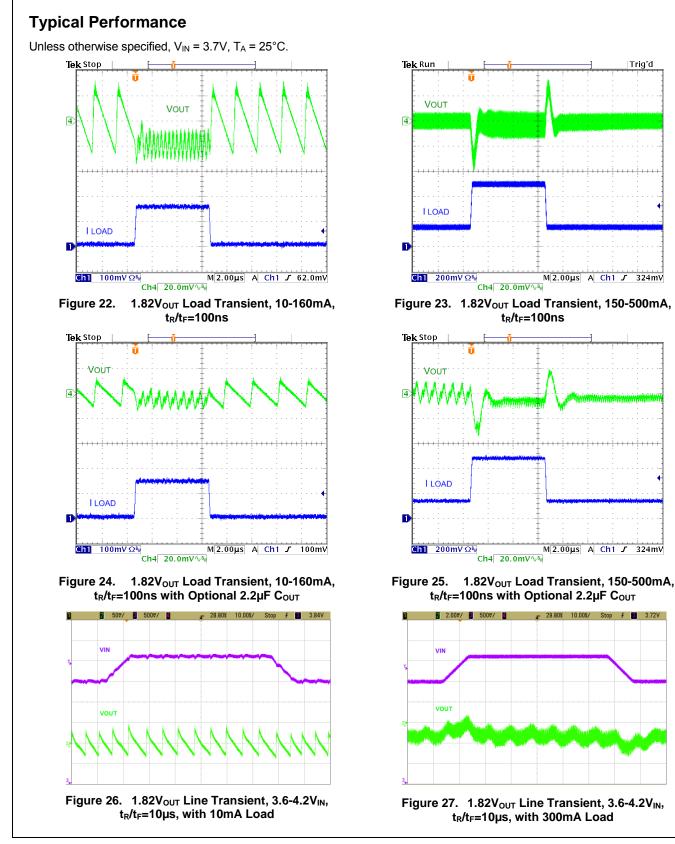
PWM

## **Typical Performance**

Unless otherwise specified,  $V_{IN}$  = 3.7V,  $T_A$  = 25°C.



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### **Operation Description**

The FAN4603 is a 500mA, step-down, switching-voltage regulator that delivers a fixed output from an input voltage supply of 2.3V to 5.5V. Using a proprietary architecture with synchronous rectification, the FAN4603 is capable of delivering a peak efficiency of >90%, while maintaining efficiency over 80% at load currents as low as 1mA. The regulator operates at a nominal PWM frequency of 6MHz.

#### **Control Scheme**

The FAN4603 uses a proprietary, non-linear, fixed-frequency PWM modulator to deliver a fast load transient response, while maintaining a constant switching frequency over a wide range of operating conditions. Regulator stability is not dependent on output capacitor ESR, which allows the use of ceramic capacitors. Although this type of operation normally results in a switching frequency that varies with input voltage and load current, an internal frequency loop holds the switching frequency constant over a large range of input voltages and load currents.

For very light loads, FAN4603 incorporates a discontinuous current (DCM) single-pulse PFM mode, which produces lower output ripple when compared with other PFM architectures. Transition between PWM and PFM is seamless, with a glitch of less than 20mV at  $V_{OUT}$  during the transition between DCM and CCM modes.

Combined with exceptional transient response characteristics, the very low quiescent current of the controller  $(35\mu A)$  maintains high efficiency; even at very light loads, while preserving fast transient response for applications requiring tight output regulation.

### **Enable and Soft-Start**

When EN is LOW, all circuits in FAN4603 are off and the IC draws ~100nA of current. When EN is HIGH and V<sub>IN</sub> is above its UVLO threshold, the regulator begins a soft-start cycle. The output ramp during soft-start is a fixed slew rate of 50mV/ $\mu$ s from 0 to 1 V<sub>OUT</sub>, then 12.5mV/ $\mu$ s until the output reaches its setpoint.

PWM mode operation is prohibited during the soft-start cycle to prevent  $C_{\text{OUT}}$  from being discharged. This allows glitchless starting into a pre-charged output.

### Startup into Large Cout

The IC may fail to start if heavy load is applied during startup and a large external  $C_{OUT}$  is present. This is due to the current-limit fault response, which protects the IC in an over-current condition during soft-start.

