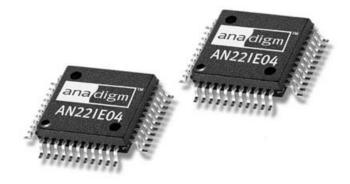


## **AN221E04** Datasheet

## Dynamically Reconfigurable FPAA With Enhanced I/O



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#### PRODUCT AND ARCHITECTURE OVERVIEW

The AN221E04 device consists of a 2x2 matrix of fully Configurable Analog Blocks (CABs), surrounded by a fabric of programmable interconnect resources. Configuration data is stored in an on-chip SRAM configuration memory. Compared with the first-generation FPAAs, the Anadigmvortex architecture provides a significantly improved signal-to-noise ratio as well as higher bandwidth. These devices also accommodate nonlinear functions such as sensor response linearization and arbitrary waveform synthesis.

The AN221E04 device features an advanced input/output structure that allows the FPAA to be programmed with up to six outputs – or triple the number provided by the ANx20E04 devices. The AN221E04 devices have four configurable I/O cells and two dedicated output cells. For I/O-intensive applications, this means a single FPAA can now be used to process multiple channels of analog signals where two or more such devices were previously needed.

In addition, the AN221E04 devices allow designers to implement an integrated 8-bit analog-to-digital converter on the FPAA, eliminating the potential need for an external converter. Using this new device, designers can route the digital output of the A/D converter off-chip using one of the dedicated output cells.

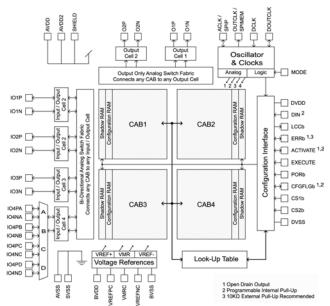


Figure 1: Architectural overview of the AN221E04 device

With dynamic reconfigurability, the functionality of the AN221E04 can be reconfigured in-system by the designer or on-the-fly by a microprocessor. A single AN221E04 can thus be programmed to implement multiple analog functions and/or to adapt on-the-fly to maintain precision operation despite system degradation and aging.

#### **PRODUCT FEATURES**

- Dynamic reconfiguration
- Four configurable I/O cells, two dedicated output cells
- 8-bit SAR analog-to-digital converter
- Fully differential architecture
- Fully differential I/O buffering with options for single ended to differential conversion
- Low input offset through chopper stabilized amplifiers
- 256 Byte Look-Up Table (LUT) for linearization and

www.DataSheet41.com arbitrary signal generation

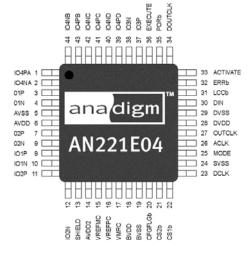
- 4:1 Input multiplexer
- Typical Signal Bandwidth: DC-2MHz (Bandwidth is CAM dependent)
- Signal to Noise Ratio:
  - Broadband 80dB
  - Narrowband (audio) 100dB
- Total Harmonic Distortion (THD): 80dB
- DC offset <100µV</li>
- Package: 44-pin QFP (10x10x2mm)
  - Lead pitch 0.8mm
- Supply voltage: 5V

### **ORDERING CODES**

AN221E04-QFPSP	Dynamically reconfigurable FPAA
AN221E04-QFPTY	Sample Pack Dynamically reconfigurable FPAA Tray (96 pcs)
AN221E04-QFPTR	Dynamically reconfigurable FPAA Tape & Reel (1000 pcs)
AN221D04-EVAL AN221D04-DEVLP	AN221E04 Evaluation Kit AN221E04 Development Kit

#### **APPLICATIONS**

- Real-time software control of analog system peripherals
- Intelligent sensors
- Adaptive filtering and control
- Adaptive DSP front-end
- Adaptive industrial control and automation
- Self-calibrating systems
- Compensation for aging of system components
- Dynamic recalibration of remote systems
- Ultra-low frequency signal conditioning
- Custom analog signal processing



[For more detailed information on the features of the AN221E04 device, please refer to the AN121E04/AN221E04 User Manual]

## ELECTRICAL CHARACTERISTICS

#### Absolute Maximum Ratings

Parameter	Symbol	Min	Тур	Max	Unit	Comment
DC Power Supplies	AVDD(2) BVDD DVDD	-0.5	-	5.5 V	V	AVSS, BVSS, DVSS and SVSS all held to 0.0 V $^{\rm a}$
xVDD to xVDD Offset		-0.5		0.5	V	Ideally all supplies should be at the same voltage
Package Power Dissipation	Pmax 25°C Pmax 85°C	-	-	1.8 0.73	W	Still air, No heatsink, 4 layer board, 44 pins. θja = 55°C/W
Analog and Digital Input Voltage	Vinmax	Vss-0.5	-	Vdd+0.5	V	
Ambient Operating Temperature	Тор	-40	-	85	°C	
Storage Temperature	Tstg	-65		150	С°	

<sup>a</sup> Absolute Maximum DC Power Supply Rating - The failure mode is non-catastrophic for Vdd of up to 7 volts, but will cause reduced operating life time. The additional stress caused by higher local electric fields within the CMOS circuitry may induce metal migration, oxide leakage and other time/quality related issues.

#### **Recommended Operating Conditions**

Parameter	Symbol	Min	Тур	Max	Unit	Comment
DC Power Supplies	AVDD(2) BVDD DVDD	4.75	5.00	5.25	V	AVSS, BVSS, DVSS and SVSS all held to 0 V
Analog Input Voltage.	Vina	VMR-1.9	-	VMR+1.9	V	VMR is 2.0 volts above AVSS
Digital Input Voltage	Vind	0	-	DVDD	V	
Junction Temp	Tj	-40	-	125	°C	Assume a package θja = 55°C/W <sup>b</sup>

<sup>b</sup> In order to calculate the junction temperature you must first empirically determine the current draw (total ldd) for the design. Once the current consumption established then the following formula can be used; Tj = Ta + Idd x Vdd x 55 °C/W, where Ta is the ambient temperature. The worst case θja of 55 °C/W assumes no air flow and no additional heatsink of any type.

