

Asahi KASEI

AK2347A

Two-way Radio Audio & Sub-Audio Processor

Features

Audio processing

- TX and RX amplifier
- Pre/De-emphasis circuit
- · Compressor and Expander with no external components
- Scrambler and De-scrambler in frequency inversion type (3.388kHz or 3.290kHz)
- · Limiter with level adjuster
- Splatter filter for wide and narrow band
- Digital controlled amplifier for microphone, modulator and demodulator sensitivity
- Sub-Audio filter with level adjuster for CTCSS and DCS
- Low power supply operation: 2.7 to 3,3V
- Wide range operating for temperature: -40 to 85 °C
- Oscillator circuit for 3.6864MHz and 3.579545MHz crystal
- Serial control interface operation
- Compact plastic packaging, 24-pin QFNJ (4.0x4.0x0.75mm 0.5mm pitch)

Description

AK2347A includes audio filter, limiter, splatter filter, compandor, scrambler, which is highly integrated two-way radio baseband functions for FRS and LMR.

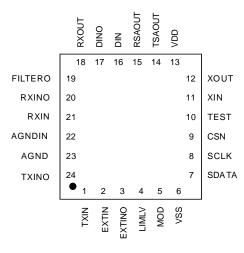
Audio high-pass filter shows a high attenuation in magnitude response characteristics under 250Hz that supports to eliminate a subaudio tone clearly.

TX limiter for deviation control has a limiting level adjuster by applying a DC voltage via external components. Splatter filter has the magnitude response for narrowband(fc=2.55kHz) and wideband(3.0kHz) to meet various regulatory agencies in the world wide.

Compandor is no adjustment type because it includes all parametric components inside the chip. Scrambler circuit is composed of frequency inversion circuit by double balanced mixer that has 3.388kHz and 3.290kHz carrier clock.

Sub-Audio filter with level adjuster is available for pre- or post-filter for CTCSS and DCS. There are four signal level adjusters for microphone, modulator and demodulator sensitivity by digital controlled amplifier (volume).

• Pin Assignment (Top view)



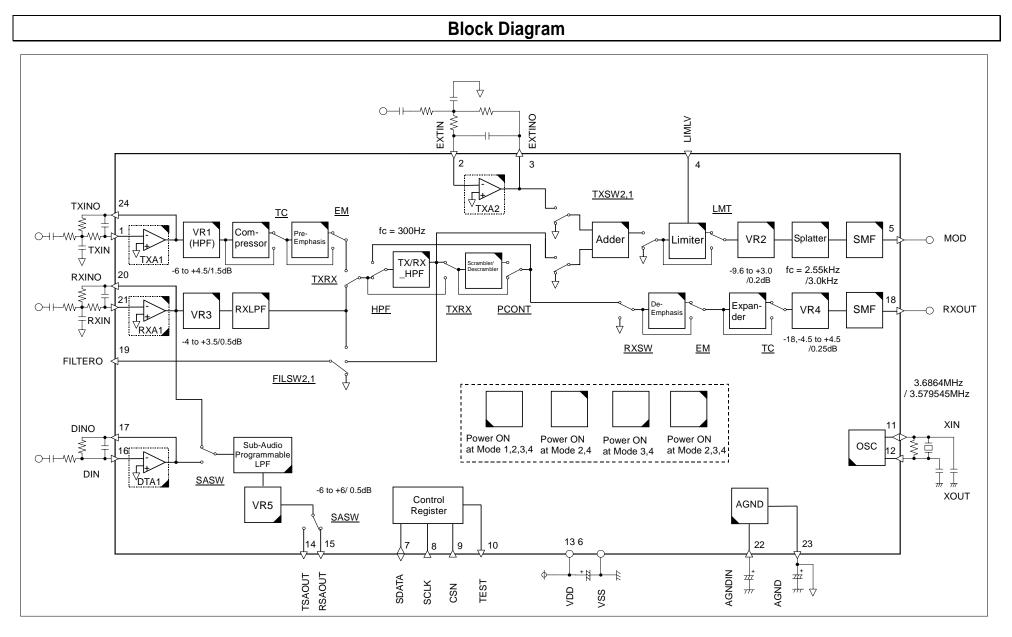
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MS0946-E-00



Block Functions

Block Fun	ction
TXA1	Operational amplifier for gain adjustment of transmit audio signal and for the filter for preventing aliasing noise of the SCF circuit in the subsequent stage. Use external resistors and capacitors to set the gain to 30dB or less and the cut-off frequency to around 10kHz.
VR1 (HPF)	This circuit controls the volume for adjusting the input level of transmit audio signal. Setting registers: VR12 to VR10, adjustment range: -6.0dB to +4.5dB in 1.5dB steps
Compressor	This circuit compresses the amplitude of transmit audio signal by 1/2 in dB scale. Cross-point: -10dBx. This circuit is turned on and off by the TC register.
Pre-emphasis	This circuit emph asizes the high-frequency com ponent of transmit au dio signal to improve the S/N ratio of the modulation signal.
TX/RX_HPF	High-pass filter to eliminate low-frequency components lower than 250Hz which are included in transmit and receive audio signals. This circuit is turned on and off by the HPF register.
Scrambler/ Descrambler	This circuit in verts the spectrum distribution of transmit and re ceive audio si gnals with respect to the carrier frequency. The carrier frequency is 3.388kHz or 3.290kHz. The Scrambl er/Descrambler or emp hasis ci rcuit ca n be sele cted using the EM and PCONT registers. These circuits cannot be used at the same time.
TXA2	Operational amplifier for gain adju stment of external tone signal. Use external re sistors and capacitors to set the gain to 0dB or less and the cut-off frequency to around 10kHz.
Adder	This circuit adds together the audio signal and external tone in put signal. This circuit is controlled by the TXSW2 and TXSW1 register.
Limiter	Amplitude limiting circuit to suppress frequency deviation in the modulation signal. The limit level can be adjusted by a pplying a DC voltage to the LIMLV pin. When the pin is left open, the level pred etermined within the device is output. This circuit is turned on and off by the LMT register.
VR2	This circuit controls the volume for adj usting the output level on the MOD pin. Setting registers: VR25 to VR20, adjustment ran ge: -3.2dB to +3.0dB in 0.2dB step s. For coarse adjustment, switching between -6.4dB and 0dB is possible.
Splatter	Low-pass filt er to eli minate high -frequency components hig her than 3 kHz which a re included in the limiter output signal. The cut-off frequency can be adjusted with the SPL register.
SMF	Smoothing filter to eliminate the high-frequency and clock components generated in the SCF circuit.
RXA1	Operational amplifier for gain adjustment of the receive demodulation signal and for the filter for preventing aliasing noise in the SCF circuit in the subsequent stage. Use external resistors and capacitors to set the gain to 20dB or less and the cut-off frequency to around 40kHz.
VR3	This circuit controls the volume for adjusting the input level of the re ceive demodulation signal. Setting registers: VR33 to VR30, adjustment range: -4.0dB to +3.5dB in 0.5dB steps
RXLPF	Low-pass filt er to eli minate high -frequency components hig her than 3 kHz which a re included in the receive demodulation signal.
De-emphasis	This circuit restores the origin al state of the signal of which high -frequency component has been emphasized by the Pre-emphasis.
Expander	This circuit e xpands the signal compressed twice by the Comp ressor in dB scale to restore the original signal state. Cross-point: -10dBx. The Expander is turned on and off with the TC register.
VR4	This circuit controls the volume for adjusting the RX output level. Setting registers: VR45 to VR40 Adjustment range: -18.0dB, -4.5dB to +4.5dB in 0.25dB steps



