Current Mode PWM Control Circuit

The CS52843 provides all the necessary features to implement off-line fixed frequency current-mode control with a minimum number of external components.

The CS52843 incorporates a new precision temperature-controlled oscillator to minimize variations in frequency. An undervoltage lockout ensures that V_{REF} is stabilized before the output stage is enabled. In the CS52843 turn on is at 8.4 V and turn off at 7.6 V.

Other features include low start-up current, pulse-by-pulse current limiting, and a high-current totem pole output for driving capacitive loads, such as gate of a power MOSFET. The output is low in the off state, consistent with N-channel devices.

Features

- Optimized for Off-Line Control
- Internally Temperature Compensated Oscillator
- PIERSER PRESENTATIVE OBSOLITIES • V_{REF} Stabilized before Output Stage is Enabled
- Very Low Start–Up Current 300 μA (typ)
- Pulse-by-Pulse Current Limiting
- Improved Undervoltage Lockout
- Double Pulse Suppression
- 2.0% 5.0 Volt Reference
- High Current Totem Pole Output



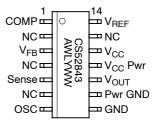
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ŜΟ. D SUFFIX CASE 751A

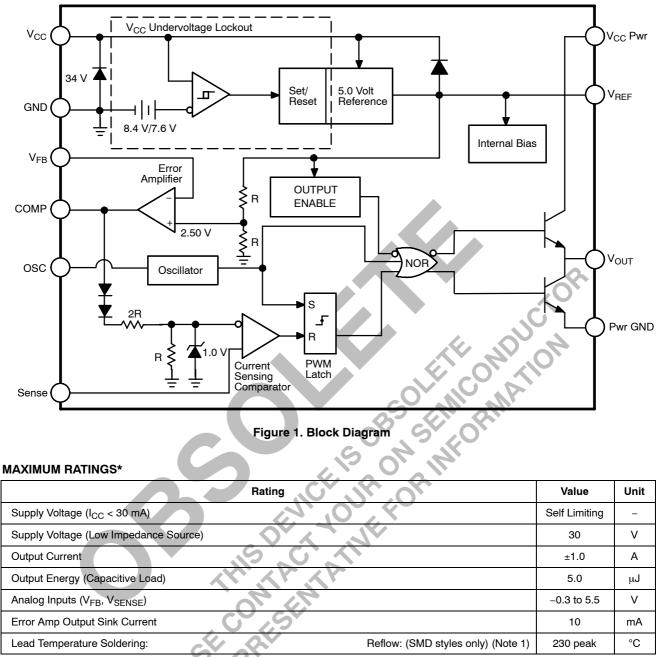
PIN CONNECTIONS AND MARKING DIAGRAMS



4	= Assembly Location
NL, L	= Wafer Lot
YY, Y	= Year
NW, W	= Work Week

ORDERING INFORMATION

Device	Package	Shipping		
CS52843ED8	SO-8	95 Units/Rail		
CS52843EDR8	SO-8	2500 Tape & Reel		
CS52843ED14	SO-14	55 Units/Rail		
CS52843EDR14	SO-14	2500 Tape & Reel		



1. 60 second maximum above 183°C.

*The maximum package power dissipation must be observed.

ELECTRICAL CHARACTERISTICS	$(-40^{\circ}C ≤ T_A ≤ 85^{\circ}C; V_{CC} = 15 V$ (Note 2.); $R_T = 680 \Omega$; C_T] = 0.022 µF for triangle
mode, R _T = 10 kΩ; C _T	node; unless otherwise specified.)