#### General Digital I/O Characteristics (Vdd = 5v +/- 10%, -40 to 85 deg.C)

-	•			-	•	
Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Voltage Low	Vih	0	-	30	-	% of DVDD
Input Voltage High	Vil	70	-	100	-	% of DVDD
ww. <u>Coutput</u> VoltageLow	Vol	0	-	20	-	% of DVDD
Output Voltage High	Voh	80	-	100	-	% of DVDD
Input Leakage Current	lil	-	-	±1.0	μA	All pins except DCLK
Input Leakage Current	lil		±12.0	_	μA	DCLK if a crystal is connected and
		_	12.0	_	μΛ	the on-chip oscillator is used
Max. Capacitive Load	Cmax	-	_	10	pF	The maximum load for a digital
	Onlax			10	P	output is 10 pF // 10 Kohm
Min. Resistive Load	Rmin	10	-	_	Kohm	The maximum load for a digital
		10				output is 10 pF // 10 Kohm
DCLK Frequency	Fmax	_	_	40	MHz	For MODE = 1, Max DCLK is
	Тпах			40		16 MHz
ACLK Frequency	Fmax	_	_	40	MHz	Divide down to <8 MHz prior to use
	i max	_		40		as a CAB clock
Clock Duty Cycle	-	45	-	55	%	All clocks

## Detailed Digital I/O Interface Characteristics: Vdd = 5.0volts

### LCCb

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Output Voltage Low	Vol	Vss	-	150	mV	Load 20pF//50Kohm to Vss
Output Voltage High	Voh	4.5	-	Vdd	V	Load 20pF//50Kohm to Vss
Max. Capacitive Load	Cmax	-	-	20	pF	Maximum load 20 pF // 50 Kohm
Min. Resistive Load	Rmin	50	-	-	Kohm	Maximum load 20 pF // 50 Kohm
Current Sink	Isnkmax	-	-	15	mA	LCCb pin shorted to Vdd
Current Source	Isrcmax	-	-	4	mA	LCCb pin shorted to Vss

### CFGFLG, ACTIVATE

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Voltage Low	Vil	0		30	%	% of DVDD
Input Voltage High	Vih	70		100	%	% of DVDD
Output Voltage Low	Vol	Vss	-	85	mV	Pin load = Internal pullup + 20pF//50K to Vss
Output Voltage High	Voh	4.5	-	Vdd	V	Pin load = Internal pullup + 20pF//50K to Vss
Output Voltage Low	Vol	Vss	-	200	mV	Pin Load = External 5K ohm pullup + 20pF//50K to Vss
Output Voltage High	Voh	4.5	-	Vdd	V	Pin Load = External 5Kohm pullup + 20pF//50K to Vss
Max. Capacitive Load	Cmax	-	-	50	pF	Maximum load 50 pF // 50 Kohm
Min. Resistive Load	Rmin	50	-	-	Kohm	Maximum load 50 pF // 50 Kohm
Current Sink	Isnkmax	-	-	2.5	mA	Pin shorted to Vdd
Current Source	Isrcmax	-	-	200	μA	Pin shorted to Vss
External Resistive Pullup	Rpullupext	5	7.5	10	Kohm	Use only if internal pullup is deselected

#### ERRb

wParameter U.com	Symbol	Min	Тур	Max	Unit	Comment
Input Voltage Low	Vil	0		30	%	% of DVDD
Input Voltage High	Vih	70		100	%	% of DVDD
Output Voltage Low	Vol	Vss	-	50	mV	
Output Voltage High	Voh	4.9	-	Vdd	V	
Max. Capacitive Load	Cmax	-	-	50	pF	Maximum load 50 pF // 50 Kohm
Min. Resistive Load	Rmin	50	-	-	Kohm	Maximum load 50 pF // 50 Kohm
Current Sink	Isnkmax	-	-	10	mA	
Current Source	Isrcmax	-	-	0	μA	
External Resistive Pullup	Rpullupext	10	10	10	Kohm	

### DCLK,Mode,DIN,EXECUTE,PORb,CS1b,CS2b

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Voltage Low	Vil	0	-	30	%	% of DVDD
Input Voltage High	Vih	70	-	100	%	% of DVDD

## OUTCLK/SPIMEM,DOUTCLK

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Output Voltage Low	Vol	0	-	20	%	% of DVDD
Output Voltage High	Voh	80	-	100	%	% of DVDD
Max. Capacitive Load	Cmax	-	-	50	pF	Maximum load 50 pF // 50 Kohm
Min. Resistive Load	Rmin	10	-	-	Kohm	Maximum load 50 pF // 50 Kohm
Current Sink	Isnkmax	-	-	17	mA	
Current Source	Isrcmax	-	-	4	mA	

#### ACLK/SPIP

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Voltage Low	Vil	0	-	30	%	% of DVDD
Input Voltage High	Vih	70	-	100	%	% of DVDD
Output Voltage Low	Vol	0	-	20	%	% of DVDD
Output Voltage High	Voh	80	-	100	%	% of DVDD
Max. Capacitive Load	Cmax	-	-	50	pF	Maximum load 50 pF // 50 Kohm
Min. Resistive Load	Rmin	10	-	-	Kohm	Maximum load 50 pF // 50 Kohm
Current Sink	Isnkmax	-	-	15	mA	
Current Source	Isrcmax	-	-	4	mA	

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## AN221E04 Datasheet – Dynamically Reconfigurable FPAA With Enhanced 1/O **Analog Inputs General**

Parameter	Symbol	Min	Тур	Max	Unit	Comment
High Precision Input Range <sup>c</sup>	Vina	0.5	-	3.5	V	VMR +/- 1.5v
Standard precision Input Range <sup>d</sup>	Vina	0.1	-	3.9	V	VMR +/- 1.9v
High Precision Differential Input <sup>c</sup>	Vdiffina	0	-	+/-3.0	V	Common mode voltage = 2 V
Standard Precision Differential Input <sup>d</sup>	Vdiffina	0	-	+/-3.8	V	Common mode voltage = 2 V
Common Mode Input Range	Vcm	1.8	2.0	2.2	V	
Input Offset	Vos	-	5	15	mV	Non-chopper stabilized input
Input Frequency	Fain	0	<2	8	MHz	Max value is clock, CAM and input stage dependant. Input frequency is limited to approx <2MHz due to CAM signal processing which is based on sampled data architectures.

<sup>c.</sup> High precision operating range provides optimal linearity and dynamic range. <sup>d.</sup> Standard precision operating range provides maximum dynamic range and reduced linearity.