Block Fun	ction
DTA1	Operational amplifier for gain adjustment of the Sub-Audio LPF input signal and for the filter for preventing aliasing noise in the SCF circuit in the subsequent stage. Use external resistors and capacitors to set the gain to 0dB or less and the cut-off frequency to around 10kHz.
Sub-Audio Programmable LPF	Low-pass filter to eliminate components of the DAT1 signal in the transmit operation and of RXA1 signal in the receive. This circuit is controlled by the SASW register for transmit or receive and by the SA5 to SA0 for cut-off frequency.
VR5	This circuit controls the v olume for adjusting the output level from the Sub-A udio LPF signal. Setting registers: VR54 to VR50, adjustment range: -6.0dB to +6.0dB in 0.5dB steps
AGND	This circuit generates the reference voltage (1/2VDD) for internal analog signals.
OSC	This circuit g enerates a 3 .6864MHz or 3.5795 45MHz reference clock signal from an external resistor and crystal oscillator. This circuit is controlled by the MCKSL register.
Control Register	Control registers set the switch status and volume for level a djustment inside the IC according to the serial input data consisting of 1-bit instruction and 4-bit add ress and 8-bit data. At power-up, the registers are set to the power-down values by the power-on reset circuit. This circuit has a software reset named RSTN register. (See the description of the registers)

Pin Functions

Pin number	Pin name	Pin type	Power- down status	Function
1 TXII	N	AI	Z	Transmit audio signal input pin This pin is the inverting i nput pin of TX A1. This pin, with re sistors and capacitors externally connected, forms a microphone amplifier.
2 EXT	ĪN	AI	Z	External input pin This pin is the inverting i nput pin of TX A2. This pin, with re sistors and capacitors externally connected, forms a amplifier. An external signal su ch as a t one signal other than the audio signal can be input.
3	EXTINO	AO	*3)	Output pin of TXA2 *1)
4 LIM	L V	AI	*4)	Limit level adjustment pin The limit level can be adj usted by apply ing a DC volt age to this pin. When this pin is lef t open, the limit level pred etermined within the device is set.
5	MOD	AO	Z	Modulated transmit signal output pin *2)
6 VSS	5	PWR	-	Negative power supply pin Normally, apply 0V.
7	SDATA	DB	Z	Serial data input and output control pin
8	SCLK	DI	Z	Serial data clock input pin
9 CSI	4	DI	Z	Serial data chip select input This signal is active low.
10 TES	ST	DO	L	Output pin for testing This pin is assigned to test pin for pre-delivery inspection in factory. Do not connect anything in normal operation.



Pin number	Pin name	Pin type	Power- down status	Function
11 XIN		DB	*5)	Pin for connecting a crystal oscillator A reference clock used within this IC is generated by connecting a 3.6864MHz or 3.57 9545MHz oscillat or bet ween this pin and the adjacent XOUT pin. For detailed information a bout the connection method and the method for supplying an external clock, see "Recommended External Application Circuits".
12	XOUT	DI	*5)	Pin for connecting a crystal oscillator
13 VDI	D	PWR	_	Positive power supply pin Connect this pin to a power supply ranging from 2.7V to 3.3V with less noise. Connect a bypass capacitor of $0.1 \mu\text{F}$ or higher between this pin and the VSS pin.
14	TSAOUT	AO	Z	Transmit Sub-Audio signal output pin *2)
15	RSAOUT	AO	Z	Receive Sub-Audio signal output pin *2)
16 DIN	I	AI	Z	Data input pin This pin i s the inverting in put pin of DT A1. This pin, with resi stors and capacitors externally connected, forms a amplifier. An external sign al such as a tone si gnal through CPU port can be input.
17	DINO	AO	Z	Output pin of DTA1 *1)
18	RXOUT	AO	Z	Receive audio signal output pin *2)
19 FIL	TERO	AO	Z	RXLPF or TX/RX_HPF block output pin This pin can be use d as a monitor pin for a signal su ch as a tone signal. The output si gnal on this pin in cludes a 57.6 kHz sampling-clock component. So, perform waveform processing externally as required. *2)
20	RXINO	AO	Z	Output pin of RXA1 *1)
21 RX	IN	AI	Z	Demodulated receive signal input pin Inverting input pin of RXA1. This pin, with re sistors and cap acitors externally connected, forms a pre-filter.
22 AG	NDIN	AI	*3)	Analog ground input pin This pin is connected to a capacitor to stabilize the analog ground level.
23 AG	ND	AO	*3)	Analog ground output pin This pin is connected to a capacitor to stabilize the analog ground level.
24	TXINO	AO	Z	Output pin of TXA1 *1)

Note) A: Analog, D: Digital, PWR: Power, I: Input, O: Output, B: Bi-directional, Z: High-Z, L: Low

- *1) Output load requirement: [load impedance] > $30k\Omega$, [load capacitance] < 50pF
- *2) Output load requirement: [load impedance] > $10k\Omega$, [load capacitance] < 50pF
- *3) AGND (=1/2VDD) level
- *4) AGND + 0.256(VDD-AGND) level
- *5) When XOUT pin is set to low level, XIN pin goes to High-Z. When XOUT pin is set to high level, XIN pin outputs low level.



Absolute Maximum Ratings

Parameter Sy	mbol	Min.	Max.	Units
Power Supply Voltage	VDD	-0.3	4.6	V
Ground Level	VSS	0	0	V
Input Voltage	V _{IN}	-0.3	VDD+0.3	V
Input Current (Except power supply pin)	I _{IN}	-10	+10	mA
Storage Temperature	T _{stg}	-55	130	°C

Note) All voltages are relative to the VSS pin.

Caution) If the device is used in conditions exceeding these values, the device may be destroyed. Normal operations are not guaranteed in such extreme conditions.

Recommended Operating Conditions

Parameter Symbol		Condition	Min.	Тур.	Max.	Units
Operating Temperature	Та		-40		85	°C
Power Supply Voltage	VDD		2.7	3.0	3.3	V
Analog Reference Voltage	AGND		1/2VDD			V

Note) All voltages are relative to the VSS pin.

Digital DC Characteristics

Parameter Symbol	pl	Condition	Min.	Тур.	Max.	Units	
High lovel input veltage	V _{IH1}	SDATA	0.7VDD			V	
High level input voltage	V _{IH2}	SCLK, CSN	0.8VDD			V	
	V_{IL1}	SDATA			0.3VDD	V	
Low level input voltage	V_{IL2}	SCLK, CSN			0.2VDD	V	
High level input current	I _{IH}	V _{IH} =VDD SDATA, SCLK, CSN			10	μA	
Low level input current	I _{IL}	V _{IL} =0V SDATA, SCLK, CSN	-10			μA	
High level output voltage	V _{OH}	I _{OH} =+0.2mA SDATA	VDD-0.4	VDD		V	
Low level output voltage	V _{OL}	I _{OL} =−0.4mA SDATA	0.0 0.4	ŀ		V	



Clock Input Characteristics

Parameter Symbol		Condition	Min.	Тур.	Max.	Units	Remarks
Master Clock Frequency	MCK	XIN, XOUT		3.6864 3.579545	MHz		
High level input voltage	V _{MCK1_IH} >	(IN	1.5			V	*1)
Low level input voltage	$V_{MCK1_{IL}}$	(IN			0.4	V	*1)
Input amplitude	V _{MCK2} X	IN	0.2		1.0	V _{PP} *2)	

*1) When directly connects to XIN pin, refer to "Recommended External Application" 7) Oscillator circuit Fig. 7.