Parameter	Parameter Test Conditions			Max	Unit	
Reference Section						
Output Voltage	T _J = 25°C, I _{REF} = 1.0 mA	4.9	5.0	5.1	V	
Line Regulation	12 ≤[¥ _{CC} ≤ 25 V	-	6.0	20	mV	
Load Regulation	1.0 ≤ I _{REF} ≤ 20 mA	-	6.0	25	mV	
Temperature Stability	Note 2.	-	0.2	0.4	mV/°C	
Total Output Variation	Line, Load, Temp. Note 2.	4.82	-	5.18	V	
Output Noise Voltage	10 Hz ≤ f ≤ 10 kHz, T _J = 25°C, Note 2.	-	50	-	μV	
Long Term Stability	T _A = 125°C, 1000 Hrs. Note 2.	-	5.0	25	mV	
Output Short Circuit	$T_A = 25^{\circ}C$	-30	-100	-180	mA	
Oscillator Section						
Initial Accuracy	Sawtooth Mode, $T_J = 25^{\circ}C$, Note 2. Triangle Mode, $T_J = 25^{\circ}C$	47 44	52 52	57 60	kHz kHz	
Voltage Stability	12 ≤[¥ _{CC} ≤ 25 V	-	0.2	1.0	%	
Temperature Stability	Sawtooth Mode $T_{MIN} \le T_A \le T_{MAX}$ Triangle Mode $T_{MIN} \le T_A \le T_{MAX}$, Note 2.		5.0 8.0		% %	
Amplitude	V _{OSC} (peak to peak)	× C	1.7	-	V	
Discharge Current	$T_{J} = 25^{\circ}C$ $T_{MIN} \le T_{A} \le T_{MAX}$	7.3 6.8	8.3 -	9.3 9.8	mA mA	
Error Amp Section	0, 9	20				
Input Voltage	V _{COMP} = 2.5 V	2.42	2.50	2.58	V	
Input Bias Current	V _{FB} = 0 V	_	-0.3	-2.0	μA	
Avol	2.0 ≤[V _{OUT} ≤ 4.0 V	65	90	-	dB	
Unity Gain Bandwidth	Note 2.	0.7	1.0	-	MHz	
PSRR	12 ≤[V _{CC} ≤ 25 V	60	70	-	dB	
Output Sink Current	V _{FB} = 2.7 V, V _{COMP} = 1.1 V	2.0	6.0	-	mA	
Output Source Current	$V_{FB} = 2.3 V, V_{COMP} = 5.0 V$	-0.5	-0.8	-	mA	
V _{OUT} HIGH	V_{FB} = 2.3 V, R_L = 15 k Ω to GND	5.0	6.0	-	V	
V _{OUT} LOW	V_{FB} = 2.7 V, R_L = 15 k Ω to V_{REF}	-	0.7	1.1	V	
Current Sense Section	SY OK					
Gain	Notes 3 & 4.	2.85	3.0	3.15	V/V	
Maximum Input Signal	V _{COMP} = 5.0 V, Note 3.	0.9	1.0	1.1	V	
PSRR	12 ≤[¥ _{CC} ≤ 25 V, Note 3.	-	70	-	dB	
Input Bias Current	V _{SENSE} = 0 V	-	-2.0	-10	μA	

2. These parameters, although guaranteed, are not 100% tested in production.

T_J = 25°C, Note 2.

3. Parameter measured at a trip point of latch with $V_{\mbox{FB}}$ = 0.

Delay to Output

4. Gain defined as: A =
$$\frac{\Delta V_{COMP}}{\Delta V_{SENSE}}$$
; 0 ≤ V_{SENSE} ≤ 0.8 V

150

_

300

ns

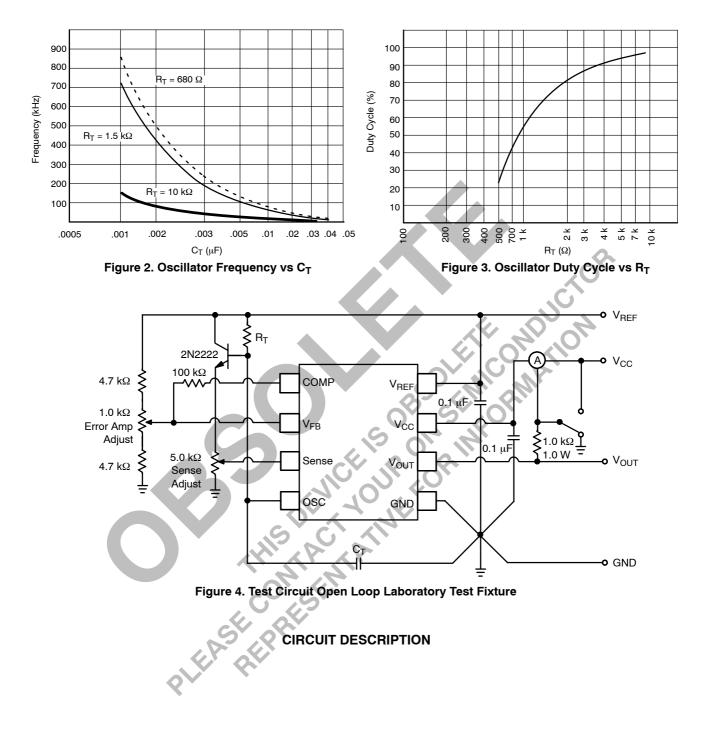
Parameter	Test Conditions	Min	Тур	Max	Unit
Output Section	· · · ·				
Output Low Level	I _{SINK} = 20 mA I _{SINK} = 200 mA		0.1 1.5	0.4 2.2	V V
Output High Level	I _{SOURCE} = 20 mA I _{SOURCE} = 200 mA	13 12	13.5 13.5	-	V V
Rise Time	T _J = 25°C, C _L = 1.0 nF, Note 5.	-	50	150	ns
Fall Time	$T_J = 25^{\circ}C, C_L = 1.0 \text{ nF}, \text{ Note 5.}$	-	50	150	ns
Output Leakage	UVLO Active V _{OUT} = 0	-	-0.01	-10	μΑ
Total Standby Current					
Start-Up Current	-	-	300	500	μΑ
Operating Supply Current	V _{FB} = V _{SENSE} = 0 V, R _T = 10 kΩ; C _T	-	11	17	mA
V _{CC} Zener Voltage	I _{CC} = 25 mA	-	34	0-	V
Undervoltage Lockout Sectior			Ĵ.		
Start Threshold	-	7.8	8.4	9.0	V
Min. Operating Voltage	After Turn On	7.0	7.6	8.2	V