#### Input Differential Amplifier ON and filter OFF

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See ar	nalog inpu	t above		Usable input range will be reduced by the effective gain setting
Gain Setting	Ginamp	16	-	128		
Gain Accuracy		-	1.0	2.5	%	
Gain Drift (Temperature, Supply Voltage zand Time)	Dist	-	-	1.0	%	
Equivalent Input Offset Voltage	Vos	-	3	12	mV	Non-chopper stabilized input When the input amplifier and filter are used in combination Vos contribution comes only from the input amplifier
Offset Voltage Temperature Coefficient	Voffsettc	-	1	10	μV/°C	from -40°C to 125°C
Input Frequency <sup>c</sup>	Fain	0	-	2	MHz	
Input Frequency <sup>d</sup>	Fain	0	<2	8	MHz	
Power Supply Rejection Ratio	PSRR	65	-	-	dB	d.c. Amp Gain =16 a.c. See graphs page 18
Common Mode Rejection Ratio	CMRR	-	67	-	dB	
Large Signal Harmonic Distortion	Dist	-	-65	-	dB	0.4v p-p Differential input at 660Hz Gain setting = 16
Input Resistance	Rin	10		-	Mohm	
Input Capacitance	Cin	-		5.0	pF	
Input Referred Noise Figure	NF	-	0.1	-	µV/sqrtHz	Input cell Gain = 16 Applies to audio frequency range (400Hz to 30KHz). See graphical data on page 18
Signal-to Noise Ratio and Distortion	SINAD	-	75	-	dB	Input signal = 285 mV p-p diff, audio frequency range See graphical data on page 18
Spurious Free Dynamic Range	SFDR	-	73	-	dB	Input signal = 100 mV p-p diff See graphical data on page 18

<sup>c.</sup> High precision operating range provides optimal linearity and dynamic range.
 <sup>d.</sup> Standard precision operating range provides maximum dynamic range and reduced linearity.

## Input Differential Chopper Amplifier on and filter OFF

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See ana	alog input ab	ove		Usable input range will be reduced by the effective gain setting
Gain Setting	Ginamp	16	-	128		
Gain Accuracy		-	1.0	2.5	%	
Gain Drift, (Temperature, Supply Voltage and Time)		-	-	1.0	%	
Chopper Frequency Clock Range	Fch	Fc/260100	-	>250	KHz	Fc = master clock frequency Set Fch as slow as possible Fch > 250KHz will result in som signal attenuation
Equivalent Input Offset Voltage	Vos	-	<100	200	μV	Chopper stabilized amplifier The maximum value of 200µV is guaranteed by production test This is a tester limitation
Offset Voltage Temperature Coefficient	Voffsettc	-	0.5	2.0	µV/°C	from -40°C to 125°C
Power Supply Rejection Ratio	PSRR	65	-	-	dB	d.c. a.c. See graphs on page 18
Common Mode Rejection Ratio	CMRR	-	102	-	dB	
Large Signal Harmonic Distortion	Dist	-	-40	-	dB	0.4v p-p Differential input at 660Hz Gain setting = 16
Input Frequency	Fain	0	Fch/20	Fch/2	KHz	Fch=Chopper clock frequency The chopper frequency and input frequency should be chosen such that subsequent low pass filtering can remove th chopper stage frequency elements
Input Resistance	Rin	10		-	Mohm	Input to filter or chopper
Input Capacitance	Cin	-		5.0	pF	
Input Referred Noise Figure	NF	_	0.09	-	µV/sqrtHz	Input cell Gain = 16 Applies to Audio frequency range Chopper clock Fch = 250KHz See graphical data on page 18
Signal-to Noise Ratio and Distortion v.DataSheet4U.com	SINAD	-	75	-	dB	Input signal = 285 mV p-p differential, Audio frequency range See graphical data on page 18
Spurious Free Dynamic Range	SFDR	-	74	-	dB	Input signal =100 mV p-p differential See graphical data on page 18

## Input Differential Amplifier OFF and filter ON

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See ar	nalog input	above		
Equivalent Input Offset	Vos	-	8	32	mV	Non-chopper stabilized input, Filter corner frequency =470KHz
Offset Voltage Temperature Coefficient	Voffsettc	-	0.05 <sup>1</sup>	1.0 <sup> "</sup>	mV/°C	from -40°C to 125°C I. measured at filter corner=470Khz II. maximum at Filter corner=76KHz
Input Frequency	Fain	-	-	-	MHz	Input filter frequency will define the maximum frequency Input filter is recommended to be >30x higher than the max input frequency, for 80dB distortion performance

Common Mode Rejection Ration	CMRR	-	60	-	dB	
Power Supply Rejection Ratio	PSRR	68	-	-	dB	d.c. a.c. See graphical data on page 19
Large Signal Harmonic Distortion	Dist	-	-82	-	dB	4v p-p Differential input at 660Hz Filter corner frequency 470KHz
Input Low Pass Filter (Anti-Alias) Corner Frequency Settings	Ffiltcorner	76	-	470	KHz	
Input Resistance	Rin	10	-	-	Mohm	Input to filter or chopper
Input Capacitance	Cin	-		5.0	pF	
Input Referred Noise Figure	NF	-	0.17	-	µV/sqrtHz	Input cell filter corner Fc = 470KHz Applies to Audio frequency range See graphical data on page 18
Signal-To Noise Ratio and Distortion	SINAD	-	84	-	dB	Input signal = 1400 mV p-p diff, Audio frequency range See graphical data on page 18
Spurious Free Dynamic Range	SFDR	-	90	-	dB	Input signal =1400 mV p-p differential See graphical data on page 18

## Input Differential Voltage mode, Amplifier OFF, Filter OFF and Unity Gain stage ON

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See an	alog inpu	t above	V	
Equivalent Input Offset	Vos	-	5	15	mV	Non-chopper stabilized input
Offset Voltage Temperature Coefficient	Voffsettc	-	20	50	μV/°C	from -40°C to 125°C
Input Frequency	Fain	-	-	1.0	MHz	Gain Bandwidth limited by input impedance
Power Supply Rejection Ratio	PSRR	60	-	-	dB	d.c. a.c. See graphs on page 18
Common Mode Rejection Ratio	CMRR	-	60	-	dB	
Large Signal Harmonic Distortion	Dist	-	-80	-	dB	4v p-p Differential input at 660Hz
Large Signal Harmonic Distortion	Dist	-	-80	-	dB	3v p-p single ended signal at 660Hz
Input Resistance	Rin	-	126	-	Kohm	Input to unity gain stage
Input Capacitance	Cin	-	2.0	5.0	pF	
Input Referred Noise Figure	NF	-	0.16	-	µV/sqrtHz	Applies to Audio frequency range See graphical data on page 18
Signal-To Noise Ratio and Distortion	SINAD	-	84	-	dB	Input signal = 1400 mV p-p diff, Audio frequency range See graphical data on page 18
Spurious Free Dynamic Range	SFDR	-	90	-	dB	Input signal =1400 mV p-p differenti See graphical data on page 18