*2) When connects to XIN pin via capacitor, refer to "Recommended External Application" 7) Oscillator circuit Fig. 8.

Current Consumption

Parameter Sym	bol	Condition	Min.	Тур.	Max.	Units
Current Consumption	IDD0	Mode 0 OSC:OFF, Audio: OFF, Sub-Audio: OFF		0.08	0.14	
	IDD1	Mode 1 OSC: ON , Audio: OFF, Sub-Audio: OFF		0.7	1.0	
	IDD2	Mode 2 OSC: ON , Audio: ON , Sub-Audio: OFF		4.5	5.8	mA
	IDD3	Mode 3 OSC: ON , Audio: OFF, Sub-Audio: ON		1.6	2.1	
	IDD4	Mode 4 OSC: ON , Audio: ON , Sub-Audio: ON		5.3	6.7	



Analog Characteristics

Unless otherwise specified, the following apply: MCK=3.6864MHz, f=1kHz, Emphasis: on, Compandor: on, Scrambler: off, VR1=VR2=VR3=VR4=0dB, HPF=LMT=1 with the external circuit shown in page.28 to 32.

"dBx" is a standardized notation to match the operating voltage and is defined b y equation 0dBx = -5+20log(VDD/2)dBm. 0dBm=0.775Vrms.

Parameter Con	dition	Min.	Тур.	Max.	Units	Remarks
Standard input level	@TXINO, EXTINO		-10		dBx	
Absolute gain	TXINO to MOD	-1.5	0	+1.5	dB	
-	EXTIN to MOD	-1.5	0	+1.5	dB	
Distortion	EXTIN to MOD, EXTINO=-3dBx When LMT is set to 0 30kHz Low-pass filtering			-35	dB	-
Limit level	EXTIN to MOD Without external R adjustment With external R adjustment	-8.6 -7.	6	-6.6 -6.6	dBx	
Compressor linearity	TXINO to MOD TXINO=-44dBx TXINO=-50dBx Relative value to 0dB fo r MOD level of -10dBx TXINO.	-20.0 -24.0	-17.0 -20.0	-14.0 -16.0 dE	3	
Compressor distortion	TXINO to MOD TXINO=-10dBx 30kHz Low-pass filtering			-35 dB		
Noise le vel w ith no signal input	TXINO to MOD C-Message filtering			-36.5 dE	ßm	_
VR1 Attenuation error	TXINO to MOD -6.0 dB to 4.5dB, 1.5dB/step	-1.5		+1.5	dB	-
VR2 ATT error (VR24,23,22,21,20)	TXINO to MOD -3.2dB to +3.0dB, 0.2dB/step	-0.2		+0.2	dB	
VR2 ATT error (VR25=0)	TXINO to MOD Relative value when -6.4/0dB is set	-6.8 -6.	4 -6.0		dB	

1) TX Audio system characteristics

2) RX Audio system characteristics

Parameter Con	dition	Min.	Тур.	Max.	Units	Remarks
Standard Input Level	@RXINO		-10		dBx	
Absolute gain	RXINO to FILTERO	-1.5	0	+1.5	dB	
	RXINO to RXOUT	-1.5	0	+1.5	dB	
Expander linearity	RXINO to RXOUT RXINO=-25dBx RXINO=-30dBx Relative value to 0dB for RXOUT level of -10dBx RXINO	-33.0 -45.0	-30.0 -40.0	-27.0 -35.0	dB	
Expander distortion	RXINO to RXOUT RXINO=-5dBx 30kHz Low-pass filtering			-35	dB	
Noise le vel w ith no signal input	RXINO to RXOUT C-Message Filtering			-70	dBm	



Parameter Con	dition	Min.	Тур.	Max.	Units	Remarks
VR3	RXIN0 to RXOUT	-0.5		+0.5	dB	
Attenuation error	-4.0dB to +3.5dB, 0.5dB/step					
VR4	RXIN0 to RXOUT	-0.25		+0.25	dB	
Attenuation error	-4.5 to +4.5dB, 1.5dB/step	-0.25		+0.25	чD	
VR4 ATT error	RXIN0 to RXOUT					
(VR4540=0,0,0,0,0,	Relative value when –18/0dB	-20 -18	-16		dB	
0)	is set					

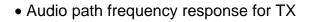
3) Audio Filter Characteristics

3.1) Emphasis: on, Compandor: off, Scrambler: off

Parameter Con		dition	Min.	Тур.	Max.	Units	Remarks
TX overall	TXINO to MOD	250Hz		-57	-40	dB	
characteristics		300Hz	-12.5		-9.5		
		2.5kHz	+6.0		+9.0	٩D	SPL=0
		3.0kHz	+4.5		+8.5	dB	fc=2.55
	Relative val ue	6.0kHz		-23	-18		K
	to gain at 1kHz	300Hz	-12.5		-9.5		
		2.5kHz	+6.0		+9.0	dB	SPL=1
		3.0kHz	+7.0		+10.5	uБ	fc=3.0K
		6.0kHz		-17	-12		
RX overall	RXINO to RXAF	250Hz		-38	-26		
characteristics		300Hz	+8.5		+11.5	dB	
	Relative val ue	3.0kHz	-11.5		-8.5	UD	
	to gain at 1kHz	6.0kHz		-52	-40		

3.2) Emphasis: off, Compandor: off, Scrambler: off (Des ign target values)									
Parameter Con	Parameter Con dition				Max.	Units	Remarks		
TX overall TXINO to MOD 250Hz				-50	-38	dB			
characteristics		300Hz to 2.0kHz 2.5kHz	-1.0 -1.5		+1.0 +1.0	ם ו	SPL=0		
	Relative val ue to gain at 1kHz	3.0kHz 6.0kHz	-4.0	-32	-1.0 -28	dB	fc=2.55 K		
		300Hz to 2.5kHz 3.0kHz 6.0kHz	-1.0 -1.5	-26	+1.0 +1.0 -22	dB	SPL=1 fc=3.0K		
RX overall characteristics	RXINO to RXAF	250Hz 300Hz	-1.5	-49	-38 +1.0	dB			
	Relative val ue to gain at 1kHz	350Hz to 3.0kHz 6.0kHz	-1.0	-38	+1.0 -28	UD .			





