APW7511



24V 3A 800kHz Synchronous Buck Converter With Adj. Soft Start

Features

- Wide Input Voltage from 4.5V to 24V
- 3A Continuous Output Current
- Adjustable Output Voltage from 0.807V to 13V
- Integrated Low R_{DS(ON)} MOSFETs
- Fixed 800kHz Switching Frequency
- Stable with Low ESR Ceramic Output Capacitors
- Power-On-Reset Detection
- Over-Temperature Protection
- Current-Limit Protection with HICCUP Mode
- Programmable Soft-Start
- Small TSOT-23-8A Package
- Lead Free and Green Devices Available
 (RoHS Compliant)

Applications

- Notebook Systems and I/O Power
- Digital Set-Top Boxes
- Flat-Panel Television and Monitors
- Distributed Power Systems

Simplified Application Circuit



APW7511 is a 3A synchronous buck converter with integrated low R_{DS(ON)} power MOSFETs. The APW7511 design with a current-mode control scheme, can convert wide input voltage of 4.5V to 24V to the output voltage adjustable from 0.807V to 13V to provide excellent output voltage regulation.

The APW7511 is equipped advance asynchronous modulation mode operation. Increase efficiency at light load. The APW7511 is also equipped with Power-on-reset, soft

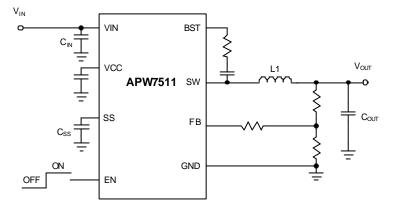
start, and whole protections (under-voltage,

over-temperature, and current-limit) into a single package. This device, available TSOT-23-8A, provides a very compact system solution external components and PCB area.

Pin Configuration

| SS 1 🗖 | 8 FB |
|--------|-------|
| VIN 2 | |
| SW 3 🔲 | 6 EN |
| GND 4 | 5 BST |

APW 7511 TSOT-23-8A

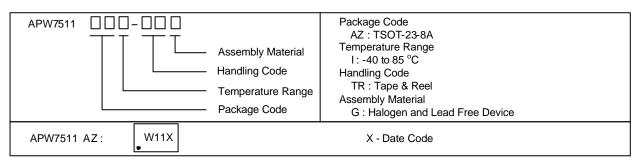


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Ordering and Marking Information



Note: ANPEC lead-free products contain molding compounds/die attach materials and 100% matte tin plate termination finish; which are fully compliant with RoHS. ANPEC lead-free products meet or exceed the lead-free requirements of IPC/JEDEC J-STD-020D for MSL classification at lead-free peak reflow temperature. ANPEC defines "Green" to mean lead-free (RoHS compliant) and halogen free (Br or Cl does not exceed 900ppm by weight in homogeneous material and total of Br and Cl does not exceed 1500ppm by weight).

Absolute Maximum Ratings (Note 1)

| Symbol | Parameter | Rating | Unit |
|---------------------|--|-----------|------|
| VIN, VEN | VIN Supply to GND Voltage and EN to GND Voltage | -0.3 ~ 27 | V |
| V _{SW} | SW to GND Voltage | -0.3 ~ 27 | V |
| V _{BST-SW} | BST to SW Voltage | -0.3 ~ 6 | V |
| | SS, VCC and FB to GND Voltage | -0.3 ~ 6 | V |
| TJ | Junction Temperature | 150 | °C |
| T _{STG} | Storage Temperature | -65 ~ 150 | °C |
| T _{SDR} | Maximum Lead Soldering Temperature (10 Seconds) | 260 | °C |

Note1: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability

Thermal Characteristics

| I | Symbol | Parameter | Typical Value | Unit |
|---|--------|---|---------------|------|
| | θια | Junction-to-Ambient Resistance in Free Air (Note 2) | 100 | °C/W |

Note 2: θ_{JA} is measured with the component mounted on a high effective thermal conductivity test board in free air.