### Input Differential Voltage mode, Amplifier OFF, Filter OFF and Unity Gain stage OFF

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See an	alog inpu	t above	V	
Equivalent Input Offset	Vos	N/A	N/A	N/A	mV	See CAM Op Amp
Offset Voltage Temperature Coefficient	Voffsettc	N/A	N/A	N/A	μV/°C	See CAM Op Amp. from -40°C to 125°C
Input Frequency	Fain	-	-	8	MHz	Dependant upon CAM
Power Supply Rejection Ratio	PSRR	N/A	N/A	N/A	dB	See CAM Op Amp
Large Signal Harmonic Distortion	Dist	-	-85	-	dB	See CAM Op Amp
Input Resistance	Rin	-	-	-	Mohm	Input to CAM directly (Input cell bypass mode). This variable is influenced by CAB capacitor size, CAB clock frequency and CAB architecture
Input Capacitance	Cin	-	-	-	pF	Input to CAM directly (Input cell bypass mode) This variable is influenced by CAB capacitor size, CAB clock frequency and CAB architecture

#### **Analog Outputs**

(See "Output Cell" section in the AN120E04/AN220E04 user manual for more details)

Parameter	Symbol	Min	Тур	Max	Unit	Comment
High Precision Output Range <sup>c</sup>	Vouta	0.5	-	3.5	V	VMR +/- 1.5v
Standard Precision Output Range <sup>d</sup>	Vouta	0.1	-	3.9	V	VMR +/- 1.9v
High Precision Differential Output <sup>c</sup>	Vdiffouta	-	-	+/-3.0	V	Common mode voltage = 2 V
Standard precision Differential Output <sup>d</sup>	Vdiffouta	-	-	+/-3.8	V	Common mode voltage = 2 V
Common Mode Voltage	Vcm	1.9	2.0	2.1	V	

<sup>c</sup>. High precision operating range provides optimal linearity and dynamic range. <sup>d</sup>. Standard precision operating range provides maximum dynamic range and reduced linearity.

#### Output Voltage mode and filter ON, corner frequency 470KHz

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See ana	alog input	above	V	
Equivalent Input Offset	Vos	-	5	15	mV	
Offset Voltage Temperature Coefficient	Voffsettc		0.05	1.0 "	mV/°C	from -40°C to 125°C <sup>I</sup> measured at filter corner: 470Khz <sup>II</sup> maximum at filter corner: 76KHz
Output Frequency	Faout	-	-	-	MHz	Output filter frequency will define the maximum frequency Input filter is recommended to be >30x higher then the max input frequency, for good distortion performance
Power Supply Rejection Ratio	PSRR	60	-	-	dB	d.c. a.c. See graphical data on page 19
Large Signal Harmonic Distortion	Dist	-	-82	-	dB	4v p-p Differential input at 660Hz Filter corner frequency 470KHz
Input Low Pass Filter (Anti-Alias) Corner Frequency Settings	Ffiltcorner	76	-	470	KHz	
Output Load <sup>ce</sup>	Rload	0.1	-	-	Mohm	
Output Load <sup>ce</sup>	Cload	-	-	50	pF	
Output Load <sup>de</sup> w.DataSheet4U.com	Rload	1	10	-	Kohm	Additional loading causes internal voltage drops across output stage and series resistances The output stage has a small signal output impedance of approx 10ohm
Output Load <sup>d e</sup>	Cload	-	-	100	pF	
Common Mode Rejection Ratio	CMRR	-	56	-	dB	
Input Referred Noise Figure	NF	-	0.22	-	µV/sqrtHz	Output filter corner fc = 470KHz Applies to Audio frequency range See graphical data on page 18
Signal-To Noise Ratio and Distortion	SINAD	-	82	-	dB	Input signal = 1400 mV p-p diff, Audio frequency range See graphical data on page 18
Spurious Free Dynamic Range	SFDR	-	90	-	dB	Input signal =1400 mV p-p diff See graphical data on page 18

<sup>c</sup>. High precision operating range provides optimal linearity and dynamic range. <sup>d</sup>. Standard precision operating range provides maximum dynamic range and reduced linearity.

<sup>e</sup>. The maximum load for an analog output is 50 pF // 100 Kohms. This load maybe with respect to analog ground VMR or AVSS.

#### Output Voltage mode and filter off (bypass mode)

Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range	Vina Vdiffina	See a	See analog input above			
Equivalent Input Offset	Vos	N/A	N/A	N/A	mV	See CAM Op Amp
Offset Voltage Temperature Coefficient	Voffsettc	N/A	N/A	N/A	mV/°C	See CAM Op Amp
Output Frequency ce	Faout	-	-	4	MHz	
Output Frequency <sup>d†</sup>	Faout	-	-	8	MHz	The realizable output frequency is limited to approx <2MHz due to CAM signal processing which is based on sampled data architectures.
Power Supply Rejection Ratio	PSRR	N/A	N/A	N/A	dB	See CAM Op Amp
Large Signal Harmonic Distortion	Dist	-	-85	-	dB	
Output Load	Rload	N/A	N/A	N/A	Mohm	See CAM Op Amp
Output Load	Cload	N/A	N/A	N/A	pF	See CAM Op Amp

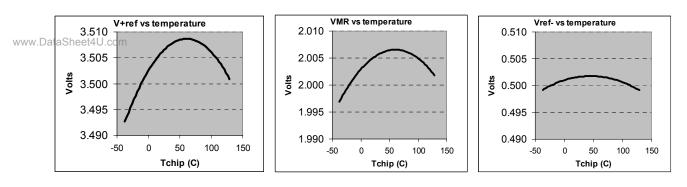
Standard precision operating range provides maximum dynamic range and reduced linearity.

e. The maximum load for an analog output is 50 pF // 100 Kohms. This load maybe with respect to analog ground VMR or AVSS.

f. The maximum load for an analog output is 100 pF // 100 Kohms. This load must be differential and with respect to analog ground(VMR).