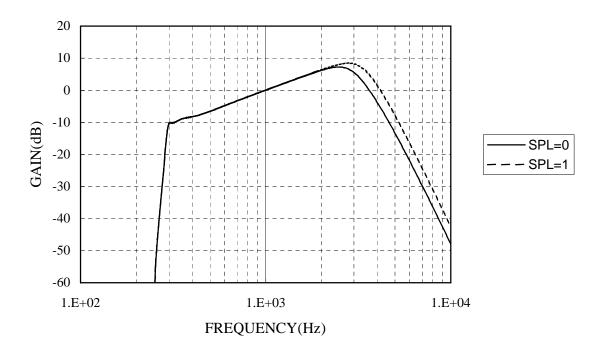
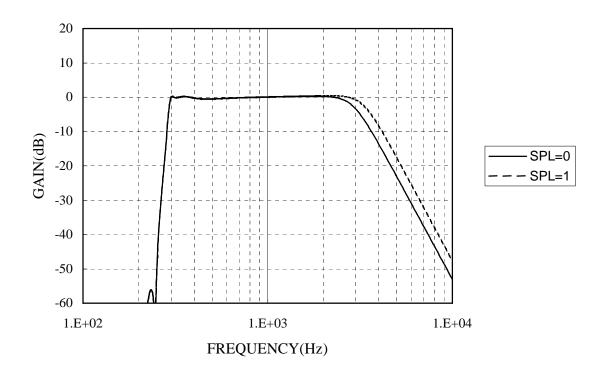
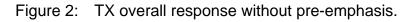


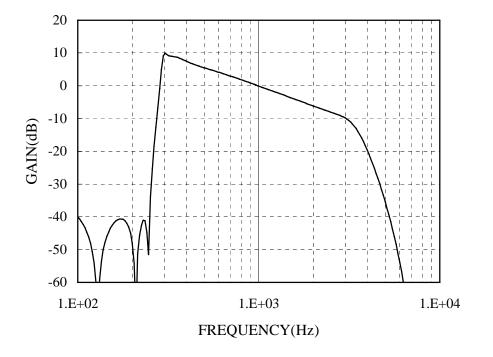
Figure 1: TX overall response with pre-emphasis.







Audio path frequency response for RX





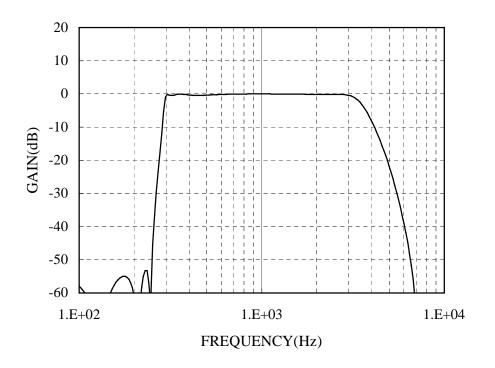


Figure 4: RX overall response without de-emphasis.



4) Scrambler Characteristics
 4.1) Scrambler: on, Emphasis: off, Compandor: off, MCKSL=1, SCSL=0/1...3.388kHz MCKSL=0,

MCKSL=0,		5	CSL=0	3.290KHZ		
Parameter Con	dition	Min.	Тур.	Max.	Units	Remarks
Carrier frequency			3.388 3.290	kH	Z	
Modulated output	TXINO to MOD, RXINO to RXOUT Input level 1.0kH z -1 0dBx Measuring-freq. 2.388kHz (3.388kHz) Measuring-freq. 2.290kHz (3.290kHz)	-12 -10		-8	dBx	
High frequency rejection	TXINO to MOD, RXINO to RXOUT Input level 1.0kH z -1 0dBx Measuring-freq. 4.388kHz (3.388kHz) Measuring-freq. 4.290kHz (3.290kHz)			-50	dBx	
Carrier signal leakage	TXINO to MOD, RXINO to RXOUT Input level No signal Measuring-freq. 3.388kH z Measuring-freq. 3.290kH z			-50	dBx	
Original signal leakage	TXINO to MOD, RXINO to RXOUT Input level 1.0kH z -1 0dBx Measuring-freq. 1.0kH z			-50	dBx	

	4.2) Scrambler: on, Em	nphasis: off, Compandor: off, MCk	KSL=0, SCSL=1	(Design target values)
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Parameter Con	dition	Min.	Тур.	Max.	Units	Remarks
Carrier frequency		3.39	0		kHz	
Modulated output	TXINO to MOD, RXINO to RXOUT Input level 1.0kH z -1 0dBx Measuring-freq. 2.390kH z	-12 -10		-8	dBx	
High frequency rejection	TXINO to MOD, RXINO to RXOUT Input level 1.0kH z -1 0dBx Measuring-freq. 4.390kH z			-50	dBx	
Carrier signal leakage	TXINO to MOD, RXINO to RXOUT Input level No signal Measuring-freq. 3.390kH z			-50	dBx	
Original signal leakage	TXINO to MOD, RXINO to RXOUT Input level 1.0kH z -1 0dBx Measuring-freq. 1.0kH z			-25	dBx	



5) Sub-Audio filter Characteristics

Únless otherwise specified, the following apply: MCKSL=1, SA5=1, SA4=1, SA3=0, SA2=0, SA1=0, SA0=1(fc=260.9Hz), VR5=0dB, 250.3Hz sinusoidal wave.

5.1) Analog characteristics

Parameter Con	dition	Min.	Тур.	Max.	Units	Remarks
Standard input level	@ TSAOUT	-10			dBx	
Transmit CTCSS signal gain	DINO to TSAOUT	-2 0		+2	dB	
Transmit CTCSS signal distortion	DINO to TSAOUT 250.3Hz,Duty50%, 585mVp-p(@3V)rectangular wave 30kHz Low-pass filtering	-37		-32	dB	
Standard input level	@ RSAOUT	-10			dBx	
Receive CTCSS signal gain	RXINO to RSAOUT	-2	0	+2	dB	
Receive CTCSS signal distortion	RXINO to RSAOUT RXINO=-10dBx input 30kHz Low-pass filtering	-37		-32	dB	
VR5 Attenuation error	RXINO to RSAOUT -6.0 to +6.0dB, 0.5dB/step	-0.5		+0.5	dB	

5.2) Filter characteristics

Parameter Con	dit	Min.	Тур.	Max.	Units	Remarks	
Overall characteristics @fc=260.9Hz	DINO to TSAOUT Relative value to gain at 100Hz	50~240Hz 250Hz 300Hz	-1.0 -1.5		+1.0 +1.0 -38	dB	

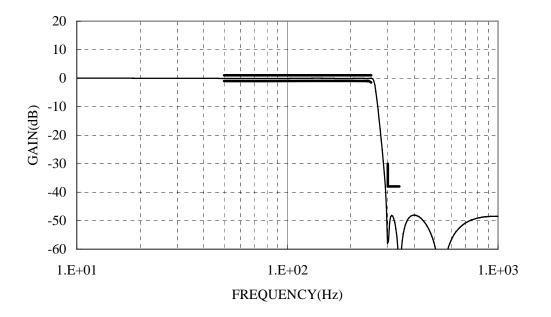
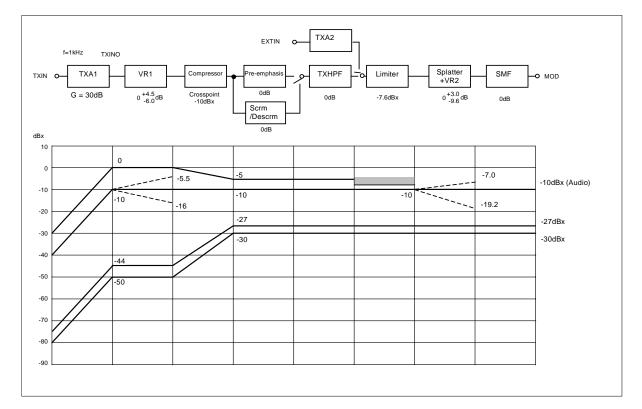


Fig.5 Sub-Audio response characteristics

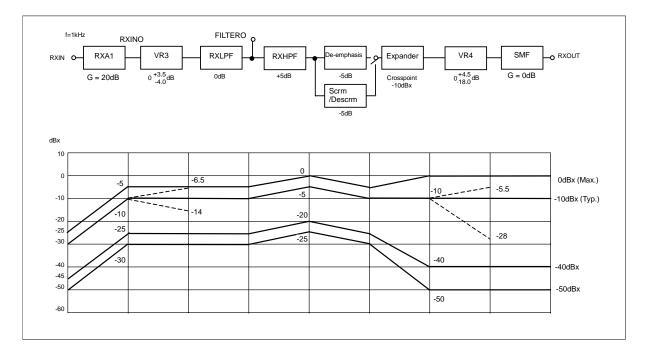


Level Diagram

1) TX system: TXRX=0



2) RX system: TXRX=1



Note) dBx is a standardized notation to match the operating voltage and is defined by equation 0dBx = -5 + 20log(VDD/2)dBm.