Recommended Operating Conditions (Note 3)

| Symbol | Parameter | Range | Unit |
|------------------|--------------------------|----------|------|
| V _{IN} | VIN Supply Voltage | 4.5 ~ 24 | V |
| V _{EN} | EN Input Voltage | 0 ~ 24 | V |
| V _{OUT} | Converter Output Voltage | 0.807~13 | V |
| Ιαπ | Converter Output Current | 0 ~ 3 | А |



Recommended Operating Conditions (Cont.) (Note 3)

| Symbol | Parameter | Range | Unit |
|----------------|------------------------------|-----------|------|
| Соит | Converter Output Capacitance | 22 ~ 47 | μF |
| L1 | Inductance | 1.5 ~ 10 | μH |
| T _A | Ambient Temperature | -40 ~ 85 | °C |
| TJ | Junction Temperature | -40 ~ 125 | °C |

Note 3 : Refer to the typical application circuit.

Electrical Characteristics

Unless otherwise specified, these specifications apply over V_{IN}=12V, V_{EN}=3V and T_A= -40 to 85 °C. Typical values are at T_A=25°C.

| | Parameter | Tako ku | | APW7511 | | |
|---------------------|----------------------------------|---|-----|---------|------|-------------------|
| Symbol | | Test Conditions | Min | Тур | Max | Unit |
| SUPPLY | CURRENT | | | 1 | | |
| I _{VIN} | VIN Supply Current | V _{FB} =0.9V, SW=NC | - | 1 | 1.2 | mA |
| I _{VIN_SD} | VIN Shutdown Supply Current | V _{EN} =0V | - | - | 10 | μA |
| POWER- | ON-RESET (POR) | • | | • | | • |
| | VIN POR Voltage Threshold | V _{IN} Rising | 3.7 | 3.9 | 4. 1 | V |
| | VIN POR Hysteresis | | - | 0.6 | - | V |
| REFEREI | | | | • | - | • |
| V_{REF} | Reference Voltage | | - | 0.807 | - | V |
| | Output Voltage Accuracy | T _J =25°C, I _{OUT} =10mA | -1 | - | +1 | % |
| ۱ _в | FB input current | | - | 10 | 50 | nA |
| V _{VCC} | VCC Regulator | | - | 5 | - | V |
| | VCC Load Regulation | I _{VCC} =3mA | - | 3 | - | % |
| OSCILLA | TOR AND DUTY CYCLE | | | • | - | |
| F_{SW} | Switching Frequency | | 700 | 800 | 900 | kHz |
| D _{AMX} | Maximum Duty Cycle | | - | 93 | - | % |
| | Minimum on-time | | - | - | 100 | ns |
| POWER | MOSFET | | | • | | |
| | High Side MOSFET Resistance | | - | 100 | - | mΩ |
| | Low Side MOSFET Resistance | | - | 40 | - | mΩ |
| | High Side Switch Leakage Current | V _{EN} =0V, V _{IN} =24V, V _{SW} =0V | - | - | 1 | μΑ |
| | Low Side Switch Leakage Current | V _{EN} =0V, V _{IN} =24V, V _{SW} =24V | - | - | 1 | μA |
| PROTEC | TIONS | | | | • | |
| I _{LIM} | High Side MOSFET Current-Limit | | 4 | 5 | 6 | A |
| | Under-Voltage Protection (UVP) | | 40 | 50 | 60 | %V _{REF} |
| | FB Over Voltage Protection | | 120 | 125 | 130 | %V _{REF} |
| | FB OVP Hysteresis | | - | 20 | - | %V _{REF} |
| | Hiccup Delay time | | - | 1 | - | T _{SS} |
| TOTR | OTP Rising Threshold (Note 4) | | - | 150 | - | °C |
| | OTP Hysteresis (Note 4) | | - | 30 | - | °C |

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Electrical Characteristics (cont.)

Unless otherwise specified, these specifications apply over V_{IN} =12V, V_{EN} =3V and T_{A} = -40 to 85 °C. Typical values are at T_{A} =25°C.