#### VMR (voltage Mid Rail) and VREF (Reference Voltage) Ratings

Parameter	Symbol	Min	Тур	Max	Unit	Comment
VMR Output Voltage	Vvmr	1.925	2.01	2.075	V	At 25°C, Vdd=5.00 volts
VREF+ Output Voltage	Vref+	3.4	3.51	3.6	V	At 25°C, Vdd=5.00 volts
VREF- Output Voltage	Vref-	0.45	0.505	0.55	V	At 25°C, Vdd=5.00 volts
Output Voltage Deviation VREF+, VMR, VREF-	Vrefout	-	0.5	1	%	Over process and supply voltage corners
Voltage Temperature Coefficient VREF+, VMR, VREF-	Vreftc	-	-	-	-	See typical graphical data below -40°C to 125°C <sup>f</sup>
Power Supply Rejection Ratio, VMR	PSSR	60	-	-	dB	
Power Supply Rejection Ratio Vref+ and Vref-	PSSR	75	-	-	dB	
Start Up Time	Tstart	-	-	1	ms	Assuming recommended capacitors



## CAB (Configurable Analog Block) Differential Operational Amplifier

Parameter	Symbol	Min	Тур	Max	Unit	Comment
High Precision Input/Output Range <sup>c</sup>	Vinouta	0.5	-	3.5	V	VMR +/- 1.5v
Standard Precision Input/Output Range <sup>d</sup>	Vinouta	0.1	-	3.9	V	VMR +/-1.9v
High Precision. Differential Input/Output <sup>c</sup>	Vdiffioa	-	-	+/-3.0	V	Common mode voltage = 2 V
Standard Precision Differential Input/Output <sup>d</sup>	Vdiffioa	-	-	+/-3.8	V	Common mode voltage = 2 V
Common Mode Input Voltage Range <sup>d</sup>	Vcm	0	2.0	4	V	
Common Mode Output Voltage Range	Vcm	1.9	2.0	2.1	V	
Equivalent Input Voltage Offset.	Voffset	0.1	5	15	mV	Some CAMs (Configurable Analog Modules) can inherent compensate
Offset Voltage Temperature Coefficient	Voffsettc	-	1	10	μV/°C	from -40°C to 125°C some CAMs (Configurable Analog Modules) can inherently compensate
Power Supply Rejection Ratio	PSSR	-	80	-	dB	Variation between CAMs is expected because of variation in architecture
Common Mode Rejection Ratio	CMRR	-	77	-	dB	Example 1 GainInv CAM CAM clock = 1MHz CAM parameter settings Gain = 1
Common Mode Rejection Ratio	CMRR	-	60	-	dB	Example 2 Filterbiquad Setting = Low pass filter CAM clock = 1MHz CAM parameter settings Gain = 1, Corner frequency = 50KHz Quality Factor = 0.707
Differential Slew Rate, Internal	Slew	-	50	-	V/µsec	Applicable when the OpAmp load is internal to the FPAA
Differential Slew Rate, External	Slew	-	10	-	V/µsec	Applicable when the OpAmp driving signal out of the FPAA package
Unity Gain Bandwidth, Full Power Mode.	UGB	-	50	-	MHz	Applicable when sourcing and loading the OpAmp with a load internal to the FPAA
Input Impedance, Internal	Rin	10	-	-	Mohm	
Output Impedance, Internal	Rout	-	-	-	Ohms	The OpAmp output is designe to drive all internal nodes, the are dominantly capacitive load
Output Impedance, External	Rout	-	-	-	Ohms	Output to an FPAA output pin (ouput cell bypass mode). This variable is influenced by CAB capacitor size, CAB clock frequency and CAB architectu
Output Load, External <sup>c e</sup>	Rload	0.1	-		Mohm	
Output Load, External ce	Cload	-	-	50	pF	
Output Load, External <sup>def</sup>	Rload	1	10	-	Kohm	Additional loading causes internal voltage drops across output stage and series resistances The output stage has a small signal output impedance of approx 100hm
Output Load, External <sup>def</sup>	Cload	-	-	50	pF	
Noise Figure <sup>9</sup>	Noise	-	0.13	-	μV/sqrtHz	Example1 <b>GainInv CAM</b> CAM clock = 1MHz Gain = 1

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Signal-To Noise Ratio and Distortion <sup>g</sup>	SINAD	-	80	-	dB	Input signal=1400 mV p-p differential Audio frequency range Example. <b>GainInv CAM</b> CAM clock = 1MHz Gain = 1
Spurious Free Dynamic Range <sup>9</sup>	SFDR	-	92	-	dB	Input signal=1400 mV p-p differential, Audio frequency range Example. <b>GainInv CAM</b> CAM clock = 1MHz Gain = 1

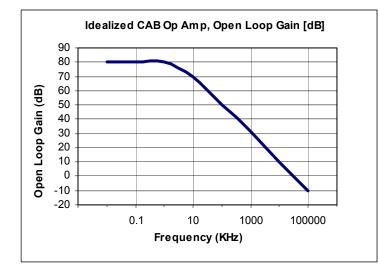
<sup>c</sup>. High precision operating range provides optimal linearity and dynamic range.

<sup>d</sup>. Standard precision operating range provides maximum dynamic range and reduced linearity.

<sup>e</sup>. The maximum load for an analog output is 50 pF || 100 Kohms. This load may be with respect to analog ground VMR or AVSS.

<sup>f</sup>. Using the FPAA with CAB Op Amp's driving directly off-chip, requires care, full characterization of the performance of each application circuit by the circuit designer is necessary.

<sup>g</sup>. This specification parameter can only be characterized when a circuit topology is configured onto the CAB differential amplifier architecture. The figure provided here is an representative on the performance of one specific CAM, as specified in the comments.



The idealized open loop gain plot is provided for information only. This information is associated with the FPAA in full power mode of operation. The FPAA operation amplifier open loop gain cannot be observed nor used when associated with external connections to the device. Internal reprogrammable routing impedances and switched capacitor circuit architecture using this operational amplifier limit the effective usable bandwidth of a circuit realized in the FPAA to less than 2MHz.