Digital AC timing

1) Serial Interface Timing

AK2347A is connected to a CPU by three-wired interface through CSN, SCLK and SDATA pins, which can make reading and writing data for control registers.

Serial data named SDATA is consist of 1-bit read and write instruction(R/W), 4-bit address (A3 to A0) and 8-bit data(D7 to D0) in one frame.

Write mode

CSN		
SCLK		
SDATA (Input)	$ \begin{array}{c} \hline \\ \hline $	
SDATA	Hi-Z	
(Output)		

Read mode

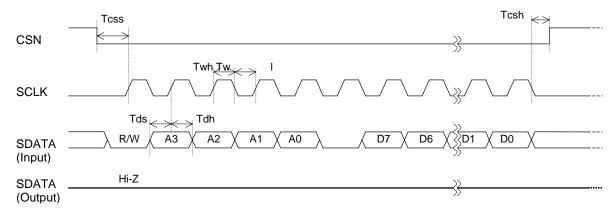
CSN		
SCLK		
SDATA		-
(Input) SDATA (Output)	Hi-Z D7 \ D6 \ D5 \ D4 \ D3 \ D2 \ D1 \ D0	<u> </u>

- R/W : Instruction bit controls to write data to AK2347A or read back from it. When set to low, AK2347A is in write mode. When set to high, AK2347A is in read mode.
- A3 to A0: Register address to be accessed.
- D7 to D0: Write or read date to be accessed.
- (1) CSN(Chip select) is normally selected high for disable. When CSN is set to low, serial interface becomes active.
- In write mode, instruction, address and data input from SDATA pin are synchronized and latched with the rising edge of 14 iterations of SCLK clock. Set to low between address A0 and data D7.
 In read mode, instruction and address are synchronized and latched with the rising edge of 5 iterations of SCLK clock. And the register data are output from SDATA pin synchronized with the falling edge of 9 iterations of SCLK clock. The date between address A0 and data D7 is unstable.
 A CPU port to SDATA pin is fixed to High-Z during the interval that SDATA outputs the read data.
- (3) AK2347A assumes that write and read is set by 14 iterations SCLK clock while CSN sets to Low. If SCLK iterations are less or more than 14 clocks, serial data would not set properly.

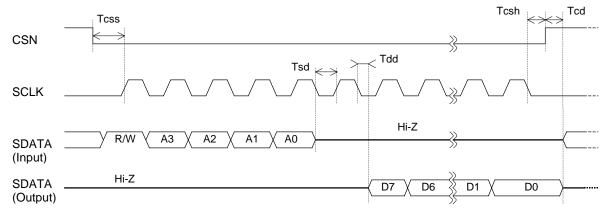


2) Detail Timing Chart

Write mode

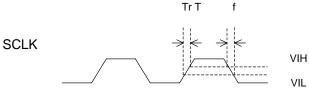


Read mode



f

Rising and falling time



Parameter Symbo	I	Condition	Min.	Тур.	Max.	Unit
CSN setup time	Tcss		100			ns
SDATA setup time	Tds		100			ns
SDATA hold time	Tdh		100			ns
SCLK high time	Twh		500			ns
SCLK low time	Twl		500			ns
CSN hold time	Tcsh		100			ns
SDATA Hi-Z setup time	Tsd		500			ns
SCLK to SDATA	Tdd 20	oF load			500	ns
delay time	100 20				500	115
CSN to SDATA	Tcd 20p	F load		100		ns
delay time				100		115
SCLK rising time	Tr				100	ns
SCLK falling time	Tf				100	ns

Note) In digital input timing, rising time is relative to VIH and falling time is relative to VIL. In digital output timing, rising time is relative to VOH and falling time is relative to VOL.



Register Function Description

	Add	ress	s Da	ta Function								
A3	A2	A1	A0	Function	D7	D6 D5	D4		D3	D2	D1	D0
0	0	0	0	Control register 1	BS3 BS	2	BS1	TXRX	TXSW2	TXSW1	RXSW	MCKSL
0	0	0	1	Control register 2	TC EM		PCONT	SPL	SCSL	LMT	HPF	SASW
0	0	1	0	Volume register 1	VR54 V	R53	VR52	VR51	VR50 V	R12 VR11		VR10
0	0	1	1	Volume register 2	FILSW2	FILSW1	VR25	VR24	VR23 V	R22 VR21		VR20
0	1	0	0	Volume register 3	-	_	-	-	VR33	VR32 V	R31	VR30
0	1	0	1	Volume register 4	_	_	VR45	VR44	VR43 V	R42 VR41		VR40
0 1		1	0	Sub-Audio frequency	_	_	SA5	SA4	SA3 SA	2 SA1		SA0
0 1		1	1	Software-reset & Revision register	-	_	-	RSTN REVNUM[3:0]				
1	0/1 0	V1	0/1	Reserved	Х	ХХХ			Х	Х	Х	х

1) Register configuration

Note1) The mark "-" means that a write to those bits does not have any influence on the LSI operation and read back the writing data.

- Note2) All registers except address 0111 are write and readable registers.
- Caution) Never access the mark "X" test register and unlisted bits of VR33 to VR30, VR45 to VR40 and SA5 to SA1. If an access is made to these addresses inadvertently, the LSI operation is not guaranteed.



2) Descriptions of registers

2.1) Control register 1

	Address Data							
A3	A2	A1	A0	D7 D6 D5 D4 D3 D2			D1	D0
0	0	0	0	BS3 BS2 BS1	TXRX	TXSW2 TX SW1	RXSW	MCKSL
	When power-down			000111			1	1

2.1.1) Operation mode setting

BS3 BS	S2 BS1		Mode name	OSC and AGND system	TX and RX audio system	Sub-Audio system
0	0	0	Mode 0 (Power-down)	OFF OFF OF	F	
0	0	1	Mode 1 (standby)	ON OFF		OFF
010			Mode 2	ON	ON	OFF
011			Mode 3	ON	OFF	ON
1	0/1	0/1	Mode 4	ON	ON	ON

2.1.2) TX and RX setting

Data Ite	n	Func	Remarks		
Data ite		0	1	Remains	
TXRX	TX-RX switch	TX operation *1)	RX operation *2)	*3)	
RXSW	RX audio mute	Mute	Normal operation	*4)	
MCKSL	Master clock frequency	3.579545MHz 3.6864	MHz		

2.1.3) TX path setting

TXSW2	TXSW1	Function	Remarks
11		Mute	
		$(AGND \rightarrow Limiter \rightarrow Splatter)$	
0.1		Audio system operation	
0 1		(HPF \rightarrow Limiter \rightarrow Splatter)	
1.0		External signal operation	
10		(EXTIN pin \rightarrow Limiter \rightarrow Splatter)	
0.0		Audio signal and external signal added together	
0 0		(HPF+EXTIN pin \rightarrow Adder \rightarrow Limiter \rightarrow Splatter)	

*1) When TXRX is set to 0 and RXSW is set to 1, the signal input from the TXIN pin can be output to the RXOUT pin. In this case, because use of the Scrambler/Descrambler is inhibited, be sure to set PCONT to 1.