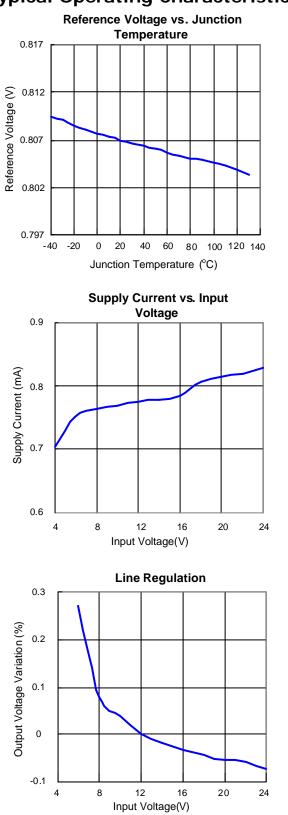
| Cumhal | Parameter | Test Conditions | | l lmit | | |
|---------|---------------------------------|---------------------|-----|--------|-----|------|
| Symbol | Symbol Test Conditions | | Min | Тур | Max | Unit |
| SOFT-ST | ART, ENABLE | | -1 | | | 1 |
| Iss | Soft-Start Charge Current | | - | 11 | - | μA |
| | EN Rising Threshold Voltage | | 1.2 | 1.4 | 1.6 | V |
| | EN Falling Threshold Hysteresis | | - | 0.2 | - | V |
| | EN turn off delay | | - | 8 | - | μs |
| | EN Input Current | V _{EN} =2V | - | 2 | - | μΑ |

Note4: Guaranteed by design.

Pin Description

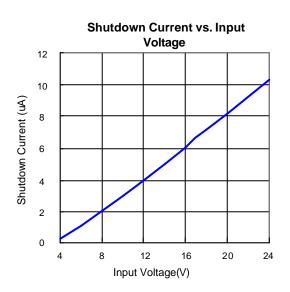
| Р | IN | FUNCTION |
|-----|------|---|
| NO. | NAME | FUNCTION |
| 1 | SS | Soft Start. Connect a capacitor to GND to set the soft start interval. |
| 2 | VIN | Power Input. VIN supplies the power to the control circuitry, gate driver. Connecting a ceramic bypass capacitor and a suitably large capacitor between VIN and GND eliminates switching noise and voltage ripple on the input to the IC. |
| 3 | SW | Power Switching Output. SW is the Source of the N-Channel power MOSFET to supply power to the output LC filter. |
| 4 | GND | Signal and power ground. |
| 5 | BST | High-Side Gate Drive Boost Input. BS supplies the voltage to drive the high-side N-channel MOSFET. At least 10nF capacitor should be connected from SW to BST to supply the high side switch. |
| 6 | EN | Enable Input. EN is a digital input that turns the regulator on or off. EN threshold is 1.4V with 0.2V hysteresis. |
| 7 | VCC | Bias Supply. Decouple with a 0.1uF capacitor or higher is recommended. |
| 8 | FB | Output feedback Input. The IC senses the feedback voltage via FB and regulates FB voltage at 0.807V. Connecting FB with a resistor-divider from the converter's output to set the output voltage. |

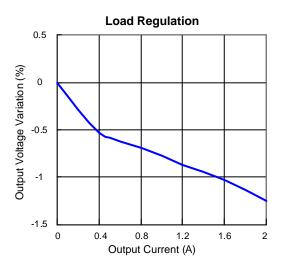




Typical Operating Characteristics







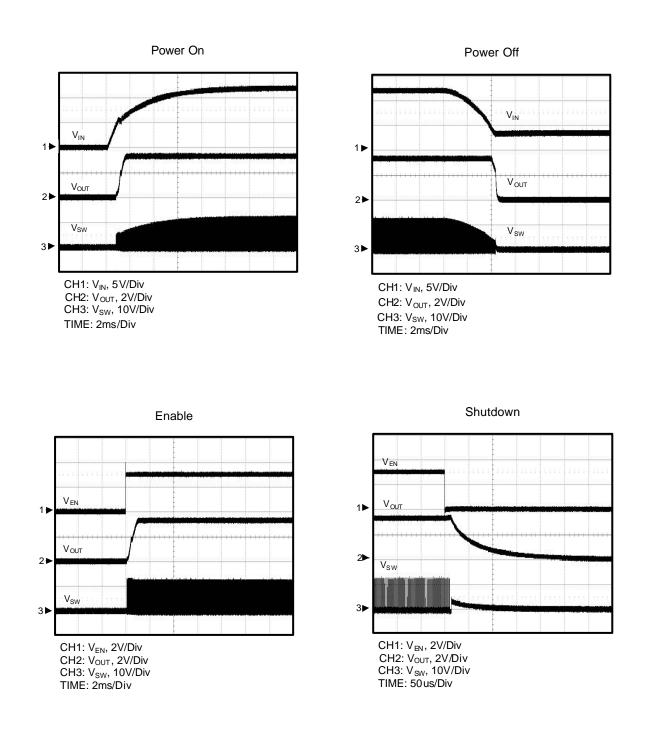


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Operating Waveforms

Refer to the typical application circuit. The test condition is $V_{IN}=12V$, $T_A=25^{\circ}C$ unless otherwise specified.