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Parameter	Symbol	Min	Тур	Max	Unit	Comment
Input Range, Internal	Vina	0.1	_	3.9	V	
Input Range, External	Vina	0.0	-	Vdd	V	
Differential Input, Internal	Vdiffina	-	-	+/-3.8	V	Common mode voltage = 2 V
Differential Input, External	Vdiffina	+/- 0.0	-	+/- Vdd	V	<b>U</b>
Common Mode Output Voltage Range, Internal <sup>c</sup>	Vcm	1.9	2.0	2.1	V	
Common Mode Input Voltage Range, External <sup>c</sup>	Vcm	0	2.0	4	V	
Common Mode Input Voltage, External <sup>d</sup>	Vcm	0	-	5	V	The comparator will function correctly
Differential Output	Voutdiff	-	-	+/-5	V	
Single Pin Output (Ox1P)	Vout	0	-	5	V	
Input Voltage Offset	Voffcomp	-	2	10	mV	Zero hysterisis
Offset Voltage Temperature Coefficient	Voffsettc	-	1	10	µV/°C	from -40°C to 125°C, Zero Hysterisis
Setup Time, Internal	Tsetint	-	-	125	nsec	
Setup Time, External	Tsetext	-	-	500	nsec	
Delay Time	Tdelay	½Td+25	-	1½Td+25	nsec	Td = 1/Fc Fc = master clock frequency
Output Load	Rload	10	-	-	Kohm	Applies if comparator drive off chip with output cell in bypass mode
Output Load	Cload	-	-	50	pF	Applies if comparator drive off chip with output cell in bypass mode
Differential Variable Reference Voltage Settings	CompVref	0	-	+/-4.0	V	
Differential Hysteresis	Hysta1	-	Voffcomp	-	mV	Hysteresis setting = zero
Differential Hysteresis	Hysta2	-	20	-	mV	Hysteresis setting = 10mV
Differential Hysteresis	Hysta3	-	40	-	mV	Hysteresis setting = 20mV
Differential Hysteresis	Hysta4	-	80	-	mV	Hysteresis setting = 40mV
Hysteresis Setting Accuracy	Hystb	-	25	-	%	-
Hysteresis Temperature Coefficient	Hysttc1	-	5	-	µV/°C	Hysteresis setting = zero
Hysteresis Temperature Coefficient	Hysttc2	-	50	-	µV/°C	Hysteresis setting = 10mV
Hysteresis Temperature Coefficient	Hysttc3	-	100	-	µV/°C	Hysteresis setting = 20mV
Hysteresis Temperature Coefficient	Hysttc4	-	200	-	µV/°C	Hysteresis setting = 40mV

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<sup>c</sup>. High precision operating range provides optimal linearity and dynamic range. <sup>d</sup>. Standard precision operating range provides maximum dynamic range and reduced linearity.

#### **ESD** Characteristics

Pin Type	Human Body Model	Machine Model	Charged Device Model
Digital Inputs	4000V	250V	4kV
Digital Outputs	4000V	250V	4kV
Digital Bidirectional	4000V	250V	4kV
Digital Open Drain	4000V	250V	4kV
Analog Inputs	2000V	200V	4kV
Analog Outputs	1500V	100V	4kV
Reference Voltages	1500V	100V	4kV

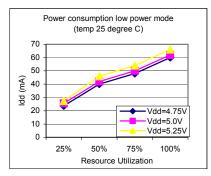
The AN221E04 is an ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000V readily accumulate on the human body and test equipment and can discharge without detection. Although the AN221E04 device features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high-energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

#### Power Consumption – Low Power Mode

Parameter	Symbol	Min	Тур	Мах	Unit	Comment
Minimum Power <sup>1a</sup>	ldd	-	0.2	-	mA	Vdd=5.00 volts, Tj=25°C
Nominal 25% Power <sup>1b</sup>	ldd	-	25	30	mA	Vdd=5.00 volts, Tj=25°C
Nominal 50% Power <sup>1c</sup>	ldd	-	42	47	mA	Vdd=5.00 volts, Tj=25°C
Nominal 75% Power <sup>1d</sup>	ldd	-	50	55	mA	Vdd=5.00 volts, Tj=25°C
Maximum Power <sup>1e</sup>			60	-		Vdd=4.75 volts, Tj=85°C
	ldd	-	63	68	mA	Vdd=5.00 volts, Tj=25°C
			66	-		Vdd=5.25 volts, Tj= -40°C
Temperature Coefficient	-	-	-2	-10	µA/°C	

1a. External clock, all analog function disabled, memory active.

- 1b FPAA active elements Two core op-amps (low power mode), one comparator, one input (bypass mode), one output filter and differential to single-ended converter (low power mode).
- FPAA active elements Four core op-amps (low power mode), two comparators (one using SAR), two inputs (bypass mode), two output filters and two differential to singleended converters (low power mode).
- 1d FPAA active elements Six core op-amps (low power mode), three comparators (two using SAR), three inputs (bypass mode, two output filters and two differential to single-ended converters (low power mode).
- 1e FPAA active elements Eight core op-amps (low power mode), four comparators (two using SAR), four inputs (bypass mode), two output filters and two differential to singleended converters (low power mode).



#### **Power Consumption – Full Power Mode**

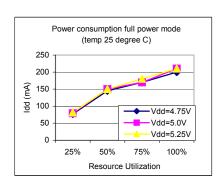
Parameter	Symbol	Min	Тур	Max	Unit	Comment
Full Power Mode Minimum Power <sup>2a</sup>	ldd	-	1.5	-	mA	Vdd=5.00 volts, Tj=25°C
Full Power Mode Nominal 25% Power <sup>2b</sup>	ldd	-	80	90	mA	Vdd=5.00 volts, Tj=25°C
Full Power Mode Nominal 50% Power <sup>2c</sup>	ldd	-	150	160	mA	Vdd=5.00 volts, Tj=25°C
Full Power Mode Nominal 75% Power <sup>2d</sup>	ldd	-	170	190	mA	Vdd=5.00 volts, Tj=25°C
Full Power Mode Maximum Power <sup>2e</sup>	ldd	-	200 210 220	- 230 -	mA	Vdd=4.75 volts, Tj=85°C Vdd=5.00 volts, Tj=25°C Vdd=5.25 volts, Tj= -40°C

2a. AN220E04 Crystal Oscillator, all analog functions disabled, memory active.

2b. FPAA active elements – Two core op-amps, one comparator, one input filter and chopper amplifier, one output filter and differential to single-ended converter.

 FPAA active elements – Four core op-amps, two comparators (one using SAR), two Input filters and two chopper amplifiers, two output filters and two differential to single-ended converters.

- 2d FPAA active elements Six core op-amps, three comparators (two using SAR), three input filters and three chopper amplifiers, two output filters and two differential to single-ended converters.
- 2e FPAA active elements Eight core op-amps, four comparators (two using SAR), four input filters and two chopper amplifiers, two output filters and two differential to single-ended converters.