The current required to charge  $C_{\text{OUT}}$  during soft-start, referred to as "displacement current," is given as:

$$I_{\text{DISP}} = C_{\text{OUT}} \bullet \frac{dV}{dt}$$
(1)

where  $\frac{dV}{dt}$  refers to the soft-start slew rate.

To prevent shutdown during soft-start, the following condition must be met:

$$I_{\text{DISP}} + I_{\text{LOAD}} < I_{\text{MAX(DC)}}$$
(2)

where  $I_{MAX(DC)}$  is the maximum load current the IC is guaranteed to support (500mA).

Table 1 shows combinations of external  $C_{OUT}$  that allow the IC to start successfully with the minimum  $R_{LOAD}$  that can be supported at each.

Table 1. Minimum  $R_{\text{LOAD}}$  Values for Soft-Start with Various External  $C_{\text{OUT}}$  Values

C <sub>OUT</sub>	Minimum R <sub>LOAD</sub>
2.2μF, 0402	V <sub>OUT</sub> / 0.50
4.7μF, 0402	V <sub>OUT</sub> / 0.50
10μF, 0603	V <sub>OUT</sub> / 0.45

Multiple soft-start cycles may be required for  $C_{\rm OUT}{>}10\mu f$  (15 $\mu f$  with no load). The IC shuts down for 85 $\mu s$  when  $I_{\rm DISP}{+}$   $I_{\rm LOAD}$  exceeds  $I_{\rm LIMIT}$  for more than 21 $\mu s$  of current limit. The IC then begins a new soft-start cycle. Subsequent soft-start cycles begin with any charge retained by  $C_{\rm OUT}$  while the IC is off, allowing  $V_{\rm OUT}$  to incrementally reach regulation over multiple soft-start attempts.

### **Current Limit, Fault Shutdown, and Restart**

A heavy load or short circuit on the output causes the current to increase until a maximum current threshold is reached. Upon reaching this point, the high-side switch turns off, preventing high currents from causing damage. The regulator continues to limit the current cycle-by-cycle. After 21 $\mu$ s of current limit, the regulator triggers an over-current fault, causing the regulator to shut down for about 85 $\mu$ s before attempting an automatic restart.

If the fault is caused by short circuit, the soft-start circuit attempts to restart and produces an over-current fault after about  $32\mu s$ , which results in a duty cycle of less than 30%, limiting power dissipation.

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(4)

### **Under-Voltage Lockout (UVLO)**

When EN is HIGH, the under-voltage lockout keeps the part from operating until the input supply voltage rises high enough to properly operate. This ensures no misbehavior of the regulator during startup or shutdown.

### **Thermal Shutdown**

When the die temperature increases, due to a high load condition and/or a high ambient temperature, the output switching is disabled until the temperature on the die has fallen sufficiently. The junction temperature at which the thermal shutdown activates is nominally 150°C with 15°C hysteresis. After cooling, the IC automatically restarts, with a soft-start cycle.

### **Reducing PFM Output Ripple**

PFM output ripple amplitude can be reduced by adding external  $C_{\text{OUT}},$  with negligible impact on efficiency.

Reduced output ripple also results in less DC voltage excursion at very light loads. Maximum PFM ripple occurs at no load and is  $V_{\rm IN}$  and  $V_{\rm OUT}$  proportional.

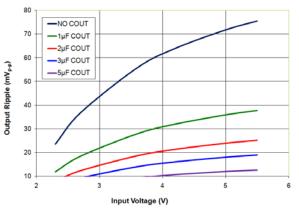


Figure 28. Typical  $1.82V_{OUT}$  No Load PFM Ripple vs.  $V_{IN}$  Note:

3. Ripple is less for lower  $V_{OUT}$  levels.

The effective value of external  $C_{\mbox{\scriptsize OUT}}$  for a desired ripple amplitude can be determined using:

$$C_{OUT.}(\mu F) = \left(\frac{(V_{IN} - V_{OUT}) \bullet V_{OUT} \bullet 62}{V_{IN} \bullet V_R}\right) - 1$$
(3)

where  $V_{\mathsf{R}}$  is the desired output ripple amplitude in mV and  $V_{\mathsf{IN}}$  and  $V_{\mathsf{OUT}}$  are in Volts.