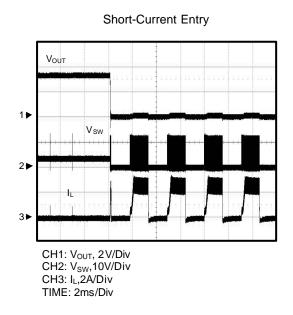


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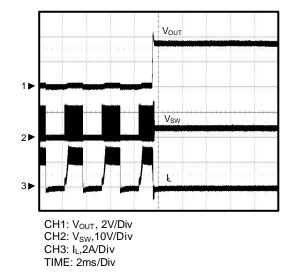


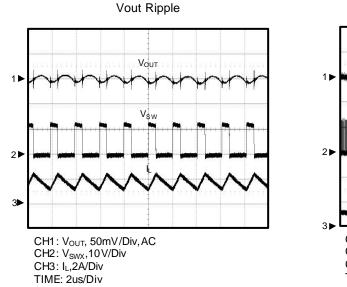
Operating Waveforms

Refer to the typical application circuit. The test condition is $V_{IN}=12V$, $T_A=25^{\circ}C$ unless otherwise specified.

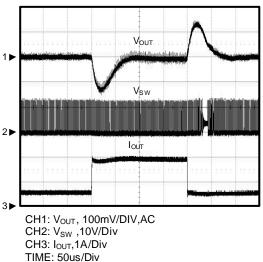


Short-Current Recovery





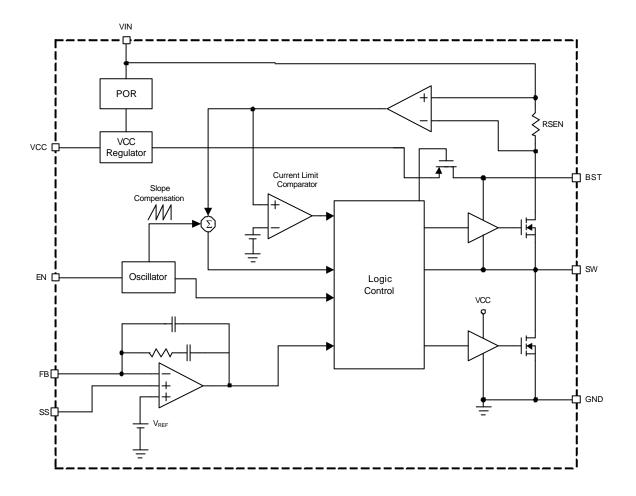




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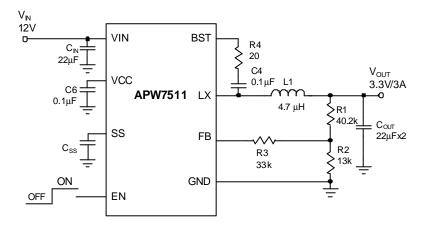


Block Diagram



APW7511

Typical Application Circuit



Dual Power Input : VIN Pre-existing & VOUT setting less than VIN POR

Single Power Input : VIN divided to EN & VOUT setting more than VIN POR

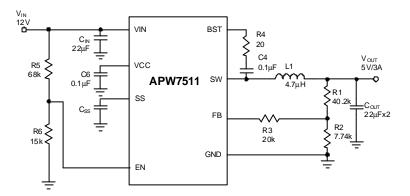


Table 1. Componments Selection for Different Output Voltage

| V _{оυт} (V) | R1 (kW) | R2 (kW) | R3 (k W) | R4 (W) | L (uH) | С _{оит} (uF) |
|----------------------|---------|---------|------------------|--------|--------|-----------------------|
| 1.05 | 20.5 | 68.1 | 100 | 20~60 | 1.5 | 44 |
| 1.2 | 20.5 | 42.1 | 75 | 20~60 | 1.8 | 44 |
| 3.3 | 40.2 | 13 | 33 | 20~60 | 4.7 | 44 |
| 5 | 40.2 | 7.74 | 20 | 20~60 | 4.7 | 44 |