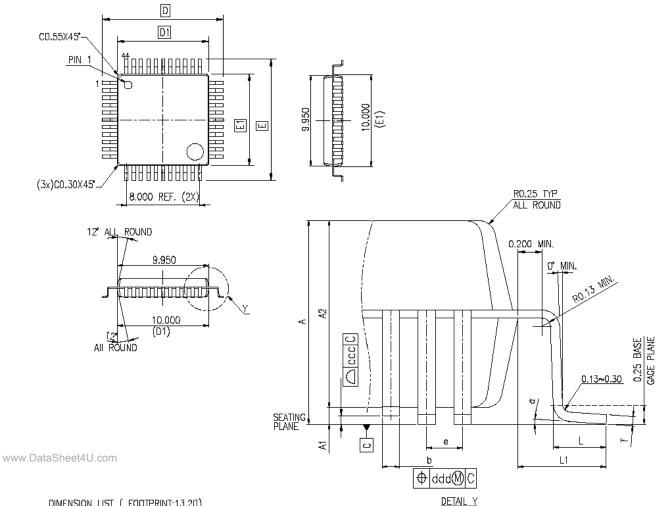
## PINOUT

Pin Numb er	Pin Name	Pin Type	Comments
1	I4PA	Analog IN+	
2	I4NA	Analog IN-	
3	01P	Analog OUT+	
4	01N	Analog OUT-	
5	AVSS	Analog Vss	
6	AVDD	Analog Vdd	
7	O2P	Analog Vuu Analog OUT	
8	O2N	Analog OUT	
9	I1P	Analog IN+	
10	I1N	Analog IN-	
11	I2P	Analog IN+	
12	12N	Analog IN-	
13	SHIELD	Analog Vdd	Low noise Vdd bias for capacitor array n-wells
14	AVDD2	Analog Vdd	Analog power
15	VREFMC	Vref	Attach filter capacitor for VREF-
16	VREFPC	Vref	Attach filter capacitor for VREF+
17	VMRC	Vref	Attach filter capacitor for VMR (Voltage Main Reference)
18	BVDD	Analog Vdd	Analog power for bandgap Vref Generators
19	BVSS	Analog Vss	Analog ground for bandgap Vref Generators
20	CFGFLGb	Digital IN	In multi-device systems
20	CFGFLGD	Digital IN	In multi-device systems     O, Ignore incoming data (unless currently addressed)     1, Pay attention to incoming data (watching for address)     O, Device is being configured
21	CS2b	Digital IN	Z, Device is not being configured (if internal pullup is selected) 0. Chip is selected
		-	0, Allow configuration to proceed
22	CS1b	Digital IN (during config) Digital IN	1, Hold off configuration Passes read-back data through to LCC_B pin
23	DCLK	(after config)_ Digital IN	
24	SVSS	Digital Vss	Digital ground, substrate tip
			Digital ground - substrate tie
25	MODE	Digital IN	0, Synchronous serial interface 1, SPI EPROM Interface
26	ACLK / SPIP	Digital IN Digital OUT	MODE = 0, analog clock < 40 MHz MODE = 1, SPI EPROM or serial EPROM clock
27	OUTCLK / SPIMEM	Digital OUT Digital OUT	During power-up, sources SPI EPROM initialization command string After power-up, sources any of the four internal analog clocks
28	DVDD	Digital Vdd	
29	DVSS	Digital Vss	
30	DIN	Digital IN	Serial configuration data input
31	LCCb	Digital OUT	<ol> <li>Local configuration is needed. Once configuration is completed, it is a registered versio CS1b or if the device is addressed for read, it serves as serial data out port</li> </ol>
132taShe	et <b>EBRb</b> om	Digital IN (monitored OUT)	0, Initiate reset 1, No action 0. Error condition
00		Digital OUT	Z, No error condition (external pullup required)
33	ACTIVATE	Digital IN	0, Hold off completion of configuration Rising Edge, Allow completion of configuration O.D. Output 0, device has not yet completed primary configuration Z, Device has completed primary configuration (if internal pullup is selected)
34	DOUTCLK / TEST	Digital OUT Digital IN	A buffered version of DCLK. (Factory reserved test input. Float if unused)
35	PORb	Digital IN	0, Chip held in reset state Rising edge, re-initiates power on reset sequence To initiate a POR reset cycle, the minimum pulse width required on the PORb pin is 25ns.
36	EXECUTE	Digital IN	0, No action 1, Transfer shadow RAM into configuration RAM
37	I3P	Analog IN+	
38 39	I3N I4PD	Analog IN- Analog IN+	Analog multiplexer input signals.
			The multiplexer can accept 4 differential inputs or 8 single ended inputs
40 41	I4ND I4PC	Analog IN- Analog IN+	
42	I4NC	Analog IN-	
	I4PB	Analog IN+	
43	1420	/ thatog into	

#### **MECHANICAL AND HANDLING**

The AN221E04 comes in the industry standard 44 lead QFP package.

Dry pack handling is recommended. The package is qualified to MSL3 (JEDEC Standard, J-STD-020A, Level 3). Once the device is removed from dry pack, 30°C at 60% humidity for not longer than 168 hours is the maximum recommended exposure prior to solder reflow. If out of dry pack for longer than this recommended period of time, then the recommended bake out procedure prior to solder reflow is 24 hours at 125°C.



#### DIMENSION LIST ( FOOTPRINT: 13.20)

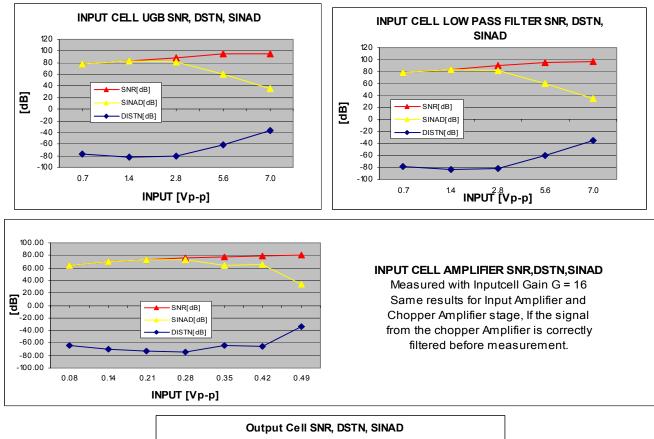
S/N	SYM	DIMENSIONS	REMARKS
1	А	MAX. 2.450	OVERALL HEIGTH
2	A1	MAX. 0.5DO	STANDOFF
3	A2	2.000±0.200	PKG THICKNESS
4	D	13.200±0.250	LEAD TIP TO TIP
5	D1	10.000±0.100	PKG LENGTH
6	Е	13.200±0.250	LEAD TIP TO TIP
7	E1	10.000±0.100	PKG WIDTH
8	L	0.880±0.150	FODT LENGTH
9	L1	1.600 REF.	LEAD LENGTH
10	Т	0.130~0.230	FRAME THICKNESS
11	a	0~7°	LEAD FLAT ANGLE
12	Ь	0.300~0.450	LEAD WIDTH
13	е	0.800 BASE	LEAD PITCH
14	000	0.100	FOOT PLANARITY
15	ddd	0.100	FOOT POSITION