When RXSW is set to 0, the RXOUT pin output is muted.

- *2) When TXRX is set to 1 and TXSW2 and TXSW1 are set to 0 and 1 respectively, the signal input from the RXIN pin can be output to the MOD pin. In this case, because use of the Scrambler/Descrambler is inhibited, be sure to set PCONT to 1. When TXSW2 and TXSW1 are set to 1 and 1 respectively, the MOD pin output is muted.
- *3) Set the gain level for each circuit block properly according to the level diagrams on page 15.
- *4) If RXSW is set to 0, the FILTERO pin output is not muted.



2.2) Control register 2

Address Data											
A3	A2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	1	тс	EM	PCONT	SPL	SCSL	LMT	HPF	SASW
When power-down			1	1	1	1	0	1	1	1	

Data Ite		Fur	nction	Remarks		
Data ite	2111	0 1		Remarks		
TC	Compandpr	OFF (Bypass)	ON (Active)			
SPL	Splatter Cut-off frequency	2.55kHz 3.0kHz				
SCSL	Scrambler carrier frequency MCKSL is set to 0	3.290kHz	3.390kHz			
	MCKSL is set to 1	3.388kHz 3.388	kHz			
LMT	Limiter	OFF (Bypass)	ON (Active)			
HPF	TX/RX HPF	OFF (Bypass)	ON (Active)			
SASW	Sub-Audio operation	DIN pin \rightarrow TSAOUT pin	RXIN pin \rightarrow RSAOUT pin			

EM PC	ONT	Function	Remarks
1	1	Emphasis: ON (Active), Scrambler/Descrambler: OFF (Bypass)	
0	1	Emphasis: OFF (Bypass), Scrambler/Descrambler:OFF (Bypass)	
0/1	0	Emphasis: OFF (Bypass), Scrambler/Descrambler:ON (Active)	



2.3) Volume Register 1

	Address Data										
A3	A2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
0	0	1	0	VR54	VR53	VR52	VR51	VR50	VR12	VR11	VR10
	When power-down			0	1	1	0	0	1	0	0

VR54 VR53	VR52 VR51	VR50	VR5 gain (dB)
0 0	0 0	0	-6.0
0 0	0 0	1	-5.5
0 0	0 1	0	-5.0
0 0	0 1	1	-4.5
0 0	10	0	-4.0
0 0	10	1	-3.5
0 0	11	0	-3.0
0 0	11	1	-2.5
0 1	0 0	0	-2.0
0 1	0 0	1	-1.5
0 1	0 1	0	-1.0
0 1	0 1	1	-0.5
0 1	10	0	0.0
0 1	10	1	+0.5
0 1	1 1	0	+1.0
0 1	1 1	1	+1.5
1 0	0 0	0	+2.0
1 0	0 0	1	+2.5
1 0	0 1	0	+3.0
10	0 1	1	+3.5
10	10	0	+4.0
10	10	1	+4.5
10	11	0	+5.0
10	11	1	+5.5
11	0 0	0	+6.0

VR12 VR11	VR10	VR1 gain (dB)
0 0	0	-6.0
0 0	1	-4.5
0 1	0	-3.0
0 1	1	-1.5
10	0	0.0
10	1	+1.5
11	0	+3.0
11	1	+4.5



2.4) Volume Register 2

	Address Data										
A3	A2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
0	0	1	1	FILSW2	FILSW1	VR25	VR24	VR23	VR22	VR21	VR20
	When power-down			1	1	1	1	0	0	0	0

FILSW2 F	IL SW1	Function	Remarks
1	1	FILTERO pin output is muted.	
0	1	RXLPF circuit signal is output on FILTERO pin.	
0/1	0	TX/RX_HPF circuit signal is output on FILTERO pin.	

VR25	VR2 gain (dB)
0	-6.4
1	0.0

VR24 VR23		VR22 VR21		VR20	VR2 gain (dB)
0	0	0	0	0	-3.2
0 0		0 0		1	-3.0
0 0		0 1		0	-2.8
0 0		0 1		1	-2.6
0 0		10		0	-2.4
0 0		10		1	-2.2
0 0		11		0	-2.0
0 0		11		1	-1.8
0 1		0 0		0	-1.6
0 1		0 0		1	-1.4
0 1		0 1		0	-1.2
0 1		0 1		1	-1.0
0 1		10		0	-0.8
0 1		10		1	-0.6
0 1		11		0	-0.4
0 1		11		1	-0.2
1	0	0	0	0	0.0
1	0	0	0	1	+0.2
1 0		0 1		0	+0.4
10		0 1		1	+0.6
1 0		10		0	+0.8
1 0		10		1	+1.0
1 0		11		0	+1.2
1 0		11		1	+1.4
1 1		0 0		0	+1.6
1 1		0 0		1	+1.8
1 1		0 1		0	+2.0
1 1		0 1		1	+2.2
1 1		10		0	+2.4
1 1		10		1	+2.6
1 1		11		0	+2.8
1 1		11		1	+3.0



2.5) Volume Register 3

	Add	ress Dat			а							
A3	A2	A1	A0	D7 D6 D5 D4 D3				D2	D1	D0		
0 1		0	0	_	_	_	- VR	33	VR32	VR31	VR30	
	When power-down				_	_	-	1	0	0	0	

VR33 VR32	VR31 VR30	VR3 gain (dB)
0 0	0 0	-4.0
0 0	0 1	-3.5
0 0	10	-3.0
0 0	1 1	-2.5
0 1	0 0	-2.0
0 1	0 1	-1.5
0 1	1 0	-1.0
0 1	1 1	-0.5
10	0 0	0.0
10	0 1	+0.5
10	1 0	+1.0
1 0	1 1	+1.5
1 1	0 0	+2.0
1 1	0 1	+2.5
1 1	1 0	+3.0
1 1	1 1	+3.5



2.6) Volume Register 4

1.0) VOI		egister 4 dress Dat						а			
A3	A2	A1	A0	D7 D6	D5 D4 D3				D2	D1	D0
0 1		0	1	_	- VR45	5	VR44	VR43	VR42	VR41	VR40
	When po	ower-down		_	_	010			0	1	1
	_								t		
VR4	5			42 VR41	VR40					VR4 ga	
0		000								-18	
0		000								-4.	
0		000								-4.2	
0		000								-4.(
0		001								-3.	
0		001								-3.2	
0		001								-3.(
0		010								-2.7	
0		010								-2.	
0		010								-2.2	
0		010								-2.0	
0		011								-1.7	75
0		011	01							-1.	50
0		011	10							-1.2	25
0		011	1 1							-1.(00
0		100	00							-0.	75
0		100	01							-0.	50
0		100								-0.2	25
0		100								0.0	00
0		101								+0.	
0		101								+0.	50
0		101								+0.	
0		101								+1.	
0		110								+1.	
0		110								+1.	
0		110								+1.	
0		110								+2.	
0		111								+2.	
0		111								+2.	
0		<u> </u>								+2.	
1		000								+3.	
1		000								+3.	
1		000								+3.	
1		000								+4.	
1		000								+4.	
1		001								+4.	