ELECTRICAL CHARACTERISTICS (continued) (-40°C ≤ $T_A ≤ 85^{\circ}C$; $V_{CC} = 15 \text{ V}$ (Note 2.); $R_T = 680 \Omega$; $C_T \oplus 0.022 \mu$ F for triangle mode, $R_T = 10 \text{ k}\Omega$; $C_T \oplus 3.3 \text{ nF}$ sawtooth mode; unless otherwise specified.)

PACKAGE PIN DESCRIPTION

KAGE PIN DESCRIPTION					
Package Lead Number					
SO-8	SO-14	Lead Symbol	Function		
1	1	СОМР	Error amp output, used to compensate error amplifier.		
2	3	V _{FB}	Error amp inverting input.		
3	5	SENSE	Noninverting input to Current Sense Comparator.		
4	7	OSC	Oscillator timing network with capacitor to ground, resistor to VREF.		
5	8	GND	Ground.		
-	9	Pwr GND	Output driver ground.		
6	10	Vout	Output drive pin.		
-	11	V _{CC} Pwr	Output driver positive supply.		
7	12	V _{CC}	Positive power suppy.		
8	14	V _{REF}	Output of 5.0 V internal reference.		
_	2, 4, 6, 13	NC	No Connection.		

TYPICAL PERFORMANCE CHARACTERISTICS



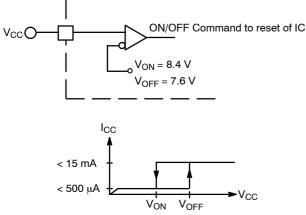


Figure 5. Startup Voltage for the CS52843

Undervoltage Lockout

During Undervoltage Lockout (Figure 5), the output driver is biased to sink minor amounts of current. The output should be shunted to ground with a resistor to prevent activating the power switch with extraneous leakage currents.

PWM Waveform

To generate the PWM waveform, the control voltage from the error amplifier is compared to a current sense signal which represents the peak output inductor current (Figure 6). An increase in V_{CC} causes the inductor current slope to increase, thus reducing the duty cycle. This is an inherent feed-forward characteristic of current mode control, since the control voltage does not have to change during changes of input supply voltage.

When the power supply sees a sudden large output current increase, the control voltage will increase allowing the duty cycle to momentarily increase. Since the duty cycle tends to exceed the maximum allowed to prevent transformer saturation in some power supplies, the internal oscillator waveform provides the maximum duty cycle clamp as programmed by the selection of oscillator timing components.