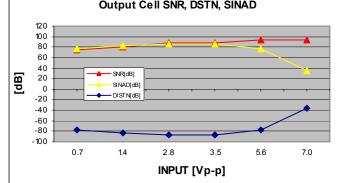
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S/N	DESCRIPT	SPECIFICATION	
1	GENERAL TOLERANCE.	DISTANCE	±0.100
		ANGLE	±2.5°
2	MATTE FINISH ON PACKAGE BODY SURFACE Ro1.5~2.5um EXPECT EJECTION AND PIN 1 MARKING.		
3	ALL MOLDED BODY SHARP CORNER RADII MAX. R0.200 UNLESS OTHERWISE SPECIFIED.		
4	PACKAGE/LEADFRAME MISALIGNMENT ( X, Y ): MAX. 0.127		
5	TOP/BTM PACKAGE MISALIGNMENT ( X, Y ): MAX. 0.127		
6	DRAWING DOES NOT INCLUDE PLASTIC OR METAL PROTRUSION OR CUTTING BURR.		
7	COMPLIANT TO JEDEC STANDARD: MS-022		

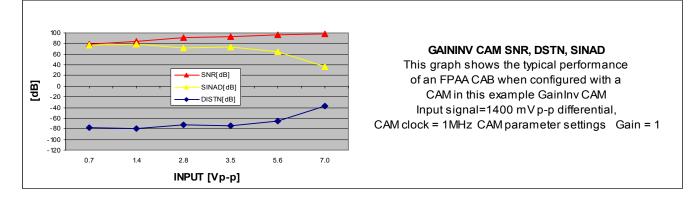
#### **Distortion, SINAD and SNR Measurements**

The following plots give an indication of the Distortion, SINAD and SNR for some representative CAMs.



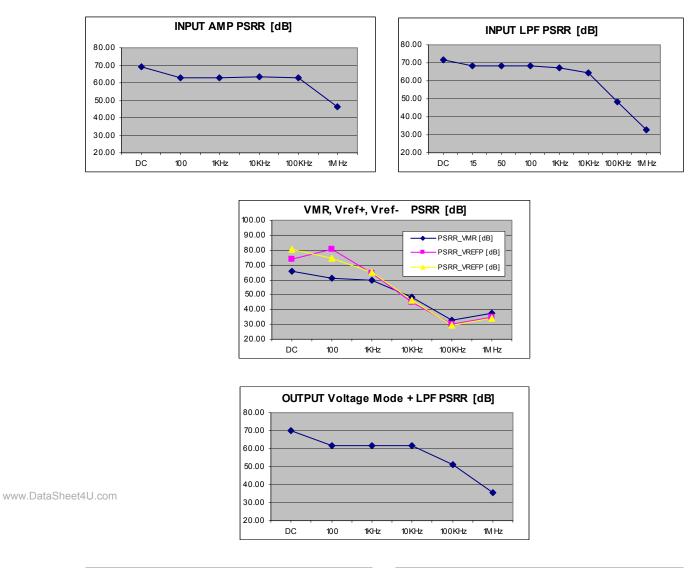


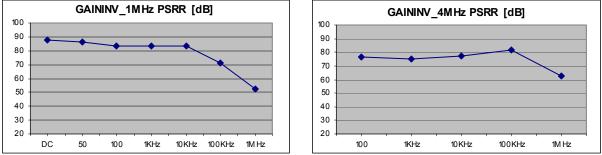
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#### Power Supply Rejection Ratio (PSRR) Measurements

The following plots give an indication of the PSRR for some representative CAMs. AVDD to Power Supply (PS): 5v +/- 0.25v sinusoidal waveform (100 kHz to 1 MHz)



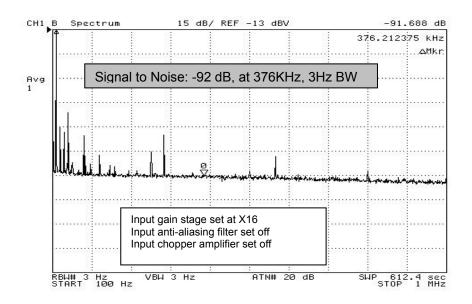


The following is provided for information only, as and when additional characterization data is collected 'noise measurements' will be added formally to the datasheet.

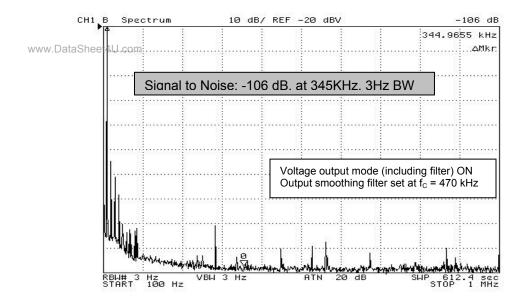
#### **Noise And Distortion Observations**

The following plots give an indication of the noise characteristics of Anadigm<sup>®</sup>'s AN221E04 FPAA device. These were done using a simple set-up and in many cases reflect the noise limit of the setup. Actual device noise margins are expected to be better.

Signal and Noise for the Input Cell (input signal - 50mVp-p differential to the FPAA at 10 kHz)



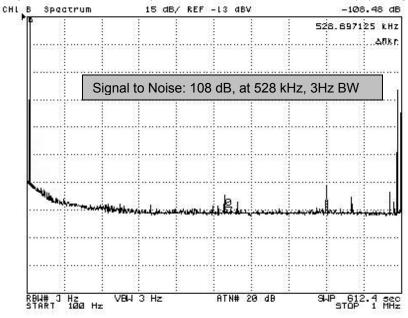
Signal and Noise for the Output Cell (with a differential input 4V p-p, 660Hz)



Measured THD for input and output cells (with a differential input 4V p-p, 660Hz)

Settings	Distortion in dB
Input cell with anti-aliasing filter set at f <sub>c</sub> = 470 kHz	81.6
Output cell with differential to single ended converter and output smoothing filter set at $f_c = 470 \text{ kHz}$	82

Signal and Noise for a representative CAM – Gaininv CAM (input signal of 700mV p-p differential at 10 kHz)



THD for a representative CAM – Gaininv CAM (with a differential input 4V p-p, 660Hz)

CAM Clock	Distortion
Frequency	(dB)
250 KHz	80.00
1 MHz	72.83
2 MHz	69.22
4 MHz	73.48