The simplified equation above is representative of nominal component values and does not account for device tolerances. The bias level effects associated with case size, voltage rating, and dielectric type of ceramic capacitors should be considered when selecting  $C_{\text{OUT}}$ .

### Minimum Off-Time Effect on Switching Frequency

 $t_{OFF(MIN)}$  is 50ns. This imposes constraints on the maximum

 $\frac{V_{OUT}}{V_{IN}}$  that the FAN4603 can provide or the maximum output

voltage it can provide at low  $V_{\text{IN}},$  while maintaining a fixed switching frequency in PWM mode.

When V<sub>IN</sub> is LOW, fixed switching is maintained as long as  $\frac{V_{OUT}}{1 - t_{OTT}} \le 1 - t_{OTT} \le 0.7$ 

$$\frac{1}{V_{IN}} \le 1 - t_{OFF(MIN)} \bullet f_{SW} \approx 0.7$$

The switching frequency drops when the regulator cannot provide sufficient duty cycle at 6MHz to maintain regulation.

The calculation for reduced switching frequency to maintain regulation is given by:

$$f_{SW} = \frac{1}{t_{SW(MAX)}}$$

where:

$$t_{SW(MAX)} = 50ns \bullet \left(1 + \frac{V_{OUT} + 0.36 \bullet I_{OUT}}{V_{IN} - V_{OUT} - 0.52 \bullet I_{OUT}}\right)$$
(5)

where V and I terms are in Volts and Amperes, respectively.

## **PCB Layout Guideline**

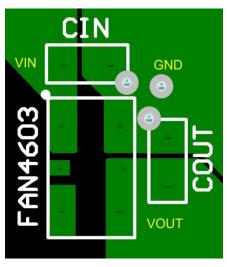


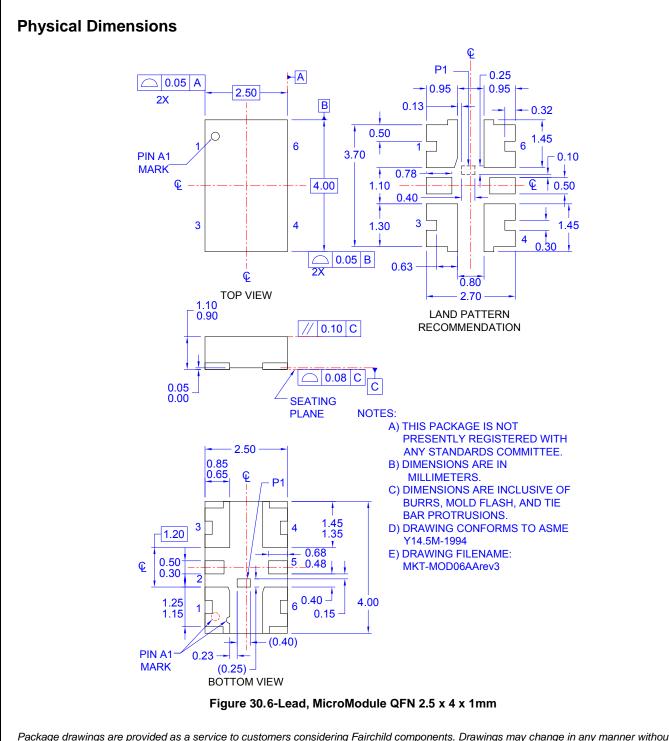
Figure 29. Recommended PCB Layout

FB (pin 5) must be directly connected to VOUT (pin 4).

Figure 29 shows the recommended locations of optional  $C_{\text{IN}}$  and  $C_{\text{OUT}},$  shown as 0603 size devices.

Pad P1 shown in Figure 30 "Land Pattern" is for reference only and should not be connected on PCB.

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#### Definition of Terms

Semicon of Terms					
Advance Information Formative / In Design Preliminary First Production No Identification Needed Full Production		Definition			
		Datasheet contains the design specifications for product development. Specifications may change in any manner without notice.			
		Datasheet contains preliminary data; supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve design.			
		Datasheet contains final specifications. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve the design.			
Obsolete	Not In Production	Datasheet contains specifications on a product that is discontinued by Fairchild Semiconductor. The datasheet is for reference information only.			

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