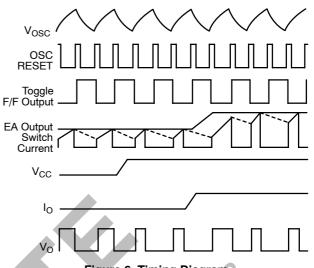


Figure 6. Timing Diagram

Setting the Oscillator

The times t_c and t_d can be determined as follows:

$$t_{c} = R_{T}C_{T}ln\left(\frac{V_{REF} - V_{LOWER}}{V_{REF} - V_{UPPER}}\right)$$

$$\mathbf{d} = \mathsf{R}_{\mathsf{T}}\mathsf{C}_{\mathsf{T}}\mathsf{In}\left(\frac{\mathsf{V}_{\mathsf{REF}} - \mathsf{I}_{\mathsf{d}}\mathsf{R}_{\mathsf{T}}\mathsf{V}_{\mathsf{LOWER}}}{\mathsf{V}_{\mathsf{REF}} - \mathsf{I}_{\mathsf{d}}\mathsf{R}_{\mathsf{T}}} - \mathsf{V}_{\mathsf{UPPER}}\right)$$

Substituting in typical values for the parameters in the above formulas:

$$EF = 5.0 V, VUPPER = 2.7 V,$$

$$V_{LOWER} = 1.0 V, I_{d} = 8.3 mA$$

$$t_{C}\approx 0.5534R_{T}C_{T}$$

$$t_{d} = R_{T}C_{T} \ln \left(\frac{2.3 - 0.0083R_{T}}{4.0 - 0.0083R_{T}}\right)$$

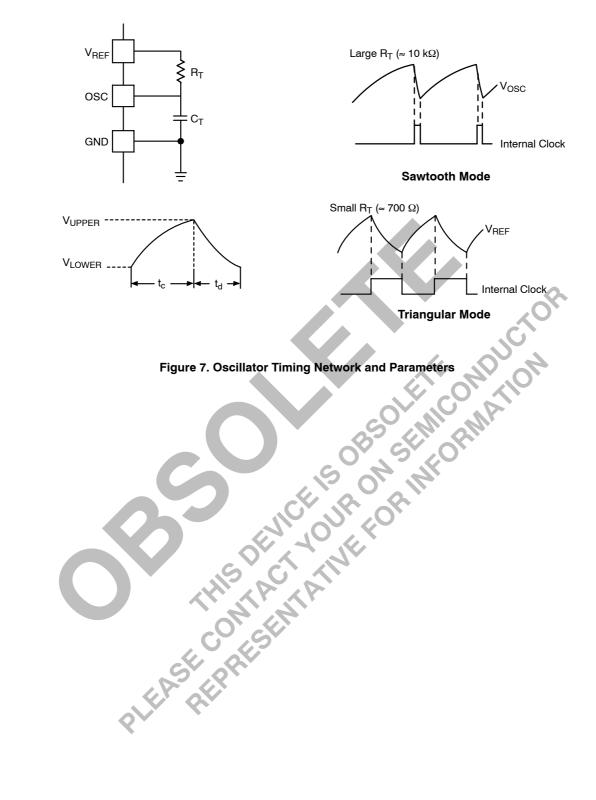
For better accuracy R_T should be $\ge 10 \text{ k}\Omega$.

Grounding

High peak currents associated with capacitive loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected close to GND in a single point ground.

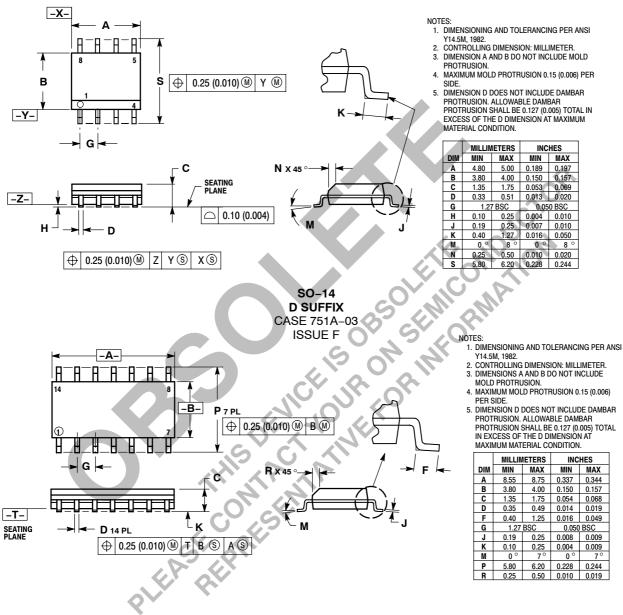
The transistor and 5.0 k Ω potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to Sense.

VIN



PACKAGE DIMENSIONS

SO-8 DF SUFFIX CASE 751-07 ISSUE W



PACKAGE THERMAL DATA

Parameter		SO-8	SO-14	Unit
$R_{\Theta JC}$	Typical	45	30	°C/W
$R_{\Theta JA}$	Typical	165	125	°C/W



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