Function Description

Main Control Loop

The APW7511 is a constant frequency, synchronous rectifier and current-mode switching regulator. In normal operation, the internal upper power MOSFET is turned on each cycle. The peak inductor current at which ICMP turn off the upper MOSFET is controlled by the voltage on the COMP node, which is the output of the error amplifier (EAMP). An external resistive divider connected between VOUT and ground allows the EAMP to receive an output feedback voltage V_{FB} at FB pin. When the load current increases, it causes a slightly decrease in V_{FB} relative to the 0.807V reference, which in turn causes the COMP voltage to increase until the average inductor current matches the new load current.

Enable/Shutdown

Driving EN to the ground places the APW7511 in shutdown mode. When in shutdown, the internal power MOSFETs turn off, all internal circuitry shuts down and the quiescent supply current reduces to $1 \mu A$ typical.

Under Voltage Lockout (UVLO)

An under-voltage lockoutfunction prevents the device from operating if the input voltage on VIN is lower than approximately 3.9V. The device automatically enters the shutdown mode if the voltage on VIN drops below approximately 3.9V. This under-voltage lockout function is implemented in order to prevent the malfunctioning of the converter.

Over-Current-Protection and Hiccup

The APW7511 has a cycle-by-cycle over-current limitwhen the inductor current peak value exceeds the set current limit threshold. Meanwhile, the output voltage drops until FB is below the Under-Voltage (UV) threshold below the reference. Once UV is triggered, the APW7511 enters hiccup mode to periodically restart the part. This protection mode is especially useful when the output is dead-shorted to ground. The average short circuit current is greatly reduced to alleviate thermal issues and to protect the regulator. The APW7511 exits the hiccup mode once the over-current condition is removed.

Over-Temperature Protection (OTP)

The over-temperature circuit limits the junction temperature of the APW7511. When the junction temperature exceeds 150°C, a thermal sensor turns off the both power MOSFETs, allowing the devices to ∞ ol. The thermal sensor allows the converters to start a soft-start process and regulate the output voltage again after the junction temperature ∞ ols by 30°C. The OTP is designed with a 30°C hysteresis to lower the average Junction Temperature (T_j) during continuous thermal overload conditions, increasing the lifetime of the device.

Over-Voltage Protection (OVP)

The over-voltage function monitors the output voltage by FB pin. Once the FB voltage exceeds 125% of the reference voltage, the over-voltage protection comparator forces the low-side MOSFET on. This action actively pulls down the output voltage to prevent the end device be damage. As soon as the output voltage is below 105% of the reference voltage, the low-side MOSFET off and the OVP comparator is disengaged. The chip restores its normal operation.

Frequency Foldback

The foldback frequency is controlled by the FB voltage. When the output is short to the ground, the frequency of the oscillator will be reduced to 0.25 x F_{sw} . This lower frequencyallows the inductor current to safely discharge, there by preventing current runaway. The oscillator's frequency will gradually increase to its designed rate when the feedback voltgae on FB again approaches 0.807V.

Soft-Start

The APW7511 provides the programmed soft-start function to limit the inrush current. The soft-start time can be programmed by the external capacitor between SS and GND. Typical charge current is 11μ A, and the soft-start time can be calculated by the following formula:



Function Description

Soft-Start (cont.)

 $T_{SS} (ms) = \frac{C_{SS}(nF) \times 0.807(V)}{11(\mu A)}$

The APW7511 integrates soft-start circuits to ramp up the output voltage of the converter to the programmed regulation set point at a predictable slew rate. The slew rate of output voltage is internally controlled to limit the inrush current through the output capacitors during soft start process. When the EN pin is pulled above the rising EN threshold voltage, the device initiates a soft-start process to ramp up the output voltage.