2.7) Sub-Audio LPF frequency

2.7) 30											
	Address Dat a										
A3	A2	A1	A0	D7 D6 D5 D4 D3 D2 D1 D0						D0	
0	1	1	0	-	_	SA5	SA4	SA3	SA2	SA1	SA0
	When power-down				_	1	1	0	0	0	1

					MCKSL=0(3	3.579545M) M	CKSL=1(3.686	4MHz)
SA5	SA4 SA3 SA2 S	A1	SA0	Divide	Cut-off	Target	Cut-off	Target
				*1)	Frequency	CTCSS	Frequency	CTCSS
					(Hz)	(Hz)	(Hz)	(Hz)
000		0 0	0	640	59.4	-	61.1	-
000		0 0	1	630	60.3	-	62.1	-
000		0 1	0	620	61.3	-	63.1	-
000		0 1	1	610	62.3	-	64.1	-
000		10	0	600	63.3	-	65.2	-
000		10	1	590	64.4	-	66.3	-
000		11	0	580	65.5	-	67.5	-
000		11	1	570	66.7	-	68.6	-
001		0 0	0	560	67.8	-	69.9	-
001		0 0	1	550	69.1	-	71.1	-
001		0 1	0	540	70.3	-	72.5	67.0
001		0 1	1	530	71.7	67.0	73.8	-
001		10	0	520	73.1	-	75.2	69.3
001		10	1	510	74.5	69.3	76.7	71.9
001		11	0	500	76.0	-	78.3	-
001		11	1	490	77.6	71.9	79.8	74.4
010		0 0	0	480	79.2	-	81.5	-
010		0 0	1	470	80.6	74.4	83.3	77.0
010		0 1	0	460	82.6	77.0	85.1	79.7
010		0 1	1	450	84.4	-	87.0	-
010		10	0	440	86.3	79.7	89.0	82.5
010		10	1	430	88.4	82.5	91.0	85.4
010		11	0	420	90.4	-	93.2	-
010		11	1	410	92.7	85.4	95.5	88.5
011		0 0	0	400	95.0	88.5	97.8	91.5
011		0 0	1	390	97.4	91.5	100.3	-
011		0 1	0	380	100.0	-	103.0	94.8
011		0 1	1	370	102.7	94.8	105.8	97.4
011		10	0	360	105.6	97.4	108.7	100.0
011		10	1	350	108.6	100.0	111.8	103.5
011		11	0	340	111.8	103.5	115.1	107.2
011		11	1	330	115.1	107.2	118.6	110.9



						MCKSL=0(3.579545M) M	CKSL=1(3.686	4MHz)
SA5	SA4 SA	A3 SA2 SA	A 1	SA0	Divide *1)	Cut-off	Target	Cut-off	Target
					1)	Frequency	CTCSS	Frequency	CTCSS
						(Hz)	(Hz)	(Hz)	(Hz)
100			00	0	320	118.8	110.9	122.3	114.8
100			00	1	310	122.5	114.8	126.2	118.8
100			0 1	0	300	126.6	118.8	130.5	123.0
100			0 1	1	290	131.0	123.0	135.0	127.3
100			10	0	280	135.7	127.3	139.8	131.8
100			10	1	270	140.7	131.8	144.9	136.5
100			11	0	260	146.2	136.5	150.5	141.3
100			11	1	250	152.0	141.3	156.5	146.2
101			0 0	0	240	158.3	146.2 151.4	163.0	151.4
101			0 0	1	230	165.2	156.7	170.1	156.7 162.2
101			0 1	0	220	172.7	162.2	177.9	167.9
101			0 1	1	210	180.9	167.9 173.8	186.4	173.8
101			10	0	200	190.0	179.9	195.6	179.9 186.2
101			10	1	190	200.0	186.2 192.8	205.9	192.8
101			11	0	180	211.1	203.5	217.3	203.5 210.7
101			11	1	170	222.9	210.7	229.6	218.1
110			0 0	0	160	237.5	218.1 225.7	244.5	225.7 233.6
110			0 0	1	150	253.2	233.6 241.8	260.9	241.8 250.3
110			0 1	0	140	271.3	250.3 (254.1)	279.5	(254.1) (268.8)
110			0 1	1	130	292.4	(268.8)	300.9	-
110			10	0	120	316.7	-	326.0	-
110			10	1	110	345.3	-	355.8	-
110			11	0	100	379.9	-	391.3	_
110			1 1	1	90	422.2	(403.2)	434.7	(403.2)
111			0 0	0	80	475.0	-	489.0	-
111			0 0	1	70	542.7	-	559.1	-

*1) Divide = 10 x [64 - (register setting value)] This equation states that Divide is divided number of master clock.



2.8) Software reset & revision register

	Add	ress Data									
A3	A2	A1	A0	D7	D6	D5	D4	D3	D2	D1	D0
0	1	1	1	_	_	_	RSTN		REVN	JM[3:0]	
	When po	wer-down		_	_	_	1	0	0	1	0

2.8.1) Software rest

When D4: RSTN data is set to 0, software reset is executed and all register data is set to power-down status and. This register is a write only register and set to 1 automatic after completing software reset.

2.8.2) Revision register

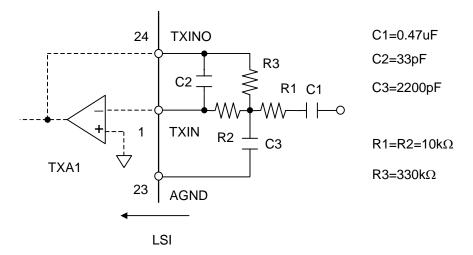
When D3 to D0 data is accessed, users can read the number of mask revision. This register is a read only register.



Recommended External Application Circuits

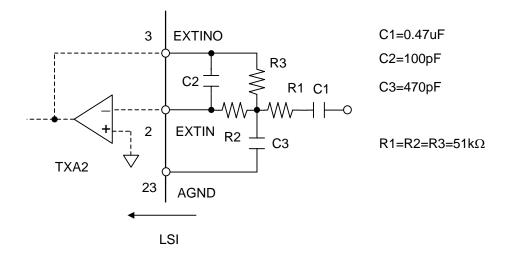
1) TXA1 amplifier

This circuit can be used as the TX microphone amplifier. Set the gain to 30dB or less. If there is a possibility that a high frequency noise component over 100kHz is input, form a first or second order anti-aliasing filter. The following gives a sample configuration of a second order LPF with a gain of 30dB and cut-off frequency of 10kHz:



2) TXA2 amplifier

This amplifier is used for adjusting the gain of the tone signal. Set the gain to 0dB or less. For high frequency noise over 100kHz, form an anti-aliasing filter. The following gives a sample configuration of a second order LPF with a gain of 0dB and cut-off frequency of 13kHz:

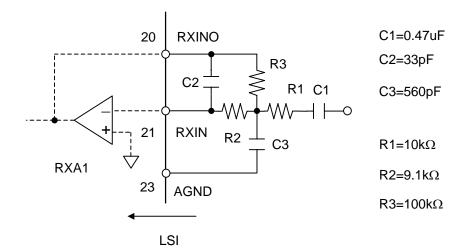






3) RXA1 amplifier

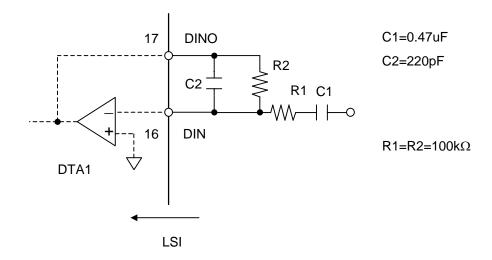
This amplifier is used for adjusting the gain of the RX signal. Set the gain to 20dB or less. For high frequency noise over 100kHz, form an anti-aliasing filter. The following gives a sample configuration of a second order LPF with a gain of 20dB and cut-off frequency of 39kHz:



4) DTA1

amplifier

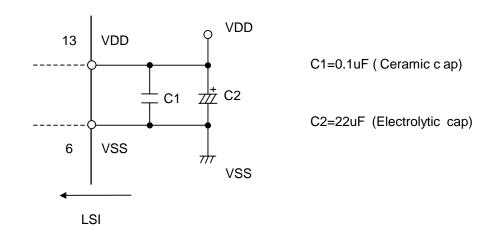
This amplifier is used for adjusting the gain of the signal to Sub-Audio Programmable LPF. Set the gain to 0dB or less. For high frequency noise over 100kHz, form an anti-aliasing filter. The following gives a sample configuration of a second order LPF with a gain of 0dB and cut-off frequency of 7.2kHz:





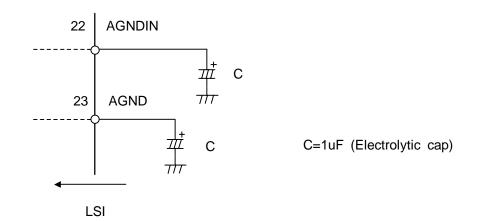
5) Power supply stabilizing capacitors

Connect capacitors between VDD and VSS pins to eliminate ripple and noise included in power supply. For maximum effect, the capacitors should be placed at a shortest distance between the pins.



6) AGND stabilizing capacitors

It is recommended that capacitors with 0.3μ F or lager be connected between VSS and the AGND and AGNDIN pins to stabilize the AGND signal. The capacitors must be placed as close to the pins as possible.





7) Oscillator circuit

When the built-in oscillator circuit is to be used, connect a 3.6864MHz or 3.579545MHz crystal oscillator and a capacitor as shown in Fig. 6. The internal buffer is designed to allow stable oscillation of a crystal oscillator for the electrical equivalent circuitry with a resonance resistance of 150Ω (Max.) and a shunt capacitance of 5pF (Max.).

It is recommended that 22pF capacitors be connected externally so that the total load capacitance is $\underline{16pF}$ (5pF + 22pF//22pF) or less. Place the oscillator, resistor, and capacitors as close to the XIN and XOUT pins as possible.

When an external clock is to be supplied, connect the clock line as shown in Fig. 7 or Fig. 8 according to the clock amplitude level.

The circuit in the first stage of the XIN pin has a constant threshold voltage (0.8V). Therefore, if the high level of the input clock is 1.5V or higher and the low level is 0.4V or lower, connect the clock signal as shown in Fig. 7. If the input clock amplitude (p-p value) is between 0.2V and 1.0V, connect the clock signal as shown in Fig. 8.

When the clock is to be shared with peripheral ICs, the clock must be input and output on the XIN pin. The clock amplitude must not exceed the absolute maximum ratings.

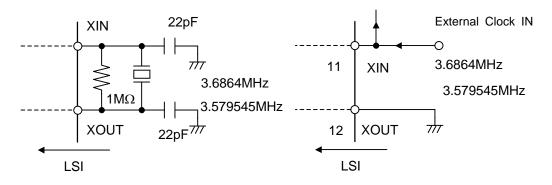


Fig. 6



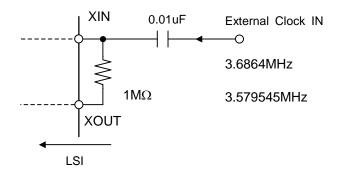


Fig. 8



8) LIMLV pin

The LIMLV pin is used for adjusting the limit level of the limiter circuit. This pin may be left open or may be used by connecting resistors as shown in the figure below.

When the pin is left open, a predetermined limit level can be obtained. The limit level is expressed as follows:

 $HVref = 0.256 \times (VDD - AGND)$ [Vo-p]

For example, let VDD be 3V. The limit level is calculated as follows:

Hvref = $0.256 \times (3.0 - 1.5) = 0.384$ Vo-p Then, 1.5 ± 0.384 V is the typical value of the limit level.

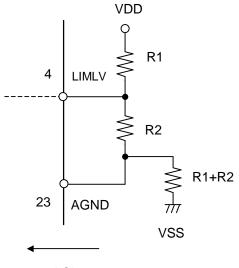
When a DC voltage higher than the AGND voltage level (= 1/2VDD) is applied to the pin through resistors, the limit level can be adjusted. The limit level is the difference between LIMLV and AGND and is expressed as AGND ±(LIMLV - AGND).

Let VDD be 3V. The limit level is calculated as follows:

LIMLV=1.6V	\rightarrow 1	.5 ±0.1V
1.7V	\rightarrow 1.5	±0.2V
1.8V	\rightarrow 1.5	±0.3V
1.9V	\rightarrow 1.5	±0.4V
1.933V	$\rightarrow 1.5$	+0.433V (ed

1.933V \rightarrow 1.5 \pm 0.433V (equivalent to -6.6dBx (Max.)) Then, the above values are obtained as the typical limit levels.

Because AGND level is used as the reference level for the limiter circuit operation as mentioned above, when resistors are connected, they should be connected so that VDD and AGND are separated by these resistors to supply a DC level to the LIMLV pin. In addition, make adjustments so that the sum of resistance (R1 + R2) is around $51k\Omega$.

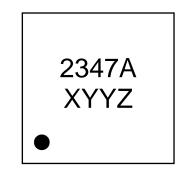


LSI



Packaging

Marking



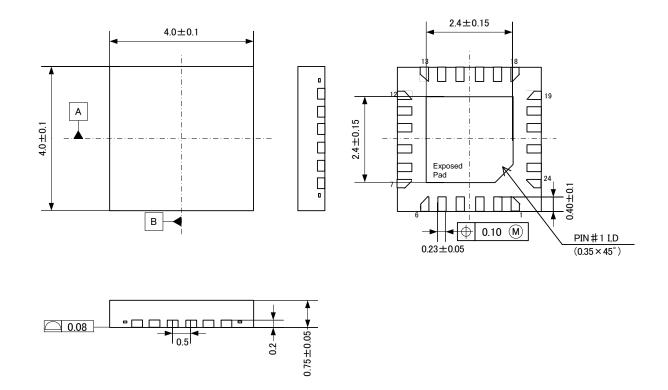
[Contents of XXXYZ]

XXX : Date of manufacturing

Last digit of the year, 2 digits of week number

- Y : Production lot number
- Z : Assembly plant code

• 24-pin QFNJ outline dimension



Note: The exposed pad at the center of the back of the package must be connected to VSS or opened.



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