Application Information

Input Capacitor Selection

Because buck converters have a pulsating input current, a low ESR input capacitor is required. This results in the best input voltage filtering, minimizing the interference with other circuits caused by high input voltage spikes. Also, the input capacitor must be sufficiently large to stabilize the input voltage during heavy load transients. For good inputvoltage filtering, usuallya 22μ F input capacitor is sufficient. It can be increased without any limit for better input-voltage filtering. Ceramic capacitors show better performance because of the low ESR value, and they are less sensitive against voltage transients and spikes compared to tantalum capacitors. Place the input capacitor as close as possible to the input and GND pin of the device for better performance.

Inductor Selection

For high efficiencies, the inductor should have a low DC resistance to minimize conduction losses. Especially at high-switching frequencies, the core material has a higher impact on efficiency. When using small chip inductors, the efficiency is reduced mainly due to higher inductor core losses. This needs to be considered when selecting the appropriate inductor. The inductor value determines the inductor ripple current. The larger the inductor value, the smaller the inductor ripple current and the lower the conduction losses of the converter. Conversely, larger inductor values cause a slower load transient response. A reasonable starting point for setting ripple current, ΔI_L , is 40% of maximum output current. The recommended inductor value can be calculated as below:

$$L \geq \frac{V_{OUT} \left(1 - \frac{V_{OUT}}{V_{IN}}\right)}{F_{SW} \cdot \Delta I_{L}}$$
$$I_{L(MAX)} = I_{OUT(MAX)} + \frac{1}{2} \Delta I_{L}$$

To avoid saturation of the inductor, the inductor should be rated at least for the maximum output current of the converter plus the inductor ripple current.

Output Voltage Setting

In the adjustable version, the output voltage is set by a resistive divider. The external resistive divider is connected to the output, allowing remote voltage sensing as shown in "Typical Application Circuits". The output voltage can be calculated as below:

$$V_{OUT} = V_{REF} \cdot \left(1 + \frac{R1}{R2}\right) = 0.807 \cdot \left(1 + \frac{R1}{R2}\right)$$

If VOUT setting less than VIN POR, suggest using Dual Power Input of Typical Application Circuit.

If VOUT setting more than VIN POR, suggest using Single Power Input of Typical Application Circuit.

Output Capacitor Selection

The current-mode control scheme of the APW7511 allows the use of tiny ceramic capacitors. The higher capacitor value provides the good load transients response. Ceramic capacitors with low ESR values have the lowest output voltage ripple and are recommended. If required, tantalum capacitors may be used as well. The output ripple is the sum of the voltages across the ESR and the ideal output capacitor.

$$\Delta V_{OUT} \cong \frac{V_{OUT} \cdot \left(1 - \frac{V_{OUT}}{V_{IN}}\right)}{F_{SW} \cdot L} \cdot \left(ESR + \frac{1}{8 \cdot F_{SW} \cdot C_{OUT}}\right)$$

When choosing the input and output ceramic capacitors, choose the X5R or X7R dielectric form ulations. These dielectrics have the best temperature and voltage characteristics of all the ceramics for a given value and size.

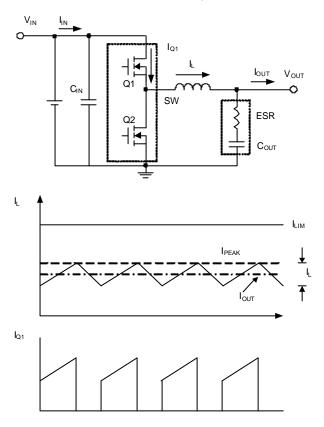
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Application Information

Output Capacitor Selection

When choosing the input and output ceramic capacitors, choose the X5R or X7R dielectric formulations. These dielectrics have the best temperature and voltage characteristics of all the ceramics for a given value and size.



Layout Consideration

For all switching power supplies, the layout is an important step in the design; especially at high peak currents and switching frequencies. If the layout is not carefully done, the regulator might show noise problems and duty cycle jitter.

1. The input capacitor should be placed close to the VIN and GND. Connecting the capacitor and VIN/GND with short and wide trace without any via holes for good input voltage filtering. The distance between VIN/GND to capacitor less than 2mm respectively is recommended.

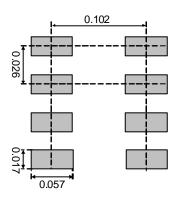
2. To minimize copper trace connections that can inject noise into the system, the inductor should be placed as close as possible to the SW pin to minimize the noise coupling into other circuits.

Copyright © ANPEC Electronics Corp. Rev. A.3 - Oct., 2015 3. The output capacitor should be place closed to converter VOUT and GND.

4. Since the feedback pin and network is a high impedance circuit the feedback network should be routed away from the inductor. The feedback pin and feedback network should be shielded with a ground plane or trace to minimize noise coupling into this circuit.

5. A star ground connection or ground plane minimizes ground shifts and noise is recommended.

Recommended Minimum Footprint



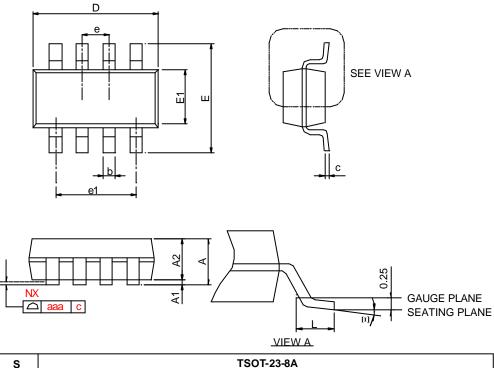
TSOT-23-8A

APW7511



Package Information

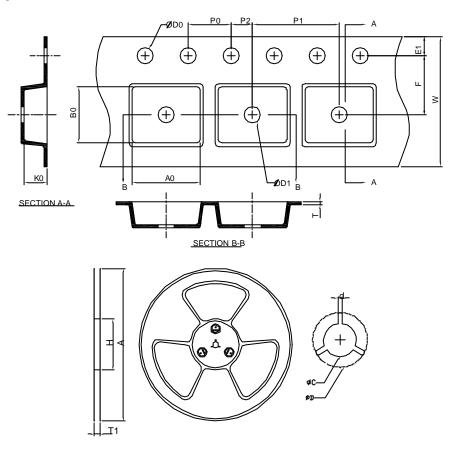
TSOT-23-8A



| Ş | | TSOT | T-23-8A | | |
|-----------------|-------------|------|---------|-------|--|
| S≻ <u>M</u> BOL | MILLIMETERS | | INCI | HES | |
| l C | MIN. | MAX. | MIN. | MAX. | |
| А | 0.70 | 1.00 | 0.028 | 0.039 | |
| A1 | 0.01 | 0.10 | 0.000 | 0.004 | |
| A2 | 0.70 | 0.90 | 0.028 | 0.035 | |
| b | 0.22 | 0.40 | 0.009 | 0.016 | |
| с | 0.08 | 0.20 | 0.003 | 0.008 | |
| D | 2.70 | 310 | 0.106 | 0.122 | |
| E | 2.60 | 3.00 | 0.102 | 0.118 | |
| E1 | 1.40 | 1.80 | 0.055 | 0.071 | |
| е | 0.65 | BSC | 0.026 | BSC | |
| e1 | 1.95 | BSC | 0.077 | BSC | |
| L | 0.30 | 0.60 | 0.012 | 0.024 | |
| Ð | 0° | 8° | 0° | 8° | |
| aaa | 0 | .10 | 0.0 | 004 | |



Carrier Tape & Reel Dimensions



| Application | Α | Н | T1 | С | d | D | W | E1 | F |
|-------------|-------------------|-------------------|-------------------|--------------------|----------|-------------------|------------|--------------------|------------|
| | 178.0 ±2.00 | 50 MIN. | 8.4+2.00 -0.00 | 13.0+0.50 -0.20 | 1.5 MIN. | 20.2 MIN. | 8.0 ±0.30 | 1.75 ±0.10 | 3.5 ±0.05 |
| TSOT-23-8A | P0 | P1 | P2 | D0 | D1 | Т | A0 | B0 | K0 |
| | 4.0 ± 0.10 | 4.0 ± 0.10 | 2.0 ±0.05 | 1.5+0.10 -0.00 | 1.0 MIN. | 0.6+0.00 -0.40 | 3.20 ±0.20 | 3.10 ± 0.20 | 1.20 ±0.20 |

(mm)

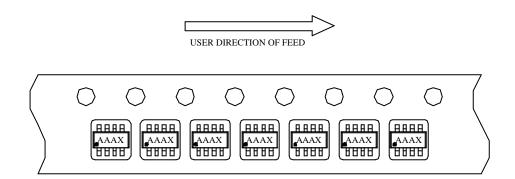
Devices Per Unit

| Package Type | Unit | Quantity |
|--------------|-------------|----------|
| TSOT-23-8A | Tape & Reel | 3000 |

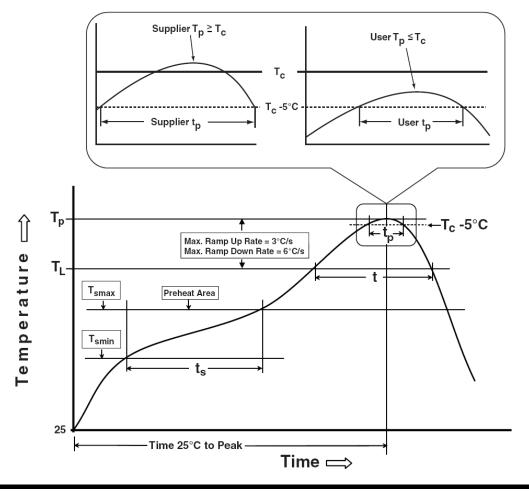


Taping Direction Information

TSOT-23-8A



Classification Profile



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Classification Reflow Profiles

| Profile Feature | Sn-Pb Eutectic Assembly | Pb-Free Assembly |
|---|---|------------------------------------|
| Preheat & Soak Temperature min (T _{smin}) Temperature max (T _{smax}) Time (T _{smin} to T _{smax}) (t _s) | 100 °C 150 °C 60-120 seconds | 150 °C 200 °C 60-120 seconds |
| Average ramp-up rate (T _{smax} to T _P) | 3 °C/second max. | 3°C/second max. |
| Liquidous temperature (T∟) Time at liquidous (t∟) | 183 °C 60-150 seconds | 217 °C 60-150 seconds |
| Peak package body Temperature (T _P)* | See Classification Temp in table 1 | See Classification Temp in table 2 |
| Time $(t_P)^{**}$ within 5°C of the specified classification temperature (T_c) | 20** seconds | 30** seconds |
| Average ramp-down rate $(T_p to T_{smax})$ | 6 °C/second max. | 6 °C/second max. |
| Time 25°C to peak temperature | 6 minutes max. | 8 minutes max. |
| * Tolerance for peak profile Temperat | ure (T_p) is defined as a supplier minimu | m and a user maximum. |

** Tolerance for time at peak profile temperature (t_p) is defined as a supplier minimum and a user maximum.

Table 1. SnPb Eutectic Process – Classification Temperatures (Tc)

| Package | Volume mm ³ | Volume mm ³ |
|-----------|------------------------|------------------------|
| Thickness | <350 | ³ 350 |
| <2.5 mm | 235 °C | 220 °C |
| ≥2.5 mm | 220 °C | 220 °C |

Table 2. Pb-free Process – Classification Temperatures (Tc)

| Package Thickness | Volume mm ³ <350 | Volume mm ³ 350-2000 | Volume mm ³ >2000 |
|----------------------|--------------------------------|------------------------------------|---------------------------------|
| <1.6 mm | 260 °C | 260 °C | 260 °C |
| 1.6 mm – 2.5 mm | 260 °C | 250 °C | 245 °C |
| ≥2.5 mm | 250 °C | 245 °C | 245 °C |

Reliability Test Program

| Test item | Method | Description |
|---------------|--------------------|------------------------------|
| SOLDERABILITY | JESD-22, B102 | 5 Sec, 245°C |
| HOLT | JESD-22, A108 | 1000 Hrs, Bias @ Tj=125°C |
| PCT | JESD-22, A102 | 168 Hrs, 100%RH, 2atm, 121°C |
| ТСТ | JESD-22, A104 | 500 Cycles, -65°C~150°C |
| НВМ | MIL-STD-883-3015.7 | VHBM 2KV |
| MM | JESD-22, A115 | VMM 200V |
| Latch-Up | JESD 78 | 10ms, 1 ¹ 100mA |



Customer Service

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